

Devices Connected/Referenced

ADV7612	Dual Port Xpressview 225 MHz HDMI Receiver
ADV7511	225 MHz, High Performance HDMI Transmitter with ARC

Quad HDMI Input, Fast Switching Multiplexer

Using the **ADV7612** Receiver with Extended Temperature Range

EVALUATION AND DESIGN SUPPORT

Design and Integration Files

[Schematics](#), [Layout Files](#), [Bill of Materials](#)

CIRCUIT FUNCTION AND BENEFITS

The **ADV7612** is a dual port Xpressview™ 225 MHz HDMI® receiver that allows fast switching between two inputs. The circuit shown in Figure 1 shows the use of two **ADV7612**'s as a quad-input fast switching HDMI receiver.

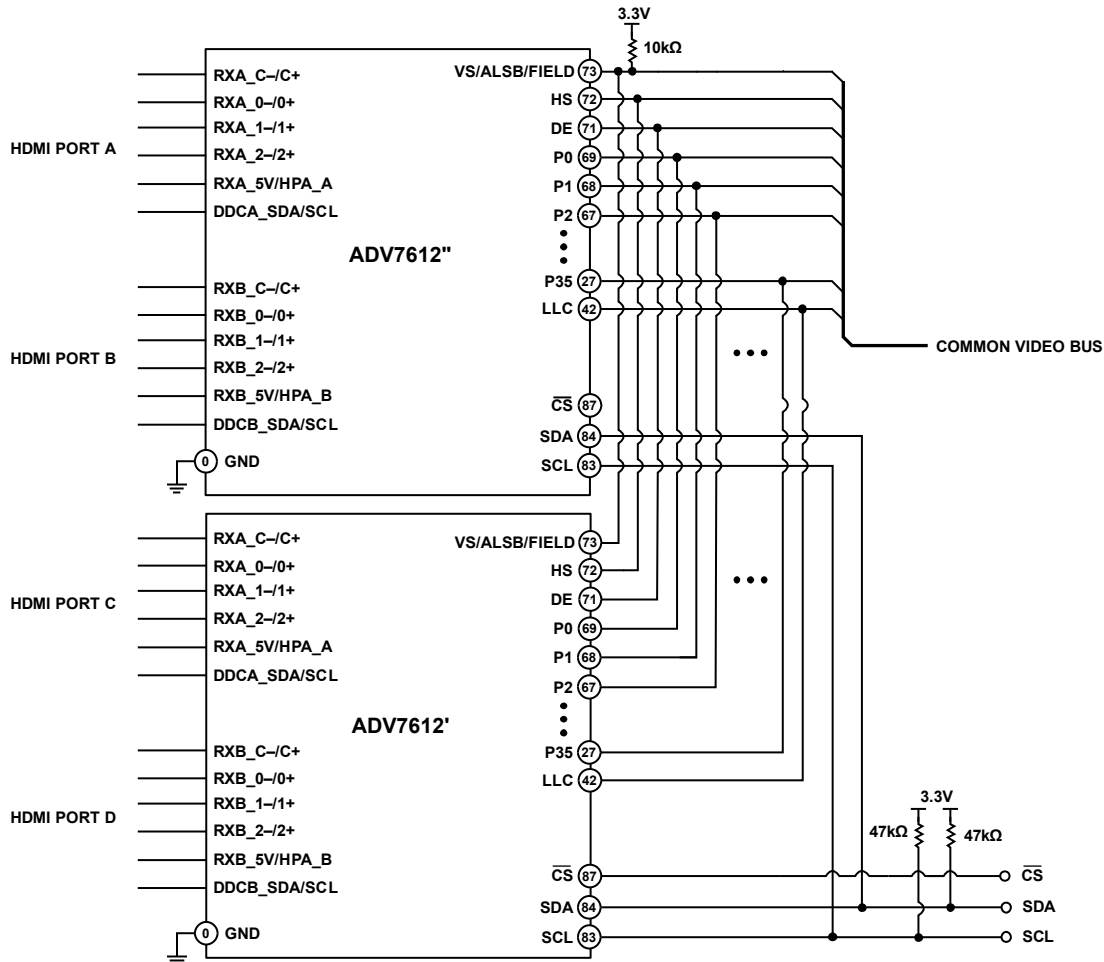


Figure 1. Dual **ADV7612** Circuit (Simplified Schematic: Decoupling, Terminations, Resets, and All Connections Not Shown)

Rev.0

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This circuit shows the expandability of the [ADV7612](#) in applications requiring four multiplexed HDMI inputs of up to 225 MHz TMDS (1080p60, 12 bits per channel; 148.5 MHz LLC pixel clock) or UXGA (1600 × 1200, 10 bits per channel; 162 MHz LLC pixel clock). The circuit offers a cost effective solution to this application and operates over the extended industrial temperature range of -40°C to $+85^{\circ}\text{C}$.

CIRCUIT DESCRIPTION

The [ADV7612](#) provides a receiving solution for two HDMI inputs. Figure 1 shows how to connect two [ADV7612](#)'s in parallel on a common shared video and audio bus, thereby providing multiplexing of four HDMI inputs. How to set up the I²C communications without bus conflicts and how to switch between the sources will be shown. A software package is available showing how to handle communication and authentication in an HDMI repeater application (see <http://ez.analog.com/community/video>).

In order for multiple [ADV7612](#) devices to share the same bus, we need to consider the output state of the devices, capabilities of tri-stating buses, and the electrical parameters of the load on the bus. Additionally, the devices must be controlled from an I²C bus in a non-conflicting manner. The board layout of this circuit is critical and should follow a straight line using controlled impedances to reduce risk of reflections and cross-coupling. Complete PCB layouts are contained in the design support package downloadable at www.analog.com/CN0224-DesignSupport. An [ADV7511](#) HDMI transmitter was used as a backend device.

Bus Output States

After resetting, the [ADV7612](#) tri-states pins P0-P35, HS, VS/FIELD/ALSB, DE, LLC, AP0...AP5, SCLK/INT2, and MCLK/INT2. These pins can be set to the active state using registers TRI_PIX, TRI_SYNCS, TRI_LLC, TRI_AUDIO as described in the UG-216 Hardware User Guide, available at <http://ez.analog.com/docs/DOC-1751>.

Video and Audio Bus Loading

Only one [ADV7612](#) can access the AV buses at a time; the second must remain tri-stated. Assuming an output driver resistance (P0...P35) of 10 Ω to 20 Ω (highest drive strength) and a trace characteristic impedance of 75 Ω , a series resistor of 55 Ω to 65 Ω is required to match the characteristic impedance of the trace. The maximum capacitance of a tri-stated output bus driver on the [ADV7612](#) is 20 pF (refer to Electrical Specifications in the [ADV7612 data sheet](#)).

Layout and Termination Considerations

For this design, it is important to make sure the transmission line is properly terminated and has controlled impedance. Otherwise, reflections (which may occur on longer lines) can have a negative impact on transmitted data.

For pixel lines (P0...P35), video synchronization (VS/FIELD/ALSB, HS, DE), and audio lines (AP0, AP1/12S_TDM, AP2...AP5, MCLK/INT2, SCLK/INT2)—other than LLC—it is suggested to use series termination resistors of 51 Ω at the [ADV7612](#) driver side, and tracks having a characteristic impedance of 75 Ω .

The line locked clock (LLC) line has the same characteristic impedance of 75 Ω and should have no series resistor, but should be terminated at the far end with a symmetrical termination (150 Ω to +3.3 V and to 150 Ω to GND), as shown in Figure 2.

Even though theory suggests a best termination value between 50 Ω and 60 Ω , it was observed during tests that a symmetrical 75 Ω ($2 \times 150 \Omega$) termination increases the swing and centers the signal around mid-supply (1.65 V), which is desirable. The [ADV7511](#) HDMI transmitter is included on the board and is used to transmit the multiplexed output of the two [ADV7612](#)'s.

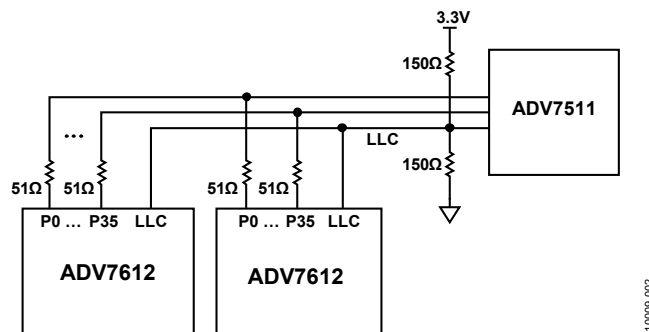


Figure 2. Terminations for P0...P35 Data Lines and LLC Traces

Figure 3 to Figure 6 show waveforms for various terminations. In each case, a symmetrical LLC termination was placed at the far end (close to the [ADV7511](#)), and series termination resistors as close as possible to the two [ADV7612](#) devices, as shown in Figure 2.

Measurements were taken on the [ADV7511](#) pins with Tektronix P6243 FET probes (1 M Ω resistance, 1 GHz bandwidth, less than 1 pF capacitance) and a Tektronix TDS5104B scope.

From the waveforms we can see that using $2 \times 150 \Omega$ terminations on the LLC line ensures a maximum swing of 3.3 V.

Using 75 Ω on the data lines slows the edges too much. 33 Ω and 15 Ω on data lines caused undershoots on falling edges (Figure 5 and Figure 6) and overshoots on rising edges (not illustrated). Therefore, $2 \times 150 \Omega$ was chosen for LLC, and 51 Ω was used on data lines, which is illustrated in the eye diagrams shown in Figure 9 and Figure 10.

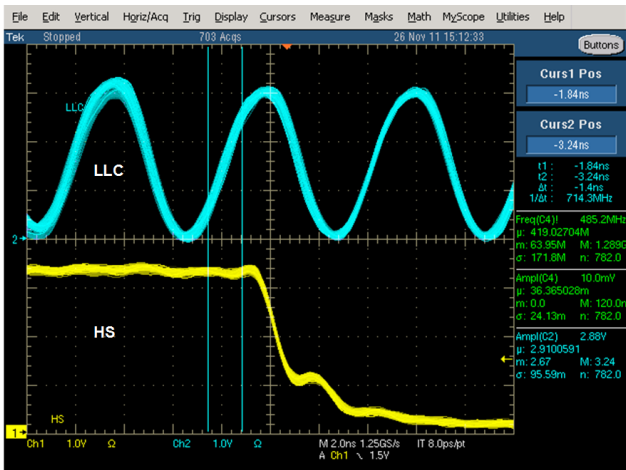


Figure 3. Termination: Symmetrical $2 \times 150 \Omega$ on LLC line, 75Ω on Data Lines (HS). Vertical Scale: 1 V/div, Horizontal Scale: 2 ns/div

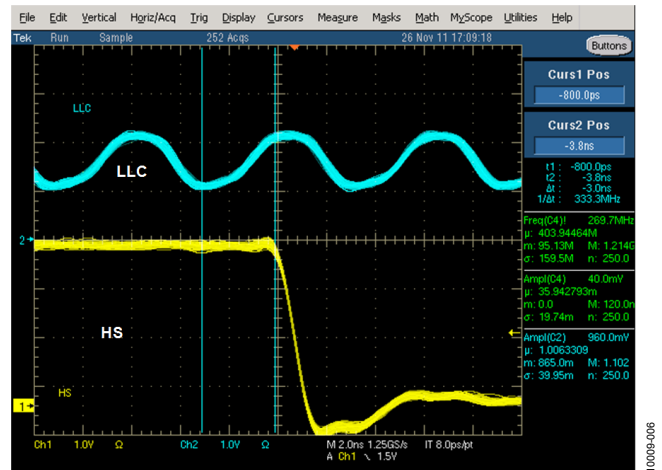


Figure 6. Termination: Symmetrical $2 \times 33 \Omega$ on LLC, Series 15Ω Termination on Data Lines (HS). Note 1 V Undershoot. Vertical Scale: 1 V/div, Horizontal Scale: 2 ns/div

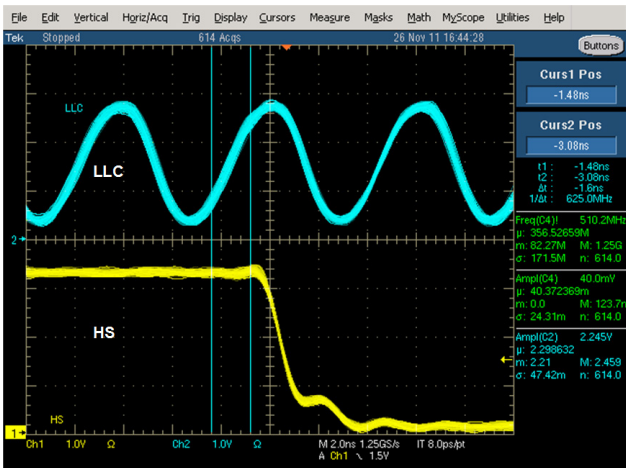


Figure 4. Termination: Symmetrical $2 \times 100 \Omega$ on LLC line, 51Ω on Data Lines (HS). Vertical Scale: 1 V/div, Horizontal Scale: 2 ns/div

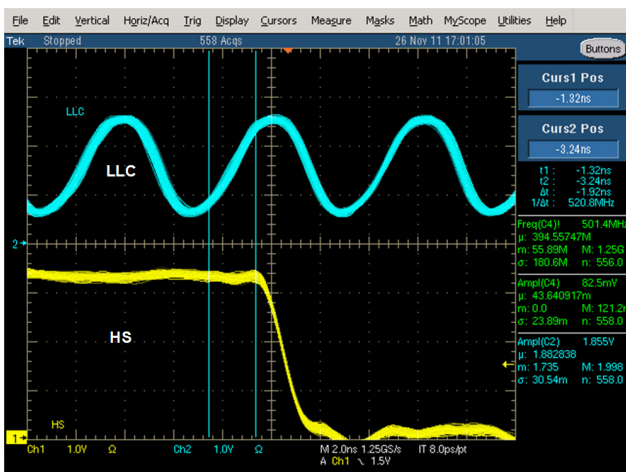


Figure 5. Termination: Symmetrical $2 \times 68 \Omega$ on LLC, Series 33Ω Termination on Data Lines (HS). Note 0.5 V Undershoot. Vertical Scale: 1 V/div, Horizontal Scale: 2 ns/div

I²C Access

After power up, both **ADV7612** devices will have the same I²C address on the main map, which may lead to conflicts. A $\overline{\text{CS}}$ pin is provided on both parts, which allows selecting one of the two devices. When the $\overline{\text{CS}}$ line is pulled low, I²C communication is enabled.

When the $\overline{\text{CS}}$ line is pulled high, I²C communication is disabled.

A simple inverter reduces the resources required on the microcontroller side, as shown in Figure 7.

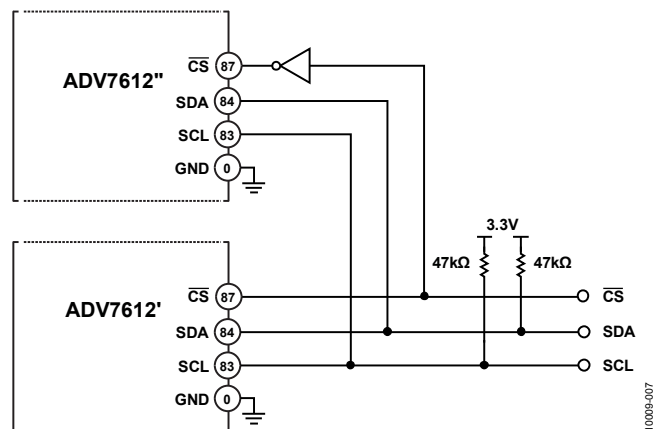


Figure 7. I²C Access

CEC

CEC implementation is not mandatory on the board, and it is up to end user to implement it. If CEC is not desired, CEC pins should be left floating (as described in [UG-216](#), Appendix B: Recommended Unused Pin Configuration). This user guide is included at <http://ez.analog.com/docs/DOC-1751>.

In the other case, separate engines should be used to handle CEC commands.

XTAL_N, XTAL_P

There are two ways of driving the [ADV7612](#) clock. Both parts may have separate crystals connected to XTAL_N and XTAL_P pins, or they can share same signal clock. In the circuit, the 1.8 V signal clock from the oscillator is provided to pins XTAL_P of

both parts. In this configuration, XTAL_N must be left floating. It is critical to ensure proper layout routing and grounding to eliminate coupling between sensitive lines. The length of each trace of the bus should be kept equal.

Interrupts

Interrupts from both devices must be considered. The [ADV7612](#) has two possible interrupts : INT1 (INT1 pin) and INT2 (available through SCLK/INT2, MCLK/INT2 or HPA_A/INT2).

It is advised not to use INT2 via pins MCLK/INT2 or SCLK/INT2, as tri-stating the audio bus with the TRI_AUDIO register will also tri-state these pins.

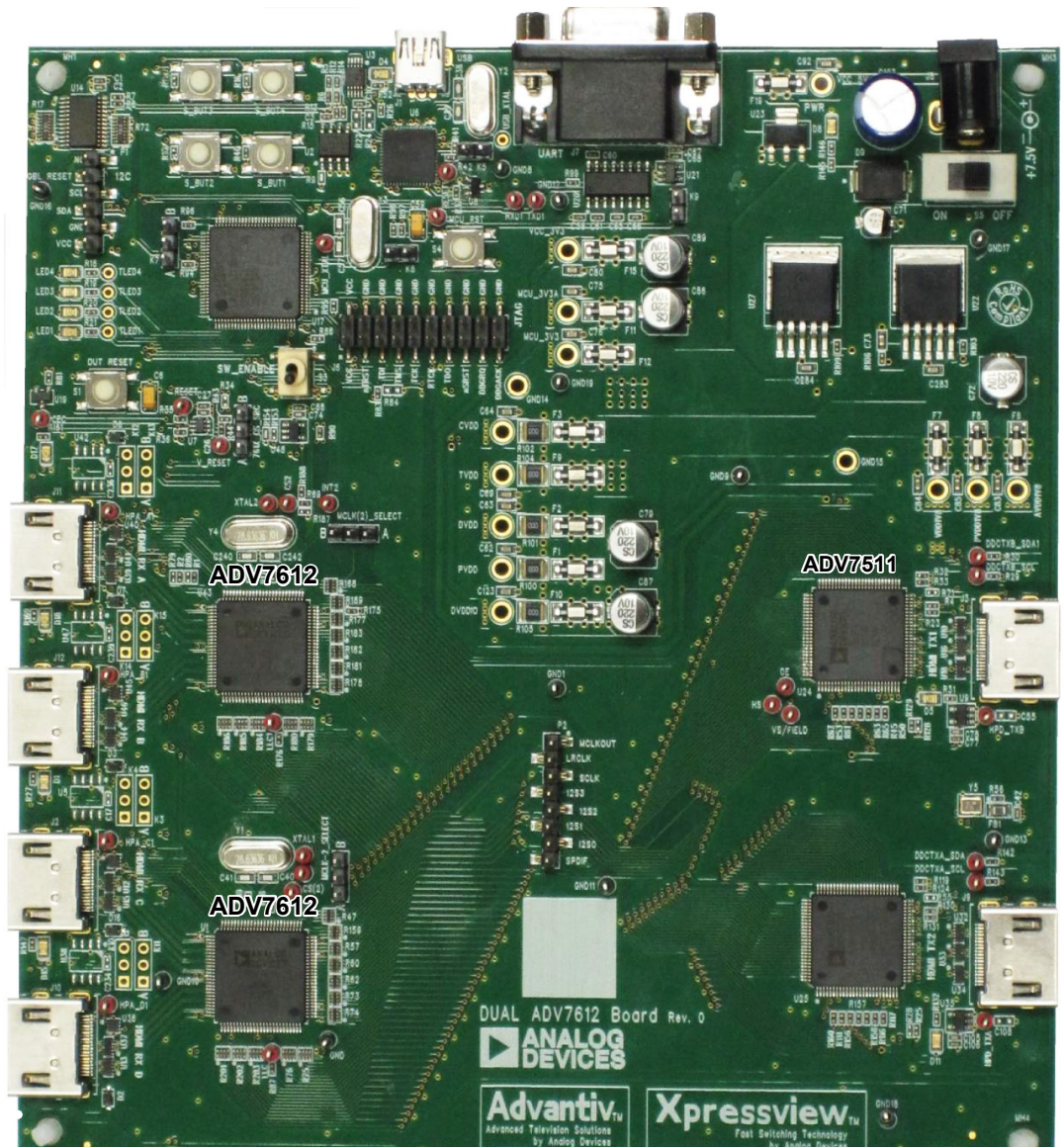


Figure 8. Dual [ADV7612](#) Board Solution with [ADV7511](#)

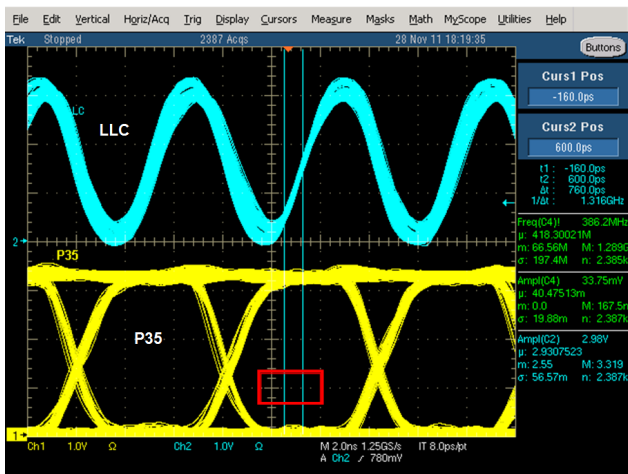


Figure 9. Screenshot from Scope. Signal Driven from **ADV7612-U43**. LLC Line (162 MHz) Measured at **ADV7511**'s Input and Pixel Line P35. Red Rectangle Shows Eye Mask for **ADV7511**. $2 \times 150 \Omega$ Symmetrical Termination on LLC and 51Ω Series Resistors on Data Lines.

Vertical Scale: 1 V/div, Horizontal Scale: 2 ns/div

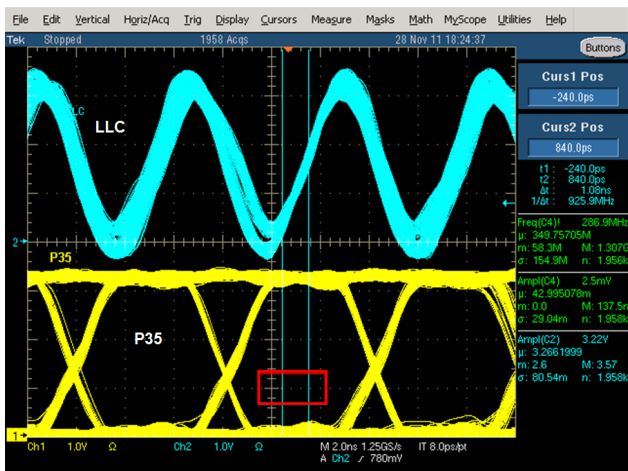


Figure 10. Screenshot from Scope. Signal Driven from **ADV7612-U1**. LLC Line (162 MHz) Measured at **ADV7511**'s Input and Pixel Line P35. Red rectangle Shows Eye Mask for **ADV7511**. $2 \times 150 \Omega$ Symmetrical Termination on LLC and 51Ω Series Resistors on Data Lines.

Vertical Scale: 1V/div, Horizontal Scale: 2ns/div

Layout Considerations

Layout should consist of very short traces. In the ideal case, traces connecting two pins of the same function between the two **ADV7612**'s should be as short as possible and should share a common series termination resistor placed as close as possible to both devices and then connected to the bus. In practice, this is not possible due to layout constraints; therefore, each device requires its own series termination resistor (see Figure 2). The video traces should be kept as close to the same length as possible for delay matching.

Evaluation and Test

The circuit was evaluated using two video generators (Quantum Data 882) to generate UXGA 1600×1200 pixels, 30-bits, and 1080p60, 36 bits (Samsung2 and MoirèX patterns). As an HDMI sink (output from **ADV7511**), an Astro VA-1831 video analyzer was used. Additionally, video signals from the **ADV7612** (LLC and P35) were observed on the **ADV7511** pins with a P6243 (1 pF, 1 M Ω , 1 GHz) probe attached to a Tektronix TDS5104B oscilloscope. The resulting waveforms are presented as eye diagrams with **ADV7511** eye masks and are shown in Figure 9 and Figure 10 for UXGA (162 MHz, 30 bits) with the Samsung2 pattern. The MoirèX pattern used during tests also showed a similar safety margin.

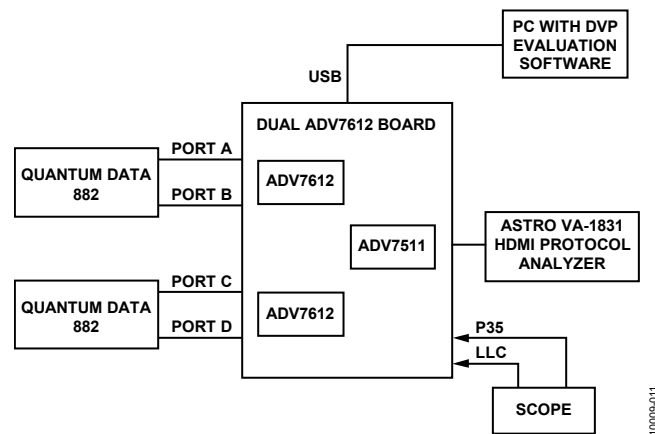


Figure 11. Test Setup

Video timing measured by the Astro VA-1831 showed no anomalies. Even and odd vertical lines of MoirèX were analyzed (1080p60 36-bit and UXGA 30-bit), and it showed that all bits toggle properly at the same time without any leakage between lines (MoirèX pattern). The Astro VA-183 also showed stable HDMI synchronization signals and packets having correct CRC checksum. This indicates properly received clock and synchronization information by the backend **ADV7511**.

Test Steps

1. Prepare test configuration as shown in Figure 11.
2. Power up board and measuring equipment; start DVP Evaluation Software.
3. In DVP Evaluation Software, load **ADV7612** board.
4. Run INIT_PARTS_AND_SET_PORT_A_ver4.py script in DVP Evaluation Software (refer to design resources).
5. Press Init button. Once board is initialized, click one of the buttons: Port A, Port B, Port C, or Port D to select desired input (see Figure 12).

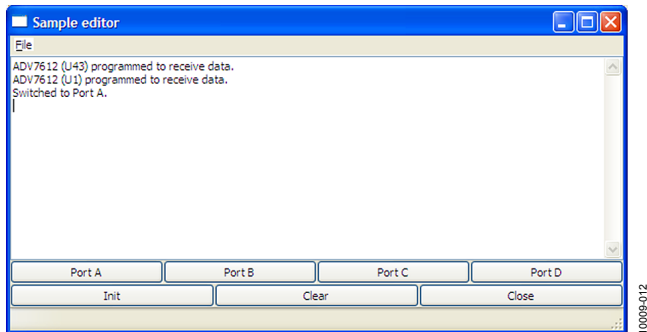


Figure 12. Software Used for Selecting HDMI Input

LEARN MORE

CN-0224 Design Support Package:

www.analog.com/CN0224-DesignSupport

ADV7612 Design Support Files on Engineer Zone:

<http://ez.analog.com/docs/DOC-1751>

Ardizzoni, John. *A Practical Guide to High-Speed Printed-Circuit-Board Layout*, Analog Dialogue 39-09, September 2005.

MT-031 Tutorial, *Grounding Data Converters and Solving the Mystery of "AGND" and "DGND"*, Analog Devices.

MT-101 Tutorial, *Decoupling Techniques*, Analog Devices.

Howard Johnson, Martin Graham, *High-Speed Digital Design*, Prentice Hall, ISBN-10: 0133957241, ISBN-13: 978-0133957242.

Howard Johnson, Martin Graham, *High Speed Signal Propagation*, Prentice Hall, ISBN-10: 013084408X, ISBN-13: 978-0130844088.

Data Sheets and Evaluation Boards

ADV7612 Data Sheet.

ADV7511 Data Sheet.

UG-216, ADV7612 Hardware User Guide:

<http://ez.analog.com/docs/DOC-1751>

REVISION HISTORY

12/11—Revision 0: Initial Release

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