

# 12-Bit Video **Analog-to-Digital Converter**

**FFATURES** 12-Bit Resolution 20MHz Word Rate Single 35-In<sup>2</sup> PC Board **ECL Compatible No External Support Circuits** 

**APPLICATIONS** Radar Digitizing **Medical Instrumentation** Digital Signal Processing Spectrum Analysis

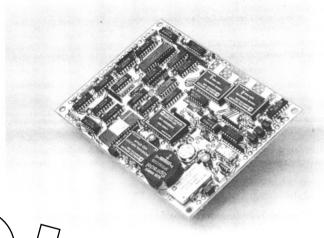
ransient Analysis

Analog Devices model CA A/D converter is an outstanding combination of 12-bit resolution and small size. The unit is capable of solving a multitude of high-speed digitizing problems. Its design is based on concepts pioneered in the MOD-1020 and MOD-1205 and taken to an even higher level of achievement in CAV-1210.

It is pin-for-pin compatible with the other units in the MOD and CAV series of A/D converters. But it doubles the word rate of its predecessor CAV-1210, making it possible for system designers to offer options or upgrade their high-resolution systems without new layouts.

This remarkable converter includes a track-and-hold, along with encoding and timing circuits. The CAV-1220 is an ideal choice for the designer who needs state-of-the-art performance in highresolution, ultra-high speed A/D conversion.

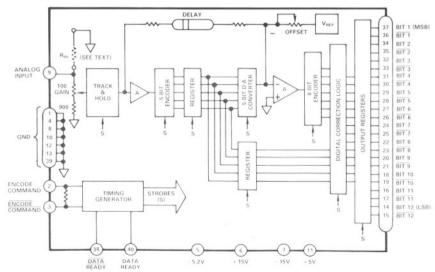
For radar applications, 12-bits of resolution increase the dynamic



the converter, making it possible to detect weaker signals than would be possible with lower resolution characteristics. The high-word rates enhance ranging resolution, thereby increasing system effe

tions, the CAV-1220 incre and/or color resolution of systems in which it is used. Its high-word rates increase spatial resolution, and this combination of high resolution and high speed can materially improve performance.

All digital inputs and outputs are ECL compatible; optimum analog input impedance can be selected by the user. The unit requires only an encode command and external power supplies for operation. The CAV-1220 is repairable and backed by Analog Devices' limited one-year warranty.



CAV-1220 Block Diagram

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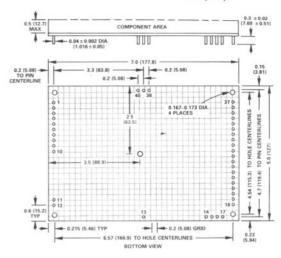
Texas 214/231-5094

# **SPECIFICATIONS** (typical @ +25°C with nominal power supplies unless otherwise noted)

Parameter  RESOLUTION (FS = Full Scale)  LSB WEIGHT 2.048V p-p FS  ACCURACY (Including Linearity) @ dc Monotonicity Nonlinearity vs. Temperature Offset vs. Temperature Gain Error Adjustable to Zero with On-Card Potentic Gain vs. Temperature	Units Bits (% FS)  mV  % FS ± 1/2LSB  ppm/°C (max)  ppm/°C (max)	CAV-1220 H 12(0.024) 0.5 0.0125 Guaranteed 10(15)
LSB WEIGHT 2.048V p-p FS  ACCURACY (Including Linearity) @ dc Monotonicity Nonlinearity vs. Temperature Offset vs. Temperature Gain Error Adjustable to Zero with On-Card Potentic	mV  % FS ± 1/2LSB  ppm/°C (max)  ppm/°C (max)	0.5 0.0125 Guaranteed
2.048V p-p FS  ACCURACY (Including Linearity) @ dc Monotonicity Nonlinearity vs. Temperature Offset vs. Temperature Gain Error Adjustable to Zero with On-Card Potentic	% FS ± 1/2LSB  ppm/°C (max)  ppm/°C (max)	0.0125 Guaranteed
ACCURACY (Including Linearity) @ dc Monotonicity Nonlinearity vs. Temperature Offset vs. Temperature Gain Error Adjustable to Zero with On-Card Potentic	% FS ± 1/2LSB  ppm/°C (max)  ppm/°C (max)	0.0125 Guaranteed
(Including Linearity) @ dc Monotonicity Nonlinearity vs. Temperature Offset vs. Temperature Gain Error Adjustable to Zero with On-Card Potentic	ppm/°C (max) ppm/°C (max)	Guaranteed
Monotonicity Nonlinearity vs. Temperature Offset vs. Temperature Gain Error Adjustable to Zero with On-Card Potentic	ppm/°C (max) ppm/°C (max)	Guaranteed
Nonlinearity vs. Temperature Offset vs. Temperature Gain Error Adjustable to Zero with On-Card Potenti	ppm/°C(max)	
Gain Error Adjustable to Zero with On-Card Potentic		
Adjustable to Zero with On-Card Potentic		220 (250)
	%FS	2
Gain vs. Temperature		150
	ppm/°C, max	130
DYNAMIC CHARACTERISTICS In-Band Harmonics <sup>1</sup>		
540kHz Input	dB Below FS, min	70
2.3MHz Input	dB Below FS, min	65
9.3MHz Input	dB Below FS, min	50
Conversion Time <sup>2</sup>	ns (max)	$1 \operatorname{Clock} \operatorname{Period} + 155 \operatorname{ns} (\pm 10)$
Aperture Uncertainty (Jitter)	MHz, max ps, rms max	20 25
Effective Aperture Delay Time <sup>3</sup>	ns (max)	25/+25)
Signal to Noise Ratio (SNR) <sup>4</sup>	dB, (min)	66(65) 64(63)
Noise Power Ratio (NPR)5	dB	52
Transient Response	(is	100
Overvoltage Recovery?  Input Bandwidth	nb un	200
Small Signal, JdB <sup>8</sup>	MHz	40
Large Signal, 3dB <sup>9</sup>	MHz	35
Two-Tone Linearity (@ Input Frequencies)10		11//
(60kHz; 62kHz)	dB Below FS	70 / / /
(2.496MHz; 2.498MHz)	dB below FS	65
(4.996MHz; 4.998MHz)	dB below FS	00
ANALOG INPUT Voltage Range <sup>11</sup>		
Operating	V, FS	± 1.024
Maximum Without Damage	V, max	± 2
Input Type		Bipolar
Impedance	Ω	1000
Offset12	mV	Adjustable to Zero with On-Card Potentiometer
ENCODE COMMAND INPUT <sup>13</sup>		On-Card Fotentionicter
Logic Levels, ECL-Compatible	V	"0" = -1.7
(Balanced Input)	v	"1" = $-0.9$
Impedance	$\Omega$ , max	100
Rise and Fall Times	ns, max	5
Width Min		10
Max	ns 70% of Encode Command	10 d Period
Frequency <sup>14</sup>	MHz	dc to 20
DIGITAL OUTPUT		
Format	Data Bits	12 Parallel; NRZ
	Data Ready and	
	Data Ready	2; RZ
Logic Levels, ECL-Compatible (Balanced Output)	V	"0" = -1.7
Drive (Line-to-Line)	V Ω, min	"1" = -0.9
Time Skew	ns, max	5
Coding	,	Binary (BIN);
		2's Complement (2SC)
DATA READY OUTPUT		
Logic Levels, ECL-Compatible	V	"0" = $-1.7$
(Balanced Output)		"1" = -0.9
Drive (Line-to-Line) Rise and Fall Time	Ω, min	75 5
Duration	ns, max ns (max)	22 (±3)
POWER REQUREMENTS <sup>15</sup>	113 (IIIIA)	22(-3)
+ 15V ± 5%	mA (max)	174 (192)
	mA (max)	157 (173)
$-15V \pm 5\%$	mA (max)	174(192)
+ 5V ± 5%		2.78(3.06)
+ 5V ± 5% - 5.2V ± 5%	A (max)	
+ 5V ± 5% - 5.2V ± 5% Power Consumption		20.3 (22.3)
+ 5V ± 5% - 5.2V ± 5% Power Consumption TEMPERATURE RANGE	A(max) W(max)	20.3 (22.3)
+ 5V ± 5% - 5.2V ± 5% Power Consumption TEMPERATURE RANGE Operating	A (max) W (max)	20.3 (22.3) 0 to +70
+ 5V ± 5% - 5.2V ± 5% Power Consumption TEMPERATURE RANGE	A(max) W(max)	20.3 (22.3) 0 to +70 -55 to +85
+ 5V ± 5% - 5.2V ± 5% Power Consumption TEMPERATURE RANGE Operating Storage	A (max) W (max) °C °C	20.3 (22.3) 0 to +70 -55 to +85 500
+ 5V ± 5% - 5.2V ± 5% Power Consumption TEMPERATURE RANGE Operating Storage Cooling Air Requirements	A (max) W (max)  °C °C LFPM	20.3 (22.3) 0 to +70 -55 to +85 500
+ 5V ± 5% - 5.2V ± 5% Power Consumption TEMPERATURE RANGE Operating Storage Cooling Air Requirements	A (max) W (max)  °C °C LFPM	20.3 (22.3) 0 to +70 -55 to +85 500
+ 5V ± 5% - 5.2V ± 5% Power Consumption  TEMPERATURE RANGE Operating Storage Cooling Air Requirements  CONSTRUCTION	A (max) W (max)  °C °C LFPM (Linear Feet Per Minute)	20.3 (22.3) 0 to +70 -55 to +85 500

#### **OUTLINE DIMENSIONS**

Dimensions shown in inches and (mm).



## PIN DESIGNATIONS

-	$\sim$		
PIN	FUNCTION	PIN	FUNCTION
	GROUND	40	DATA READY
2	ENCODE COMMAND	39	GROUND
3	ENCODE COMMAND	38	DATA READY
4	GROUND	37	B/T 1 (MSB)
5	-5.2V	36	BIT 1 (MSB)
6	+ 15V	35	BIT 2
7	- 15V	34	BIT 2
8	GROUND	33	BIT 3
9	ANALOG INPUT	32	BIT 3
10	GROUND	31	BIT 4
11	+5V	30	BIT 4
12	GROUND	29	BIT 5
13	GROUND	28	BIT 5
14	BIT 12 (LSB)	27	BIT 6
15	BIT 12 (LSB)	26	BIT 6
16	BIT 11	25	BIT 7
17	BIT 11	24	BIT 7
18	BIT 10	23	BIT 8
19	BIT 10	22	BIT 8
20	BIT 9	21	BIT 9

#### NOTES

<sup>1</sup>In-Band Harmonics expressed in terms of spurious in-band signals and harmonics generated at 20MHz encode rate.

<sup>2</sup>Measured from leading edge Encode Command to trailing edge Data Ready; use trailing edge to strobe output data into external circuits

See text for description of Effective Aperture Delay Time

<sup>4</sup>Rms signal to rms noise ratio with full-scale 540kHz analog input (see Figure 3)

<sup>5</sup>Dc to 8.2MHz white noise bandwidth with slot frequency of 3.886MHz; and encode rate of 20MHz.

<sup>6</sup>For full-scale step input, 12-bit accuracy attained in specified time

Recovers to 12-bit accuracy in specified time after 2 × FS input overvoltage.

<sup>8</sup>With analog input 40dB below FS.

With FS analog input. (Large-signal bandwidth flat within 0.2dB, dc to 10MHz).

<sup>10</sup>Both frequencies applied at level 7dB below full scale

"Standard bipolar input is adjustable +5% with on-card potentiometer (see text and Figure 2). Unipolar 0 to +2V input range is available on

special order; consult factory for details.

12 Adjustable ± 15mV without performance degradation (see text and Figure 2).

13 Digital "0" to digital "1" transition initiates encoding

14Encode rate specified by customer; see Ordering Information. Units operated outside ± 10% of specified frequency (up to maximum 20MHz) must be returned to factory for recalibration. For operation at word rates below 500kHz, consult factory.

15 ± 15V must be equal and opposite within 200mV and track over temperature.

Specifications subject to change without notice.

For Applications Help, Call Computer Labs Division @ (919) 668-9511.

#### THEORY OF OPERATION

Refer to the block diagram of the CAV-1220.

Analog input signals to be digitized are applied to a track-and-hold (T/H) amplifier, which is normally operating as a buffer amplifier in the "track" mode, following all changes in analog input as they occur. The user of the CAV-1220 determines the point at which the analog signal is to be digitized by applying an Encode Command.

The leading edge of the ECL-compatible encode command causes the track-and-hold to switch momentarily to the "hold" mode of operation, "freezing" the analog input signal long enough to begin the digitizing process. The instant this switching action occurs is affected by one of the parameters of the CAV-1220, called our as Effective Aperture Delay Time in the Specifications table.

Basically, effective aperture delay time is a measure of the difference between the converter's digital and analog delays  $(t_d-t_a)$  and can assume a zero, positive, or negative value depending on the comparative lengths of the two delays. In the CAV 1220, the analog delay  $(t_a)$  is less than the digital delay  $(t_d)$ , and causes effective aperture delay to be typically  $\lambda 5 \, \mathrm{ns}$ .

The "held" value of analog signal at the output of the TH is applied to a 5-bit encoder. It is also applied to an analog delay circuit, whose time delay is equal to the interval required for the first step of the digitizing/reconstruction process.

The digitized signal is applied to a 5-bit D/A converter which has 12-bit accuracy. Via registers, the same digital signal is directed to the digital correction logic circuits. The stored data will represent Bits 1–5 of the 12-bit digital output of the CAV-1220.

The reconstructed output of the D/A converter becomes one input to an operational amplifier; its other input is the delayed analog signal from the delay line. The output of the wideband, fast-settling op amp represents the residue which remains after a 5-bit digital representation of the analog input has been subtracted from that input.

This residue, or error, signal is encoded by a second encoder and is applied as 8-bit digital information to the digital correction logic circuits which contain Bits 1–5.

The correction circuits combine the 5-bit and 8-bit bytes of data to compensate for possible nonlinearities and other errors to assure the final 12-bit output of the CAV-1220 is 12-bit accurate.

Expressed in its simplest terms, the digital correction logic circuits use the information in the 8-bit signal to determine what modifications of Bits 1-5 may be necessary. The value of the MSB in the 8-bit byte establishes whether the 5-bit data are passed "as is" or whether they are increased by a value of binary "1". The remaining bits (2-8) of the 8-bit byte become Bits 6-12 of the CAV-1220 digital output.

Digitally corrected subranging (BCS), the innovative technique described here, helps compensate for a wide range of potential errors which could otherwise be avoided only if the CAV-1220 design included expensive, high precision components.

The use of 13 bits to obtain an accurate 12 bits of output cannot prevent gain error, track/hold droop error, linearity error, offset error, or any of the other inherent characteristics of "real-world" A/D converters. But DCS can, and does help nullify their effects and makes it economically feasible to accomplish high speed, high-resolution digitizing of analog signals.

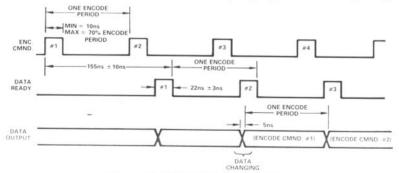


Figure 1. CAV-1220 Timing Diagram

#### CAV-1220 TIMING

Refer to Figure 1, the CAV-1220 Timing Diagram.

The intervals shown represent a continuous update rate of approximately 10MHz, which is considerably below the maximum capabilities of the CAV-1220. But that frequency helps illustrate the "pipeline delay" characteristic of the converter.

At this word rate, spacing between encode commands is approximately 100 nanoseconds; and three encode commands have occurred before the data associated with the first command are valid. In Figure 1, this pipeline delay has a total time of approximately 255 nanoseconds (155ns + 100ns). This interval will be different at other word rates, but will always include 155ns; depending upon the update rate, either more or fewer encode commands may occur before the first data are available.

After the initial delay, valid data will be available at the word rate dictated by encode commands. Note the spacing between Encode Command #1 and Encode Command #2 is equal to one encode period. This is the same spacing as that between Data Ready #1 and Data Ready #2; and is also the spacing between the first and second groups of valid data.

System timing can be adjusted as necessary to take into account the pipeline delay effects and assure that the data of interest are strobed out of the converter at the appropriate time.

Figure 1 also illustrates why the trailing edge of the Data Ready pulse is recommended as the strobe for output data. Typically, data begin changing 5ns after the leading (rising) edge of each Data Ready pulse; they will be fully settled at the time of the trailing (falling) edge and available for use in external circuits.

Another possibility for strobing the output data is to use the  $\overline{DATA}$  READY pulse. Its leading edge occurs at the same time as the trailing edge of the DATA READY signal, but is a rising edge, which may facilitate its use as a strobe.

#### ANALOG INPUT IMPEDANCE

Refer again to the block diagram of the CAV-1220 and note the resistor shown in dashed lines and designated as  $R_{\rm IN}$ .

This resistor value is chosen by the user to allow the analog input impedance of the CAV-1220 to be matched to the characteristic impedance of the analog signal source.

Without an added resistor, the input impedance of the unit is 1,000 $\Omega$ ; this is the series total of the GAIN control and the 900 $\Omega$  resistor shown in the block diagram.

When a resistor is added, it is in parallel with the internal impedance of the CAV-1220; various values of resistors can be used to obtain standard impedances:

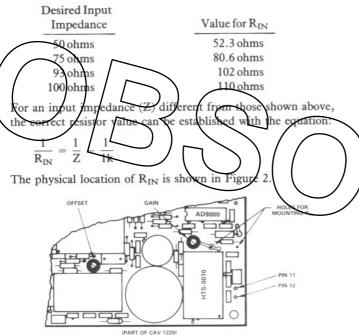


Figure 2. CAV-1220 Adjustment Controls

### OFFSET AND GAIN ADJUSTMENTS

The design and manufacture of the CAV-1220 A/D converter are innovative and precise, and have resulted in a high-performance converter which is virtually adjustment-free. This elimination of variable controls helps make the unit less susceptible to performance degradation caused by vibration, shock, or inadvertent and/or incorrect adjustment.

Despite the complexity of the circuits required to obtain highresolution digitizing at high speeds, there are only two control settings used in the unit. Factory adjustments during final calibration use selected fixed resistors to assure optimum performance without a need for "tweaking" by the user.

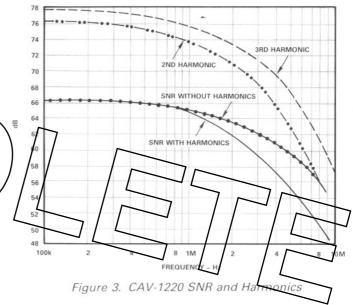
Only OFFSET and GAIN controls are available, and even these are sealed at the factory before shipment. In those rare instances where they may require readjustment, the procedure outlined below is one which should be used.

Refer to Figure 2, the CAV-1220 Adjustment Controls.

When adjusting offset and gain of the CAV-1220 in the system, the OFFSET control should be adjusted first. The adjustment sequence is:

- 1. Apply to the analog input a precise ( $\pm 0.25 \text{mV}$ ) dc level corresponding to midscale of the desired input range. (For standard units with  $\pm 1 \text{V}$  range, this is 0 V input.)
- 2. Adjust OFFSET control while observing MSB (Bit 1); adjust for MSB "toggling" between digital "0" and digital "1".

- Apply a precise (±0.25mV) dc level corresponding to the most negative excursion of the desired input range. (For standard units, this is -1V input.)
- 4. Adjust GAIN control while observing LSB (Bit 12); adjust for output of Bits 1–11 solid "0" with LSB "toggling".
- 5. Apply a precise  $(\pm 0.25 \text{mV})$  dc level corresponding to the most positive excursion of the desired input range. (For standard units, this is +1V input.)
- Check digital output to assure Bits 1-11 are solid "1" with LSB "toggling".
- Adjust OFFSET and GAIN controls alternately as necessary to obtain analog input range to tolerance of ±1/2LSB.



#### DYNAMIC PERFORMANCE

Figure 3 shows typical performance on some of the dynamic characteristics which play an important role in the performance of systems using the CAV-1220 A/D converter.

The A/D was calibrated in final test for an encode rate of 20MHz. As shown, signal-to-noise ratio (SNR) with harmonics is typically 66dB at an input frequency of 100kHz; and remains greater than 50dB for full-scale inputs of 8MHz. As expected, SNR without harmonics is better and is typically 56dB at 8MHz.

The level of 2nd and 3rd harmonics at a word rate of 20MHz is also depicted; in these characteristics, too, the CAV-1220 displays exceptional performance.

### ORDERING INFORMATION

For standard CAV-1220 units, order by model number CAV-1220-XXX; XXX is specified by the customer to indicate the desired optimized word rate. The decimal place is assumed (but not shown) between the second and third places. CAV-1220-150, for example, indicates final calibration and optimum performance at 15MHz.

Optimum performance will be achieved within a band of frequencies approximately  $\pm 10\%$  around the selected word rate; but the maximum rate of 20MHz must be considered. If later applications require word rates beyond the limits of the original optimum frequency, the unit must be returned to the factory for calibration; there is a nominal charge for this service.

Mating sockets for the CAV-1220 converters are model number MSB-2 (thru hole) or MSB-3 (closed end). These are individual solder-type pin sockets for mounting the A/D on PC boards.