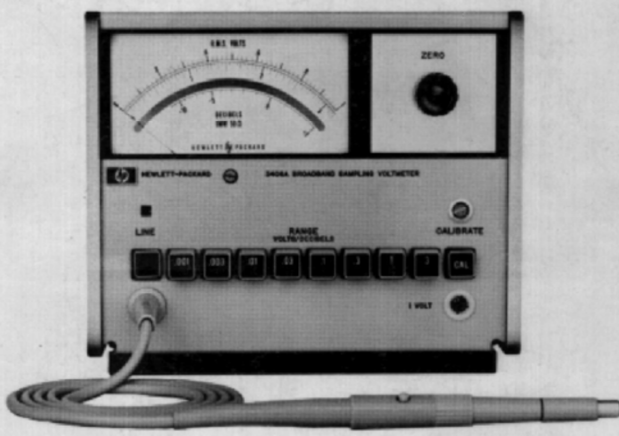
HP 3406A

OPERATING AND SERVICE MANUAL

BROADBAND SAMPLING VOLTMETER

3406A



HEWLETT hp PACKARD

HP 34

OPERATING AND SERVICE MANUAL

(HP Part NO. 03406-90002)

MODEL 3406A BROADBAND SAMPLING VOLTMETER

SERIALS PREFIXED: 942-

APPENDIX C, Manual Backdating Changes, adapts manual to serials prefixed: 841-, 830-, 625-, 606-.

P.O. Box 301, Loveland, Colorado, 80537 U.S.A.

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TABLE OF CONTENTS

TABLE OF CONTENTS (Cont'd)

LIST OF TABLES

LIST OF ILLUSTRATIONS

Table 1-1. Specifications

Voltage Range:
1 mV to 3 volts full scale in eight ranges; decibels from - 50 to + 20 dBm (0 dBm = 1 mW into 50 ohms); absolute average-reading instrument calibrated to rms value of sine wave.

Frequency Range:
10 kHz to 1.2 GHz; useful sensitivity from 1 kHz to beyond 2 GHz.

Full-Scale Accuracy (%) with Appropriate Accessory:
(after probe is properly calibrated)

10 kHz – 20 kHz	20 kHz – 25 kHz	25 kHz – 100 kHz	100 kHz – 100 MHz	100 MHz – 700 MHz	700 MHz – 1 GHz	1 GHz – 1.2 GHz
±13	±8	±5	±3	±5	±8	±13

Input Impedance:
Input capacity and resistance will depend upon accessory tip used. 100,000 ohms shunted by less than 2.1 pF at 100 kHz with bare probe; less than 10 pF with 11072A isolator tip supplied.

Sample Hold Output:
Provides ac signal whose unclamped portion has statistics that are narrowly distributed about the statistics of the input, inverted in sign, (operating into greater than 200 kilohm load with less than 1000 pF).

Sample Hold Noise:
less than 175 uV rms.

Accuracy (after probe is properly calibrated):
0.01 V Range and Above: Same as Full Scale Accuracy of instrument.
0.001 V to 0.003 V Range: Value of input signal can be computed by taking into account the residual noise of the instrument.

Jitter:
Meter indicates within ± 2% peak of reading, 95% of the time (as measured with an HP Model 3400A true rms voltmeter).

RMS Crest Factor: 0.001 V to 0.3 V: 20 dB.
1 V: 13 dB.
3 V: 3 dB.

Meter

Meter Scales:
Linear voltage, 0 to 1 and 0 to 3; decibel, - 12 to + 3. Individually calibrated taut band meter.

Response Time:
Indicates within specified accuracy in less than 3 sec.

Jitter:
± 1% peak (of reading).

General

DC Recorder Output:
Adjustable from zero to 1.2 mA into 1000 ohms at full scale, proportional to meter deflection.

Overload Recovery Time:
Meter indicates within specified accuracy in less than 5 sec (30 V p-p max).

Maximum Input:
± 100 V dc, 30 V p-p.

RFI:
Conducted and radiated leakage limits are below those specified in MIL-6181D and MIL-1-16910C except for pulses emitted from probe. Spectral intensity of the pulses are nominally 50 $nV/\sqrt{Hz}$; spectrum extends beyond 2 GHz.

Temperature Range:
Instrument: 0 to + 55° C.
Probe: + 10° C to + 40° C.

Power:
115 or 230 volts ± 10%, 50 Hz to 400 Hz, nominally less than 20 watts.

Dimensions:
Standard 1/2 module 6-1/2 in high, 8-7/8 in. wide, 11-1/2 in. deep (165x225x292 mm).

Weight:
Net 12 lbs. (5,4 kg).

SECTION I
GENERAL INFORMATION

1-1. INTRODUCTION.

1-2. This section contains general information about the Model 3406A Broadband Sampling Voltmeter. A general description and various applications of the instrument are described in the following paragraphs. The published specifications are given in Table 1-1. The accessories supplied with the instrument and accessories available are identified in Table 1-2 and 1-3. Instrument/Manual Identification is also included in this section.

1-3. DESCRIPTION.

1-4. The Model 3406A Broadband Sampling Voltmeter is an easy-to-use high-frequency voltmeter incorporating an incoherent sampling technique for accurate measurements over a wide frequency range. Using the sampling technique, the accuracy of the Model 3406A is ± 3% of full scale at frequencies between 100 kHz and 100 MHz. Eight voltage ranges, selected by front panel pushbuttons, enable the 3406A to measure voltages from 1 mV to 3 volts full scale with 20 uV resolution on the 1 mV range.

1-5. The meter features two easy-to-read linear voltage scales (0 to 1 and 0 to 3) calibrated in the rms value of a sine wave, and a dBm scale of - 12 to + 3 referenced to 50 ohms. (Zero dBm is equal to one milliwatt dissipated across 50 ohms). Each range is calibrated in 10 dBm steps for power or gain measurement from - 62 to + 23 dBm in a 50 ohm system. The meter movement is individually calibrated and average-responding on all ranges. An overload circuit within the voltmeter protects the meter movement and internal circuits during specified overload on any range. Recovery time after an overload is less than 5 seconds.

1-6. The input impedance of the Model 3406A depends upon the accessory probe tip used, and the frequency of the signal being measured. Typically, the input impedance is 100 kilohms at 100 kHz in parallel with 2 pF. Using the 11072A Isolator Tip to eliminate the effect of source resistance, the total input shunt capacitance is between 9 and 10 pF. Input impedance curves as well as frequency range and use of various probe tips are given in Section III of this manual.

1-7. The 3406A is equipped with a slenderized probe for ease of measurement. A pushbutton located on the probe case retains the meter deflection when depressed. This makes possible measurements in awkward positions where

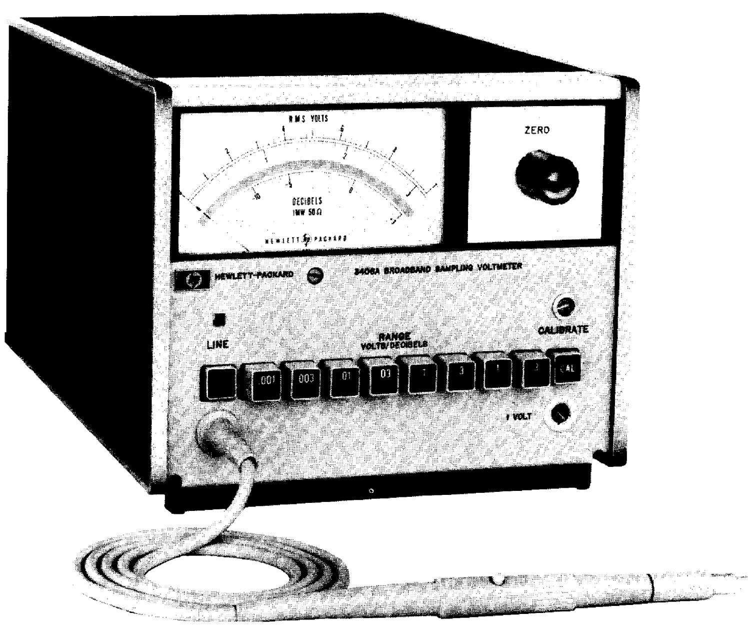


Figure 1-1. Model 3406A Broadband Sampling Voltmeter

it is difficult for the operator to place the probe in the circuit under test and at the same time read the meter.

1-8. FRONT PANEL CALIBRATION.

1-9. The Model 3406A provides a means of "calibrating" and "zeroing" the voltmeter from the front panel with no external test equipment. With the probe inserted into the 1 VOLT receptacle on the front panel and the CAL pushbutton released (out) the instrument may be zeroed in the presence of an RF field. When the CAL pushbutton is depressed an internal calibrator voltage is provided at the 1 VOLT receptacle. This voltage is used as a reference to adjust the gain of the Signal Amplifier within the 3406A for a 1 volt deflection on the meter. Since the sampling efficiency of 3406A depends upon the probe tip used, the Signal Amplifier gain is adjusted, via the front panel CALIBRATE control, each time the probe tip is changed. Refer to Section III for front panel zero and calibrate procedures. Refer to Section IV for definition of sampling efficiency.

1-10. APPLICATION.

1-11. In addition to voltage or dBm measurements on the 3406A meter, two outputs are provided on the rear panel to extend the measuring capabilities and usefulness of the Model 3406A.

1-12. RECORDER OUTPUT.

1-13. The RECORDER OUTPUT on the rear panel provides a dc output proportional to meter deflection. The dc RECORDER OUTPUT is used to drive auxiliary equipment, record measurements on a dc recorder, or digital voltmeter or it may be used as a high frequency AC-to-DC converter. The RECORDER OUTPUT may also be used to control the output level of a broadband signal generator which has dc modulation capability by monitoring the generator output with the the 3406A.

1-14. SAMPLE HOLD OUTPUT.

1-15. A SAMPLE HOLD OUTPUT is also provided at the rear of the instrument. A signal statistically equivalent to the input signal being measured is made available at this output. The frequency of the SAMPLE HOLD OUTPUT is within the audio spectrum, even for input signals up to 2 GHz. Examples of statistically equivalent signals are given in Figure 3-2.

1-16. True rms and peak voltage values, amplitude modulation envelopes, pulse height information, and probability density functions of broadband signals can be determined by observing the SAMPLE HOLD OUTPUT. Refer to Section III for additional information on sample hold output measurements and how to take them.

1-17. INSTRUMENT/MANUAL IDENTIFICATION.

1-18. Hewlett-Packard uses a two-section eight-digit serial number (000-00000). If the first three digits of the serial number on your instrument do not agree with those on the title page of this manual, change sheets supplied with the manual or Manual Backdating Changes in Section VII will

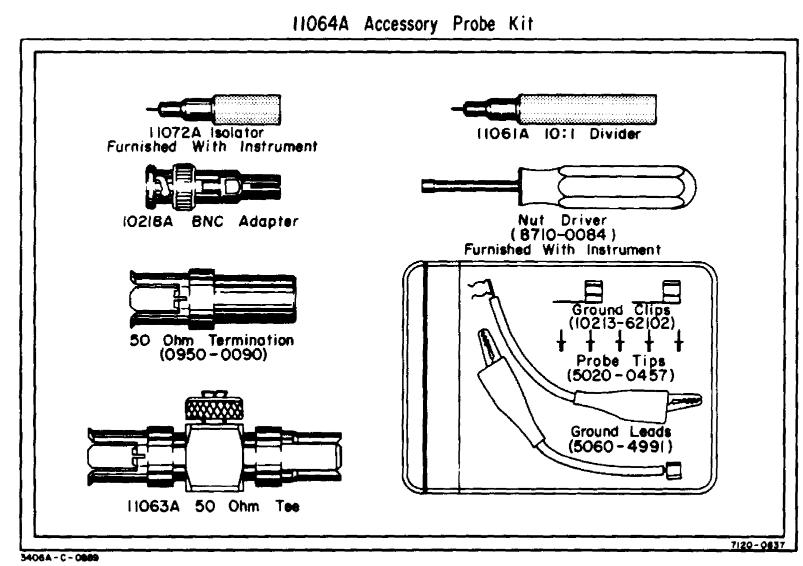


Figure 1-2. 11064A Basic Probe Kit

define differences between your instrument and the Model 3406A described in this manual.

1-19. If a letter prefixes the serial number, the instrument was manufactured outside the United States.

1-20. ACCESSORIES/EQUIPMENT SUPPLIED.

1-21. The accessories and equipment supplied with each Model 3406A are listed in Table 1-2.

1-22. ACCESSORIES AVAILABLE.

1-23. Accessories available to increase the test capabilities of the Broadband Sampling Voltmeter are described briefly in Table 1-3. The Model 3406A accessories are available in two accessory probe kits shown in Figures 1-2 and 1-3, however, each accessory may be purchased individually if desired. Contact your local -hp- Sales and Service Office for additional information (see Appendix B for office locations). To select the proper probe tip or adapter for various measurements, refer to the typical impedance curves and frequency range data given in Section III.

Table 1-2. Accessories/Equipment Supplied

DESCRIPTION	-HP- PART NO.	QUANTITY
Isolator Probe Tip	11072A	1
Nut Driver	8710-0084	1
Replacement Tips	5020-0457	10
Ground Clips	10213-62102	2
Ground Lead	5060-4991	1
Power Cord	8120-0078	1
Operating and Service Manual	03406-90002	1
Operating Note (Red)	7124-0339	1

Table 1-3. Accessories Available

MODEL	TITLE	PURPOSE
11061A	10:1 Divider	Decreases input sensitivity of Model 3406A by a factor of ten (maximum sensitivity with divider is 30 V full scale).
10218A	BNC Adapter	Permits connection of probe to female BNC connector. (May be used with Isolator Tip).
11063A	50 Ohm Tee	Permits monitoring of signal in 50 ohm system without disturbing the line or attenuating the signal on the line.
—	50 Ohm Termination	Terminates signals in a 50 ohm system (use with 50 Ohm Tee).
11073A	Pen-Type Isolator Note If 11073A is ordered individually only the alligator clip and a ground lead will be supplied with the order.	Permits the probe to be clipped on or connected to the point of measurement using one of the following accessories: 1. Alligator clip (5060-0416) 2. Pincer Jaw (5060-0417) 3. Pin Tip (5060-0418) 4. Hook Tip (5060-0419) 5. Spring Tip (5060-0420) 6. Banana Tip (1251-0013)
10219	Type 874A Adapter	Permits connection of probe to Type 874 coaxial connection (manufactured by General Radio Co.) may be used with Isolator Tip.
10220A	Microdot Adapter	Permits connection of probe to Microdot connection (manufactured by Microdot, Inc.) may be used with Isolator Tip.

Table 1-3. Accessories Available (Cont'd)

MODEL	TITLE	PURPOSE
11064A	Basic Probe Kit (See Figure 1-2)	Includes: 11063A 50 Ohm Tee 50 Ohm Termination 11061A 10:1 Divider 10218A BNC Adapter Bag of probe tips and ground leads
11071A	Complete Probe Kit (See Figure 1-3)	Includes: 11063A 50 Ohm Tee 50 Ohm Termination 11061A 10:1 Divider 10218A BNC Adapter 10219A Type 874A Adapter 10220A Microdot Adapter 11073A Pen type Isolator with alligator clip, pincer jaw, and all accessory tips. Bag of probe tips and ground leads

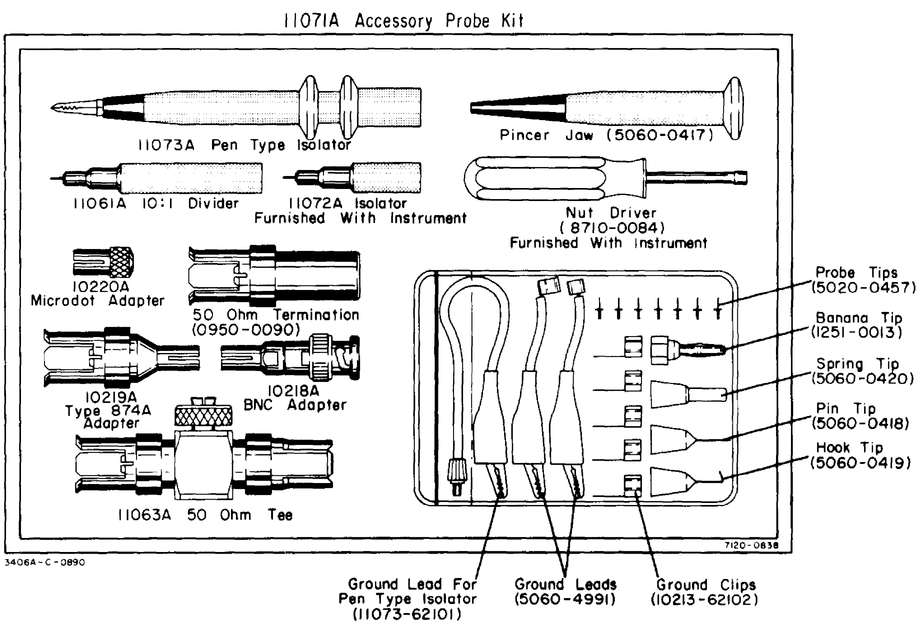


Figure 1-3. 11071A Complete Probe Kit

SECTION II
INSTALLATION

2-1. INTRODUCTION.

2-2. This section includes information and instructions necessary for the installation or shipment of this instrument. Included is information pertaining to initial inspection, installation, and repackaging for shipment.

2-3. INSPECTION.

2-4. This instrument was carefully inspected both mechanically and electrically before shipment. It should be physically free of mars or scratches and in perfect electrical order upon receipt. To confirm this, the instrument should be inspected for physical damage. Also, the electrical performance of the instrument should be tested using the procedure outlined in Section V of this manual. If there is damage or deficiency, see the warranty at the front of this manual.

2-5. POWER REQUIREMENTS.

2-6. This instrument will operate from either 115 or 230Vac, 50Hz to 400Hz. The instrument can easily be converted from 115 volt to 230 volt operation by changing the position of the slide switch located on the rear panel, so that the designation appearing on the switch matches the nominal voltage of the power source.

2-7. GROUNDING REQUIREMENTS.

2-8. To protect operating personnel, the National Electrical Manufacturers' Association (NEMA) recommends that the instrument panel and cabinet be grounded. Each Hewlett-Packard instrument is equipped with a three-conductor power cord which, when plugged into an appropriate receptacle, grounds the instrument. The offset pin on the power cord three-prong connector is the ground wire.

2-9. To preserve the protection feature when operating the instrument from a two-contact outlet, use a three-prong to two-prong adapter and connect the green pigtail on the adapter to ground.

2-10. INSTALLATION.

2-11. This instrument is fully transistorized; therefore no special cooling is required. However, the instrument should not be operated where the ambient temperature is outside the limits specified in Table 1-1.

2-12. RACK/BENCH INSTALLATION.

2-13. This instrument is initially shipped as a bench-type instrument with plastic feet and tilt stand in place. This instrument can be rack mounted by using a rack mounting kit available at your nearest -hp- Sales and Service Office.

2-14. REPACKAGING FOR SHIPMENT.

2-15. The following is a general guide for repackaging for shipment. If you have any question, contact your local -hp- Sales and Service Office. (See Appendix at the back of this manual for office location.)

NOTE

If the instrument is to be shipped to Hewlett-Packard for service or repair, attach a tag to the instrument identifying the owner and indicating the service or repair to be accomplished; include the model number and full serial number of the instrument. In any correspondence, identify the instrument by model number, serial number and serial number prefix.

2-16. Place the instrument in its original container if available. If the original container is not available, a suitable one can be purchased from your nearest -hp- Sales and Service Office.

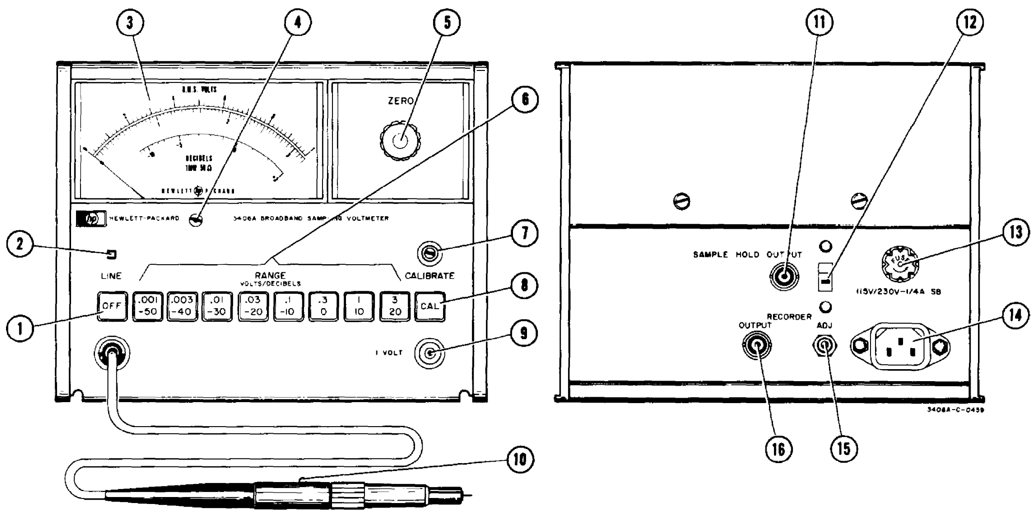
2-17. If the original container is not used:

a. Wrap the instrument in heavy paper or plastic before placing in an inner container.

b. Use plenty of packing material around all sides of the instrument and protect panel faces with cardboard strips.

c. Place the instrument and inner container in a heavy carton or wooden box and seal with strong tape or metal bands.

d. Mark the shipping container with "DELICATE INSTRUMENT," "FRAGILE" etc.



(1) LINE OFF Pushbutton: disconnects ac operating power applied to instrument when depressed.

(2) LINE Indicator: illuminates when ac operating power is applied to instrument.

(3) Meter: indicates level of applied input signal in volts or dBm.

(4) Mechanical Zero Adjustment: screwdriver adjustment for mechanically zeroing of meter movement. (See Paragraph 5-19.)

(5) ZERO Control: provides variable control of zero offset on all ranges.

(6) RANGE Pushbuttons: applies ac operating power to instrument, and selects input voltage range.

(7) CALIBRATE Control: a screwdriver adjustment for calibrating the instrument per probe tip in use. (See note under Paragraph 3-9.)

(8) CAL Pushbutton: connects internal calibrator voltage to the 1 VOLT receptacle when depressed.

(9) 1 VOLT Receptacle: provides connections for probe, with or without probe tip, when zeroing or calibrating the instrument. (See Paragraphs 3-8 and 3-9.)

(10) Meter Hold Pushbutton: retains the meter indication when depressed.

(11) SAMPLE HOLD OUTPUT: provides connection for measuring peak and true rms value of input signal above 0.5 mV (See Paragraph 3-33 for application).

(12) Line Voltage Slide Switch: selects 115 or 230 Vac source for the operating power of the instrument. Voltage selected appears on switch.

(13) FUSE: protects internal circuits of instrument (1/4 amp slow-blow for both 115 and 230 Vac operation).

(14) Power Input Connector: connects ac operating power to instrument through power cord furnished with instrument.

(15) RECORDER ADJ: screwdriver adjustment for varying the RECORDER OUTPUT between zero and 1.2 Ma into 1000 ohms at full scale.

(16) RECORDER OUTPUT: provides a dc output proportional to meter deflection with full scale value determined by the RECORDER ADJ .

Figure 3-1. Front and Rear Panel Controls, Indicators and Connectors

SECTION III

OPERATING INSTRUCTIONS

3-1. INTRODUCTION.

3-2. This section contains instructions and information necessary for operation of the Model 3406A Broadband Sampling Voltmeter. Included are turn-on procedures, operating instructions, probe tip selection, and SAMPLE HOLD OUTPUT applications. Also included in this section are front panel ZERO and CALIBRATE procedures.

3-3. FRONT AND REAR PANEL CONTROLS, INDICATORS, AND CONNECTORS.

3-4. The function of the front and rear panel controls, indicators, and connectors is identified in Figure 3-1.

3-5. TURN-ON PROCEDURE.

3-6. To turn-on, zero and calibrate the Model 3406A, proceed as follows:

3-7. TURN-ON.

a. Position the Line Voltage Switch (12, Figure 3-1) on the rear panel to the source voltage to be used. Connect the power cord.

b. Depress the 1 volt RANGE Pushbutton (6) and verify that LINE indicator (2) lights.

3-8. ZERO.

a. With the CAL Pushbutton (8) released (out) insert probe with Isolator Tip (11072A) into the 1 VOLT receptacle (9).

b. Adjust ZERO control (5) for zero indication on Meter (3).

c. The zero adjustment for the upper five ranges (.03 to 3 volt range) may be performed on any one of the upper five ranges. On the .01, .003 and .001 volt ranges zero the voltmeter on the range selected prior to taking any measurements on that range.

d. If the probe body temperature exceeds 30° C when using the .001 V range, special zeroing procedures are necessary. For readings below 1/3 scale zero the 3406A on the 1 mV range with the probe inserted into a terminated 50 ohm tee at the temperature of measurement. For readings above 1/3 scale zero the 3406A on the 1 V range with the probe inserted into a terminated 50 ohm tee at the temperature of the measurement. If a 50 ohm tee is not available, insert the probe at temperature of the measurement into the 1 V receptacle with the CAL pushbutton released and adjust ZERO as rapidly as possible.

3-9. CALIBRATE.

a. With probe in 1 VOLT receptacle, (with Isolator Tip attached) depress CAL.

b. Adjust the CALIBRATE control (7) with a screwdriver for 1 volt indication on meter.

c. Remove probe from 1 VOLT Receptacle, and release CAL Pushbutton.

NOTE

Since the sampling efficiency of the 3406A voltmeter depends on the probe tip (accessory) used, the front panel "CALIBRATE" procedure is used to compensate the meter indication for the probe tip in use. Zero and calibrate the voltmeter each time the probe tip is changed or when just the bare probe is used with or without the 50 ohm tee. Refer to Paragraph 3-27 for information on how to zero and calibrate the voltmeter per probe tip or adapter selected. Refer to Section IV for definition of sampling efficiency.

3-10. OPERATING INSTRUCTIONS.

3-11. The 3406A is designed to be used with either the Isolator Tip or one of the accessories available in the Base Probe Kit (11064A) or the Complete Accessory Kit (11071A). No other probe tip or probe kit is applicable. For proper selection of probe tip, refer to Table 1-3 and the frequency response and input impedance curves given in this section. All measurements require proper grounding of probe. Use one of the ground clips supplied with the 3406A or one of the ground leads in the Accessory Probe Kit. Connect the ground lead as close to the point of measurement as possible.

NOTE

For measurements where absolute accuracy is required, one of the following probe tips or adapters must be used:

11073A	Pen-Type Probe
11072A	Isolator Tip
11061A	10:1 Divider
11063A	50 Ohm Tee

For relative measurements with the same source resistance the bare probe may be used to achieve higher input impedance.

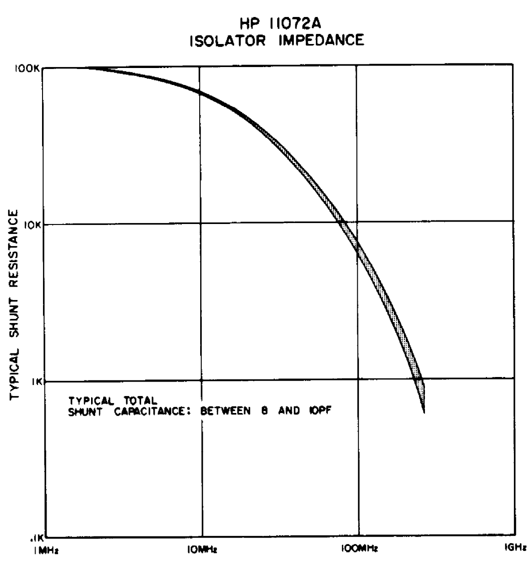
3-12. The maximum input voltage with any probe tip except the 10:1 divider is 30 V p-p AC or 100 V DC. An AC Signal, up to 5 V peak, superimposed on a DC level of 100 V may be measured if desired.

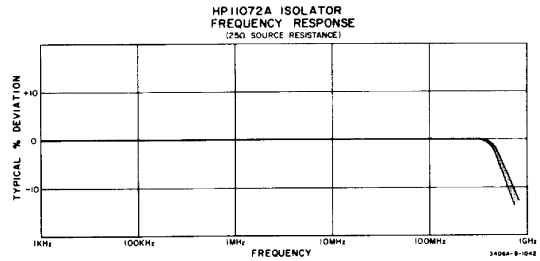
CAUTION

MEASURING AN AC VOLTAGE SUPERIMPOSED ON A DC LEVEL GREATER THAN 100 V MAY RESULT IN PROBE DAMAGE.

3-13. 11072A ISOLATOR.

3-14. The 11072A Isolator essentially eliminates effects of source impedance variations. Total input capacitance with Isolator Tip is less than 10 pF. Recommended frequency range is 10 kHz to 250 MHz. The frequency response and input impedance with Isolator Tip is shown below. For explanation of why the Isolator Tip is necessary and how it works see Paragraph 4-14.





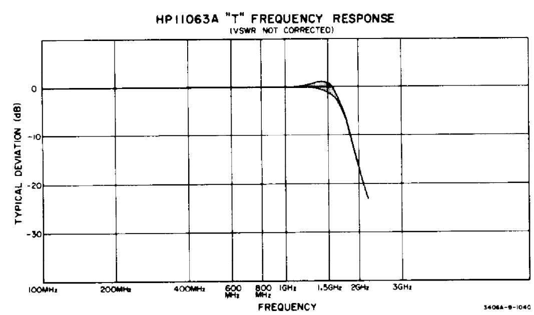
3-15. 11063A 50 OHM TEE.

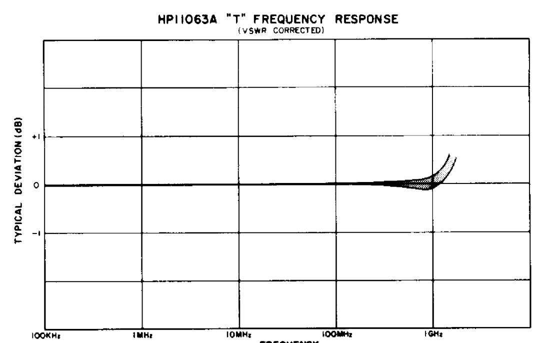
3-16. The 11063A tee is used whenever measurements are made in 50 ohm systems.

VSWR: less than 1.15 up to 1 GHz (bare probe in tee)
Insertion Power Loss: less than 1 dB up to 1 GHz

NOTE

The following curves apply only with the bare probe inserted in the tee. The frequency response of the tee with 10:1 divider is given in curve under Paragraph 3-18.

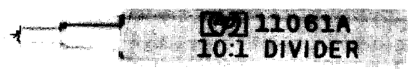


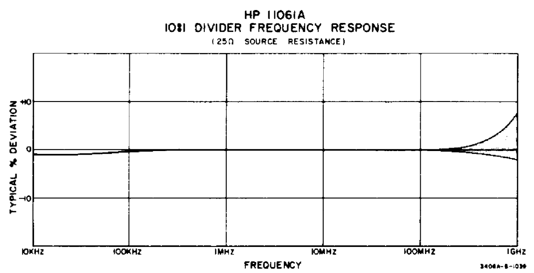


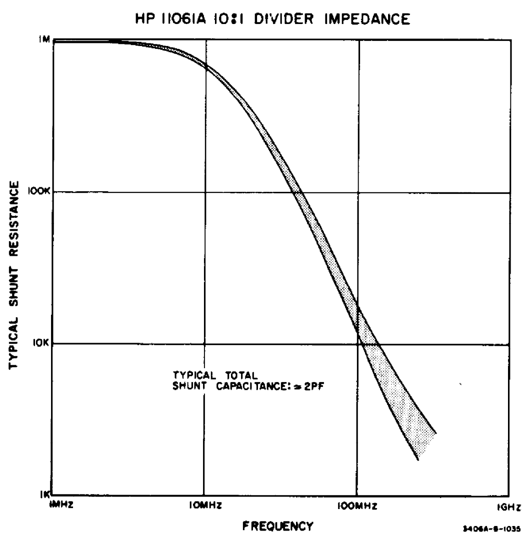
3-17. 11061A 10:1 DIVIDER.

3-18. The 11061A 10:1 Divider eliminates the effects of source impedance variations as well as divides the input voltage by a factor of ten.

Division Accuracy:	± 5% 1 kHz to 400 MHz ± 12% 400 MHz to 1 GHz
Max Input:	150 V p-p 600 Vdc







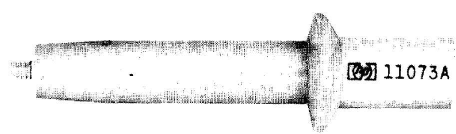
3-19. 11073A PEN TYPE ISOLATOR.

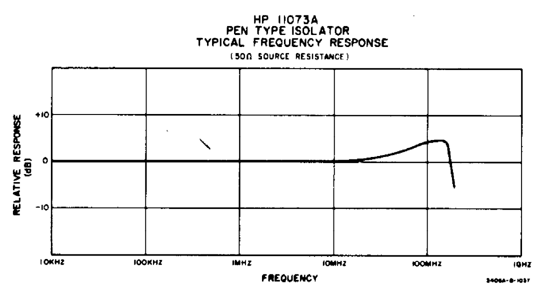
3-20. The 11073A Pen Type Isolator has a built-in isolator to eliminate the effect of source impedance variations. Various attachments to the Pen Type Isolator facilitates point-to-point measurements. The recommended frequency range is 10 kHz to 50 MHz. The added input capacitance with Pen Type Isolator is less than 8 pF. One of the following accessory tips must be used with the Pen Type Isolator:

Alligator Clip Pincer Jaw Pin Tip Hook Tip Spring Tip Banana Tip	See Figure 1-3

NOTE

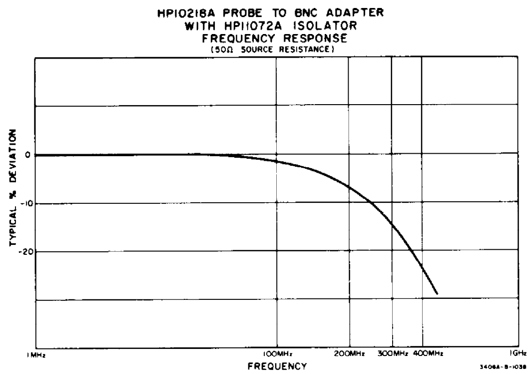
Use the alligator clip to ZERO and CALIBRATE the voltmeter when the Pen Type Isolator is used. After performing the ZERO and CALIBRATE procedure with the alligator clip any one of the six Pen Type Isolator accessories identified above may be used. Remove the plastic insulator on the front of the probe before attaching the Pen Type Isolator.





3-21. 10218A PROBE TO MALE BNC ADAPTER.

3-22. The 10218A Adapter is used to connect the probe to a female BNC connector. The recommended frequency range is 10 kHz to 250 MHz. To eliminate the effect of source impedance variations the Isolator Tip (11072A) is used with the Probe-to-BNC Adapter. The frequency response of the 10218A Adapter with the Isolator Tip is shown below.

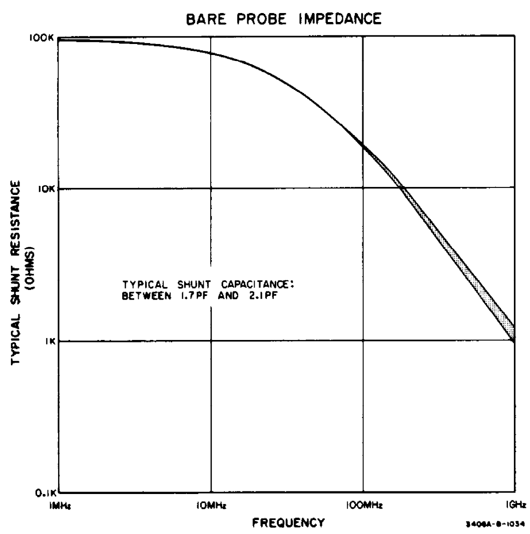


3-23. THE BARE PROBE.

3-24. The bare probe may be used for relative measurements with same source resistance. For example, checking the frequency response of a tuned tank circuit or the gain of IF stage. A higher input impedance is obtained with the bare probe.

CAUTION

TO PROTECT THE PROBE TIP WHEN THE INSTRUMENT IS NOT IN USE, INSERT THE PROBE INTO THE 1 VOLT RECEPTACLE WITH THE CAL PUSHBUTTON RELEASED.



3-25. PROBE TIP REPLACEMENT.

3-26. When replacing the probe tip on the bare probe, use the nut driver (-hp- Part No. 8710-0084) furnished with the instrument. Additional probe replacement tips may be purchased under -hp- Part No. 5020-0457.

CAUTION

WHEN REPLACING THE PROBE TIP DO NOT APPLY EXCESSIVE TORQUE; OTHERWISE DAMAGE TO COMPONENTS WITHIN THE PROBE ASSEMBLY WILL RESULT.

3-27. ZERO AND CAL THE PROBE TIP ACCESSORIES.

3-28. After selecting the proper probe tip, it is necessary to compensate the meter indication to the probe tip by performing the following procedure per probe tip or adapter selected.

a. 11063A 50 Ohm Tee: When the 50 Ohm Tee is used, remove the Isolator Tip (11072A) and perform the ZERO and CALIBRATE procedures outlined in Paragraphs 3-8 and 3-9 with the bare probe inserted into the 1 VOLT receptacle. After performing the ZERO and CALIBRATE procedures, connect the 50 Ohm Tee into circuit to be measured, and insert the bare probe into the 50 Ohm Tee. The 50 Ohm Termination (-hp- Part No. 0950-0090) is used to terminate the Tee into 50 ohms.

b. 11061A 10:1 Divider Tip: When using the 10:1 Divider Tip, remove the Isolator Tip (11072A) and perform the ZERO and CALIBRATE procedures, outlined in Paragraphs 3-8 and 3-9, with the 10:1 Divider Tip attached to the probe, and the 0.1 volt RANGE pushbutton depressed.

c. 10218A BNC Adapter: The BNC Adapter is used in conjunction with the Isolator Tip (11072A). Perform the ZERO and CALIBRATE procedures outlined in Paragraphs 3-8 and 3-9 before attaching the BNC Adapter to the Isolator Tip.

d. 11073A Pen-Type Probe: The Pen-Type Probe has a built-in isolator. To ZERO and CALIBRATE the Pen-Type Probe perform the steps outlined in Paragraphs 3-8 and 3-9 with the alligator clip attached to the Pen-Type Probe. After performing the ZERO and CALIBRATE procedures with the alligator clip, any one of the six pen probe attachments identified below may be used.

1. Alligator Clip
2. Pincer Jaw
3. Pin Tip
4. Hook Tip
5. Spring Tip
6. Banana Tip

e. 10219A Type 874A Adapter: The Type 874A Adapter is used in conjunction with the Isolator Tip (11072A). Perform the ZERO and CALIBRATE procedures outlined in Paragraphs 3-8 and 3-9 before attaching the Type 874A Adapter to the Isolator Tip.

f. 10220A Microdot Adapter: The Microdot Adapter is used in conjunction with the Isolator Tip (11072A). Perform the ZERO and CALIBRATE procedures outlined in Paragraphs 3-8 and 3-9 before attaching the Microdot Adapter to the Isolator Tip.

3-29. METER HOLD.

3-30. The Meter Hold Pushbutton (10) located on the Probe Assembly retains the meter indication for a minimum of 15 seconds when depressed. The meter Hold Pushbutton is used when making measurements in an awkward position where it is difficult to place the probe in the circuit under test and at the same time read the meter indication.

3-31. APPLICATIONS.

3-32. The Model 3406A Broadband Sampling Voltmeter is equipped with a RECORDER and SAMPLE HOLD OUTPUT in addition to the average responding direct reading rms meter. A few applications of both the RECORDER and SAMPLE HOLD OUTPUT are listed in the following paragraphs.

3-33. SAMPLE HOLD OUTPUT.

3-34. The SAMPLE HOLD OUTPUT is located on the rear panel of the instrument. A signal statistically equivalent to the input signal being measured is made available at this output. The frequency of the SAMPLE HOLD OUTPUT is within the audio spectrum, even for input signals above 2 GHz. Examples of statistically equivalent signals for three separate input signals are illustrated in Figure 3-2.

3-35. The accuracy of the SAMPLE HOLD OUTPUT is the same as the full-scale accuracy of the meter indication on the 0.01 V RANGE and above. On the 0.001 V or 0.003 V RANGE the value of input signal can be computed by taking into account the residual noise of the instrument as outlined in Paragraph 3-36 step c.

3-36. When using the SAMPLE HOLD OUTPUT the following conditions must be observed:

a. The crest factor of the input signal must not exceed 20 dB (10:1) on the 0.001 to 0.3 volt RANGE, 13 dB (4.46:1) on the 1 VOLT RANGE, and 3 dB (1.41:1) on the 3 VOLT RANGE.

b. The SAMPLE HOLD OUTPUT is valid only when the probe is connected to the circuit under test. The Meter Hold Pushbutton has no effect on the SAMPLE HOLD OUTPUT.

c. When using the SAMPLE HOLD OUTPUT to measure true rms on the .003 or .001 volt RANGE, it becomes necessary to calculate the true rms value of the input signal because of the noise present on the .003 or .001 volt RANGE using the following formula:

$$E_{rms} = \sqrt{E_s^2 - E_n^2}$$

where: E_n = Noise measured on the 3400A connected to the SAMPLE HOLD OUTPUT with the 3406A input shorted to ground and RANGE set on .003 or .001 volts.

NOTE

To ground the 3406A input, insert the probe into the 1 VOLT receptacle with the CAL pushbutton released.

E_s = 3400A indication with the input signal to be measured applied to the 3406A input.

3-37. SAMPLE HOLD OUTPUT MEASUREMENTS.

a. True RMS Measurements:
The SAMPLE HOLD OUTPUT may be used to measure the true rms value of the input signal on a True RMS Voltmeter such as the Hewlett-Packard Model 3400A. A true rms measurement may be made on any recurring waveforms within the frequency range of the Model 3406A.

NOTE

The SAMPLE HOLD OUTPUT voltage is always 0.316 volts at full-scale on all ranges because of the location of the SAMPLE HOLD OUTPUT terminal in the 3406A (refer to Block Diagram, Figure 4-4). Therefore, set the voltmeter connected to the SAMPLE HOLD OUTPUT on the 0.3 volt range and use the RANGE setting of the 3406A to determine the proper reading of the voltmeter attached to the SAMPLE HOLD OUTPUT.

b. Peak or Peak-to-Peak Measurements:
The SAMPLE HOLD OUTPUT may be used to measure the peak or peak-to-peak value of the input signal on a low frequency oscilloscope such as the Hewlett-Packard Model 130C.

NOTE

The SAMPLE HOLD OUTPUT voltage is always 0.316 volts at full-scale on all ranges because of the location of the SAMPLE HOLD OUTPUT terminal. Therefore, to determine the proper peak or peak-to-peak value of the input signal on an oscilloscope, use the settings identified in Table 3-1.

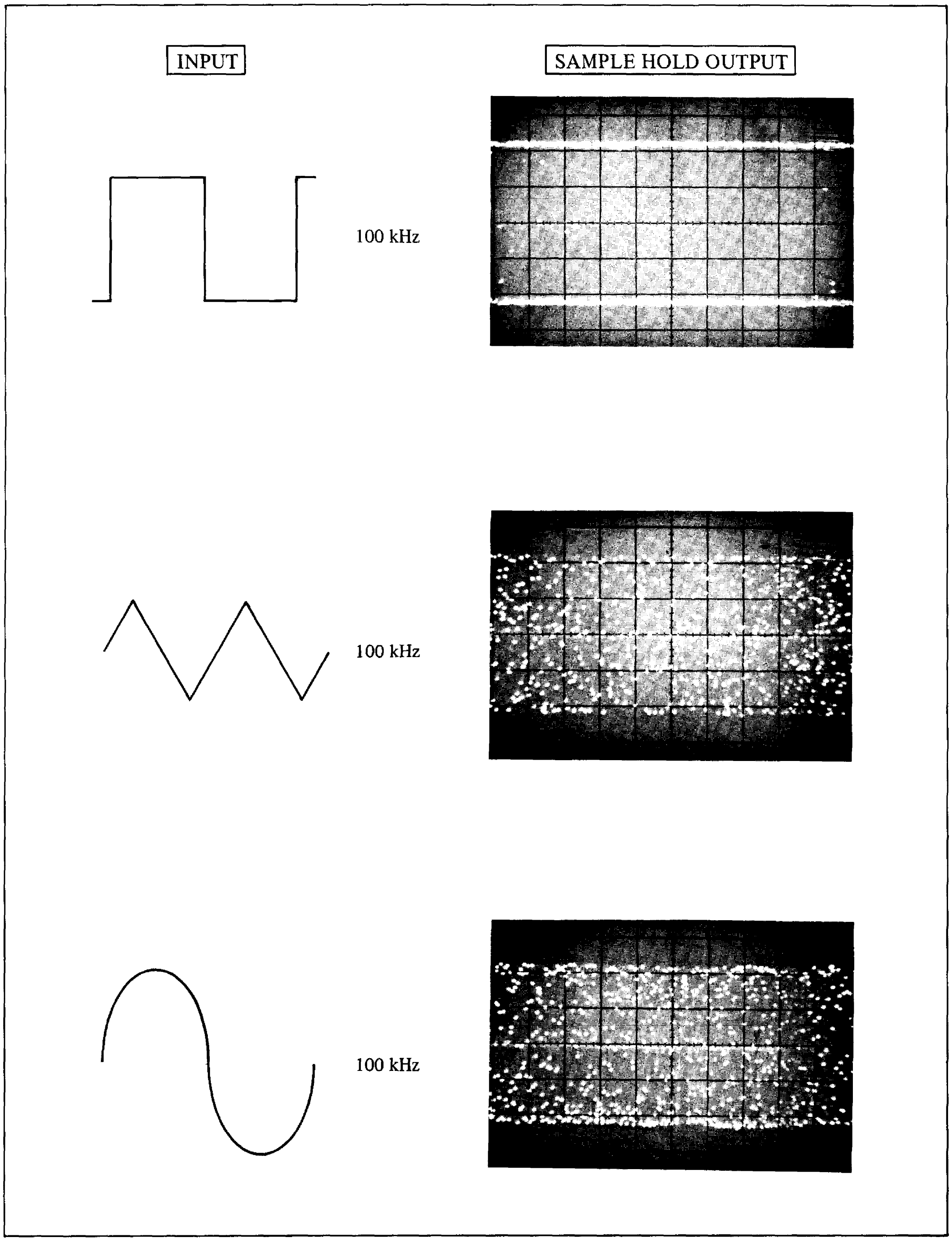
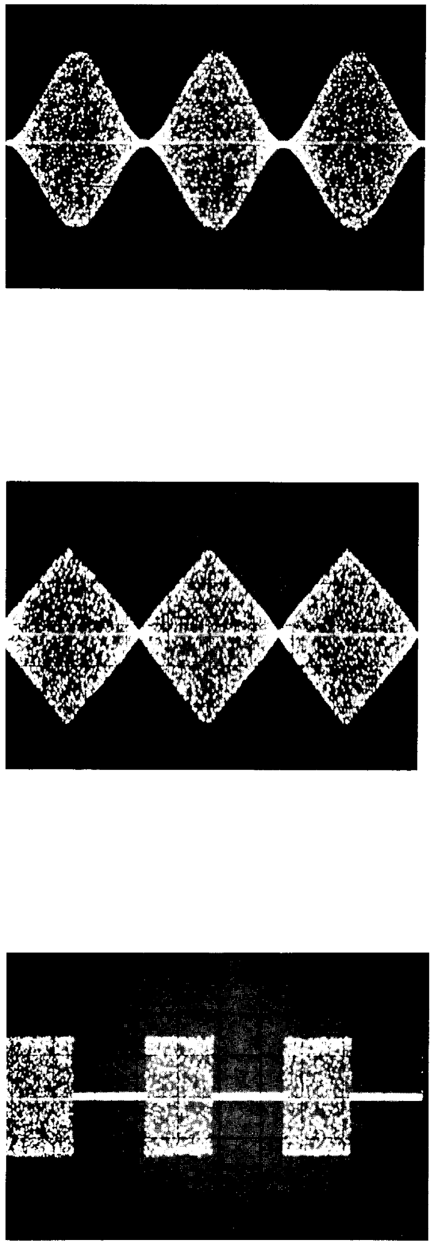


Figure 3-2. Statistically Equivalent Signals



65 MHz
CARRIER AMPLITUDE
MODULATED WITH
A SINE WAVE

65 MHz
CARRIER AMPLITUDE
MODULATED WITH
A TRIANGLE WAVE

65 MHz
CARRIER AMPLITUDE
MODULATED WITH
A SQUARE WAVE

Figure 3-3. Amplitude Modulated Signals

Table 3-1. Oscilloscope Settings

3406A RANGE	SAMPLE HOLD OUTPUT GAIN	SAMPLE HOLD OUTPUT (full scale)	OSCILLOSCOPE VERT. SENSITIVITY	
			ACTUAL SETTING	WITH RESPECT TO INPUT SIGNAL
3 Volt	0.105	0.316 Volts	.1 V/cm	1 V/cm
1 Volt	0.316	0.316 Volts	.1 V/cm	.33 V/cm
.3 Volt	1.05	0.316 Volts	.1 V/cm	.1 V/cm
.1 Volt	3.16	0.316 Volts	.1 V/cm	.033 V/cm
.03 Volt	10.5	0.316 Volts	.1 V/cm	.01 V/cm
.01 Volt	31.6	0.316 Volts	.1 V/cm	3.3 mV/cm
.003 Volt	105	0.316 Volts	.1 V/cm	1 mV/cm
.001 Volt	316	0.316 Volts	.1 V/cm	.3 mV/cm

c. Recovering Amplitude-Modulated Signals:
The SAMPLE HOLD OUTPUT may be used to monitor an AM signal with a carrier frequency within the frequency range of the 3406A, and a modulation frequency of DC to 1 kHz. Using this application hum modulation on a RF carrier may be observed. Examples of the SAMPLE HOLD OUTPUT for amplitude modulated signals are illustrated in Figure 3-3.

d. Measure true rms value of ac signal with dc component:
Since the 3406A is an ac device the SAMPLE HOLD OUTPUT can be used to measure only the rms value of the ac component. If it is necessary to include the rms value of the dc component when measuring a signal use the following formula:

NOTE

The frequency of the ac signal must be within the frequency range of the 3406A, and the dc component level must not exceed 100 Vdc.

$$E_{rms} = \sqrt{E_{ac}^2 + E_{dc}^2}$$

E_{ac} = SAMPLE HOLD OUTPUT voltage measured with 3400A.

E_{dc} = The dc component measured with the -hp- Model 414A, DC Voltmeter.

3-38. RECORDER OUTPUT.

3-39. The RECORDER OUTPUT on the rear panel provides a dc output proportional to meter deflection. The output current level is variable between zero and 1.2 mA into 1000 ohms at full scale. The output level is controlled by the RECORDER ADJ on the rear panel.

3-40. RECORDER OUTPUT APPLICATIONS.

3-41. The RECORDER OUTPUT is used to record measurements or drive auxiliary equipment. Two applications are as follows:

a. High frequency AC-to-DC converter:
The RECORDER OUTPUT may be used as an AC-to-DC converter for ac signals from 10 kHz to 2 GHz.

b. Leveling:
The RECORDER OUTPUT may be used to control the output level of a broadband signal generator which has a dc modulation capability and the necessary loop compensation circuits to prevent oscillations.

SECTION IV
THEORY OF OPERATION

4-1. INTRODUCTION.

4-2. This section contains the theory of operation for the -hp- Model 3406A Broadband Sampling Voltmeter. A discussion of the sampling technique used as well as a theory of operation is given.

4-3. SAMPLING TECHNIQUE.

4-4. The -hp- Model 3406A uses a unique "incoherent" sampling technique to translate a high frequency input signal into a low frequency equivalent. This incoherent sampling used to produce the equivalent signal is illustrated in Figure 4-1 and described in the following paragraphs.

4-5. COHERENT SAMPLING.

4-6. Waveforms A and B in Figure 4-1 illustrate the coherent sampling technique commonly used in oscilloscopes to reconstruct the waveform at a lower frequency. With coherent sampling each sample is taken at a fixed frequency with the period of time between samples remaining constant as illustrated in waveform A.

4-7. Waveform B is a connection of pulses representing each sample (1 thru 12) taken on waveform A. The relative phase (order), polarity and proportional amplitude of the input signal at the point of sample are all preserved. Therefore, if a sufficient number of samples are taken a lower frequency equivalent of the input signal will be reconstructed. This lower frequency equivalent will have the same average, peak and rms value as the original signal, waveform A.

4-8. INCOHERENT SAMPLING.

4-9. If the pulses in waveform B were "scrambled" with only the amplitude and polarity preserved a pulse train similar to waveform C may result. The pulses in waveform C, having no phase relationship with the input signal will not reconstruct the input waveform, but the statistical properties (average, peak and rms) of waveform C will be the same as waveform B. The statistical properties are the same because the same pulses are used in both waveforms. The only thing different is the order. Therefore, if a sufficient number of samples were taken in this "scrambled" fashion the resulting pulses would be statistically equivalent to the input signal with the only relative phase relationship lost.

4-10. As shown in waveform C the relative phase does not affect the statistical properties of the reconstructed pulse train; therefore, the 3406A can use a sampling technique where only the polarity and amplitude of each sample is used to determine the value of the input signal. This sampling technique is referred to as "incoherent sampling." With incoherent sampling, the interval between samples does not remain constant and the relative phase between each sample is lost as illustrated in waveforms D and E. However, the polarity and proportional amplitude of the input signal at the point of sample are still preserved; therefore, if a sufficient number of incoherent samples are taken, the average, peak and rms values of the reconstructed pulse train will be statistically equivalent to the input.

4-11. In the 3406A this statistically equivalent pulse train is amplified and applied to the SAMPLE HOLD OUTPUT for true rms or other types of measurements. The same statistically equivalent pulse train is also detected and applied to the meter circuit for a meter indication. With this reconstructed pulse train applied to the meter circuit, a small amount of meter jitter shows up. This is because the accuracy of reconstructing the input signal is proportional to number of samples taken. If an infinite number of samples could be taken, then the representation (reconstruction) of the input would be identical to the input. This, however, is impractical. Therefore, due to the finite sample rate (between 10 and 20 thousand per second) of the 3406A, there will be small movements of the meter needle (jitter) over a period of time.

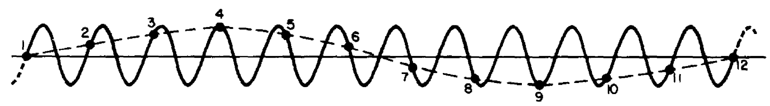
4-12. WHY INCOHERENT SAMPLING.

4-13. The 3406A uses incoherent sampling to avoid difference frequencies within the passband (0 to 10 Hz) of its metering circuit. These difference frequencies or "beat signals" could arise if the sampling frequency is held constant. For example, if the sampling frequency of the 3406A was held constant, say 10 kHz, thousands of points (frequencies) within the bandwidth of the 3406A would be harmonics of the sampling frequency. If one of these harmonics became the input signal to be measured all the samples would have exactly the same amplitude and the 3406A would see only one amplitude and one polarity (a beat signal) making it impossible to reconstruct an equivalent signal. To avoid this number in the 3406A, the sampling frequency is varied between 10 kHz and 20 kHz to produce "incoherent" sampling of the input signal.

4-14. SAMPLING EFFICIENCY.

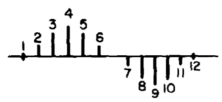
4-15. The sampling efficiency of the Model 3406A Voltmeter is determined by the source impedance of the signal being measured or the probe tip being used. Figure 4-2, illustrating the Sampling Probe Assembly A1, a source impedance Rs and a source voltage Vs, is used to explain sampling efficiency.

A



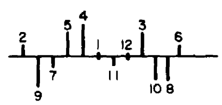
Input signal being sampled coherently.

B



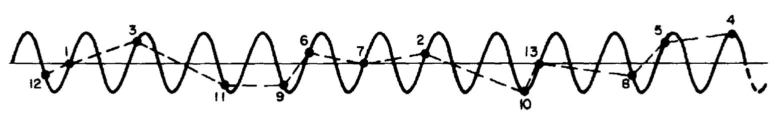
A collection of pulses reconstructing a low frequency equivalent of the input signal by preserving the height, polarity and relative phase of each sample taken.

C



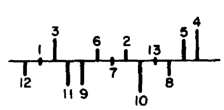
A collection of pulses "scrambled" with only the height and polarity preserved will have the same average, peak and rms values as waveform B.

D



Input signal being sampled in an incoherent fashion.

E



A collection of pulses representing the polarity and height of each sample taken incoherently. The average, peak and rms values of this waveform are the same as waveform C and B.

Figure 4-1. Sampling Technique

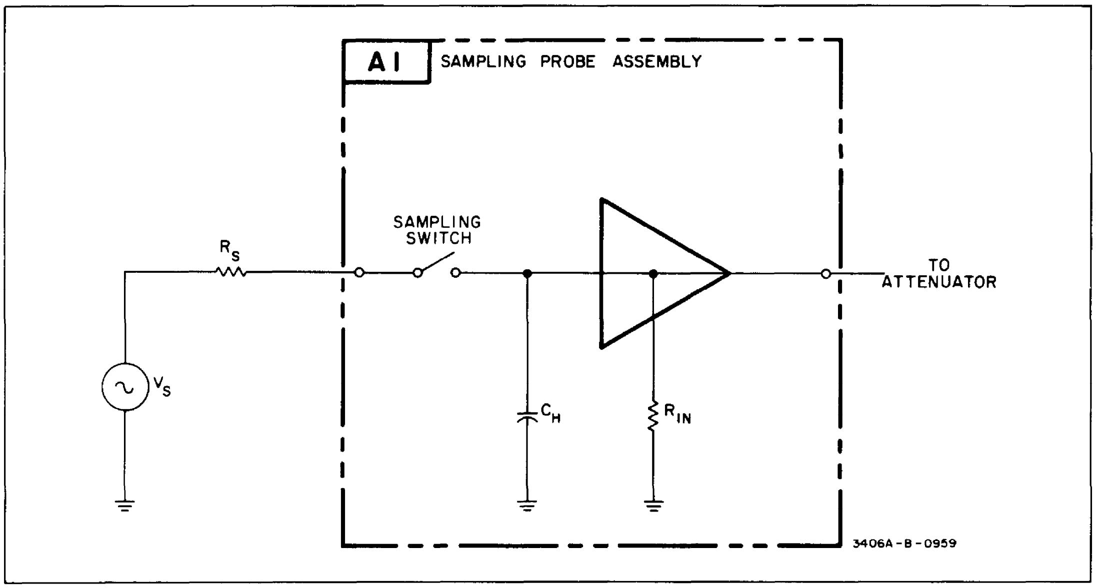


Figure 4-2. Sampling Probe Assembly

4-16. A switch within the sampling probe assembly closes for a very short period of time during each sample. When the switch closes, Holding Capacitor, Ch, will start to charge through the source resistance Rs to a value proportional to the source voltage Vs. A measurement of how close the charge on the holding capacitor Ch comes to the actual input voltage is referred to as "sampling efficiency". Therefore, sampling efficiency, in percent, can be defined by the following formula:

$$\text{Sampling Efficiency} = \frac{Vc}{Vs} \times 100$$

Where: Vc = Charge developed on capacitor during the time the switch is closed.

Vs = Actual input voltage of the source.

4-17. Since the period of time that the sampling switch remains closed is fixed (between 200 and 300 picoseconds), the source resistance Rs is the only factor determining the sampling efficiency when only the bare probe is being used. The higher the source resistance the less the charge on the holding capacitor Ch and consequently the lower the sampling efficiency. With a smaller source resistance the time constant of Rs and Ch is faster and the charge on Ch comes closer to the total source voltage Vs, and the sampling efficiency goes up. Since the charge on Ch is directly proportional to the meter deflection and SAMPLE HOLD OUTPUT level, the absolute accuracy of the voltmeter becomes a function of the sampling efficiency. This effect of source resistance on the sampling efficiency of the voltmeter is the reason that the bare probe can only be used for relative measurements with constant source resistance.

4-18. To maintain a constant sampling efficiency and eliminate the effect of varying source impedance various "isolator tips" such as the 11072A Isolator Tip, the 11073A Pen Type Isolator, the 11061A 10:1 Divider or the 50 Ohm Tee may be attached to the bare probe. These isolators compensate the effect source resistance by inserting a resistor R1 and capacitor C1 before the sampling switch as illustrated in Figure 4-3. For high source resistance (large TC for Rs - Ch) the isolator puts a capacitor C1 in parallel to aid the charge on holding capacitor Ch by discharging when the switch closes. For low source resistance (small TC for Rs - Ch) the isolator has a series resistor R1 to increase the time constant and prevent raising the percentage of charges on holding capacitor Ch. Both C1 and R1 within the isolators are selected at the factory to give a constant sampling efficiency over a wide range of source impedance. However, since the physical size and value of R1 and C1 varies between the 11072A Isolator Tip, the 11073A Pen Type Isolator and the 11061A 10:1 Divider, the actual sampling efficiency depends on the type of isolator used. For this reason a means of compensating for the difference in sampling efficiency between isolator tips is provided on the front panel.

4-19. Front panel calibrate procedure outlined in Section III adjusts the gain of the Signal Amplifier (A2Q2 thru A2Q6) to compensate for the difference in sampling efficiency between the types of isolator tips used. This is the reason the front panel CALIBRATE control must be

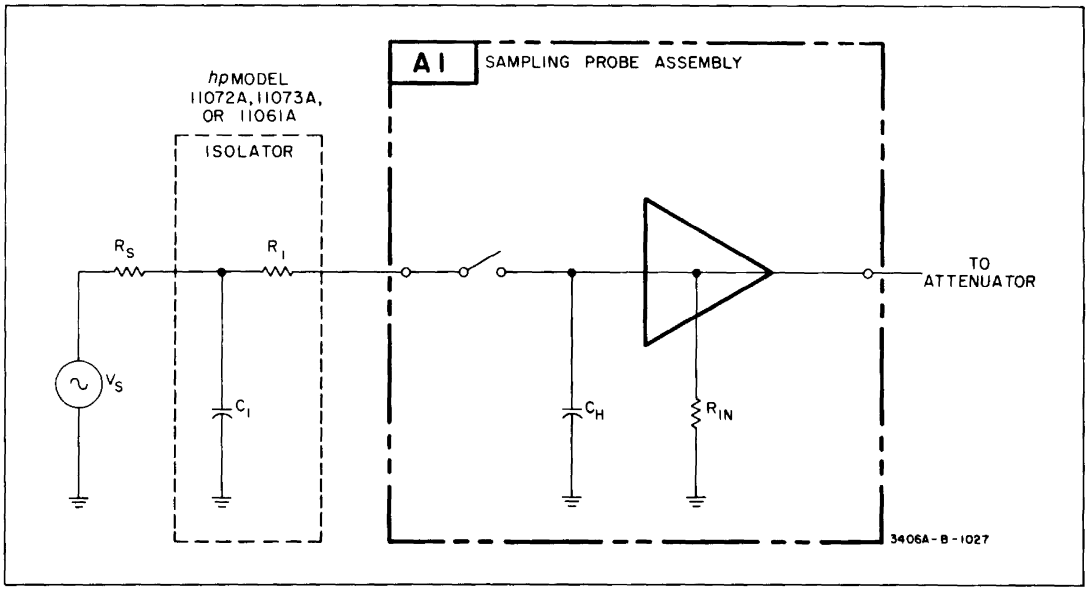


Figure 4-3. Function of Isolator Tip

adjusted each time a different probe tip is selected. Also, since the sampling efficiency of the bare probe is different than any one of the isolators, it becomes necessary to "recalibrate" the voltmeter with just the bare probe before inserting it into the 50 Ohm Tee.

4-20. GENERAL THEORY OF OPERATION.

4-21. Figure 4-4 is a block diagram illustrating the Model 3406A Broadband Sampling Voltmeter operation. All blocks shown in Figure 4-4 except the Sampling Probe Assembly and Range Control are located on two etched circuit boards (top and bottom assemblies) inside the modular cabinet of the instrument. The Sample Probe Assembly is physically located within the probe. The Range Control consists of the front panel RANGE pushbuttons and associated resistors.

4-22. Instead of continuously monitoring the input, as other RF voltmeters do, a series of samples are taken by the Sampling Probe Assembly and reconstructed into a statistical equivalent pulse train.

4-23. The input signal to be measured is applied through the appropriate probe tip to a sampling switch (see Figure 4-2) within the Sampling Probe Assembly A1. The sampling switch closes to sample the input for 200 to 300 picoseconds. The rate of closing and opening the sampling switch is controlled by Sampler Drive Pulses from the Pulse Generator p/o A3. The Sampler Drive Pulse repetition rate varies between 10 kHz and 20 kHz at a 10 Hz rate, thus producing the incoherent sampling previously discussed.

4-24. A Triangle Wave Generator and a Voltage Controlled Oscillator (all p/o A3) are used to vary the pulse repetition rate of Sample Drive Pulse. A fixed 10 Hz triangle wave from the Triangle Wave Generator is applied to the Voltage Controlled Oscillator. The voltage level of the applied triangle wave determines the output frequency of the VCO. The peak (antinode) point of the triangle wave corresponds to a 10 kHz sawtooth output and the node (0 volts) corresponds to a 20 kHz sawtooth output. Therefore, the output of the VCO becomes a sawtooth sweeping between 10 kHz and 20 kHz at the rate of the applied triangle (10 Hz). This sweeping sawtooth is applied to the Pulse Generator where a Sampler Drive Pulse is generated for each sawtooth received. Since the sawtooth is sweeping between 10 kHz and 20 kHz at a 10 Hz rate the occurance of the Sampler Drive Pulse will also vary between 10 kHz and 20 kHz at the same rate.

4-25. The Sampler Drive Pulse closes the sampling switch, and the holding capacitor Ch (see Figure 4-2) charges to a voltage proportional to the input signal. When the sampling switch reopens, after each sample, the holding capacitor completely discharges. Each charge on holding capacitor is applied to the Attenuator illustrated in Figure 4-4. Thus the sampling probe output becomes a train of pulses (one pulse per sample) whose amplitudes are directly proportional to the amplitude of the input signal at the instant the samples are taken as illustrated in Figure 4-5 waveforms A and B.

4-26. The amount of attenuation is controlled by the front panel RANGE pushbuttons. After appropriate attenuation, the pulses are amplified by the Signal Amplifier (p/o A2). Since the input signal has been translated into a lower

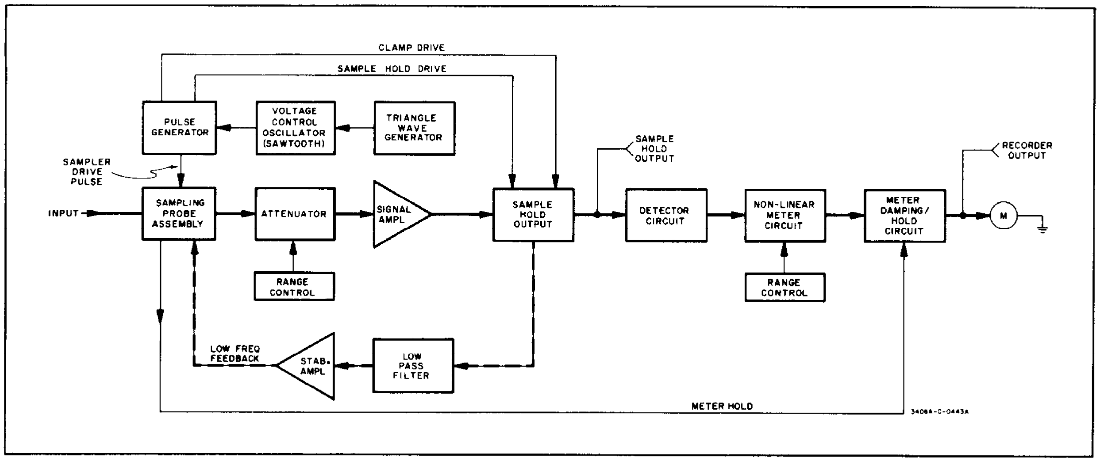


Figure 4-4. Block Diagram

frequency equivalent no high frequency compensation circuits are required in the Signal Amplifier. Three pulses from the Signal Amplifier, representing three different samples taken, are shown in Figure 4-5 waveform D. After passing through the input cable and the Signal Amplifier the sample pulses become about 5 us wide. Note, however, that the peak amplitudes of these pulses are proportional to the amplitude of the input signal at the point of sample. After the Signal Amplifier the pulses are applied to the Sample Hold Circuit.

4-27. Within the Sample Hold Circuit (p/o A2), the peak amplitude of each pulse from the Signal Amplifier is detected and held after a 2.5 usec "clamping interval" as illustrated in Figure 4-5 waveform E. Note that the peak amplitude detected is held until the next sample is initiated.

4-28. The clamping interval is the period of time (2.5 us immediately following each sample) during which the output of Sample Hold Circuit is clamped to zero volts. The clamping of each sample to zero volts establishes the dc reference level for the Signal Amplifier and the Sample Hold Output, as well as eliminating any transients that may have been caused by the Pulse Generator when the actual sample was taken.

4-29. The Clamp Drive signal from the Pulse Generator controls the starting and stopping of the clamping interval. The Clamp Drive signal is synchronized with the Sampler Drive Pulse sent to the sampling probe and occurs at the same time. Each Clamp Drive signal from the Pulse Generator, clamps the output of the Sample Hold Circuit to zero volts for 2.5 usec. During this 2.5 usec the maximum amplitude of the pulse from the signal Amplifier is reached (detected) and held as shown in Figure 4-5. The peak or maximum amplitude reached is proportional to the amplitude of the input signal at the point of sample.

4-30. The Sample Hold Drive signal, also from the Pulses Generator, is the signal that enables the Sample Hold Circuit to detect the maximum amplitude of each pulse while the output is clamped to zero volts. The Sample Hold Drive signal is also synced with the Sampler Drive Pulse and occurs at the same time.

4-31. The Sample Hold Drive signal breaks down two zener diodes, A2CR1 and A2CR2 shown in Figure 4-6, each time a sample is taken. With the two diodes conducting in reverse direction the respective pulse from the Signal Amplifier charges capacitor A2C12 to the peak value of the pulse during the 2.5 us clamping interval. The charge on capacitor A2C12 remains at this peak value even after the Sample Hold Drive and the Clamp Drive Signals are removed. This is the reason for the term "Sample-Hold". Each sample pulse is actually held at its peak amplitude until the next sample is taken, and then the Sample Hold Output is again clamped back to zero volts.

4-32. The end result of the Sample Hold Circuit is a series of square wave like pulses with amplitudes proportional to the peak amplitudes of each pulse from the Signal Amplifier, as illustrated in Figure 4-5. Note, that after a sufficient number of pulses (samples) from the Signal Amplifier, the Sample Hold Output becomes statistically equivalent to the input signal. This statistically equivalent signal is applied to the Detector Circuit and the SAMPLE HOLD OUTPUT jack on the rear panel. Refer to Section III for applications using this statistically equivalent signal available at the SAMPLE HOLD OUTPUT jack.

4-33. The Detector Circuit (p/o A2) detects the absolute average value of all the pulses from the Sample Hold Circuit. The Detector Circuit has an input-output linearity characteristic that eliminates the low level non-linear region of common rectifying type detectors. The dc output of the Detector Circuit is applied to the Non-Linear Meter Circuit (p/o A2).

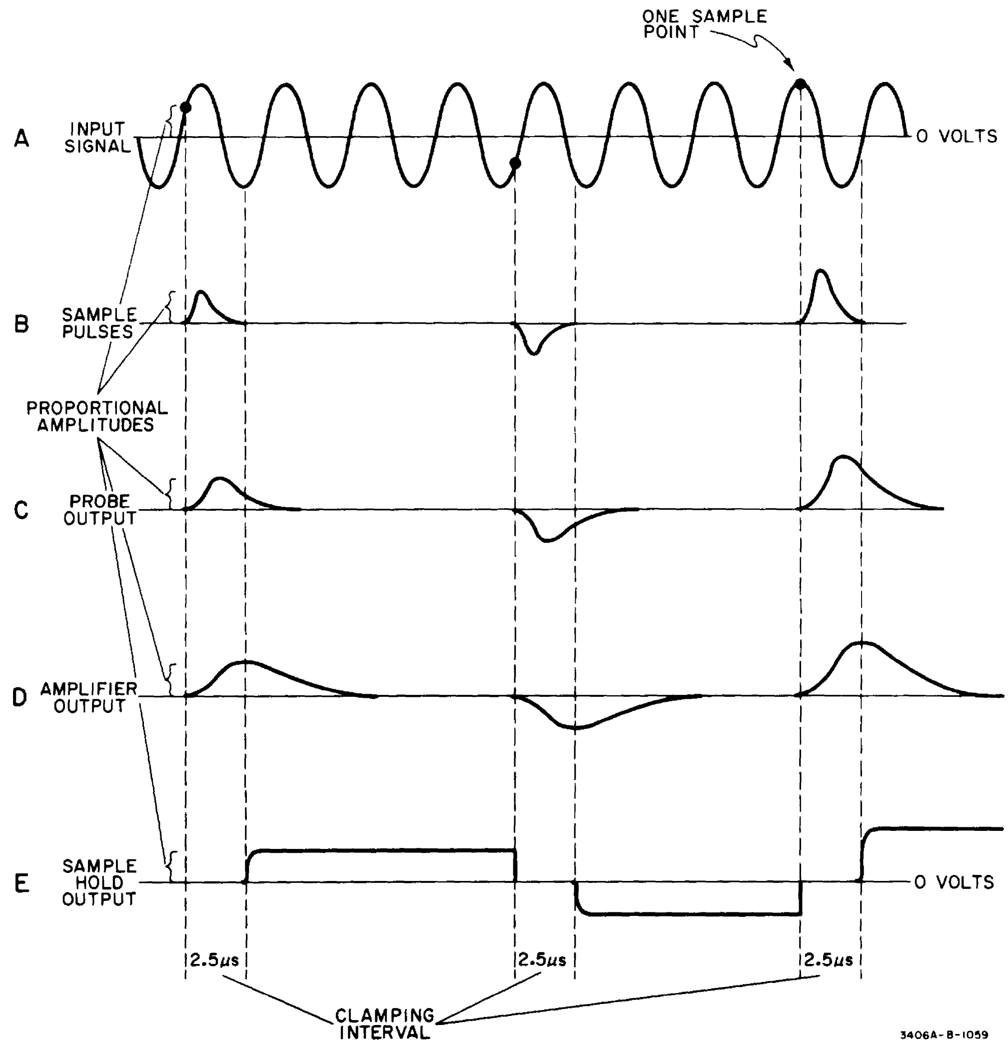


Figure 4-5. Sample Hold Output

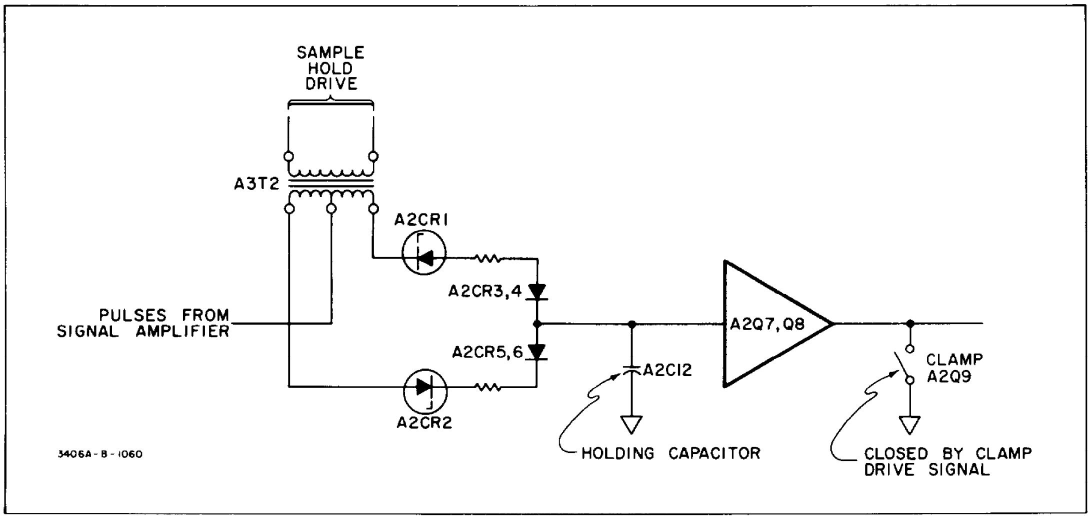


Figure 4-6. Sample Hold Circuit

4-34. Because the signal-to-noise ratio is small on the lower ranges, the noise caused by the sampling probe starts affecting the apparent gain of the detector as shown in Figure 4-7. The gain of the detector is low for small signal-to-noise ratios and approaches unity gain with larger signal-to-noise ratios. To compensate for this non-linear gain caused by the signal-to-noise ratio, a Non-Linear Meter Circuit is inserted between the detector and the final Meter Amplifier. The gain of the Non-Linear Meter Circuit is the inverse of the detector gain as shown in Figure 4-7. Therefore, the resulting gain of both circuits is unity and the effects of small signal-to-noise ratios are eliminated.

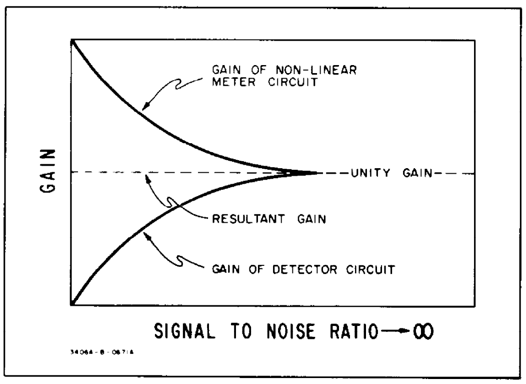


Figure 4-7. Resultant Gain

4-35. The gain of the Non-Linear Meter Circuit is controlled by negative feedback around a differential amplifier as shown in Functional Block Diagram, Figure 7-1. The amount of feedback is controlled by the front panel RANGE pushbuttons and internal "tracking" adjustments.

4-36. The output of the Non-Linear Meter Circuit is applied to the Damping/Hold Circuit. The Damping/Hold Circuit (p/o A3) impresses heavy damping on the meter movement for small changes appearing in the input voltage being measured. For large changes in input voltage, the Damping/Hold Circuit results in fast meter movements. This type of damping is used to reduce meter jitter (small meter movements) caused by the finite sampling rate of the voltmeter as described in Paragraph 4-11. The Damping/Hold Circuit also performs the meter hold operation when the pushbutton on the case of the probe is depressed.

4-37. The output of the Damping/Hold Circuit is applied to the meter, M1, for meter indication in rms, and the RECORDER OUTPUT jack on the rear panel. The RECORDER OUTPUT is used for monitoring the input signal on a dc recorder or digital voltmeter.

4-38. The Stabilizing (Stab.) Amplifier (p/o A2) and Low Pass Filter connected in the feedback path from the Sample Hold Circuit passes only low frequencies around 10 Hz and any dc drift. The Stabilizing Amplifier amplifies any dc drift and any 10 Hz component caused by the sampling frequency sweeping between 10 kHz and 20 kHz at a 10 Hz rate. The output of the Stabilizing Amplifier is applied to the Sampling Probe Assembly to stabilize the Signal Amplifier and eliminate any dc from getting into the Sample Hold Circuit.

NOTE

A Functional Block Diagram of the 3406A Voltmeter is shown in Figure 7-1. Pertinent notes and waveforms are given on this Functional Block Diagram to aid in understanding the theory of operation.

Table 5-1. Required Test Equipment

INSTRUMENT TYPE	REQUIRED CHARACTERISTICS	RECOMMENDED MODEL
Oscillator	Frequency Range: 10 kHz to 10 MHz	-hp- Model 652A Test Oscillator
VHF Oscillator	Output Frequency: 10 MHz to 400 MHz Output Level: 0 dBm into 50 ohm load	-hp- Model 608C VHF Signal Generator
UHF Oscillator	Output Frequency: 500 MHz to 1.2 GHz	-hp- Model 612A UHF Signal Generator
AC Voltmeter	Frequency Range: 1 kHz to 100 kHz Voltage Range: 25 mV to 1 V Accuracy: ± 1%	-hp- Model 400E AC Voltmeter
DC Null Voltmeter	Voltage Range: ± 10 mV Accuracy: ± 2%	-hp- Model 419A DC Null Voltmeter
DC Voltmeter	Voltage Range: 0 to 20 volts Accuracy: ± 1%	-hp- Model 414A Auto Voltmeter
DC Standard	Output Voltage: 1 V at 200 mA Accuracy: 0.1%	-hp- Model 740B DC Standard Differential Voltmeter
Power Meter	Frequency Range: 500 MHz to 1 GHz Power Range: - 10 dBm Accuracy: 0.5%	432A/478A Power Meter/ Thermistor Mount
Thermal Converter	Frequency Range: 10 kHz to 100 MHz	-hp- Model 11050A Option 02
Bucking Supply	BT1: 1.34 V battery, mercury R1: 50 ohm variable ± 10% R2: 100 ohm variable ± 10% 20-turn R3: 6500 ohm ± 1%	Mallory RM-42R -hp- Part No. 2100-0002 -hp- Part No. 2100-0973 -hp- Part No. 0811-0392
Slide Screw Tuner	Frequency Range: 500 MHz to 1.2 GHz Impedance: 50 ohms	-hp- Model 872A Coaxial Slide Screw Tuner
Slotted Line	Frequency Range: 500 MHz to 1.2 GHz Residual SWR: less than 1.04	-hp- Model 805C Slotted Line
SWR Meter	Frequency: 1000 Hz ± 2% Range: 70 dB	-hp- Model 415E SWR Meter
Oscilloscope	Frequency Range: 50 MHz Vertical Sensitivity: .005 V/cm	-hp- Model 140A/1402A Oscilloscope
Low-Pass Filter	Cutoff Frequency: 1200 MHz	-hp- Model 360B Low-Pass Filter
Power Divider	Impedance: 50 ohms Frequency Range: 500 MHz to 1.2 GHz	-hp- Model 11549A
3406A Accessory Kit	—	-hp- Model 11064A
Resistor	44.2 kilohms ± 1% 1/2 W	-hp- Part No. 0698-4936
Variable Line Transformer	Voltage Range: 103 to 127 V, 207 to 253 V Power Capability: 17 watts	Superior Type VC1M

SECTION V
MAINTENANCE

5-1. INTRODUCTION.

5-2. This section contains instructions and information necessary for the maintenance of the -hp- Model 3406A Broadband Sampling Voltmeter. Included are Performance Checks, Adjustment and Calibration Procedures, Servicing Tips, and Troubleshooting Procedures.

5-3. TEST EQUIPMENT REQUIRED.

5-4. The test equipment required to perform the operations outlined in this section is listed in Table 5-1. This table includes the type of instrument required, critical specifications and recommended model. If the model recommended is not available, equipment which meets or exceeds the required characteristics listed may be substituted.

5-5. PERFORMANCE CHECKS.

5-6. The performance checks presented in this section are front panel procedures designed to compare the Model 3406A voltmeter with its published specifications (Table 1-1). These tests may be incorporated in periodic maintenance, post repair, and incoming quality control inspection. The Performance Checks should be conducted before any attempt is made to adjust or calibrate the instrument internally. During the Performance Checks, connect the Model 3406A voltmeter to the ac source through a variable power-line transformer so that line voltage may be varied ± 10% from 115 or 230 vac to assure that the instrument operates correctly at various ac line voltages.

NOTE

Before starting the Performance Checks allow a 30-minute warmup and stabilization period. ZERO and CALIBRATE the 3406A with the bare probe inserted into 1 VOLT receptacle on the front panel as outlined in Paragraphs 3-8 and 3-9. To assure elimination of RF interference and proper operating temperature both the side and top covers must be on during all Performance Checks.

5-7. If the Model 3406A does not meet the specification test limits given in the following procedures, perform the Adjustment and Calibration Procedures outlined in Paragraph 5-16.

5-8. RANGE-TO-RANGE TRACKING AND 10 kHz TO 10 MHz ACCURACY.

CAUTION

DO NOT ALLOW INPUT VOLTAGE TO THERMAL CONVERTER TO EXCEED 1.2 V RMS. OTHERWISE THE THERMAL CONVERTER MAY BE DAMAGED.

a. The following equipment is required for the check:

Oscillator (-hp- Model 652A)
DC Standard (-hp- Model 740B)
Power Divider (-hp- Model 11549A)
DC Null Voltmeter (-hp- Model 419A)
Thermal Converter (-hp- Model 11050A Option 02)
Bucking Supply (See Table 5-1)

b. Connect the equipment as shown in Figure 5-1.

c. Adjust the dc standard for exactly 0.900 volts.

d. Adjust the bucking supply for a null indication on the null voltmeter. Do not readjust the bucking supply until step h of this check.

e. Connect the equipment as shown in Figure 5-2, using the dc standard (A) to supply the input to the power divider.

f. Increase the output of the dc standard until the null voltmeter indicates a null condition. This provides a 0.900 volt input to the thermal converter.

g. Disconnecting the thermal converter first and reconnecting it last, turn the power divider over, connecting the known 0.900 volt output to the 11063A tee and the unknown output to the thermal converter. This applies a known 0.900 volt to the 3406A.

h. Readjust the bucking supply for a null indication on the null voltmeter. Do not readjust the bucking supply for the remainder of the accuracy checks. This null condition will exist when the input to the 11063A tee is 0.900 V.

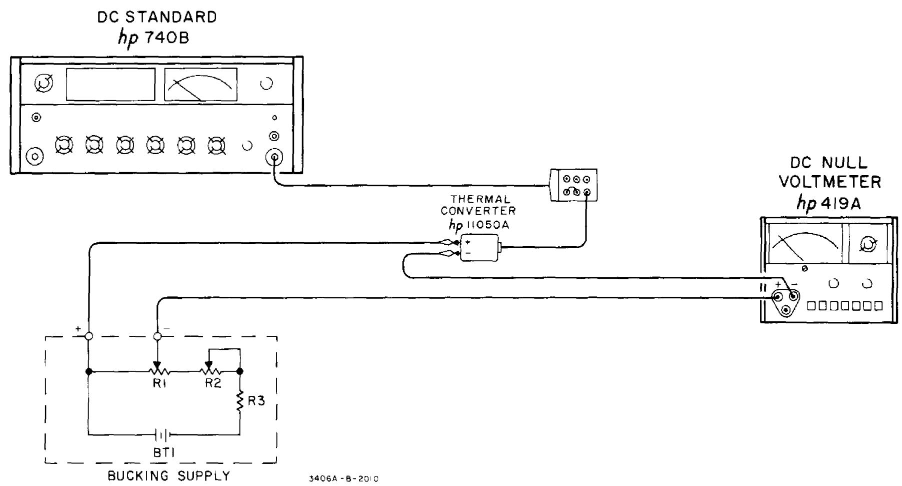


Figure 5-1. Bucking Supply Setup

NOTE

The temperature of the 50 ohm tee should be approximately equal to that of the bare probe.

i. Disconnect the dc standard and connect the oscillator (B) to the power divider. Set the oscillator frequency to 10 kHz.

j. Increase the oscillator amplitude until a null condition is indicated on the null voltmeter. Observe that the 3406A on the 1 V range indicates between 0.77 and 1.03 volts. This verifies an accuracy of ± 13% at 10 kHz.

k. Set the oscillator frequency to 20 kHz and adjust the oscillator amplitude for a null condition. Observe that the 3406A indicates between 0.82 and 0.98 volts. This verifies an accuracy of ± 8% at 200 kHz.

l. Set the oscillator frequency to 25 kHz and adjust the oscillator amplitude for a null condition. Observe that the 3406A indicates between 0.85 and 0.95 volts. This verifies an accuracy of ± 5% at 25 kHz.

m. Set the oscillator frequency to 100 kHz and adjust the oscillator amplitude for a null condition. Observe that the 3406A indicates between 0.87 and 0.93 volts. This verifies an accuracy of ± 3% at 100 kHz.

n. Repeat step m at 1 MHz and 10 MHz to verify 3% accuracy.

o. Note the exact 3406A indication at 10 MHz. Disconnect the 11063A tee from the power divider and connect the 11063A tee to the output of the oscillator. Adjust the oscillator amplitude to the exact reading noted at the beginning of this step.

p. Set a reference on the oscillator.

q. Switch the 3406A to the 3 volt range, and change the oscillator attenuator to increase the output by 10 dB (to + 20 dB setting). The 3406A should indicate between 2.76 and 2.94 volts on the 3 volt scale.

r. Decrease the oscillator output by 10 dB. The 3406A should indicate between 0.91 and 1.09 volts on the 3 volt scale.

s. Switch the 3406A to the next lowest range. The 3406A should indicate between 2.76 and 2.94 on the 3 volt scale.

t. Repeat steps r and s for all ranges of the 3406A. This verifies that range to range tracking is within the 3% accuracy specification. If the 3406A does not track properly at third scale and full scale, perform the tracking adjustment outlined in Paragraph 5-36.

5-9. 10 MHz TO 100 MHz ACCURACY CHECK.

5-10. The 10 MHz to 100 MHz accuracy check uses the following test equipment:

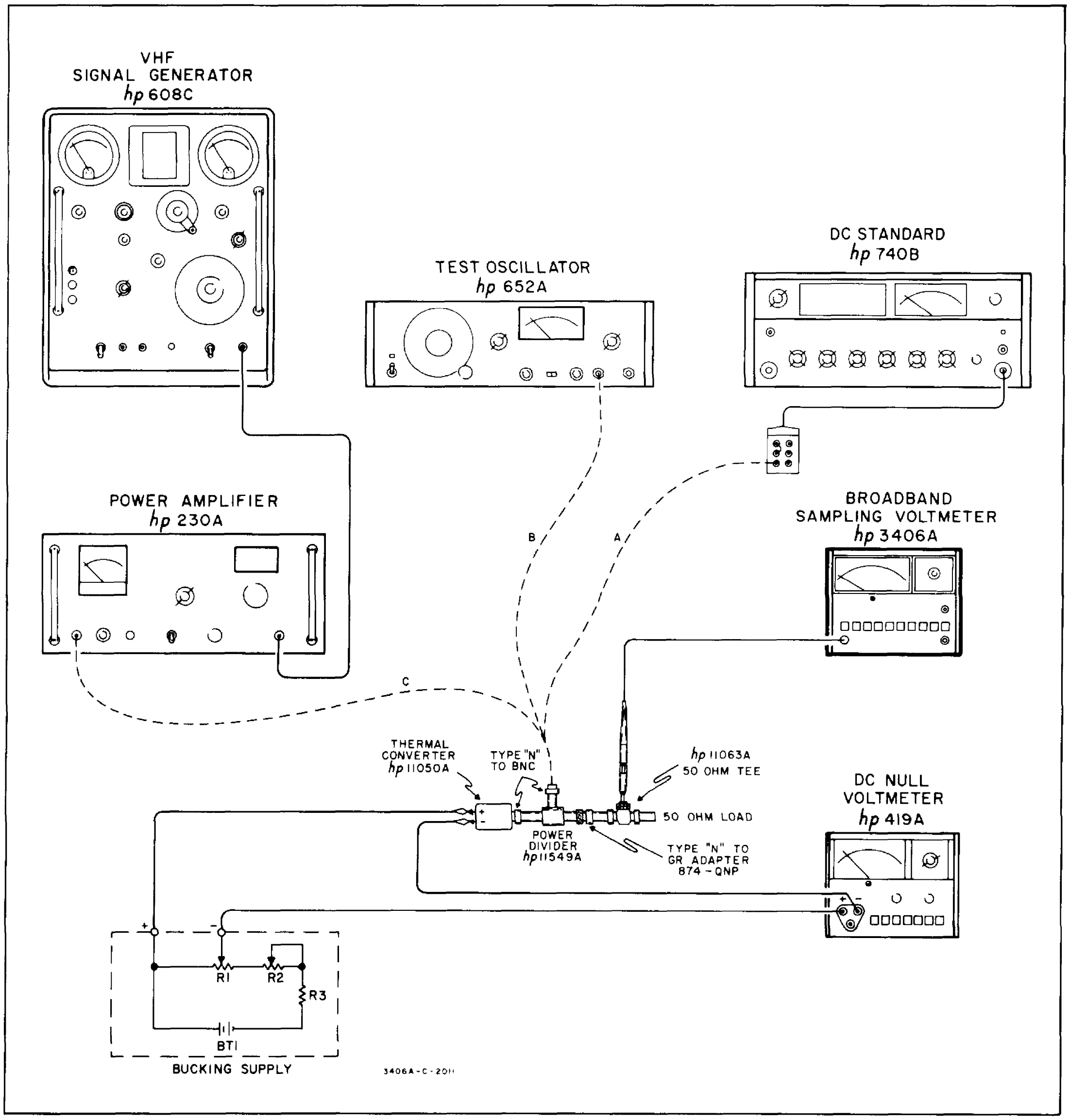


Figure 5-2. Accuracy and Range-to-Range Tracking Test Setup.

VHF Signal Generator (-hp- Model 608C)
Power Amplifier (-hp- Model 230A)
DC Null Voltmeter (-hp- Model 419A)
Power Divider (-hp- Model 11549A)
Thermal Converter (-hp- Model 11050A Option 02)
Bucking Supply (See Table 5-1)

a. Connect the test setup as shown in Figure 5-2, using the VHF Signal Generator (C) as the signal source.

b. This check assumes the bucking supply to be adjusted as in step h of Paragraph 5-8.

c. Set the VHF signal generator to 10 MHz frequency and increase the output amplitude until a null is observed on the null voltmeter. This verifies that 0.900 volts is being applied to the 11063A 50 ohm tee. The 3406A indication should be the same as was noted in step o of Paragraph 5-8.

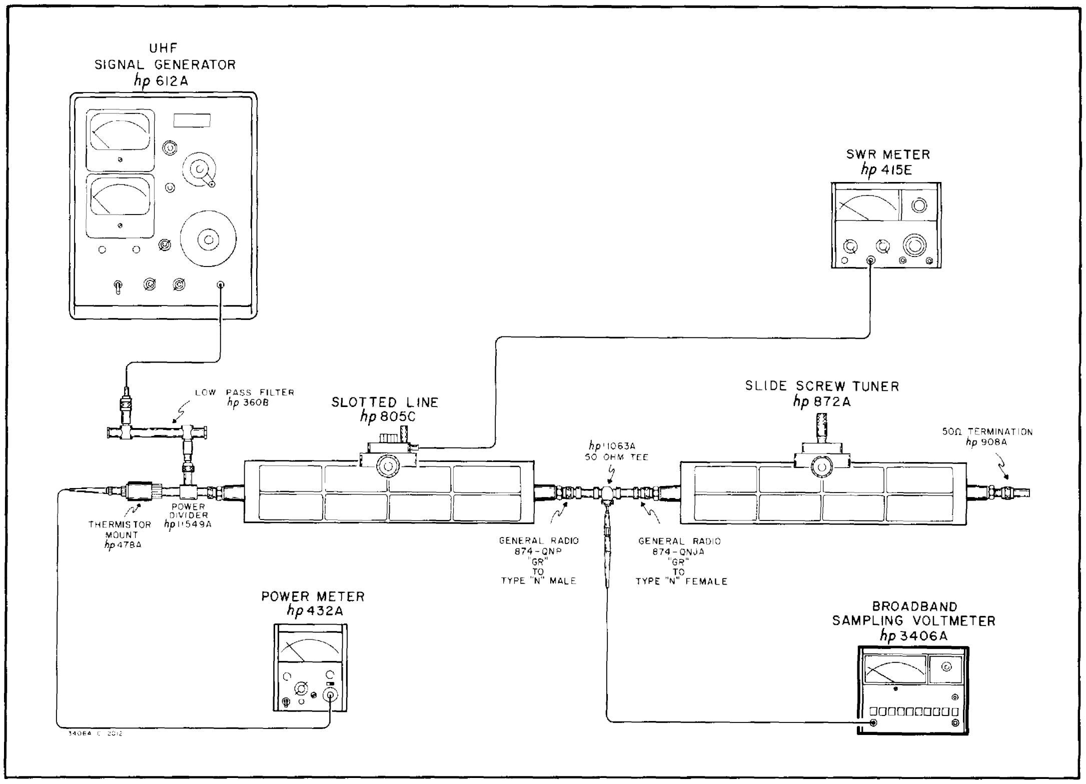


Figure 5-3. 700 MHz to 1.2 GHz Frequency Response Test Setup

d. Set the VHF signal generator frequency to 100 MHz, and adjust the output amplitude for a null indication on the null voltmeter. The 3406A should indicate between 0.87 and 0.93 volts.

NOTE

If the 3406A failed any of the preceding specification limits, perform the adjustment and calibration procedures, Paragraph 5-16.

5-11. 700 MHz TO 1.2 GHz ACCURACY CHECK.

5-12. The 700 MHz to 1.2 GHz accuracy check test setup is shown in Figure 5-3, and requires the following equipment:

Coaxial Slide Screw Tuner (-hp- Model 872A)
Low Pass Filter (-hp- Model 360B)
Slotted Line (-hp- Model 805C)
Power Divider (-hp- Model 11549A)
Thermistor Mount (-hp- Model 478A)
Power Meter (-hp- Model 432A)
SWR Meter (-hp- Model 415E)
UHF Signal Generator (-hp- Model 612A)
50 Ohm Termination (-hp- Model 908A)

a. Connect the equipment as shown in Figure 5-3.

b. Set the UHF signal generator frequency to 700 MHz.

c. Adjust the UHF signal generator output amplitude for a power meter indication of - 10 dBm.

d. Adjust the slide screw tuner for minimum SWR indication on the SWR meter. Then turn off the modulation on the signal generator.

e. Adjust the UHF signal generator output amplitude for a power meter indication of exactly - 10 dBm.

f. The 3406A should indicate between 0.0657 and 0.0757 volts on the 0.1 volt range. This verifies an accuracy of ± 5%.

g. Set the UHF signal generator frequency to 1 GHz and repeat setps c, d, and e. The 3406A should indicate between 0.0650 and 0.0764 volts on the 0.1 volt range.

h. Set the UHF signal generator frequency to 1.2 GHz and repeat steps c, d, and e. The 3406A should indicate between 0.0615 and 0.0799 volts on the 0.1 volt range.

NOTE

If the 3406A failed any of the above steps, perform the adjustment and calibration procedures, Paragraph 5-16.

5-13. INPUT IMPEDANCE CHECK.

a. A test oscillator (-hp- Model 652A) and 44.2 kilohm resistor (-hp- Part No. 0698-4936) are required for this check.

b. With 11072A Isolator attached to probe, connect the Model 3406A to 50 ohm output on oscillator through the resistor as illustrated in Figure 5-4.

c. Depress Model 3406A 1 volt RANGE pushbutton. Set oscillator to 10 kHz, and adjust output for full scale deflection on 3406A meter.

d. Increase frequency of oscillator until Model 3406A indicates .707 volts. Oscillator frequency should be above 200 kHz.

NOTE

Oscillator frequency of 200 kHz indicates total input capacitance of 10 pF. Oscillator frequency of 250 kHz indicates total input capacitance of 8 pF.

e. This verifies input resistance of 100 kilohms with total shunt capacitance of less than 10 pF with the Isolator Tip attached to probe. Section III shows typical input impedance with the Isolator Tip for frequencies above 1 MHz. For relative measurements with the same source resistance the bare probe may be used to achieve higher input impedance.

5-14. METER RESPONSE TIME CHECK.

a. Depress 3406A 1 volt RANGE pushbutton and insert probe into 1 VOLT receptacle.

b. Depress 3406A CAL pushbutton and adjust CALIBRATE control for 1 volt indication on meter.

c. Depress 3 volt RANGE pushbutton and allow few seconds for meter to stabilize.

d. Depress 1 volt RANGE pushbutton and verify meter indicates 1 volt ±3% (between 0.97 and 1.03) within 3 seconds.

e. If the 3406A did not respond within 3 sec perform the meter response adjustment outlined in Paragraph 5-26.

NOTE

In the above procedures a meter jitter of ±1% of reading may be observed. This jitter (small movements of the meter needle over a period of time) is within the specification of the Model 3406A and is caused by the sampling technique used within the 3406A. Refer to Section IV for additional information.

5-15. SAMPLE HOLD OUTPUT NOISE CHECK.

a. Connect an AC Voltmeter (-hp- Model 400E) to SAMPLE HOLD OUTPUT jack on the 3406A rear panel.

b. Position AC Voltmeter range selector to 100 mV range.

c. With CAL pushbutton on 3406A released insert probe into 1 VOLT receptacle.

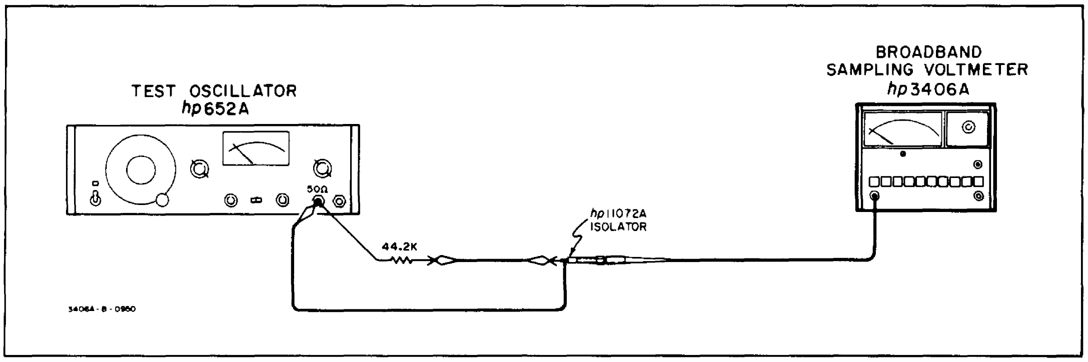
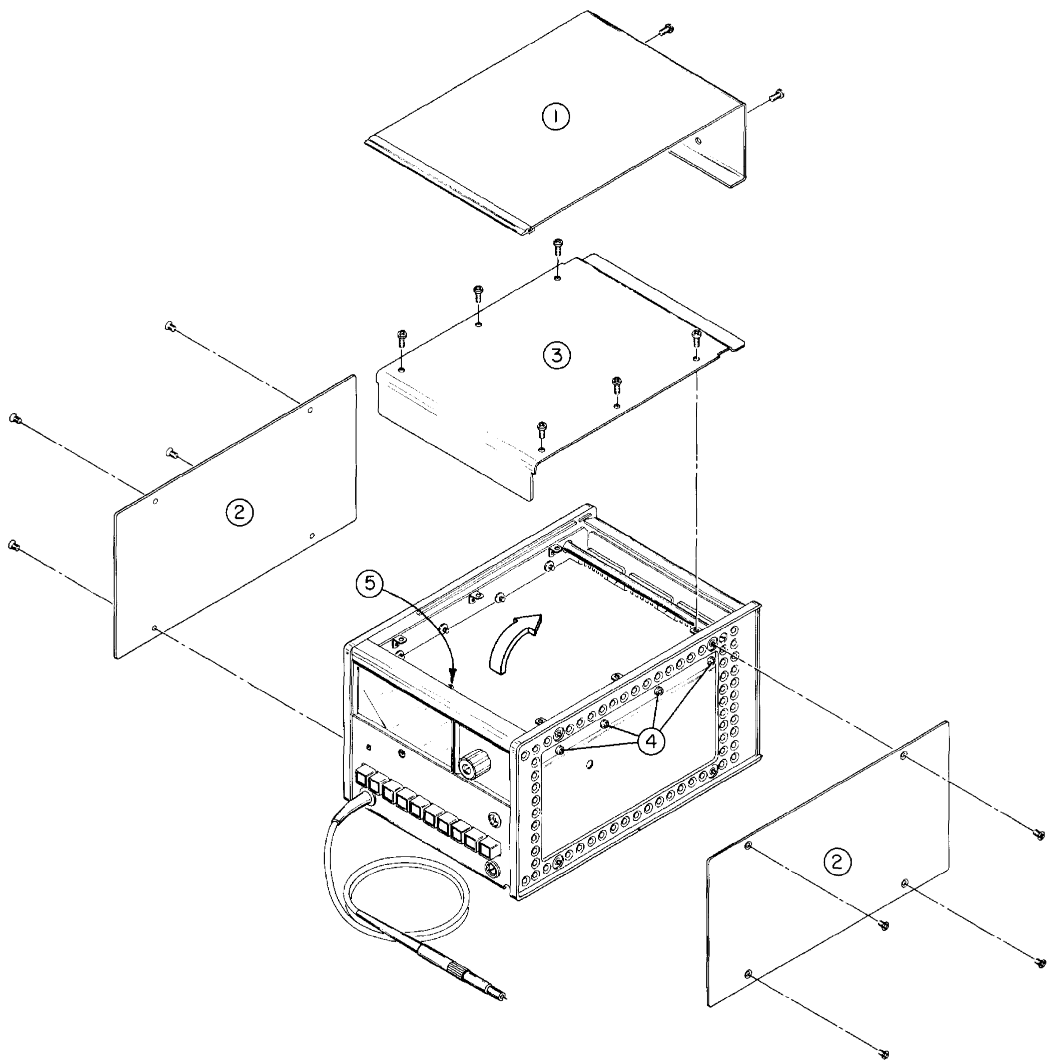


Figure 5-4. Input Impedance Test Setup



(1) Remove top cover. (2) Remove both right and left side covers. NOTE The instrument is now ready for calibration and adjustment. (3) Remove the "top shield" to make the + and - 15 V test points and Top Assembly accessible.	(4) Remove the four screws from both the right and left side gussets. (5) Remove the front-center screw on Top Assembly. NOTE The "hinged chassis" may now be lifted up to make the Bottom Assembly accessible.

Figure 5-5. Preparation for Calibration or Repair

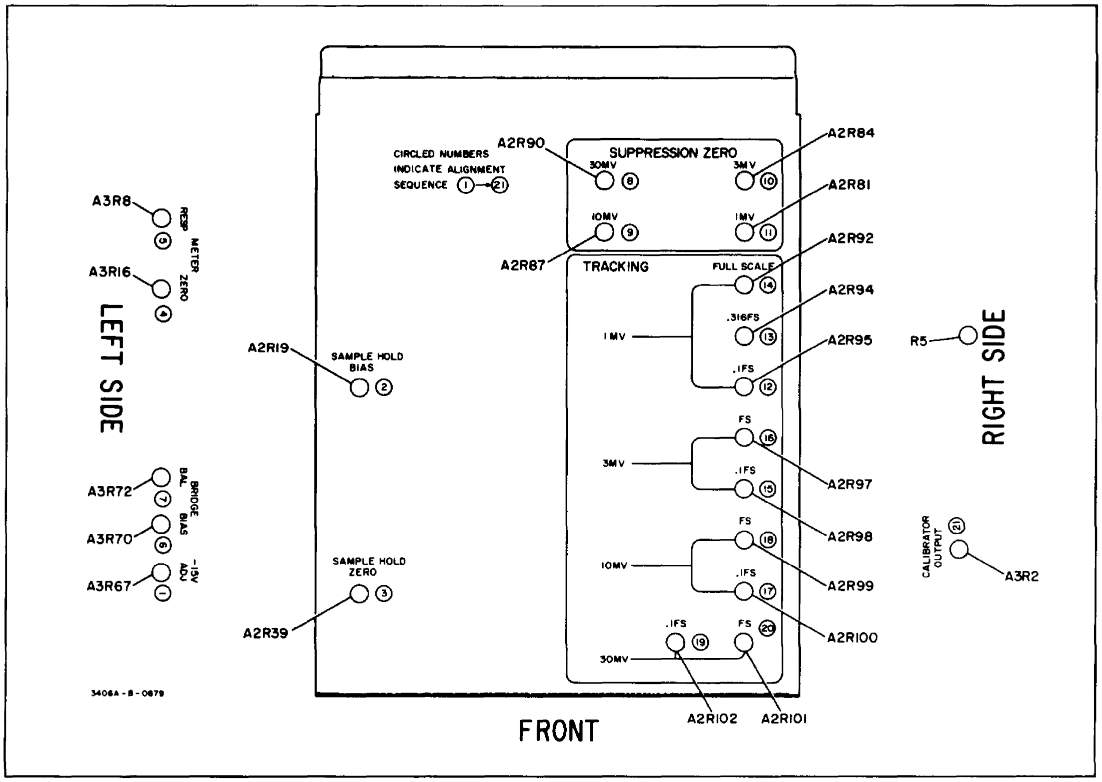


Figure 5-6. Adjust Point Location

d. Depress 3406A .001 V range pushbutton and observe meter indication of less than 55 mV on the voltmeter connected to the SAMPLE HOLD OUTPUT. This verifies the Sample Hold Output noise of less than 175 uV referred to the input.

NOTE

The Sample Hold Output voltage is 316 mV full scale for all ranges. On the 1 mV range 50 dB of gain is provided for Sample Hold Output. Therefore less than 175 uV of noise X 50 dB of gain equal less than 56 mV. See Section III for additional information concerning Sample Hold Output.

5-16. ADJUSTMENT AND CALIBRATION PROCEDURES.

The following paragraphs contain a complete adjustment and calibration procedure for the Model 3406A. This procedure should be conducted only after it has been established that the Model 3406A does not meet its published specifications. Indiscriminate adjustment of internal controls to refine readings may actually cause more difficulty. Two external adjustments, ZERO and CALIBRATE are outlined in Paragraphs 3-8 and 3-9 and should be performed prior to making any measurements or when changing the probe tip.

5-17. PREPARATION FOR CALIBRATION (COVER REMOVAL).

5-18. The 3406A is housed in a box within a box construction to eliminate RF interference. Perform the steps outlined in Figure 5-5 to prepare the instrument for adjustment and calibration.

NOTE

All adjustments are available through the top shield or side gussets as illustrated in Figure 5-6. The circled numbers indicate the alignment sequence (1)-(21). Figure 5-6 also gives the reference designator for each adjustment.

Perform steps (1) thru (5) in Figure 5-5 only if it is necessary to make the components or test points on the top or bottom assemblies accessible.

5-19. MECHANICAL METER ZERO.

a. Turn the instrument on and let it warm up for at least 20 minutes. Turn the instrument off after warmup and allow at least one minute for all capacitors to discharge.

b. Rotate the mechanical zero adjustment screw (Figure 3-1) clockwise until the meter pointer is to the left of zero and begins moving upscale toward zero; stop when the meter pointer is right on zero. If the pointer overshoots, continue turning clockwise and repeat the procedure in this step.

c. After aligning the meter pointer on zero, rotate the zero adjustment screw slightly counterclockwise to free the adjustment screw from the meter suspension. If the meter pointer moves during this step, repeat steps b and c.

5-20. POWER SUPPLY ADJUSTMENT.

5-21. A dc Voltmeter (-hp- Model 740B or 414A) is required for this adjustment.

a. Remove top shield, connect voltmeter between + 15 V and ground test points on Top Assembly A2. (See Figure 7-5 for Test Point location).

b. Adjust ①- 15 V ADJ (A3R67) for + 15 volt (± .2 volts) indication on dc voltmeter.

c. Connect voltmeter between - 15 V and ground test points on Top Assembly A2.

d. DC voltmeter should indicate between - 14.6 and - 15.4 volts.

NOTE

If the power supply is not within the test limits given above perform the regulation and ripple checks outlined in Paragraph 5-57 and 5-58.

5-22. PROBE BALANCE ADJUSTMENT.

5-23. A high frequency oscilloscope (-hp- Model 140A/1402A) is required for this adjustment.

a. Connect 3406A to oscilloscope using probe-to-BNC Adapter (-hp- Model 10218A), part of 11064A Accessory Probe Kit.

b. Set oscilloscope to most sensitive vertical gain and sweep to .1 ms/cm.

c. Depress Model 3406A 3 volt RANGE Pushbutton and adjust ②SAMPLE HOLD BIAS (A2R19) for null on oscilloscope (minimum amplitude of pulses).

d. Depress .001 volt RANGE pushbutton and adjust ①BRIDGE BAL (A3R72) for minimum noise on oscilloscope.

NOTE

Interaction exists between adjustments② and①; therefore repeat steps c and d for optimum noise on both ranges. The probe output noise must be less than 30 mV peak-to-peak.

e. Insert 3406A probe in 1 VOLT receptacle on front panel; release CAL pushbutton; set 3406A to .3 volt RANGE.

f. Connect oscilloscope to SAMPLE HOLD OUTPUT on 3406A rear panel; set oscilloscope vertical sensitivity to 0.2 V/cm and sweep to 1 ms/cm.

g. Adjust ③ SAMPLE HOLD ZERO (A2R39) for minimum amplitude pulses on oscilloscope.

5-24. METER AMPLIFIER ZERO ADJUSTMENT.

5-25. No external test equipment is required for the meter amplifier zero adjustment.

a. Remove top shield, the front, center screw on Top Assembly and lift the hinged chassis up.

b. Short A3TP2 to the ground test point located on the Bottom Assembly A3.

c. Adjust ④ METER ZERO (A3R16) for a zero reading on 3406A meter.

d. Remove short; replace center screw and shield.

5-26. METER AMPLIFIER RESPONSE ADJUSTMENT.

5-27. No external test equipment is required for the meter amplifier response adjustment.

a. Insert 3406A probe in 1 VOLT receptacle; depress CAL and 1 volt RANGE pushbuttons.

b. Adjust front panel ZERO control for 3406A meter deflection of 0.8 (reference point).

c. Release CAL pushbutton; allow 1 minute for the 3406A to stabilize.

NOTE

Meter response is properly adjusted when the meter needle moves rapidly to a point just below the reference point and then moves directly to the reference point in a few small and fast incremental steps. Misadjustment is indicated by overshoot or many incremental steps.

d. Depress CAL pushbutton; note response of meter. If misadjustment of the response is indicated, proceed with the following steps.

e. Release CAL pushbutton; allow 1 minute for the 3406A to stabilize.

f. Rotate (5) METER RESP (A3R8) a small amount in a CW direction.

g. Depress CAL pushbutton; note response of meter. If response improved from step d, repeat steps e and f. If response is worse than the response in step d, repeat steps e and f, rotating (5) METER RESP in a CCW direction.

5-28. FREQUENCY RESPONSE ADJUSTMENT.

5-29. The test setup illustrated in Figure 5-3 is required for the frequency response adjustment. The following equipment is required for this adjustment:

Coaxial Slide Screw Tuner, (-hp- Model 872A)
Low Pass Filter, (-hp- Model 360B)
Power Meter, (-hp- Model 432A/478A)
50 Ohm Terminations, (-hp- Model 908A)
UHF Signal Generator, (-hp- Model 612A)
Slotted Line, (-hp- Model 805C)
SWR Meter, (-hp- Model 415E)
Power Divider, (-hp- Model 11549A)

a. Connect test setup illustrated in Figure 5-3.

NOTE

A slotted line (-hp- Model 805C) and an SWR meter (-hp- Model 415E) must be used to adjust the slide screw tuner for a minimum standing-wave ratio. Refer to the slide screw tuner Operating Note for proper setup.

b. Set the UHF signal generator frequency to 1 GHz.

c. Adjust the UHF signal generator output amplitude for a power meter indication of - 10 dBm.

d. Adjust the slide screw tuner for minimum SWR indication on the SWR meter.

e. Adjust the UHF signal generator output amplitude for a power meter indication of exactly - 10 dBm.

f. Note the exact reading of the 3406A on the .1 volt range.

g. Set the UHF signal generator frequency to 500 MHz.

h. Repeat steps c, d, and e of this paragraph.

i. Adjust (1) Bridge Bias (A3R70) for a 3406A indication of 0.5% higher than the reading noted in step f.

NOTE

Setting Bridge Bias for 0.5% high at 500 MHz ensures a better response over the entire 3406A frequency range.

j. Repeat steps b thru e. If the 3406A indication is not within 1% of the final reading in step i, repeat steps b thru j.

5-30. SIGNAL AMPLIFIER GAIN ADJUSTMENT. (FRONT PANEL).

NOTE

This adjustment sets the Signal Amplifier gain, and verifies proper operation of the Sampling Probe Assembly A1, Pulse Generator p/o A3, Signal Amplifier p/o A2, and Sample Hold Circuit p/o A2. The SAMPLE HOLD OUTPUT is used because of its location in the circuit.

5-31. The test setup illustrated in Figure 5-2 is used for the signal amplifier gain adjustment. The following test equipment is required:

DC Standard, (-hp- Model 740B)
Oscillator, (-hp- Model 652A)
1 Volt Thermal Converter, (-hp- Model 11050A Opt 02)
Null Voltmeter, (-hp- Model 419A)
Bucking Supply, (see Table 5-1)
Power Divider (-hp- Model 11549A)

a. Set up the bucking supply as described in Paragraph 5-8, steps b thru h.

CAUTION

DO NOT ALLOW DC STANDARD OR OSCILLATOR OUTPUT TO EXCEED RATED INPUT OF THERMAL CONVERTER. ANY OVERLOAD MAY DESTROY THERMAL CONVERTER.

b. Disconnect the dc standard and connect the oscillator (B) to the power divider. Set the oscillator frequency to 1 MHz.

c. Adjust the oscillator amplitude until a null condition is indicated on the null voltmeter.

d. Connect an ac voltmeter to the SAMPLE HOLD OUTPUT.

e. Adjust CALIBRATE on 3406A front panel for a SAMPLE HOLD OUTPUT of 0.284 volts on ac voltmeter.

NOTE

This procedure establishes the 1 volt reference used to set the calibrator (1 VOLT receptacle) output voltage in Paragraph 5-38.

5-32. BRIDGE BALANCE ADJUSTMENT.

5-33. Because of interaction existing between adjustment ②and⑦, the BRIDGE BAL⑦was set in Paragraph 5-22. No additional adjustment is required.

5-34. METER SUPPRESSION ZERO ADJUSTMENTS.

5-35. To properly ground the probe for this adjustment, the 11063A Tee with 50 ohm termination is used.

a. Center the front panel ZERO control, and insert the 3406A probe into terminated 50 ohm tee with no source connected.

b. Depress 3406A 0.03 volt RANGE pushbutton and adjust ⑧ 30 MV (A2R90) for zero indication on 3406A meter.

c. Depress 0.01 volt RANGE pushbutton and adjust ⑨10 MV (A2R87) for zero indication on 3406A meter.

d. Depress 0.003 volt RANGE pushbutton and adjust ⑩ 3 MV (A2R84) for zero indication on 3406A meter.

e. Depress 0.001 volt RANGE pushbutton and adjust ⑪1 MV (A2R81) for zero indication on 3406A meter.

5-36. NON-LINEAR METER AMPLIFIER ADJUSTMENT. (Tracking).

5-37. An oscillator (-hp- Model 651B or 652A) and an ac voltmeter (-hp- Model 400E) are required for this adjustment.

a. Using the front panel ZERO control, zero the voltmeter on the .001, .003, .01, .03 and .1 volt RANGE with the bare probe inserted into terminated 50 ohm tee with no source.

b. Connect test setup illustrated in Figure 5-7; set oscillator output frequency to 1 MHz; and output voltage for .100 V indication on ac voltmeter, using - 10 dBm setting on the oscillator.

c. Using oscillator attenuator reduce output 60 dB; and depress 3406A .001 volt RANGE pushbutton.

d. Adjust ⑫ .1 FS (A2R95) for 3406A meter indication of 0.1 on 0 to 1 scale.

e. Using oscillator attenuator increase output 10 dB; adjust ⑬ .316 FS (A2R94) for 3406A meter indication of 1 on 0 to 3 scale (which is 0.316 of FS on 0 to 1 scale).

f. Using oscillator attenuator increase output 10 dB; adjust ⑭FULL SCALE (A2R92) for 3406A meter indication of 1 on 0 to 1 scale.

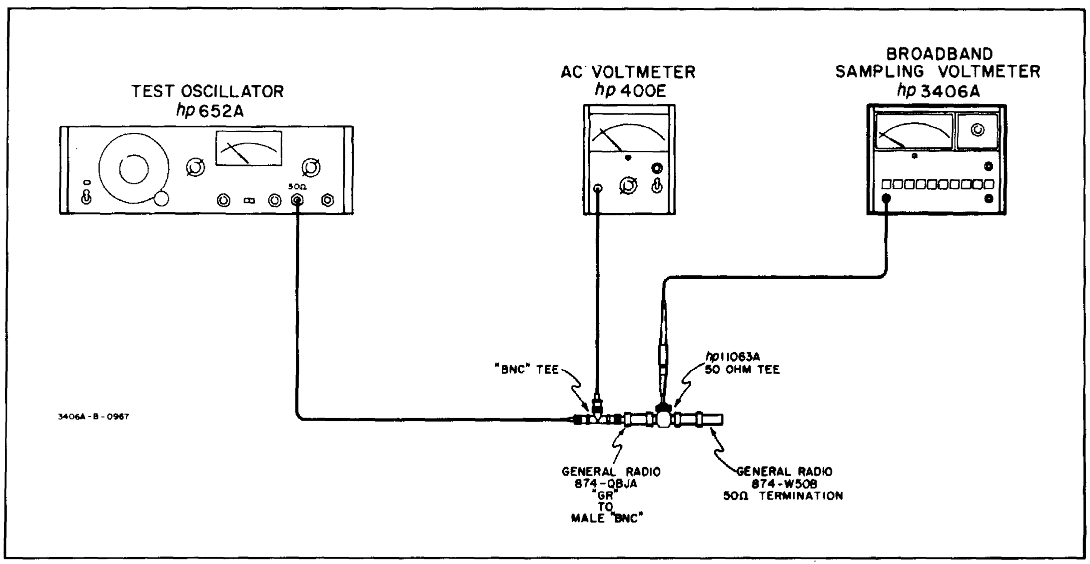


Figure 5-7. Tracking Adjustment Setup

g. Depress 3406A .003 volt RANGE pushbutton.

h. Using oscillator attenuator reduce output 10 dB; adjust (15) .1 FS (A2R98) for 3406A meter indication of 0.1 on 0 to 1 scale.

i. Using oscillator attenuator increase output 20 dB; adjust (16) FS (A2R97) for 3406A meter indication of 1 on 0 to 1 scale.

j. Depress 3406A .01 volt RANGE pushbutton.

k. Using oscillator attenuator reduce output 10 dB; adjust (17) .1 FS (A2R100) for 3406A meter indication of 1 on 0 to 1 scale.

m. Using oscillator attenuator increase output 20 dB; adjust (18) FS (A2R99) for 3406A meter indication of 1 on 0 to 1 scale.

n. Depress 3406A .03 volt RANGE pushbutton.

o. Using oscillator attenuator reduce output 10 dB; adjust (19) .1 FS (A2R102) for 3406A meter indication of 0.1 on 0 to 1 scale.

p. Using oscillator attenuator increase output 20 dB; adjust (20) FS (A2R101) for 3406A meter reading of 1 on 0 to 1 scale.

q. Depress 3406A 3 volt RANGE pushbutton and increase oscillator output by 40 dB; adjust 3 V FS (variable resistor R5 on right side of chassis) for 3406A meter reading of 1 on 0 to 1 scale.

NOTE

Interaction exists between the 3 V FS adjustment and the 1 volt range. Check full scale deflection on 1 volt range and optimize adjustment of 3 V FS if necessary.

NOTE

Because of interaction between tracking and suppression zero adjustments, repeat meter suppression zero adjustment (Paragraph 5-34) and steps a thru q of this procedure until no interaction is noted between the two adjustments.

5-38. CALIBRATOR OUTPUT ADJUSTMENT.

5-39. No external equipment is required for the calibrator output adjustment.

NOTE

The Signal Amplifier Gain Adjustment, Paragraph 5-30, establishes the 1 volt reference used to adjust the calibrator output in the following procedure. Perform the procedure outlined in Paragraph 5-30 prior to adjusting CALIBRATOR OUTPUT.

a. Insert 3406A probe with 11072A Isolator tip in 1 VOLT receptacle on front panel; depress 1 volt RANGE pushbutton; depress CAL pushbutton.

b. Adjust (21) CALIBRATOR OUTPUT (A3R2) for 3406A meter indication of 1 on 0 to 1 scale.

NOTE

The remaining steps of this procedure are not necessary unless part or all of the calibrator assembly has been replaced.

c. Note the exact reading obtained in step b after adjustment.

d. Remove the 11072A Isolator tip and insert the bare probe into the 1 VOLT receptacle, and depress CAL pushbutton.

e. Adjust hex head screw in calibrator case (bottom panel must be removed) the meter indication obtained is the same as noted in step c.

5-40. SERVICING.

5-41. The Model 3406A is housed in a box within a box construction to eliminate RFI (radio frequency interference). Within the 3406A the side gussets, the "hinged chassis" and all shields have been treated with a chromate conversion coating. Chromate conversion puts a highly conductive coating on the aluminum metal to prevent oxidation, and aid in shielding against RFI. Avoid scratching the treated surface when working with the Model 3406A.

5-42. GROUNDING.

5-43. Because of the high current pulses used within the Model 3406A and RFI, proper grounding is important. Always note the location of the "ground point" before unsoldering any ground leads, and resolder any ground leads to the same "ground point".

5-44. ETCHED CIRCUIT BOARDS.

5-45. Excluding the Sampling Probe Assembly, the Model 3406A Voltmeter contains two large etched circuit boards designated Top and Bottom Assemblies. The two boards are mounted on the "hinged chassis", illustrated in Table 7-1, with 11 screws -hp- Part No. 2390-0001. An -hp- part number is silk screened on both circuit boards to identify them. See the component location diagrams in Section VII for location of each component mounted on the boards and Section VI for component replacement information.

5-46. The etched circuit boards are a plated-through type. The electrical connections between sides of the board are made with a layer of metal plated-through the component lead holes. To prevent damage to the circuit boards and components, observe the following when soldering.

a. Use a low-heat (25 to 50 watt) small-tip soldering iron, and a small diameter rosin core solder.

b. To remove a component, clip a heat sink (long nose pliers, commercial heat sink tweezers etc.) on the component lead as close to the component as possible. Place the soldering iron directly on the component lead, and pull up on the lead. If a component is obviously damaged or faulty, clip the leads close to the component and then remove the leads from the board.

CAUTION

EXCESSIVE OR PROLONGED HEAT CAN LIFT THE CIRCUIT FOIL FROM THE BOARD OR CAUSE DAMAGE TO COMPONENTS.

c. Clean the component lead holes by heating the solder in the hole, quickly removing the soldering iron, and inserting a pointed, non-metallic object such as a toothpick.

d. To mount a new component, shape the leads and insert them in the holes. Clip a heat sink on the component, heat with the soldering iron, and add solder as necessary to obtain a good electrical connection.

e. Clip excess leads off after soldering and clean excess flux from the connection and adjoining area, using type TF Freon (-hp- Part No. 8500-0232).

5-47. To avoid surface contamination the etched circuit boards may be cleaned as outlined below:

a. Clean with a solution of "Finish Dishwashing Detergent" and warm water.

b. Rinse thoroughly with clean water and immediately dry.

CAUTION

WHEN USING ANY ANTI-HUMIDITY SPRAY, THE ETCHED CIRCUIT BOARD PINS AND/OR EXTERNAL CONNECTIONS MUST BE COVERED WITH TAPE OR EQUIVALENT MATERIAL. OTHERWISE, AN INADEQUATE ELECTRICAL CONNECTION TO THE APPROPRIATE CONNECTOR WILL RESULT.

c. When completely dry and properly masked, spray lightly with "General Electric Dry Film 88" anti-humidity spray.

d. The teflon insulators in the gate circuit of the Field Effect Transistors, A2Q7 and A3Q3 should be cleaned with fine wire brush and sprayed lightly with "General Electric Dry Film 88" anti-humidity spray.

5-48. SERVICING THE PROBE ASSEMBLY.

5-49. The Sampling Probe Assembly A1 is not a field repairable item. A complete assembly is available on exchange basis (-hp- Part No. 03406-62103) through your local Hewlett-Packard Sales and Service Office. Perform the steps outlined in Paragraph 5-52 to verify proper operation of Sampling Probe Assembly before ordering replacement probe.

5-50. TROUBLESHOOTING PROCEDURES.

5-51. The following troubleshooting procedures are designed to assist in isolating a malfunction(s) within the Model 3406A voltmeter. These procedures should be undertaken only after it has been established that the difficulty cannot be eliminated by the Adjustment and Calibration Procedures outlined in Paragraph 5-16. An investigation should also be made to ensure that the trouble is not a result of conditions external to the Model 3406A voltmeter.

NOTE

In addition to the following paragraphs, waveforms and DC voltage levels are shown on the schematic diagrams in Section VII. Use the Functional Block Diagram, Figure 7-1, to isolate the trouble to a particular block.

5-52. TROUBLESHOOTING THE PROBE ASSEMBLY.

5-53. The Sampling Probe is not a field repairable item. A complete assembly is available on an exchange basis under -hp- Part No. 03406-62103. Therefore, the first step in troubleshooting the Model 3406A is to determine if the Sampling Probe is operating properly. To verify proper operation of the probe perform the following steps:

a. Insert the probe into 1 VOLT receptacle and depress CAL and 1 volt RANGE pushbuttons.

b. Remove two Phillips head screws at the rear of the bottom cover and remove the bottom cover.

c. Remove 13 Phillips head screws from the bottom shield and remove it.

d. Remove the probe filter (FL2) shield (Refer to Figure 7-2).

A

White wire on FL2
Ac coupled
Vert = 5 V/cm
Sweep = 1 usec/cm

B

Blue wire on FL2
Ac coupled
Vert = .1 V/cm
Sweep = 2 usec/cm

NOTE

To stop the waveform from sweeping back and forth short the collector of A3Q6 on the Bottom Assembly to ground (see Figure 7-8 for location of A3Q6).

Figure 5-8. Troubleshooting the Probe

NOTE

The following identifies the function of each signal entering or leaving the Sample Probe Assembly through the Probe Filter Assembly (FL2). The color of wires connected to the feedthrough capacitors on the Probe Filter Assembly are the same on the top as they are on the bottom. (Refer to Figure 7-2).

Wire Color	Function
White	Pulse from Pulse Generator
Orange	Feedback to FET from Stabilizing Ampl
Blue	Sampling probe output
Green	Meter hold
Red	Bridge bias and balance (pos)
Brown	Bridge bias and balance (neg)
Yellow	FET drain

CAUTION

WHEN TAKING MEASUREMENTS AT THE PROBE FILTER DO NOT SHORT OUT THE WHITE WIRE. IF THE WHITE WIRE IS SHORTED DAMAGE TO THE PULSE GENERATOR WILL RESULT.

e. Connect an oscilloscope to the white wire connection on FL2 (J10 pin 6).

f. On oscilloscope observe the pulse from Pulse Generator as shown in Figure 5-8A.

NOTE

If no pulse is present check the wiring from the top of the probe filter FL2 to the Interconnecting Assembly, and troubleshoot the Pulse Generator circuit shown in Figure 7-8.

g. Connect oscilloscope to the blue wire connection on FL2, J10 pin 8, (sampling probe output).

h. Observe the "tear drop" waveform shown in Figure 5-8B.

NOTE

If the "tear drop" waveform is present the Sampling Probe is operating properly. If waveform A in Figure 5-8 (step e) is present and waveform B (step g) is not present, replace the Sampling Probe Assembly according to the following procedures.

5-54. REPLACING THE PROBE ASSEMBLY.

CAUTION

DO NOT PLUG OR UNPLUG THE PROBE FROM J10 WITH THE 3406A TURNED ON.

a. Repeat steps b, c and d in Paragraph 5-53.

b. Unplug the probe connector from its 12 pin connector J10. (See Figure 7-2).

c. Remove the cable clamp (MP17 Figure 6-2) just above the probe connector.

d. Loosen the extra-long nut (MP8 Figure 6-2) behind the front panel, and slide it off the cable. The nut is notched on the inside so that the plug on Probe Connector can fit through it.

e. Save the cable clamp and nut removed in steps c and d for installing the new probe assembly.

f. Install the new or exchange probe assembly by reversing steps d thru a. When replacing the bottom shield, it is important that all of the screws are replaced. Any open holes will degrade the RFI protection provided by the shield.

NOTE

After replacing the Probe Assembly a complete Adjustment and Calibration is required as outlined in Paragraph 5-16.

5-55. TROUBLESHOOTING THE POWER SUPPLY.

5-56. In addition to the DC voltage levels given on the power supply schematic, power supply regulation and ripple checks are given below.

5-57. POWER SUPPLY REGULATION CHECK.

a. A variable power-line transformer (Superior Type VC1M) and dc voltmeter (-hp- Model 414A) are required for this check.

b. On the Top Assembly (A2) connect dc voltmeter between + 15 volt test point and the ground test point.

c. With the variable power-line transformer vary the line voltage + and - 10% and verify that the dc voltmeter does not vary by more than 0.2 volts from the dc voltage at normal line.

5-58. POWER SUPPLY RIPPLE CHECK.

a. An oscilloscope (-hp- Model 140A/1402A) is required for this test.

b. On the top Assembly (A2) connect an oscilloscope between + 15 volt test point and the ground test point.

c. At normal-line voltage the pulses observed on oscilloscope should be less than 150 mV peak-to-peak.

5-59. TROUBLESHOOTING USING SAMPLE HOLD OUTPUT.

5-60. Because of the circuit location of SAMPLE HOLD OUTPUT Jack, it may be used to divide the instrument into two different sections and isolate the trouble to one of the two sections. Refer to Figure 6-1 for location of SAMPLE HOLD OUTPUT jack. The section prior to SAMPLE HOLD OUTPUT includes the Sample Hold Circuit, Signal Amplifier, Stabilizing Amplifier, Sampling Probe and Pulse Generator Circuits. The section after the SAMPLE HOLD OUTPUT includes the Detector Circuit, Non-linear Meter Circuit and the Meter Damping/Hold Circuit.

5-61. With a known one volt at 1 MHz into the 3406A as outlined in Paragraphs 5-30 and 5-31 the SAMPLE HOLD OUTPUT should be 0.316 ± 3%.

NOTE

With a 1 volt sine wave input the SAMPLE HOLD OUTPUT reading will read the same on a true rms voltmeter as it will on an average responding voltmeter. However, if a nonsinusoidal wave is measured the true rms voltmeter reading will differ from the average responding voltmeter.

5-62. TROUBLESHOOTING THE STABILIZING AMPLIFIER LOOP.

5-63. To verify proper operation of the Stabilizing Amplifier perform the following steps.

a. Insert the probe into 1 VOLT receptacle with the CAL pushbutton released (out).

b. Depress .3 volt RANGE pushbutton.

c. Connect oscilloscope to SAMPLE HOLD OUTPUT jack, and set vertical sensitivity to 0.2 V/cm.

d. The SAMPLE HOLD OUTPUT observed on oscilloscope should be a single trace free of any excessive 10 Hz component.

e. Turning the SAMPLE HOLD BIAS both clockwise and counterclockwise, pulses should appear but decline back to a single trace.

NOTE

If the proper results are not received in step e check the continuity of orange wire between Probe Filter FL2 and Interconnecting Assembly before troubleshooting the Stabilizing Amplifier.

5-64. FACTORY SELECTED COMPONENTS.

5-65. The value of A2R28* is factory selected to match the gain of A2Q7 to the Sampling Probe Assembly. Following is the procedure for selecting value of A2R28*: (1.5 K to 3.3 K).

a. ZERO the Model 3406A according to Paragraph 3-8.

b. Insert bare probe into 1 VOLT receptacle; depress 1 volt RANGE and CAL pushbuttons.

c. Adjust front panel CALIBRATE control from full CW to full CCW and note meter indication at both ends. Range of CALIBRATE control should be from .8 to 1.2 V.

d. Decrease value of A2R28 to obtain .8 reading; increase value of A2R28 to obtain 1.2 reading.

PERFORMANCE CHECK TEST CARD

Hewlett-Packard Model 3406A
Broadband Sampling Voltmeter
Serial No. ____________________

Tests Performed By ____________________
Date ____________________

DESCRIPTION	CHECK	
	Indication	Specification
Range-to-Range Tracking and 10 kHz to 10 MHz Accuracy.		
10 kHz	0.77 to 1.03 volts	________
20 kHz	0.82 to 0.98 volts	________
25 kHz	0.85 to 0.95 volts	________
100 kHz	0.87 to 0.93 volts	________
1 MHz	0.87 to 0.93 volts	________
10 MHz	0.87 to 0.93 volts	________
3 V range (+ 20 dB setting)	2.76 to 2.94 (3 scale)	________
3 V range (+ 10 dB setting)	0.91 to 1.09 (3 scale)	________
1 V range (+ 10 dB setting)	2.76 to 2.94 (3 scale)	________
1 V range (0 dB setting)	0.91 to 1.09 (3 scale)	________
.3 V range (0 dB setting)	2.76 to 2.94 (3 scale)	________
.3 V range (- 10 dB setting)	0.91 to 1.09 (3 scale)	________
.1 V range (- 10 dB setting)	2.76 to 2.94 (3 scale)	________
.1 V range (- 20 dB setting)	0.91 to 1.09 (3 scale)	________
.03 V range (- 20 dB setting)	2.76 to 2.94 (3 scale)	________
.03 V range (- 30 dB setting)	0.91 to 1.09 (3 scale)	________
.01 V range (- 30 dB setting)	2.76 to 2.94 (3 scale)	________
.01 V range (- 40 dB setting)	0.91 to 1.09 (3 scale)	________
.003 V range (- 40 dB setting)	2.76 to 2.94 (3 scale)	________
.003 V range (- 50 dB setting)	0.91 to 1.09 (3 scale)	________
.001 V range (- 50 dB setting)	2.76 to 2.94 (3 scale)	________
.001 V range (- 60 dB setting)	0.91 to 1.09 (3 scale)	________
10 MHz to 100 MHz Accuracy Check		
10 MHz	0.87 to 0.93 volts	________
100 MHz	0.87 to 0.93 volts	________
700 MHz to 1.2 GHz Accuracy Check		
700 MHz	0.0657 to 0.0757 volts	________
1 GHz	0.0650 to 0.0764 volts	________
1.2 GHz	0.0615 to 0.0799 volts	________
Input Impedance Check	greater than 200 kHz	________
Meter Response Time Check	less than 3 seconds	________
Sample Hold Output Noise Check	less than 55 mV	________

SECTION VI
REPLACEABLE PARTS

6-1. INTRODUCTION.

6-2. This section contains information for ordering replacement parts. Table 6-1 lists parts in alphameric order of their reference designators and indicates the description, -hp- part number of each part, together with any applicable notes, and provides the following:

a. Total quantity used in the instrument (TQ column). The total quantity of a part is given the first time the part number appears.

b. Description of the part. (See list of abbreviations below.)

c. Typical manufacturer of the part in a five-digit code. (See Appendix A for list of manufacturers.)

d. Manufacturer's part number.

6-3. Figure 6-1 illustrates the attaching hardware for Model 3406A; Figure 6-2 illustrates the -hp- Modular Cabinet, and Figure 6-3 illustrates the replaceable mechanical parts used in Model 3406A. Miscellaneous and mechanical parts not listed in one of the three figures are listed at the end of Table 6-1 under "Miscellaneous". Figure 6-4 identifies all the Model 3406A accessories in both probe kits and gives the model number or part number for each item.

6-4. ORDERING INFORMATION.

6-5. To obtain replacement parts, address order or inquiry to your local Hewlett-Packard Field Office. (See Appendix B for list of office locations.) Identify parts by their Hewlett-Packard part numbers.

6-6. NON-LISTED PARTS.

6-7. To obtain a part that is not listed, include:

a. Instrument model number.

b. Instrument serial number.

c. Description of the part.

d. Function and location of the part.

DESIGNATORS

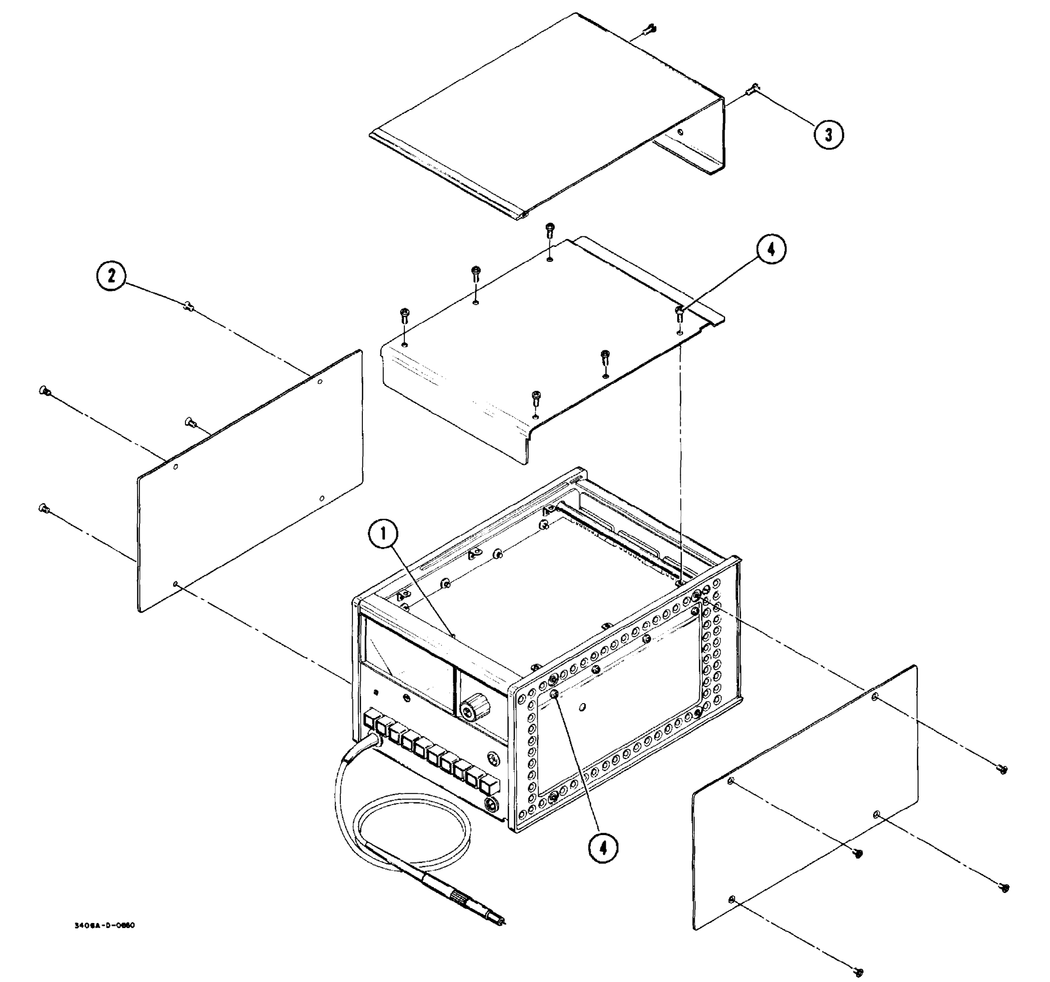
A	= assembly	F	= fuse	MP	= mechanical part	TC	= thermocouple
B	= motor	FL	= filter	P	= plug	V	= vacuum tube, neon bulb, photocell, etc.
BT	= battery	HR	= heater	Q	= transistor		
C	= capacitor	IC	= integrated circuit	QCR	= transistor-diode	W	= cable
CR	= diode	J	= jack	R	= resistor	X	= socket
DL	= delay line	K	= relay	RT	= thermistor	XDS	= lampholder
DS	= lamp	L	= inductor	S	= switch	XF	= fuseholder
E	= misc electronic part	M	= meter	T	= transformer	Z	= network

ABBREVIATIONS

Ag	= silver
Al	= aluminum
A	= ampere (s)
Au	= gold
C	= capacitor
cer	= ceramic
coef	= coefficient
com	= common
comp	= composition
conn	= connection
dep	= deposited
DPDT	= double-pole double-throw
DPST	= double-pole single-throw
elect	= electrolytic
encap	= encapsulated
F	= farad (s)
FET	= field effect transistor
fxd	= fixed
GaAs	= gallium arsenide
GHz	= gigahertz = 10^{+9} hertz
gd	= guard (ed)
Ge	= germanium
grd	= ground (ed)
H	= henry (ies)
Hg	= mercury
Hz	= hertz (cycle (s) per second)
ID	= inside diameter
impg	= impregnated
incd	= incandescent
ins	= insulation (ed)
kΩ	= kilohm (s) = 10^{+3} ohms
kHz	= kilohertz = 10^{+3} hertz
L	= inductor
lin	= linear taper
log	= logarithmic taper
m	= milli = 10^{-3}
mA	= milliampere (s) = 10^{-3} amperes
MHz	= megahertz = 10^{+6} hertz
MΩ	= megohm (s) = 10^{+6} ohms
met flm	= metal film
mfr	= manufacturer
mtg	= mounting
mV	= millivolt (s) = 10^{-3} volts
μ	= micro = 10^{-6}
μV	= microvolt (s) = 10^{-6} volts
my	= Mylar ®
nA	= nanoampere (s) = 10^{-9} amperes
NC	= normally closed
Ne	= neon
NO	= normally open
NPO	= negative positive zero (zero temperature coefficient)
ns	= nanosecond (s) = 10^{-9} seconds
nsr	= not separately replaceable
Ω	= ohm (s)
obd	= order by description
OD	= outside diameter
p	= peak
pc	= printed circuit
pF	= picofarad (s) = 10^{-12} farads
piv	= peak inverse voltage
p/o	= part of
pos	= position (s)
poly	= polystyrene
pot	= potentiometer
p-p	= peak-to-peak
ppm	= parts per million
prec	= precision (temperature coefficient, long term stability, and/or tolerance)
R	= resistor
Rh	= rhodium
rms	= root-mean-square
rot	= rotary
Se	= selenium
sect	= section (s)
Si	= silicon
sl	= slide
SPDT	= single-pole double-throw
SPST	= single-pole single-throw
Ta	= tantalum
TC	= temperature coefficient
TiO_2	= titanium dioxide
tog	= toggle
tol	= tolerance
trim	= trimmer
TSTR	= transistor
V	= volt (s)
vacw	= alternating current working voltage
var	= variable
vdcw	= direct current working voltage
W	= watt (s)
w/	= with
wiv	= working inverse voltage
w/o	= without
ww	= wirewound
*	= optimum value selected at factory, average value shown (part may be omitted)
**	= no standard type number assigned (selected or special type)

REV G

® Dupont de Nemours



INDEX NO.	DESCRIPTION	QTY.	-hp- PART NO.
1	Screw Mach 6-32 x 9/16 SS, with washer lock -hp- Part No. 2190-0008	1	2460-0028
2	Screw Mach SS 6-32 x 3/16	8	2370-0020
3	Screw Mach SS 6-32 x 1/4	2	2460-0010
4	Screw Mach SS 6-32 x 1/4 with washer-lock -hp- Part No. 2190-0008	14	2460-0010

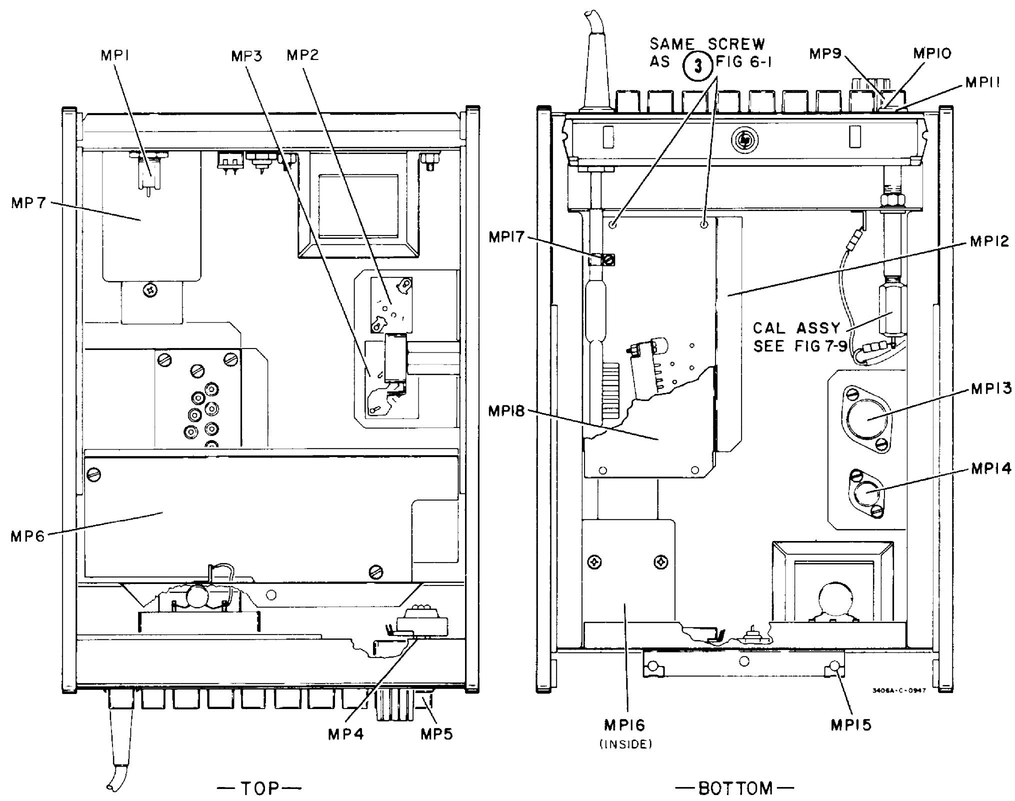
NOTE

The bottom cover is attached with two screws, Mach FH SS 6-32 x 5/16, -hp- Part No. 2370-0016. The bottom shield is attached with 13 screws the same type as No.④above.

CAUTION

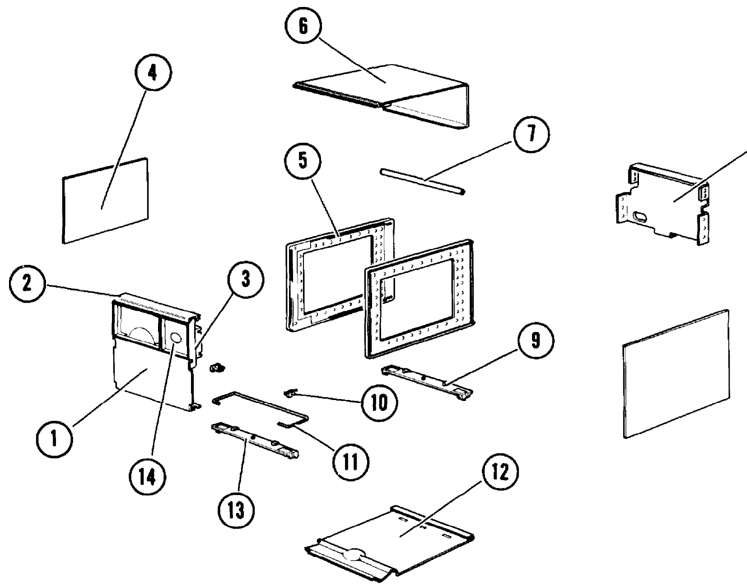
THE TOP COVER SCREWS MUST NOT EXCEED 1/4 in. IN LENGTH; OTHERWISE, INTERNAL CIRCUITS MAY BE SHORTED.

Figure 6-1. Attaching Hardware



REFERENCE DESIGNATOR	DESCRIPTION	QUANTITY	-hp- PART NO.
MP1	Fuseholder	1	1400-0084
MP2	Socket: TSTR (for Q1)	1	1200-0168
MP3	Socket: TSTR (for Q2)	1	1200-0041
MP4	Bushing: adapter (for CALIBRATE adj)	1	2950-0034
MP5	Bushing: Panel	1	1410-0052
MP6	Shield: range switch	1	03406-00603
MP7	Can: line filter	1	03406-05501
MP8	Nut: boot, probe with washer-lock -hp- Part No. 2190-0037	1	03406-22111
MP9	Knob: 3/4 in. diameter	1	0370-0025
MP10	Knob: Pushbutton	10	0370-0132
MP11	Bushing: calibrator (1 volt)	1	03406-41702
MP12	Bracket: filter, probe	1	03406-01203
MP13	Insulator: TSTR (for Q2)	1	1200-0077
MP14	Insulator: TSTR (for Q1)	1	0340-0140
MP15	Nut: sheet metal V type 6-32 (tinnerment)	2	0510-0075
MP16	Bracket: line filter	1	03406-01202
MP17	Clamp: probe cable	1	1400-0014
MP18	Shield: filter, probe	1	03406-00604

Figure 6-2. Mechanical Parts



INDEX NO.	DESCRIPTION	QUANTITY	-hp- PART NO.
1	Panel: front	1	03406-00201
2	Meter trim: half module	1	5020-5367
3	Extender: meter case	1	5040-0701
4	Side Cover: 6 x 11 SM	2	5000-0703
5	Frame assembly: 6 x 11 SM	2	5060-0703
6	Top cover assembly: 7 x 11 SM	6	5060-0718
7	Cabinet spacer: half module	1	5020-0701
8	Panel: rear	1	03406-00202
9	Foot assembly: rear 7 x 11 SM	1	241A-44A
10	Hinge	2	5040-0700
11	Stand: tilt	1	1490-0032
12	Bottom cover: 7 x 11 SM	1	5000-0717
13	Foot assembly: front 7 x 11 SM	1	5060-0728
14	Panel: Insert (ZERO)	1	03406-04301

Figure 6-3. Modular Cabinet

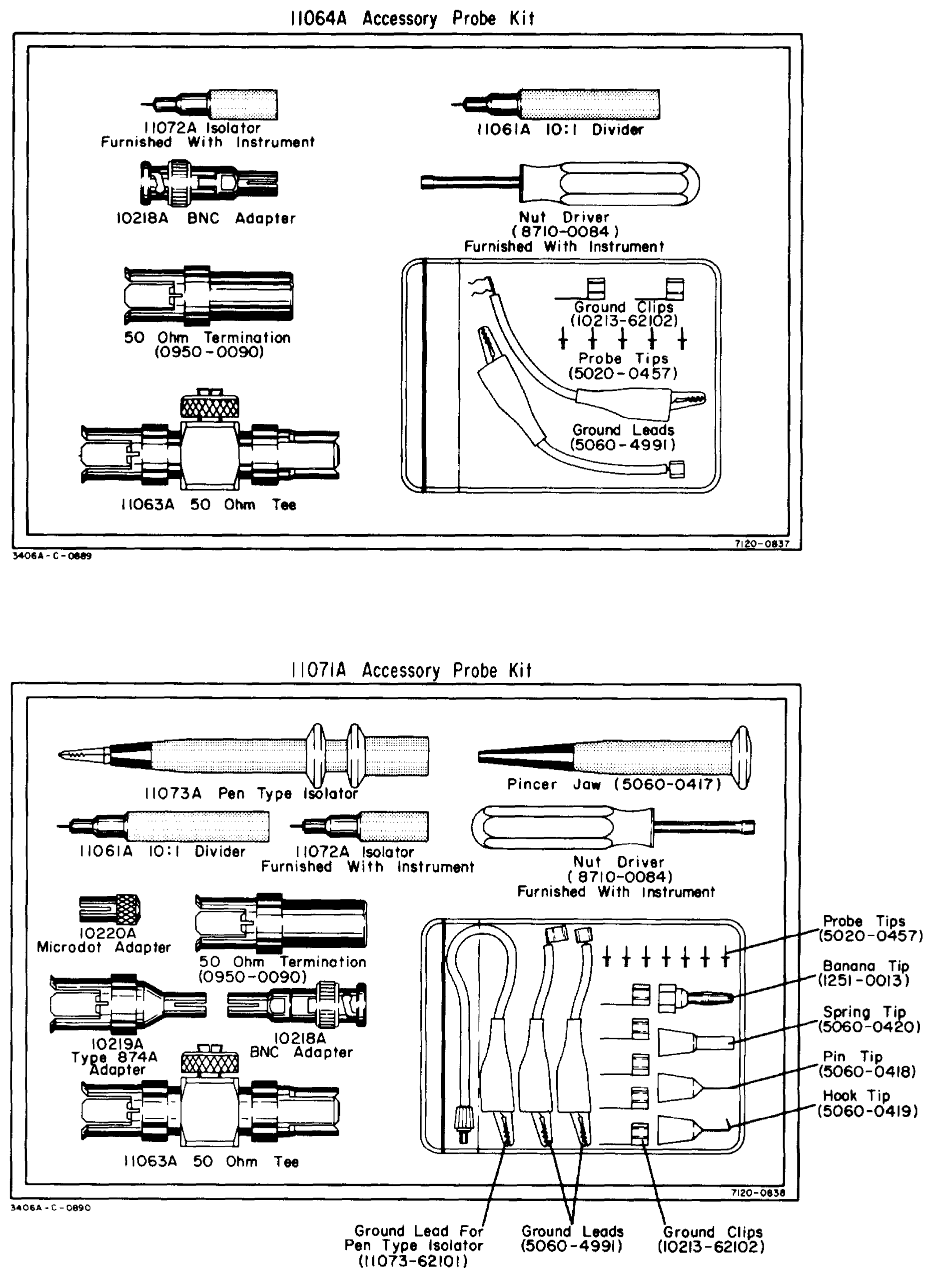


Figure 6-4. Replacement Part No's and Model No's for Accessories

Table 6-1. Replaceable Parts

REFERENCE DESIGNATOR	-hp- PART NO.	TQ	DESCRIPTION	MFR.	MFR. PART NO.
A1 (EXCHANGE)	**03406-62103**		**Assembly: Sampling Probe** **NOTE** This complete Sampling Probe Assembly (A1) is available only on an exchange basis. Refer to Section V for procedures outlining the removal and installation of the Sampling Probe Assembly.	**-hp-**	
A2	**03406-66501**		**Assembly: Board, top**	**-hp-**	
C1	0180-0100	2	C: fxd Ta 4.7 uF 10% 35 vdcw	56289	150D475X9035B2
C2	0180-0155	5	C: fxd Ta 2.2 uF 20% 20 vdcw	56289	150D225X0020A2
C3	0150-0014	3	C: fxd cer 0.005 uF 500 vdcw	04222	D1-4
C4	0180-0228	2	C: fxd Ta elect 22 uF 10% 15 vdcw	56289	150D226X9015B2
C5	0150-0084	7	C: fxd cer 0.1 uF +80% -20% 500 vdcw	56289	33C41
C6	0140-0152	3	C: fxd mica 1000 pF 5% 300 vdcw	04062	DM16F102J
C7	0180-0106	1	C: fxd Ta 60 uF 20% 6 vdcw	56289	150D606X0006B2
C8	0180-0058	8	C: fxd Al elect 50 uF +75% -10% 25 vdcw	56289	30D506G025CC2-DSM
C9	0150-0084		C: fxd cer 0.1 uF +80% -20% 500 vdcw	56289	33C41
C10,C11	0150-0096	3	C: fxd cer 0.05 uF +80% -20% 100 vdcw	72982	845-X5V-503Z
C12	0140-0152		C: fxd mica 1000 pF 5% 300 vdcw	04062	DM16F102J
C13	0140-0177	1	C: fxd mica 400 pF 1% 300 vdcw	04062	RDM15F401F3C
C14	0140-0208	2	C: fxd mica 680 pF 5% 300 vdcw	04062	RDM15F681J3C
C15	0160-0127	1	C: fxd cer 1 uF 20% 25 vdcw	56289	5C13
C16	0180-0058		C: fxd Al elect 50 uF +75% -10% 25 vdcw	56289	30D506G025CC2-DSM
C17	0180-0224	1	C: fxd Al elect 10 uF +75% -10% 15 vdcw	56289	30D106G015BA4
C18	0180-0228		C: fxd Ta elect 22 uF 10% 15 vdcw	56289	150D226X9015B2
C19	0160-0168	2	C: fxd my 0.1 uF 10% 200 vdcw	56289	192P10492
C20	0150-0084		C: fxd cer 0.1 uF +80% -20% 500 vdcw	56289	33C41
C21	0180-0058		C: fxd Al elect 50 uF +75% -10% 25 vdcw	56289	30D506G025CC2-DSM
C22,C23			Not assigned		
C24	0180-0155		C: fxd Ta 2.2 uF 20% 20 vdcw	56289	150D225X0020A2
C25	0180-0058		C: fxd Al elect 50 uF +75% -10% 25 vdcw	56289	30D506G025CC2-DSM
C26	0180-0155		C: fxd Ta 2.2 uF 20% 20 vdcw	56289	150D225X0020A2
C27			Not assigned		
C28	0180-0155		C: fxd Ta 2.2 uF 20% 20 vdcw	56289	150D225X0020A2
C29	0180-0100		C: fxd Ta 4.7 uF 10% 35 vdcw	56289	150D475X9035B2
C30	0160-0168		C: fxd my 0.1 uF 10% 200 vdcw	56289	192P10492
C31	0160-0174	1	C: fxd cer 0.47 uF +80% -20% 25 vdcw	56289	5C11B7
C32	0160-2018	1	C: fxd mica 250 pF 5% 500 vdcw	04062	RDM15F251J5S
CR1,CR2	1902-0049	4	Diode: breakdown 6.19 V 5% 400 mW	07263	obd
CR3	1901-0040	34	Diode: Si 30 mA at +1 V 30 piv 2 pF 2 ns	03877	SG5050
CR4,CR5	1901-0033	2	Diode: Si**	28480	1901-0033
CR6 thru CR12	1901-0040		Diode: Si 30 mA at +1 V 30 piv 2 pF 2 ns	03877	SG5050
CR13			Not assigned		
CR14	1901-0040		Diode: Si 30 mA at +1 V 30 piv 2 pF 2 ns	03877	SG5050
CR15			Not assigned		
CR16 thru CR22	1901-0040		Diode: Si 30 mA at +1 V 30 piv 2 pF 2 ns	03877	SG5050
L1	9170-0016	4	Ferrite Bead (for Q2)	-hp-	

Table 6-1. Replaceable Parts (Cont'd)

REFERENCE DESIGNATOR	-hp- PART NO.	TQ	DESCRIPTION	MFR.	MFR. PART NO.
A2 (Cont'd)					
Q1	1854-0071	22	TSTR: Si NPN 2N3391	24446	2N3391
Q2	1854-0087	3	TSTR: Si NPN**	-hp-	
Q3 thru Q6	1854-0071	1	TSTR: Si NPN 2N3391	24446	2N3391
Q7	1855-0073	2	TSTR: Si P-Channel FET	-hp-	
Q8	1854-0071		TSTR: Si NPN 2N3391	24446	2N3391
Q9	1854-0094	2	TSTR: Si NPN 2N3646	07263	obd
Q10 thru Q12	1854-0071		TSTR: Si NPN 2N3391	24446	2N3391
Q13	1853-0016	8	TSTR: Si PNP 2N3638	07263	2N3638
Q14,Q15	1854-0071		TSTR: Si NPN 2N3391	24446	2N3391
Q16	1854-0087	2	TSTR: Si NPN 2N3417	24446	2N3417
Q17,Q18	1854-0071		TSTR: Si NPN 2N3391	24446	2N3391
Q19,Q20	1853-0016		TSTR: Si PNP 2N3638	07263	2N3638
Q21	1854-0071		TSTR: Si NPN 2N3391	24446	2N3391
R1	0687-1031	9	R: fxd comp 10 kilohms 10% 1/2 W	01121	EB 1031
R2	0687-2721	4	R: fxd comp 2700 ohms 10% 1/2 W	01121	EB 2721
R3,R4	0687-1821	3	R: fxd comp 1800 ohms 10% 1/2 W	01121	EB 1821
R5	0687-8231	3	R: fxd comp 82 kilohms 10% 1/2 W	01121	EB 8231
R6	0687-2241	2	R: fxd comp 220 kilohms 10% 1/2 W	01121	EB 2241
R7	0687-1011	5	R: fxd comp 100 ohms 10% 1/2 W	01121	EB 1011
R8	0687-1041	1	R: fxd comp 100 kilohms 10% 1/2 W	01121	EB 1041
R9	0687-2241		R: fxd comp 220 kilohms 10% 1/2 W	01121	EB 2241
R10	0687-3331	3	R: fxd comp 33 kilohms 10% 1/2 W	01121	EB 3331
R11	0687-1831	1	R: fxd comp 18 kilohms 10% 1/2 W	01121	EB 1831
R12	0698-3138	2	R: fxd prec met flm 277.5 ohms 1/4% 1/4 W	91637	MFF-1/4 T-O obd
R13	0687-6821	7	R: fxd comp 6800 ohms 10% 1/2 W	01121	EB 6821
R14	0687-3331		R: fxd comp 33 kilohms 10% 1/2 W	01121	EB 3331
R15	0687-1821		R: fxd comp 1800 ohms 10% 1/2 W	01121	EB 1821
R16			Not assigned		
R17	0687-5631	2	R: fxd comp 56 kilohms 10% 1/2 W	01121	EB 5631
R18	0687-1031		R: fxd comp 10 kilohms 10% 1/2 W	01121	EB 1031
R19	2100-0095	2	R: var lin 100 kilohms 30%	-hp-	
R20	0687-4701	2	R: fxd comp 47 ohms 10% 1/2 W	01121	EB 4701
R21 thru R23			Not assigned		
R24,R25	0687-5601	3	R: fxd comp 56 ohms 10% 1/2 W	01121	EB 5601
R26	0687-4711	1	R: fxd comp 470 ohms 10% 1/2 W	01121	EB 4711
R27	0687-1031		R: fxd comp 10 kilohms 10% 1/2 W	01121	EB 1031
R28*	0687-1821	1	R: fxd comp 1800 ohms 10% 1/2 W	01121	EB 2221
R29,R30	0687-2721		R: fxd comp 2700 ohms 10% 1/2 W	01121	EB 2721
R31	0687-2221	6	R: fxd comp 2200 ohms 10% 1/2 W	01121	EB 2221
R32	0687-1021	7	R: fxd comp 1000 ohms 10% 1/2 W	01121	EB 1021
R33	0687-4701		R: fxd comp 47 ohms 10% 1/2 W	01121	EB 4701
R34	0687-3931	1	R: fxd comp 39 kilohms 10% 1/2 W	01121	EB 2231
R35	0687-1031		R: fxd comp 10 kilohms 10% 1/2 W	01121	EB 1031
R36	0687-2201	3	R: fxd comp 22 ohms 10% 1/2 W	01121	EB 2201
R37	0687-8231	10	R: fxd comp 82 kilohms 10% 1/2 W	01121	EB 8231
R38	0687-1531	9	R: fxd comp 15 kilohms 10% 1/2 W	01121	EB 1531
R39	2100-0151	1	R: var lin 500 ohms 20%	-hp-	
R40	0687-1531		R: fxd comp 15 kilohms 10% 1/2 W	01121	EB 1531
R41,R42	0687-1011		R: fxd comp 100 ohms 10% 1/2 W	01121	EB 1011
R43	0687-2711	4	R: fxd comp 270 ohms 10% 1/2 W	01121	EB 2711
R44	0687-2221		R: fxd comp 2200 ohms 10% 1/2 W	01121	EB 2221
R45	0687-1531		R: fxd comp 15 kilohms 10% 1/2 W	01121	EB 1531
R46	0687-2201		R: fxd comp 22 ohms 10% 1/2 W	01121	EB 2201
R47	0687-1541	1	R: fxd comp 150 kilohms 10% 1/2 W	01121	EB 1541
R48	0687-8221	5	R: fxd comp 8200 ohms 10% 1/2 W	01121	EB 8221
R49	0687-2211	4	R: fxd comp 220 ohms 10% 1/2 W	01121	EB 2211
R50	0687-8221		R: fxd comp 8200 ohms 10% 1/2 W	01121	EB 8221
R51	0687-2731	1	R: fxd comp 27 kilohms 10% 1/2 W	01121	EB 2731

Table 6-1. Replaceable Parts (Cont'd)

REFERENCE DESIGNATOR	-hp- PART NO.	TQ	DESCRIPTION	MFR.	MFR. PART NO.
A2 (Cont'd)					
R52	0686-1645	1	R: fxd comp 160 kilohms 5% 1/2 W	01121	EB 1645
R53	0686-3035	1	R: fxd comp 30 kilohms 5% 1/2 W	01121	EB 3035
R54	0687-2211		R: fxd comp 220 ohms 10% 1/2 W	01121	EB 2211
R55,R56	0687-2221		R: fxd comp 2200 ohms 10% 1/2 W	01121	EB 2221
R57	0757-0730	2	R: fxd prec met flm 750 ohms 1% 1/4 W	19701	MF6C T-O obd
R58	0687-1521	4	R: fxd comp 1500 ohms 10% 1/2 W	01121	EB 1521
R59	0687-4721	7	R: fxd comp 4700 ohms 10% 1/2 W	01121	EB 4721
R60	0757-0764		R: fxd prec met flm 33.2 kilohms 1% 1/4 W	19701	MF6C T-O obd
R61	0757-0747	2	R: fxd prec met flm 5110 ohms 1% 1/4 W	19701	MF6C T-O obd
R62	0757-0764	2	R: fxd prec met flm 33.2 kilohms 1% 1/4 W	19701	MF6C T-O obd
R63	0757-0121	1	R: fxd prec met flm 258 kilohms 1% 1/4 W	19701	MF6C T-O obd
R64	0687-6831	3	R: fxd comp 68 kilohms 10% 1/2 W	01121	EB 6831
R65,R66	0687-1531		R: fxd comp 15 kilohms 10% 1/2 W	01121	EB 1531
R67			Not assigned		
R68	0687-1531		R: fxd comp 15 kilohms 10% 1/2 W	01121	EB 1531
R69,R70	0687-4721		R: fxd comp 4700 ohms 10% 1/2 W	01121	EB 4721
R71	0698-3212	1	R: fxd prec met flm 249 kilohms 1% 1/4 W	19701	MF6C T-O obd
R72	0687-3331		R: fxd comp 33 kilohms 10% 1/2 W	01121	EB 3331
R73	0687-1031		R: fxd comp 10 kilohms 10% 1/2 W	01121	EB 1031
R74	0687-5601		R: fxd comp 56 ohms 10% 1/2 W	01121	EB 5601
R75	0757-0753	1	R: fxd prec met flm 9090 ohms 1% 1/4 W	19701	MF6C T-O obd
R76	0757-0746	2	R: fxd prec met flm 4750 ohms 1% 1/4 W	19701	MF6C T-O obd
R77	0687-1031		R: fxd comp 10 kilohms 10% 1/2 W	01121	EB 1031
R78	0757-0730		R: fxd prec met flm 750 ohms 1% 1/4 W	19701	MF6C T-O obd
R79	0757-0765	1	R: fxd prec 3650 ohms 1% 1/4 W	19701	MF6C T-O obd
R80	0757-0340	8	R: fxd prec met flm 10 kilohms 1% 1/4 W	19701	MF6C T-O obd
R81	2100-0154	3	R: var lin 1000 ohms 30%	-hp-	
R82	0698-3144	1	R: fxd prec met flm 8870 ohms 1% 1/4 W	19701	MF6C T-O obd
R83	0757-0340		R: fxd prec met flm 10 kilohms 1% 1/4 W	19701	MF6C T-O obd
R84	2100-0154		R: var lin 1000 ohms 30%	-hp-	
R85	0698-3283	2	R: fxd prec met flm 9310 ohms 1% 1/4 W	19701	MF6C T-O obd
R86	0757-0340		R: fxd prec met flm 10 kilohms 1% 1/4 W	19701	MF6C T-O obd
R87	2100-0154		R: var lin 1000 ohms 30%	-hp-	
R88	0698-3283		R: fxd prec met flm 9310 ohms 1% 1/4 W	19701	MF6C T-O obd
R89	0757-0340		R: fxd prec met flm 10 kilohms 1% 1/4 W	19701	MF6C T-O obd
R90	2100-0090	1	R: var lin 2000 ohms 30%	-hp-	
R91	0757-0340		R: fxd prec met flm 10 kilohms 1% 1/4 W	19701	MF6C T-O obd
R92	2100-0095	1	R: var lin 100 kilohms 30%	-hp-	
R93			Not assigned		
R94	2100-0092	3	R: var lin 10 kilohms 20%	-hp-	
R95	2100-0094	2	R: var lin 50 kilohms 30%	-hp-	
R96			Not assigned		
R97,R98	2100-0092		R: var lin 10 kilohms 20%	-hp-	
R99	2100-0093	1	R: var lin 20 kilohms 20%	-hp-	
R100	2100-0091	1	R: var lin 5000 ohms 30%	-hp-	
R101	2100-0094		R: var lin 50 kilohms 30%	-hp-	
R102	2100-0182	1	R: var lin 3300 ohms 10%	-hp-	
R103	0687-6821		R: fxd comp 6800 ohms 10% 1/2 W	01121	EB 6821
R104			Not assigned		
R105	0687-2221		R: fxd comp 2200 ohms 10% 1/2 W	01121	EB 2221
R106	0687-8221		R: fxd comp 8200 ohms 10% 1/2 W	01121	EB 2221
R107			Not assigned		
R108	0687-3321	2	R: fxd comp 3300 ohms 10% 1/2 W	01121	EB 3321
R109	0687-2721		R: fxd comp 2700 ohms 10% 1/2 W	01121	EB 2721
R110	0687-6821		R: fxd comp 6800 ohms 10% 1/2 W	01121	EB 6821
R111	0687-2221		R: fxd comp 2200 ohms 10% 1/2 W	01121	EB 2221
R112	0687-1031		R: fxd comp 10 kilohms 10% 1/2 W	01121	EB 1031
R113	0687-2221		R: fxd comp 2200 ohms 10% 1/2 W	01121	EB 2221
R114	0757-0746		R: fxd prec met flm 4750 ohms 1% 1/4 W	19701	MF6C T-O obd
R115	0757-0734	1	R: fxd prec met flm 1210 ohms 1% 1/4 W	91637	MFF-1/4 T-O obd

Table 6-1. Replaceable Parts (Cont'd)

REFERENCE DESIGNATOR	-hp- PART NO.	TQ	DESCRIPTION	MFR.	MFR. PART NO.
A2 (Cont'd)					
R116			Not assigned		
R117	0757-0728	1	R: fxd prec met flm 619 ohms 1% 1/4 W	19701	MF6C T-O obd
R118	0687-1521		R: fxd comp 1500 ohms 10% 1/2 W	01121	EB 1521
A3	**03406-66502**		**Assembly: Board, bottom**	**-hp-**	
C1	0150-0084		C: fxd cer 0.1 uF +80% -20% 500 vdcw	56289	33C41
C2	0140-0152	2	C: fxd mica 1000 pF 5% 300 vdcw	04062	DM16F102J
C3,C4	0160-2348	1	C: set matched my 4 uF 20% 30 vdcw	56289	Type 148P obd
C5	0150-0084		C: fxd cer 0.1 uF +80% -20% 500 vdcw	56289	33C41
C6	0180-0155		C: fxd Ta 2.2 uF 20% 20 vdcw	56289	150D225X0020A2
C7	0140-0152		C: fxd mica 1000 pF 5% 300 vdcw	04062	DM16F102J
C8	0140-0208		C: fxd mica 680 pF 5% 300 vdcw	04062	RDM15F681J3C
C9	0140-0234	1	C: fxd mica 500 pF 1% 300 vdcw	04062	RDM15F501F3C
C10	0150-0096		C: fxd cer 0.05 uF +80% -20% 100 vdcw	72982	845-X5V-503Z
C11	0150-0014		C: fxd cer 0.005 uF 500 vdcw	04222	D1-4
C12	0180-0097	1	C: fxd Al elect 47 uF 10% 35 vdcw	56289	150D476X9035S2-DVS
C13	0150-0014		C: fxd cer 0.005 uF 500 vdcw	04222	D1-4
C14	0160-0195	1	C: fxd cer 0.001 uF 20% 250 vacw	56289	19C251A
C15	0180-0353	2	C: fxd Al elect 450 uF +100% -10% 50 vdcw	56289	D38702
C16	0150-0084		C: fxd cer 0.1 uF +80% -20% 500 vdcw	56289	33C41
C17,C18	0180-0058		C: fxd Al elect 50 uF +75% -10% 25 vdcw	56289	30D506G025CC2-DSM
C19	0180-0353		C: fxd Al elect 450 uF +100% -10% 50 vdcw	56289	D38702
C20,C21	0150-0084		C: fxd cer 0.1 uF +80% -20% 500 vdcw	56289	33C41
C22,C23	0180-0058		C: fxd Al elect 50 uF +75% -10% 25 vdcw	56289	30D506G025CC2-DSM
C24,C25	0180-0039	2	C: fxd Al elect 100 uF +75% -10% 12 vdcw	56289	30D107G012CC2-DSM
C26	0160-0170	1	C: fxd cer 0.22 uF +80% -20% 25 vdcw	56289	30D107G012CC2-DSM
CR1 thru CR6	1901-0040		Diode: Si 30 mA at +1 V 30 piv 2 pF 2 ns	03877	SG5050
CR7			Not assigned		
CR8	1901-0040		Diode: Si 30 mA at +1 V 30 piv 2 pF 2 ns	03877	SG5050
CR9	1902-1169	1	Diode: Si breakdown 7.5 V 5% 250 mW	04713	obd
CR10	1901-0040		Diode: Si 30 mA at +1 V 30 piv 2 pF 2 ns	03877	SG5050
CR11	1901-0050	1	Diode: Si 200 mA at +1 V 75 V 2 pF 2 ns	07263	FD6195
CR12	1901-0040		Diode: Si 30 mA at +1 V 30 piv 2 pF 2 ns	03877	SG5050
CR13,CR14	1901-0039	2	Diode: Switching Si 50 piv 200 mA at +1 V 6 pF 8 ns	03877	
CR15 thru CR18	1901-0158	4	Diode: Rectifier Si 200 piv	04713	SR1358-3
CR19 thru CR22	1901-0040		Diode: Si 30 mA at +1 V 30 piv 2 pF 2 ns	03877	SG5050
CR23	1902-0049		Diode: breakdown 6.19 V 5% 400 mW	07263	obd
CR24	1902-0761	1	Diode: breakdown 6.2 V 0.01% IN821	04713	IN821
CR25 thru CR28	1901-0040		Diode: Si 30 mA at +1 V 30 piv 2 pF 2 ns	03877	SG5050
CR29	1902-0049		Diode: breakdown 6.19 V 5% 400 mW	07263	obd
CR30,CR31	1901-0040		Diode: Si 30 mA at +1 V 30 piv 2 pF 2 ns	03877	SG5050
CR32	1902-3190	1	Diode: breakdown 13.0 V	04713	SZ10939-215
K1	0490-0180	1	Relay: form C 12 Vdc	80089	61-146
Q1	1854-0005	2	TSTR: Si NPN 2N708	02735	obd
Q2	1854-0071		TSTR: Si NPN 2N3391	24446	2N3391
Q3	1855-0073		TSTR: Si P-Channel FET	-hp-	
Q4	1854-0071		TSTR: Si NPN 2N3391	24446	2N3391
Q5	1853-0016		TSTR: Si PNP 2N3638	07263	2N3638

Table 6-1. Replaceable Parts (Cont'd)

REFERENCE DESIGNATOR	-hp- PART NO.	TQ	DESCRIPTION	MFR.	MFR. PART NO.
A3 (Cont'd)					
Q6 thru Q8	1854-0071		TSTR: Si NPN 2N3391	24446	2N3391
Q9	1853-0016		TSTR: Si PNP 2N3638	07263	2N3638
Q10	1855-0010	1	TSTR: Si 2N2646 uni junction	24446	obd
Q11	1854-0005	1	TSTR: Si NPN 2N708	01295	2N708
Q12	1854-0431		TSTR: Si NPN 2N3646	07263	obd
Q13	1853-0016		TSTR: Si PNP 2N3638	07263	2N3638
Q14	1854-0087		TSTR: Si NPN 2N3417	24446	2N3417
Q15,Q16	1854-0071		TSTR: Si NPN 2N3391	24446	2N3391
Q17	1853-0020	1	TSTR: Si PNP	01295	SM-9143
Q18,Q19	1853-0016		TSTR: Si PNP 2N3638	07263	2N3638
Q20	1854-0003	1	TSTR: Si NPN 2N1711	81453	RT5299
R1	0757-0732	1	R: fxd prec met flm 909 ohms 1% 1/4 W	19701	MF6C T-O obd
R2	2100-2475	2	R: var lin 1000 ohms 30%	71450	XPE200RE obd
R3	0698-3208	1	R: fxd prec met flm 4990 ohms 1% 1/4 W	91637	MF-1/8-44 obd
R4	0687-1031		R: fxd comp 10 kilohms 10% 1/2 W	01121	EB 1031
R5,R6	0687-2211		R: fxd comp 220 ohms 10% 1/2 W	01121	EB2211
R7	0687-1531		R: fxd comp 15 kilohms 10% 1/2 W	01121	EB 1531
R8	2100-2475		R: var lin 1000 ohms 30%	71450	XPE200RE obd
R9	0687-6821		R: fxd comp 6800 ohms 10% 1/2 W	01121	EB 6821
R10	0687-2711		R: fxd comp 270 ohms 10% 1/2 W	01121	EB 2711
R11	0757-0344	2	R: fxd prec met flm 1 megohm 1% 1/4 W	19701	MF6C T-O obd
R12	0757-0340		R: fxd prec met flm 10 kilohms 1% 1/4 W	19701	MF6C T-O obd
R13	0687-1031		R: fxd comp 10 kilohms 10% 1/2 W	01121	EB 1031
R14	0687-6801	1	R: fxd comp 68 ohms 10% 1/2 W	01121	EB 6801
R15	0687-5621	1	R: fxd comp 5600 ohms 10% 1/2 W	01121	EB 5621
R16	2100-2548	1	R: var lin 3300 ohms 10%	71450	XPE200RE obd
R17			Not assigned		
R18	0687-4731	1	R: fxd comp 47 kilohms 10% 1/2 W	01121	EB 4731
R19	0687-6821		R: fxd comp 6800 ohms 10% 1/2 W	01121	EB 6821
R20	0687-1821		R: fxd comp 1800 ohms 10% 1/2 W	01121	EB 1821
R21	0757-0344		R: fxd prec met flm 1 megohm 1% 1/4 W	19701	MF6C T-O obd
R22	0687-1031		R: fxd comp 10 kilohms 10% 1/2 W	01121	EB 1031
R23	0757-0340		R: fxd prec met flm 10 kilohms 1% 1/4 W	19701	MF6C T-O obd
R24	0687-5611	4	R: fxd comp 560 ohms 10% 1/2 W	01121	EB 5611
R25	0757-0340		R: fxd prec met flm 10 kilohms 1% 1/4 W	19701	MF6C T-O obd
R26	0757-0747		R: fxd prec met flm 5110 ohms 1% 1/4 W	19701	MF6C T-O obd
R27			Not assigned		
R28	0687-6811	2	R: fxd comp 680 ohms 10% 1/2 W	01121	EB 6811
R29	0687-8221		R: fxd comp 8200 ohms 10% 1/2 W	01121	EB 8221
R30	0687-6831		R: fxd comp 68 kilohms 10% 1/2 W	01121	EB 6831
R31	0687-8221		R: fxd comp 8200 ohms 10% 1/2 W	01121	EB 8221
R32	0687-6831		R: fxd comp 68 kilohms 10% 1/2 W	01121	EB 6831
R33	0687-1531		R: fxd comp 15 kilohms 10% 1/2 W	01121	EB 1531
R34	0687-5631		R: fxd comp 56 kilohms 10% 1/2 W	01121	EB 5631
R35	0687-4721		R: fxd comp 4700 ohms 10% 1/2 W	01121	EB 4721
R36,R37	0686-1035	2	R: fxd comp 10 kilohms 5% 1/2 W	-hp-	
R38	0686-9125	1	R: fxd comp 9100 ohms 5% 1/2 W	01121	EB 9125
R39	0687-1021		R: fxd comp 1000 ohms 10% 1/2 W	01121	EB 1021
R40	0757-0340		R: fxd prec met flm 10 kilohms 1% 1/4 W	19701	MF6C T-O obd
R41	0687-3321		R: fxd comp 3300 ohms 10% 1/2 W	01121	EB 3321
R42	0687-5611		R: fxd comp 560 ohms 10% 1/2 W	01121	EB 5611
R43	0687-1011		R: fxd comp 100 ohms 10% 1/2 W	01121	EB 1011
R44	0687-1021		R: fxd comp 1000 ohms 10% 1/2 W	01121	EB 1021
R45	0687-4721		R: fxd comp 4700 ohms 10% 1/2 W	01121	EB 4721
R46	0687-1521		R: fxd comp 1500 ohms 10% 1/2 W	01121	EB 1521
R47 thru R49	0757-1000	3	R: fxd prec met flm 51.1 ohms 1% 1/2 W	91637	MFF 1/2 T-O obd

Table 6-1. Replaceable Parts (Cont'd)

REFERENCE DESIGNATOR	-hp- PART NO.	TQ	DESCRIPTION	MFR.	MFR. PART NO.
A3 (Cont'd)					
R50	0687-4721		R: fxd comp 4700 ohms 10% 1/2 W	01121	EB 4721
R51	0687-1021		R: fxd comp 1000 ohms 10% 1/2 W	01121	EB 1021
R52	0698-3559	1	R: fxd comp 3.9 ohms 10% 1/2 W	01121	EB 39G1
R53	0687-1521		R: fxd comp 1500 ohms 10% 1/2 W	01121	EB 1521
R54	0687-5611		R: fxd comp 560 ohms 10% 1/2 W	01121	EB 5611
R55	0687-1021		R: fxd comp 1000 ohms 10% 1/2 W	01121	EB 1021
R56	0687-1011		R: fxd comp 100 ohms 10% 1/2 W	01121	EB 1011
R57,R58	0698-4052	2	R: fxd prec met flm 6980 ohms 1% 1/4 W	19701	MF6C T-O obd
R59	0687-4721		R: fxd comp 4700 ohms 10% 1/2 W	01121	EB 4721
R60	0687-1221	1	R: fxd comp 1200 ohms 10% 1/2 W	01121	EB 1221
R61	0698-4059	1	R: fxd comp 5.6 ohms 10% 1/2 W	01121	EB 56G1
R62	0687-6811		R: fxd comp 680 ohms 10% 1/2 W	01121	EB 6811
R63	0687-1021		R: fxd comp 1000 ohms 10% 1/2 W	01121	EB 1021
R64	0687-5611		R: fxd comp 560 ohms 10% 1/2 W	01121	EB 5611
R65	0687-1011		R: fxd comp 100 ohms 10% 1/2 W	01121	EB 1011
R66	0757-0339	1	R: fxd prec met flm 3010 ohms 1% 1/4 W	19701	MF6C T-O obd
R67	2100-2547	1	R: var lin 500 ohms 30%	71540	XPE200RE obd
R68	0757-0745	1	R: fxd prec met flm 4320 kilohms 1% 1/4 W	19701	MF6C T-O obd
R69	0687-3311	2	R: fxd comp 330 ohms 10% 1/2 W	01121	EB 3311
R70	2100-2454	1	R: var lin 100 kilohms 30%	71450	XPE200RE obd
R71	0687-1531		R: fxd comp 15 kilohms 10% 1/2 W	01121	EB 1531
R72	2100-2456	1	R: var lin 10 kilohms 30% 1/8 W	71450	XPE200RE obd
R73	0687-1531		R: fxd comp 15 kilohms 10% 1/2 W	01121	EB 1531
R74,R75	0687-4721		R: fxd comp 4700 ohms 10% 1/2 W	01121	EB 4721
R76	0687-1021		R: fxd comp 1000 ohms 10% 1/2 W	01121	EB 1021
R77,R78	0687-6821		R: fxd comp 6800 ohms 10% 1/2 W	01121	EB 6821
T1	9100-1338	1	Transformer: pulse	-hp-	
T2	9100-1323	1	Transformer: pulse	-hp-	
			CHASSIS MOUNTED COMPONENTS		
C1	0180-0058		C: fxd Al elect 50 uF +75% -10% 25 vdcw	56289	30D506G025CC2-DSM
C2	0150-0012	1	C: fxd cer 0.01 uF 20% 1000 vdcw	56289	29C214A3
C3	0150-0052	1	C: fxd cer 0.05 uF 20% 400 vdcw	56289	33C17A
DS1	2140-0015	1	Lamp: neon pilot	24455	A-165
F1	2110-0201	1	Fuse: 0.25 amps 250 V slow-blow	-hp-	
FL1	**03406-66508**		**Assembly: Line Filter**	**-hp-**	
C1 thru C4	0150-0086	1	C: fxd cer 0.0047 uF 500 vdcw	04222	D1-4
C5,C6	0150-0069	2	C: fxd cer 0.001 uF 500 vdcw	72982	801-010X5G0102Z
L1,L2	9100-1331	2	Inductor: fxd 3 mH	-hp-	
L3 thru L5	9100-1491	3	Inductor	-hp-	
L6 thru L8	9170-0016	3	Bead: ferrite shielding	02114	56-590-65A1/3B

Table 6-1. Replaceable Parts (Cont'd)

REFERENCE DESIGNATOR	-hp- PART NO.	TQ	DESCRIPTION	MFR.	MFR. PART NO.
FL1 (Cont'd)					
R1,R2	0684-1021	3	R: fxd comp 1000 ohms 10% 1/4 W	01121	CB 1021
R3,R4	0687-3311		R: fxd comp 330 ohms 10% 1/2 W	01121	EB 3311
FL2	**03406-66507**		**Assembly: Probe Filter**	**-hp-**	
C1,C2	0160-2360	4	C: fxd cer 100 pF 10% 300 vdcw	01121	FUGD
C3 thru C7	0160-2361	5	C: fxd cer 470 pF 10% 300 vdcw	01121	FUGD
FL3	**03406-63202**		**Assembly: Filter Calibrator**	**-hp-**	
C1,C2	0160-2360		C: fxd cer 100 pF 10% 300 vdcw	01121	FUGD
R1	0684-1021		R: fxd comp 1000 ohms 10% 1/4 W	01121	CB 1021
J1	5020-6828	1	Connector: ac power cord receptacle	-hp-	AC3G
J2,J3	1250-0083	2	Connector: BNC	95712	30624-1
J4 thru J7	1251-0478	4	Connector: pc 12 pin	07233	64-718-6
J8,J9	1251-0472	2	Connector: pc 12 pin	07233	65-716
J10	1251-0198	1	Connector: pc 12 pin	-hp-	
M1	1120-0393	1	Meter: flush	-hp-	
Q1	1854-0072	1	TSTR: Si NPN 2N3054	86684	2N3054
Q2	1850-0190	1	TSTR: Ge PNP 2N2138	04713	2N2138
R1	2100-1801	1	R: var comp lin 25 kilohms 30% 1/3 W	71450	5VA-45
R2	2100-0407	1	R: var ww lin 5000 ohms 10%	71450	Series AW obd
R3	0684-3331	1	R: fxd comp 33 kilohms 10% 1/4 W	01121	CB 3331
R4	2100-0136	1	R: var lin 6000 ohms 20%	71450	Type 70 obd
R5	2100-0011	1	R: var 5000 ohms 20% lin taper 1/2 W	-hp-	
S1	**03406-61901**		**Assembly: Switch Range**	**-hp-**	
R1	0698-4162	1	R: fxd prec met flm 30.79 ohms 1/4% 1/4 W	91637	MFF-1/4 T-O obd
R2	0698-4163	1	R: fxd prec met flm 128.1 ohms 1/4% 1/4 W	91637	MFF-1/4 T-O obd
R3	0687-1511	1	R: fxd prec comp 150 ohms 10% 1/2 W	01121	EB 1511
R4 thru R8	0698-3139	5	R: fxd prec met flm 410.26 ohms 1/4% 1/4 W	91637	MFF-1/4 T-O obd
R9 thru R12	0698-3138		R: fxd prec met flm 277.5 1/4% 1/4 W	91637	MFF-1/4 T-O obd

Table 6-1. Replaceable Parts (Cont'd)

REFERENCE DESIGNATOR	-hp- PART NO.	TQ	DESCRIPTION	MFR.	MFR. PART NO.
S1 (Cont'd)					
R13	0698-4574	1	R: fxd prec met flm 210 ohms 1/4% 1/4 W	-hp-	
R14	0687-1831		R: fxd comp 18 kilohms 10% 1/2 W	01121	EB 1831
S2	3101-0033	1	Switch: slide DPDT (115-230 vac)	42190	4633
T1	9100-1322	1	Transformer: power	28480	9100-1322
W1	8120-0078	1	Assembly: cable power smooth black extra limp 7.5 ft long	-hp-	
			MISCELLANEOUS		
	03406-66503	1	Assembly: board interconnecting	-hp-	
	03406-63203	1	Assembly: calibrator (includes all parts for the 1 VOLT receptacle) See Figure 7-9.	-hp-	
	03406-62102	1	Assembly: housing probe	-hp-	
	5040-0235	1	Base: lampholder (for DS1)	-hp-	
	9170-0016	20	Bead: ferrite	02114	56-590-65/3B
	03406-26505	1	Board: probe connector	-hp-	
	03406-01202	1	Bracket: line filter (FL1)	-hp-	
	2950-0034	1	Bushing: adapter (for CALIBRATE ADJ)	000LF	obd
	03406-41702	1	Bushing: calibrate (on front panel)	-hp-	
	1410-0052	1	Bushing: panel (for CALIBRATE ADJ)	95264	obd
	03406-05501	1	Can: line filter (FL1)	-hp-	
	03406-00101	1	Chassis: hinged	-hp-	
	10213-62102	2	Clip: ground	-hp-	
	5000-0717	1	Cover: bottom 7 X 11 submodule	-hp-	
	5000-0703	2	Cover: side 6 X 11 submodule	-hp-	
	5060-0718	1	Cover: top 7 X 11 submodule	-hp-	
	5040-0701	1	Extender: meter case	-hp-	
	241A-44A	1	Foot Assembly: 7 X 11 submodule	-hp-	
	5060-0728	1	Foot Assembly: 7 X 11 submodule	-hp-	
	5060-0703	2	Frame Assembly: 6 X 11 submodule	-hp-	
	1400-0084	1	Fuseholder	75915	342014
	03406-01101	1	Heat sink: TSTR (for Q1 and Q2)	-hp-	
	5040-0700	2	Hinge: (for tilt stand)	-hp-	
	0340-0060	3	Insulator: cloverleaf feed-thru teflon (used on A2 and A3 assemblies)	98291	FT-E-15
	03406-22112	1	Insulator: housing probe	-hp-	
	0340-0140	1	Insulator: transistor	02735	DF 31A
	1200-0077	1	Insulator: transistor	16037	112 obd
	0370-0132	10	Knob: pushbutton (RANGE)	-hp-	
	0370-0025	1	Knob: round black 3/4 in. dia. (ZERO)	-hp-	
	5000-0252	1	Label: pushbutton (OFF)	-hp-	
	5000-3220	1	Label: pushbutton .001 (-50)	-hp-	
	5000-3221	1	Label: pushbutton .003 (-40)	-hp-	
	5000-3222	1	Label: pushbutton .01 (-30)	-hp-	
	5000-3223	1	Label: pushbutton .03 (-20)	-hp-	

Table 6-1. Replaceable Parts (Cont'd)

REFERENCE DESIGNATOR	-hp- PART NO.	TQ	DESCRIPTION	MFR.	MFR. PART NO.
			MISCELLANEOUS (Cont'd)		
	5000-3224	1	Label: pushbutton .1 (-10)	-hp-	
	5000-3225	1	Label: pushbutton .3 (0)	-hp-	
	5000-3226	1	Label: pushbutton 1 (10)	-hp-	
	5000-3244	1	Label: pushbutton 3 (20)	-hp-	
	5000-3243	1	Label: pushbutton (CAL)	-hp-	
	5060-4991	1	Lead: ground	-hp-	
	5040-0234	1	Lampholder: (for DS1)	-hp-	
	03406-90001	1	Manual: operating and service	-hp-	
	5020-0705	1	Meter trim: submodule	-hp-	
	03406-22111	1	Nut: boot probe	-hp-	
	8710-0084	1	Nut driver: hex	96508	P-3 obd
	03406-00201	1	Panel: front	-hp-	
	03406-04301	1	Panel: insert	-hp-	
	03406-00202	1	Panel: rear	-hp-	
	03406-00203	1	Panel: sub rear	-hp-	
	03406-00601	1	Shield: bottom cover	-hp-	
	03406-00604	1	Shield: probe filter cover	-hp-	
	03406-00603	1	Shield: range switch cover	-hp-	
	03406-22113	1	Shield: switch part of 03406-62102	-hp-	
	03406-00602	1	Shield: top cover	-hp-	
	1200-0168	1	Socket: TSTR (for Q1)	000LK	PTS-4
	1200-0041	1	Socket: TSTR (for Q2)	71785	133-32-10-013
	1490-0032	1	Stand: tilt submodule	91260	obd
	7124-0339	1	Tag: red operating note	-hp-	
	5020-0457	10	Tip: pin probe	-hp-	
	11072-60001	1	Tip: isolator	-hp-	

SECTION VII
CIRCUIT DIAGRAMS

7-1. INTRODUCTION.

7-2. This section contains the circuit diagrams necessary for the operation and maintenance of the Model 3406A Broadband Sampling Voltmeter. Included is Table 7-1, with photograph of the instrument, identifying the various assemblies.

7-3. COVER REMOVAL.

7-4. To prepare the instrument for calibration or to make the components or test points on the top or bottom assemblies accessible perform the steps outlined in Figure 7-1.

7-5. FUNCTIONAL BLOCK DIAGRAM.

7-6. The functional block diagram shows the relationship between the assemblies of the instrument. Signal flow between assemblies and significant portions of assemblies as well as major feedback paths are shown. The functional block diagram is used to isolate trouble to a particular block. Test points, voltage waveforms, and voltage levels are identified on the functional block diagram.

7-7. COMPONENT LOCATION DIAGRAMS.

7-8. The component location diagrams show the physical location of parts mounted on an assembly. The components mounted on the chassis of the instrument are identified in Figure 7-2.

7-9. INTERCONNECTING ASSEMBLY.

7-10. The top and bottom assemblies are connected to each other and other circuits through the interconnecting assembly shown in Figure 7-3.

7-11. SCHEMATIC DIAGRAMS.

7-12. The circuits contained within each assembly are shown in a schematic diagram. These diagrams are used to develop an understanding of the detailed theory of operation and as an aid in isolating a malfunction to a component(s). To aid in troubleshooting dc voltage levels and waveform test points are given in red on each schematic.

7-13. An explanation of terms and symbols used on schematic diagrams is given below.

REFERENCE DESIGNATIONS

PARTIAL REFERENCE DESIGNATIONS ARE SHOWN: PREFIX WITH ASSEMBLY OR SUBASSEMBLY DESIGNATION(S) OR BOTH FOR COMPLETE DESIGNATION.

ASSEMBLY	SUBASSEMBLY	COMPONENT	COMPLETE DESIGNATION
A2	NONE	Q1	A2Q1
A2	A1	R1	A2A1R1
NONE	NONE	J1	J1

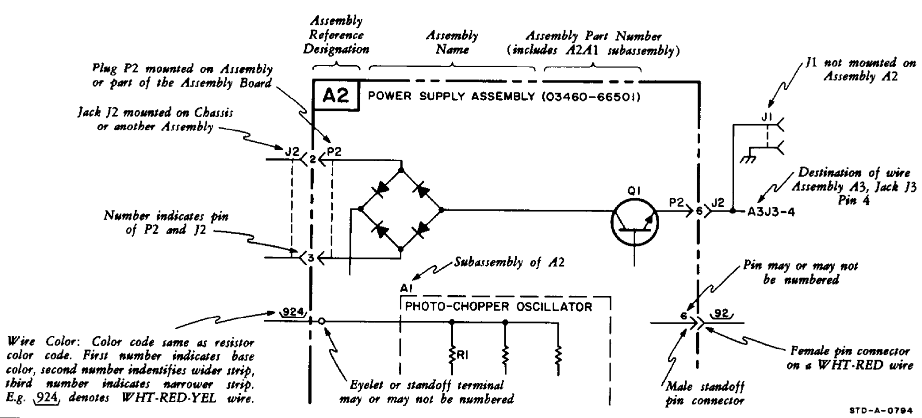


Table 7-1. Model 3406A Assembly Designations

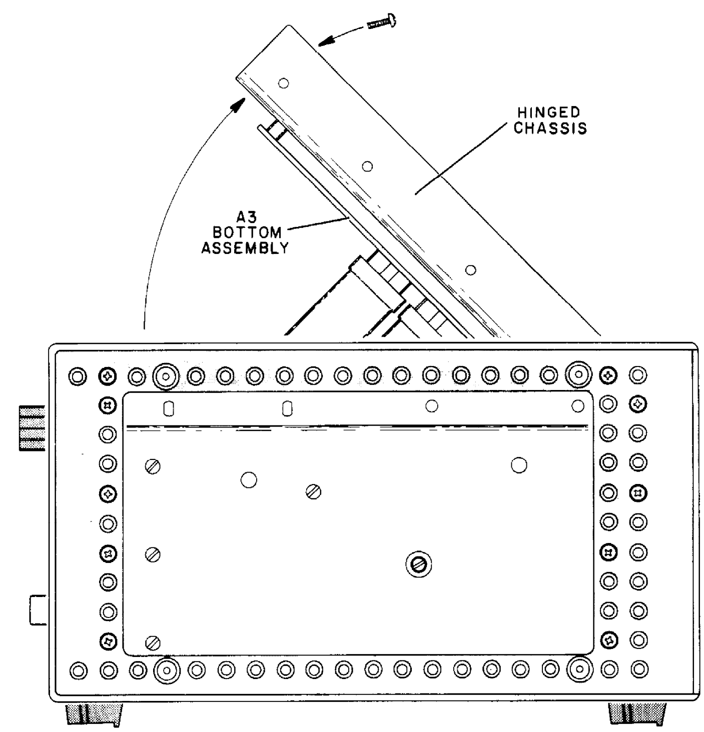
REFERENCE DESIGNATION	TITLE	-hp- PART NO.	SCHEMATIC LOCATION
A1	Sampling Probe Assembly	03406-66504	Page 1
A2	Top Assembly	03406-66501	Page 1, 2
A3	Bottom Assembly	03406-66502	Page 1, 2, 3, and 4
S1	Range Switch Assembly	03406-61901	Page 1, 2, 3, and 4

NOTE

All components not located on one of the above assemblies are considered part of the "main" assembly and are not prefixed with an assembly reference designator.

PROBE REPLACEMENT

The Sampling Probe Assembly (A1) is not a field repairable item. A complete assembly is available on an exchange basis (-hp- Part No. 03406-62103) through your local Hewlett-Packard Sales and Service office (See Appendix B for office location).



GENERAL SCHEMATIC NOTES

1. ALL DC VOLTAGE LEVELS SHOWN ON SCHEMATICS ARE TYPICAL AND MAY VARY FROM INSTRUMENT TO INSTRUMENT.
2. REFERENCE DESIGNATOR OR COMPONENTS SCREENED IN RED DENOTE BACKDATING CHANGE, SEE APPENDIX C.
3. PARTIAL REFERENCE DESIGNATIONS ARE SHOWN. PREFIX WITH ASSEMBLY OR SUBASSEMBLY DESIGNATION(S) OR BOTH FOR COMPLETE DESIGNATION.
4. COMPONENT VALUES ARE SHOWN AS FOLLOWS UNLESS OTHERWISE NOTED.

   RESISTANCE IN OHMS.

   CAPACITANCE IN MICROFARADS.
5. ⏚ DENOTES POWER LINE GROUND.
6. ▽ DENOTES GROUND ON ETCHED CIRCUIT BOARD.
7. ——— – – ——— DENOTES ASSEMBLY.
8. ———— DENOTES MAIN SIGNAL PATH.
9. — — — — DENOTES DC FEEDBACK PATH.
10. —●—●—●—● DENOTES AC FEEDBACK PATH.
11. ▭ DENOTES FRONT PANEL MARKING.
12. ⬚ DENOTES REAR PANEL MARKING.
13. DENOTES SCREWDRIVER ADJUST.
14. 924 DENOTES WIRE COLOR: COLOR CODE SAME AS RESISTOR COLOR CODE. FIRST NUMBER IDENTIFIES BASE COLOR, SECOND NUMBER IDENTIFIES WIDER STRIP, THIRD NUMBER IDENTIFIES NARROWER STRIP. (e.g. 924 = WHITE, RED, YELLOW.)
15. DENOTES COMPONENTS NOT MOUNTED ON ASSEMBLY.
16. * AVERAGE VALUE SHOWN, OPTIMUM VALUE SELECTED AT FACTORY.
17. ALL RELAYS ARE SHOWN DEENERGIZED.
18. DENOTES FERRITE BEAD.
19. —›› — DENOTES SECOND APPEARANCE OF A CONNECTOR PIN.

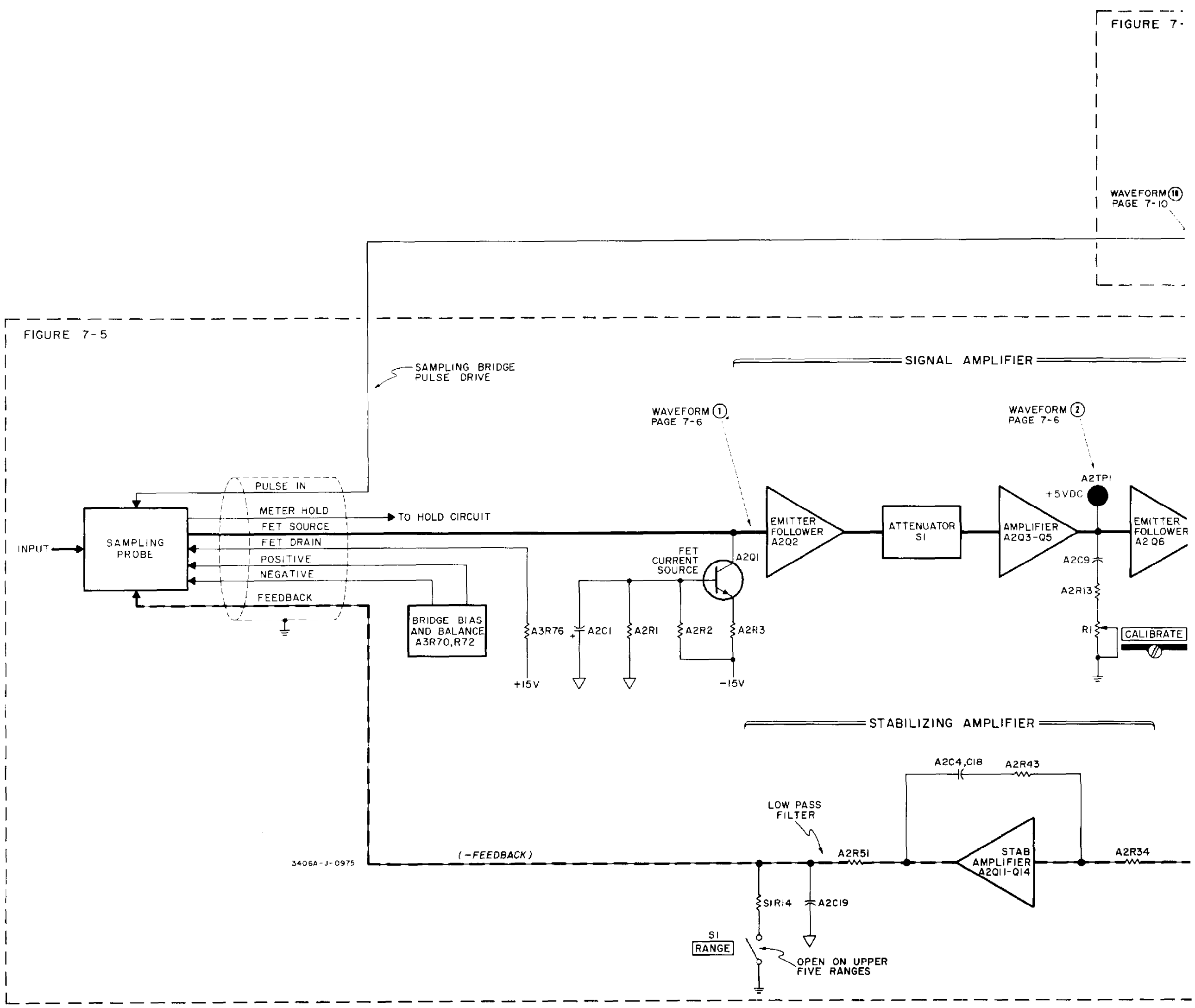


FIGURE 7-
WAVEFORM 18
PAGE 7-10
FIGURE 7-5
SAMPLING BRIDGE
PULSE DRIVE
SIGNAL AMPLIFIER
WAVEFORM 1
PAGE 7-6
WAVEFORM 2
PAGE 7-6
A2TP1
+5VDC
PULSE IN
METER HOLD
TO HOLD CIRCUIT
FET SOURCE
FET DRAIN
POSITIVE
NEGATIVE
FEEDBACK
INPUT
SAMPLING
PROBE
EMITTER
FOLLOWER
A2Q2
ATTENUATOR
S1
AMPLIFIER
A2Q3-Q5
EMITTER
FOLLOWER
A2 Q6
FET
CURRENT
SOURCE
A2Q1
A2C9
A2R13
R1
CALIBRATE
BRIDGE BIAS
AND BALANCE
A3R70,R72
A3R76
A2C1
A2R1
A2R2
A2R3
+15V
-15V
STABILIZING AMPLIFIER
A2C4,C18
A2R43
LOW PASS
FILTER
A2R51
STAB
AMPLIFIER
A2Q11-Q14
A2R34
3406A-J-0975
(-FEEDBACK)
S1R14
A2C19
S1
RANGE
OPEN ON UPPER
FIVE RANGES

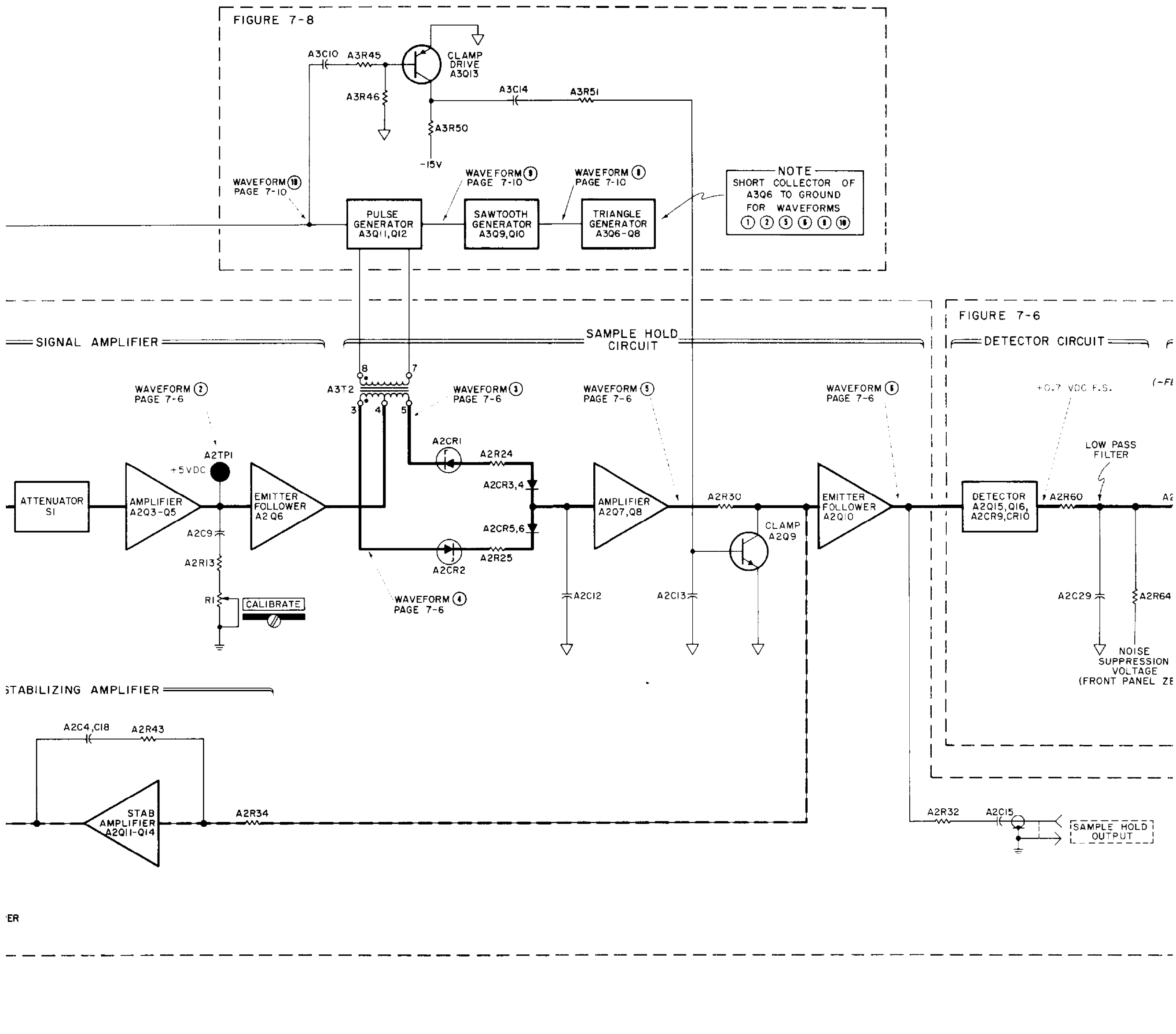


FIGURE 7-8
A3C10
A3R45
CLAMP DRIVE A3Q13
A3R46
A3C14
A3R51
A3R50
-15V
WAVEFORM (10) PAGE 7-10
WAVEFORM (9) PAGE 7-10
WAVEFORM (8) PAGE 7-10
NOTE
SHORT COLLECTOR OF A3Q6 TO GROUND FOR WAVEFORMS (1) (2) (5) (6) (8) (10)
PULSE GENERATOR A3Q11,Q12
SAWTOOTH GENERATOR A3Q9,Q10
TRIANGLE GENERATOR A3Q6-Q8
SIGNAL AMPLIFIER
SAMPLE HOLD CIRCUIT
FIGURE 7-6
DETECTOR CIRCUIT
WAVEFORM (2) PAGE 7-6
A3T2
WAVEFORM (3) PAGE 7-6
WAVEFORM (5) PAGE 7-6
WAVEFORM (6) PAGE 7-6
+0.7 VDC F.S.
A2TP1
+5VDC
A2CR1
A2R24
LOW PASS FILTER
ATTENUATOR S1
AMPLIFIER A2Q3-Q5
EMITTER FOLLOWER A2Q6
A2CR3,4
AMPLIFIER A2Q7,Q8
A2R30
EMITTER FOLLOWER A2Q10
DETECTOR A2Q15,Q16, A2CR9,CR10
A2R60
A2C9
A2CR5,6
CLAMP A2Q9
A2R13
A2CR2
A2R25
R1
CALIBRATE
WAVEFORM (4) PAGE 7-6
A2C12
A2C13
A2C29
A2R64
NOISE SUPPRESSION VOLTAGE
STABILIZING AMPLIFIER
A2C4,C18
A2R43
STAB AMPLIFIER A2Q11-Q14
A2R34
A2R32
A2C15
SAMPLE HOLD OUTPUT

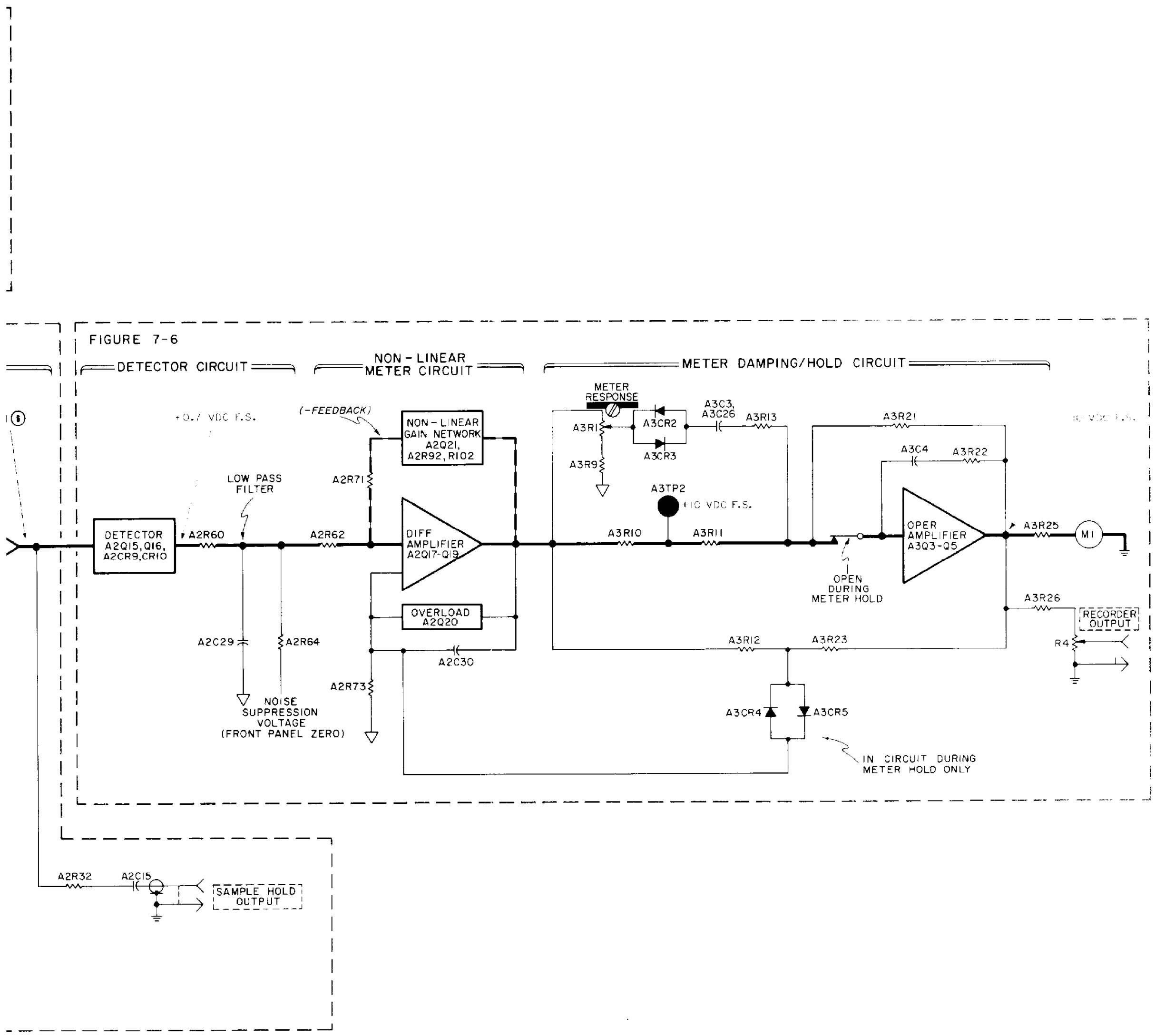


Figure 7-1. Functional Block Diagram

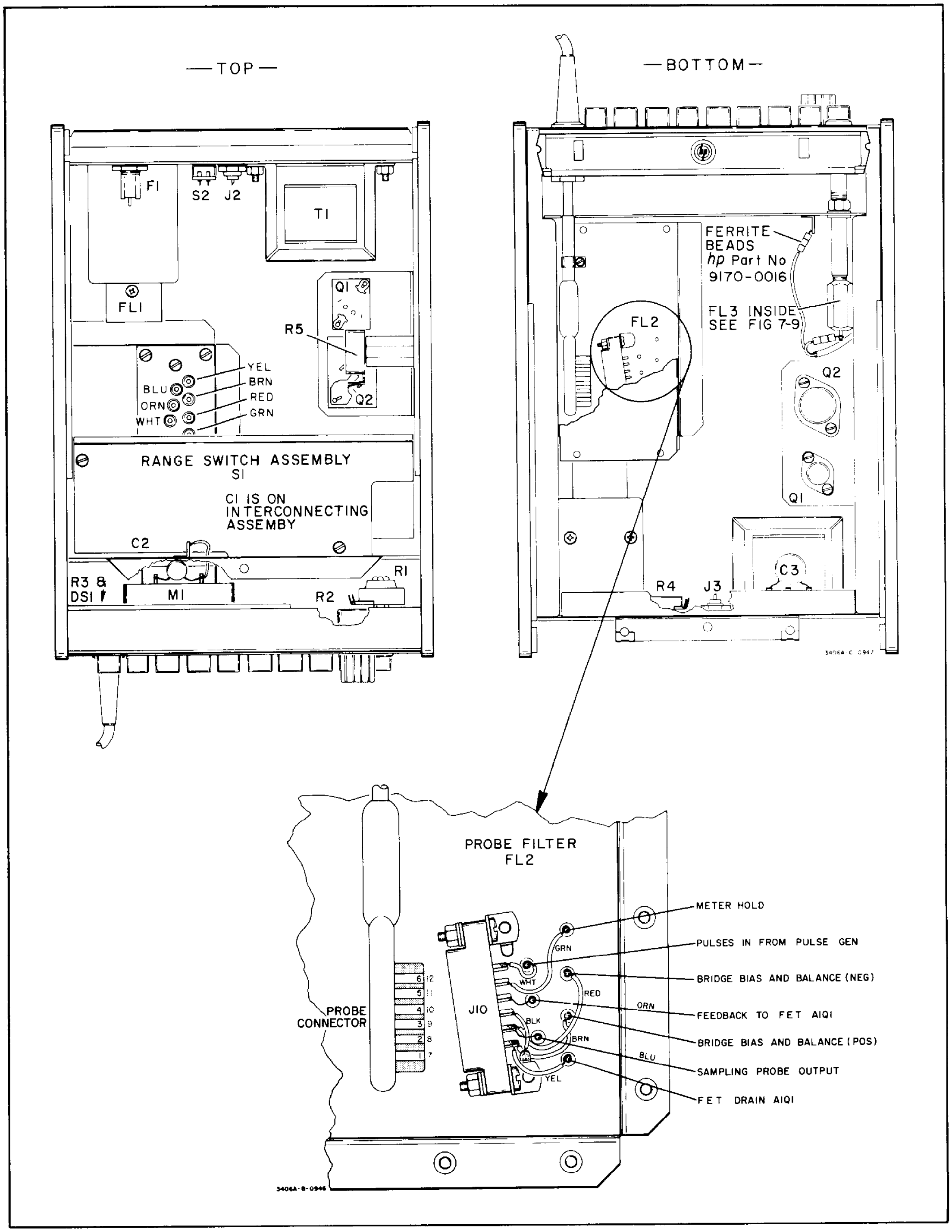
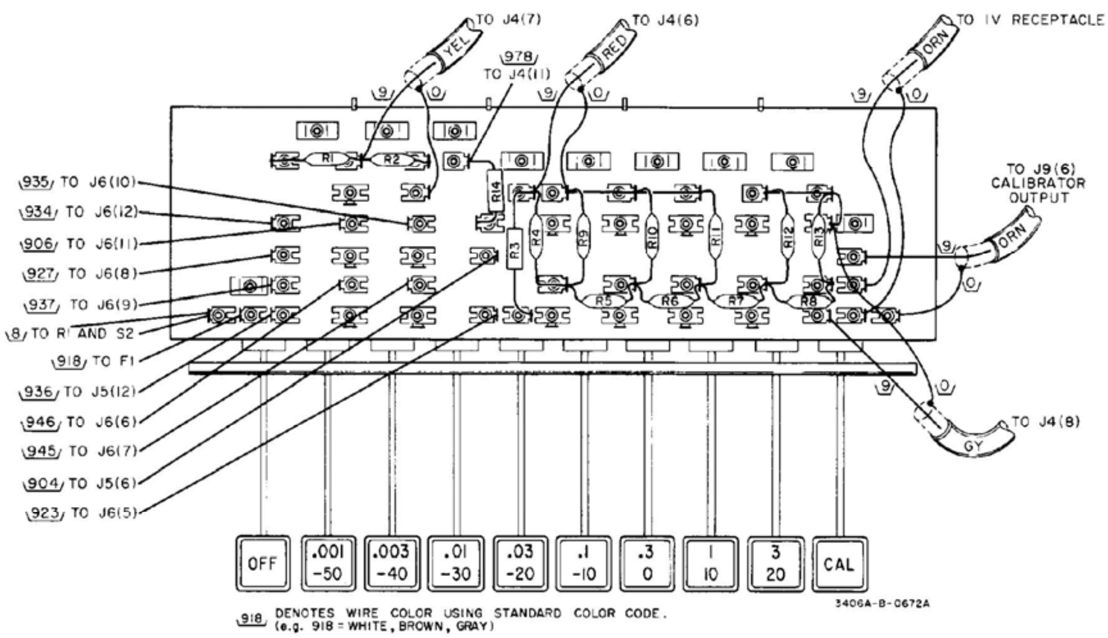
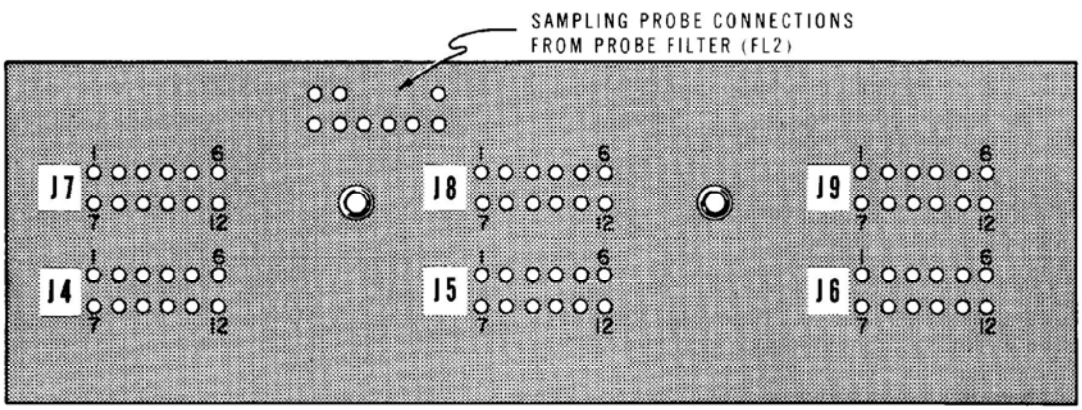


Figure 7-2. Chassis Components



RANGE SWITCH ASSEMBLY



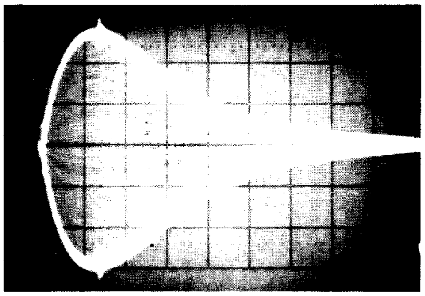
BOTTOM VIEW OF INTERCONNECTING ASSEMBLY

3406A-B-0669

Figure 7-3. Range Switch and Interconnecting Assembly

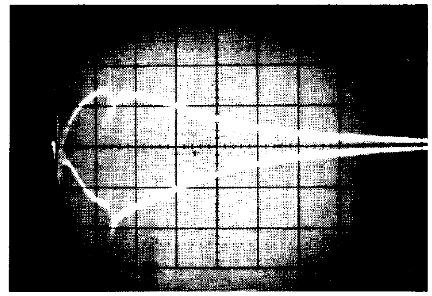
All waveforms should be observed with 10:1 Probe

1 Sampling Probe Output (Base of A2Q2)



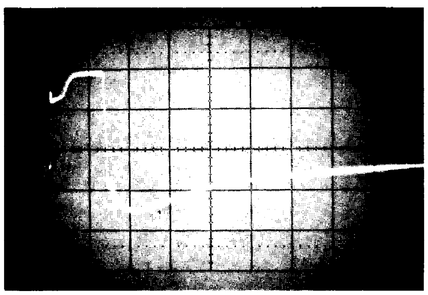
Vert = .1V/cm; sweep = 2 us/cm
AC coupled; sweeping stopped*

2 Signal Amplifier Output (A2TP1)



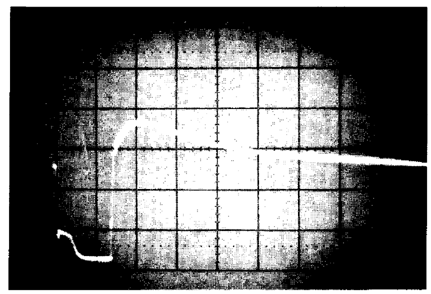
Vert = .05V/cm; sweep = 2 us/cm
AC coupled; sweeping stopped*

3 Positive Drive Pulse (Cathode of A2CR1)



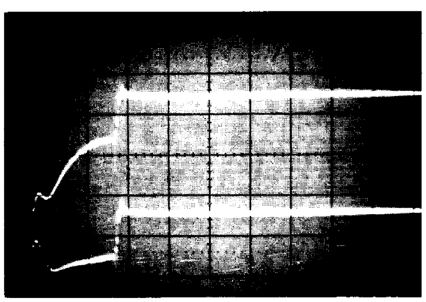
Vert = 5V/cm; sweep = 2 us/cm
AC coupled

4 Negative Drive Pulse (Anode of A2CR2)



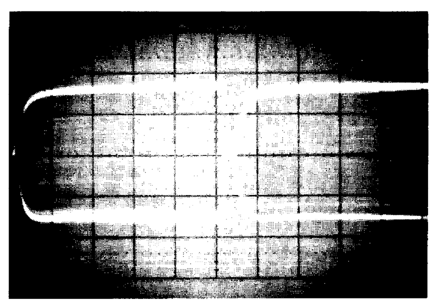
Vert = 5V/cm; sweep = 2 us/cm
AC coupled

5 (Collector of A2Q8)



Vert = .2V/cm; sweep = 1 us/cm
AC coupled; sweeping stopped*

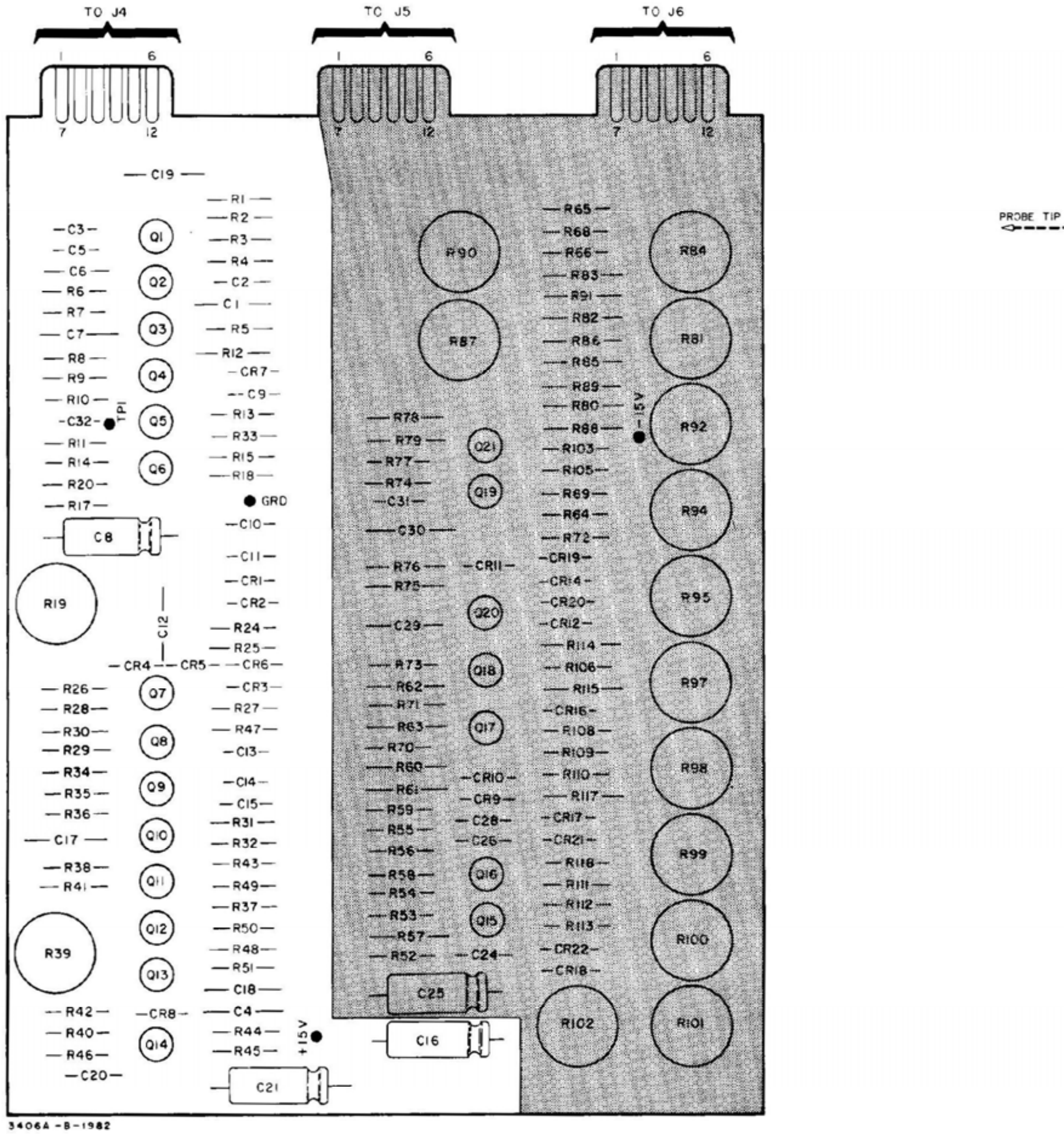
6 Sample Hold Output (Emitter of A2Q10)



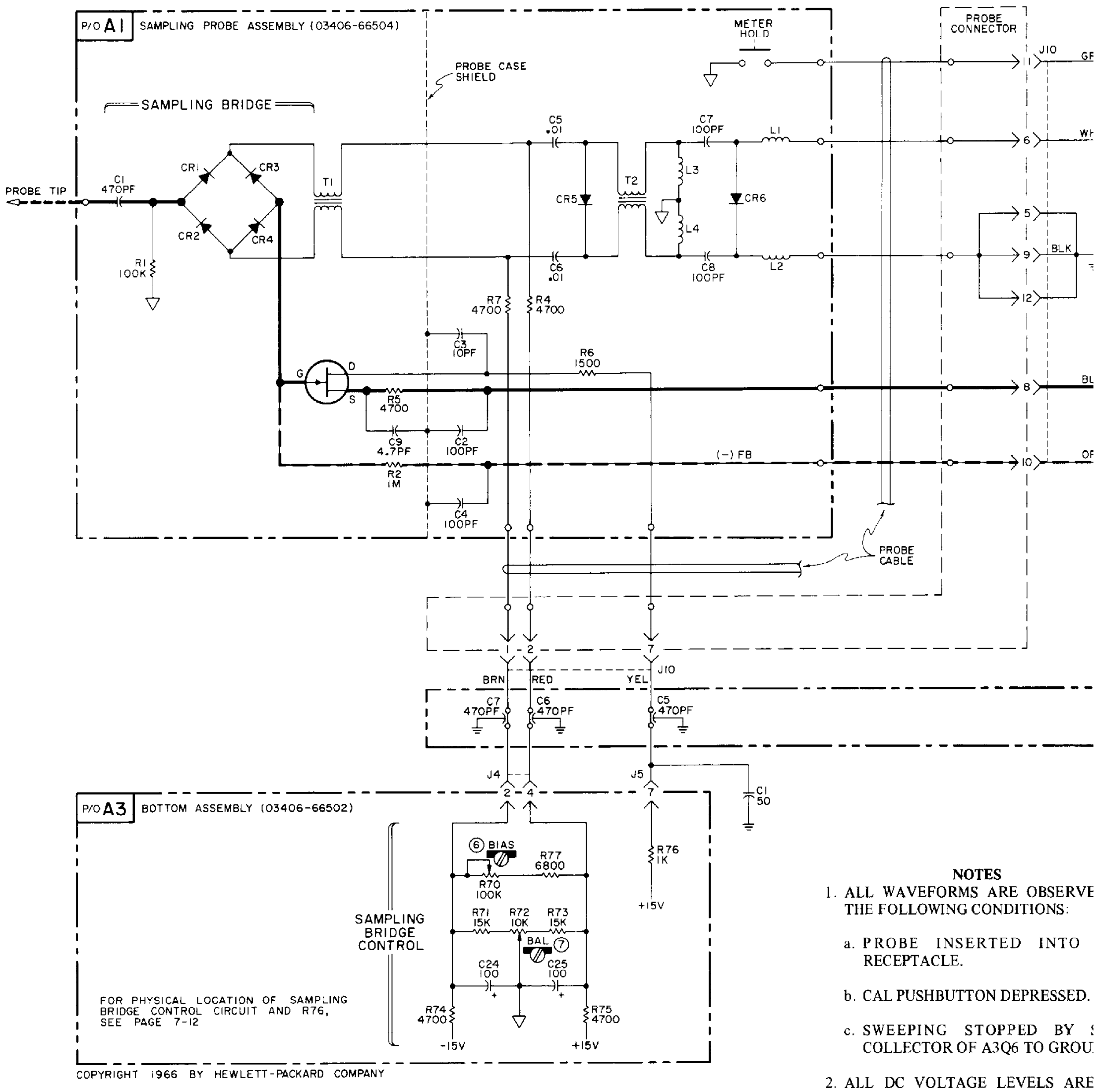
Vert = .2V/cm; sweep = 10 usec/cm
AC coupled; sweeping stopped*

*To stop sweeping, short collector of A3Q6 to ground (Figure 7-8)

Figure 7-4. Input Circuit Waveforms



A2
hp Part No. 03406-66501
Rev A



NOTES

1. ALL WAVEFORMS ARE OBSERVE
THE FOLLOWING CONDITIONS:

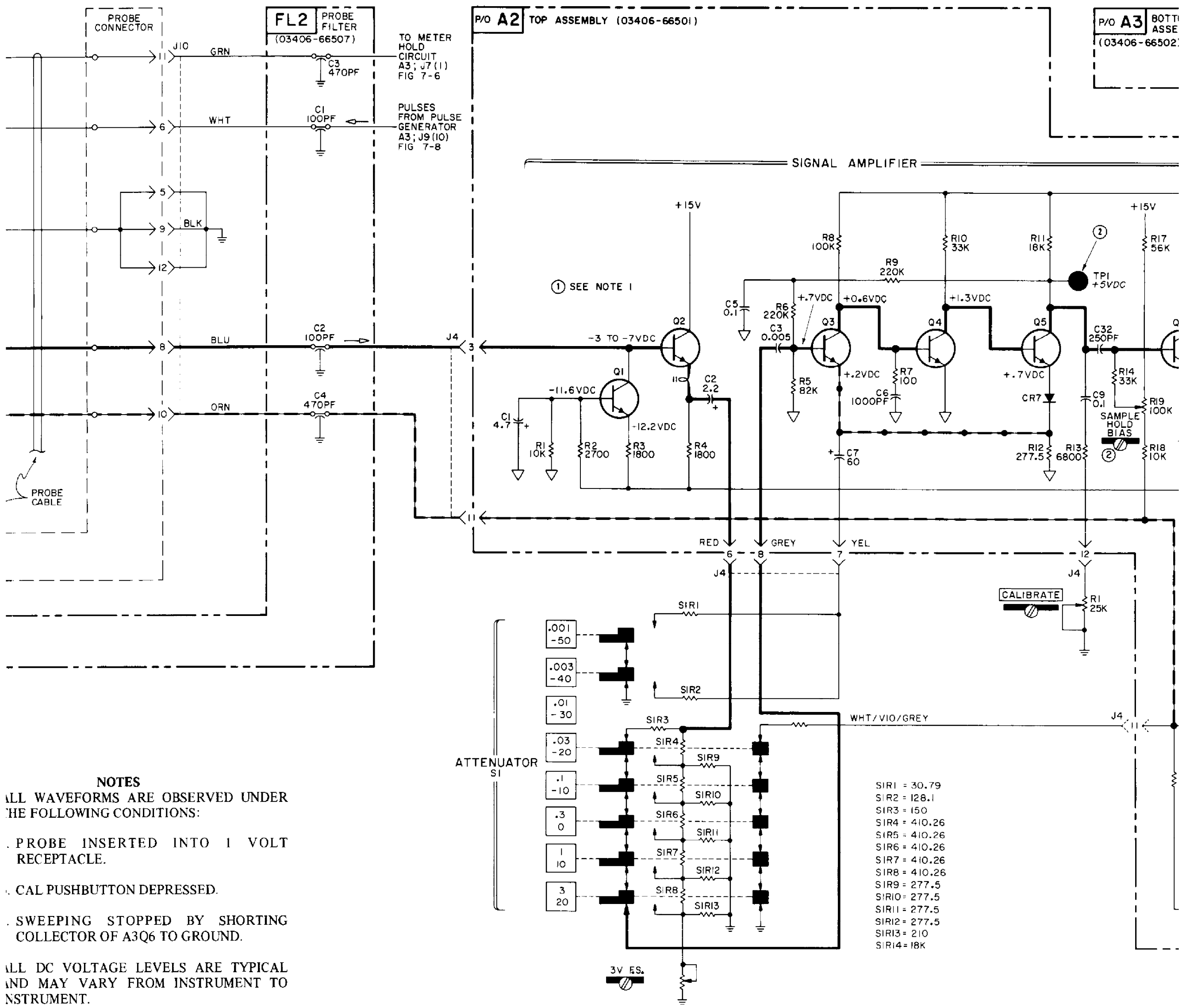
   a. PROBE INSERTED INTO
   RECEPTACLE.

   b. CAL PUSHBUTTON DEPRESSED.

   c. SWEEPING STOPPED BY
   COLLECTOR OF A3Q6 TO GROU

2. ALL DC VOLTAGE LEVELS ARE
AND MAY VARY FROM INSTRU
INSTRUMENT.

3. REFERENCE DESIGNATOR AND
SHOWN IN RED DENOTES BA
CHANGE. SEE APPENDIX C.



NOTES

ALL WAVEFORMS ARE OBSERVED UNDER THE FOLLOWING CONDITIONS:

. PROBE INSERTED INTO 1 VOLT RECEPTACLE.

. CAL PUSHBUTTON DEPRESSED.

. SWEEPING STOPPED BY SHORTING COLLECTOR OF A3Q6 TO GROUND.

ALL DC VOLTAGE LEVELS ARE TYPICAL AND MAY VARY FROM INSTRUMENT TO INSTRUMENT.

REFERENCE DESIGNATOR AND VALUES SHOWN IN RED DENOTES BACKDATING CHANGE. SEE APPENDIX C.

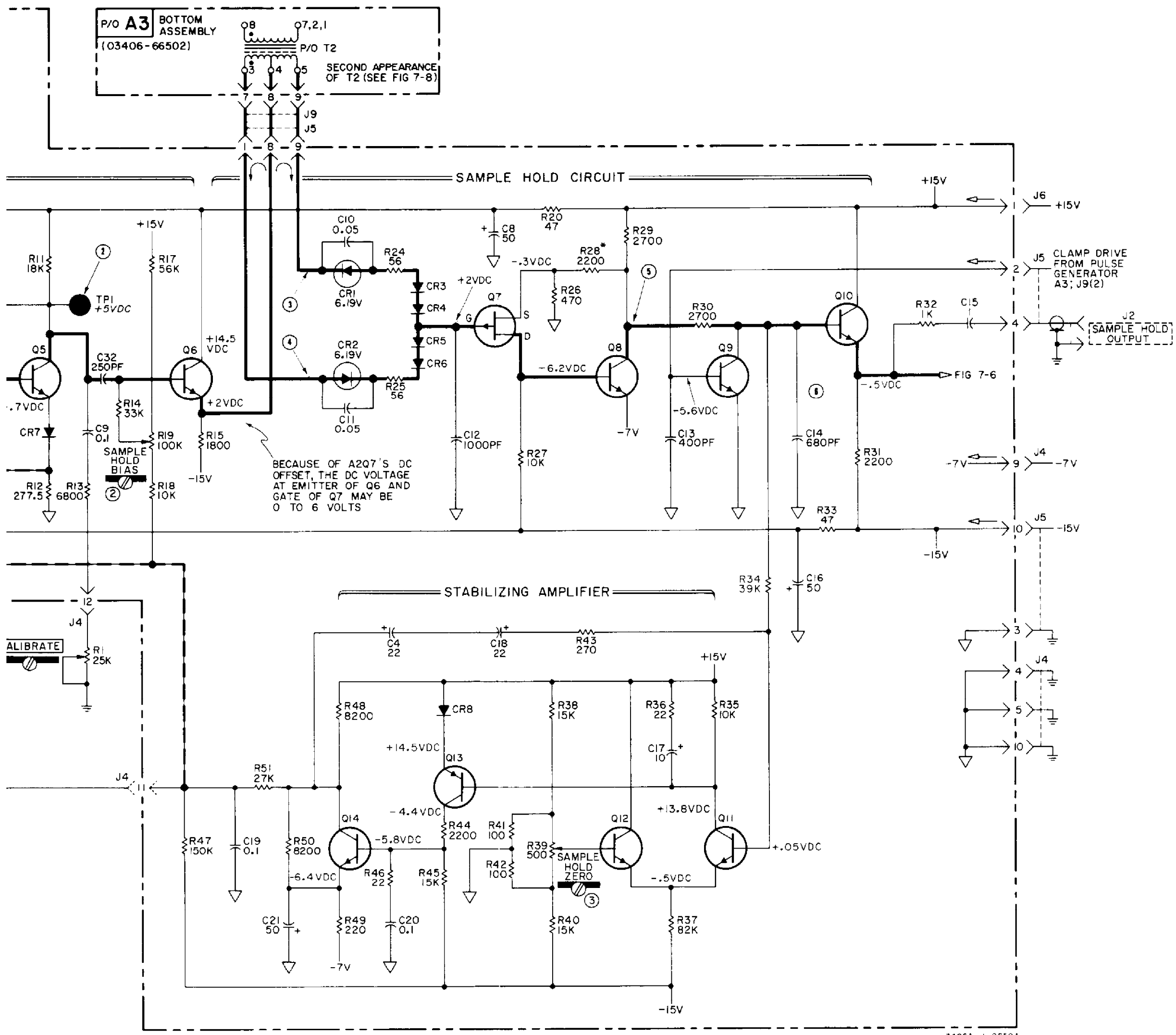
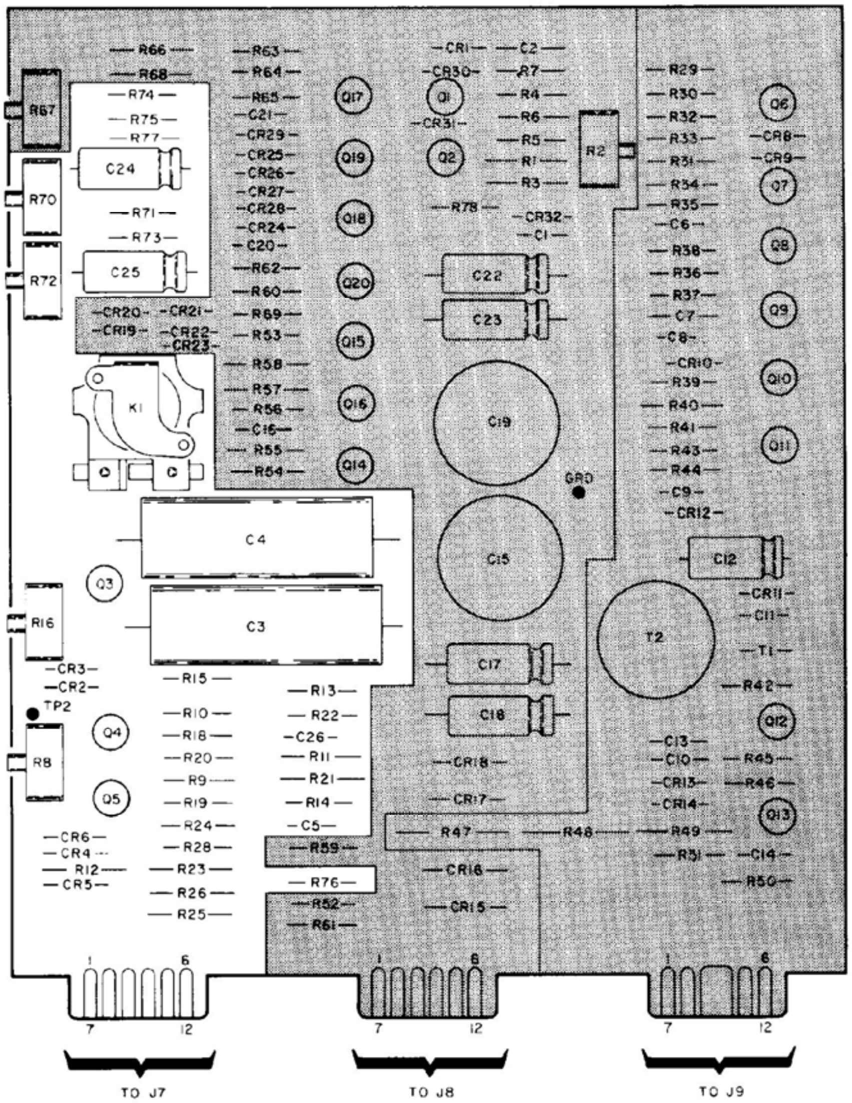
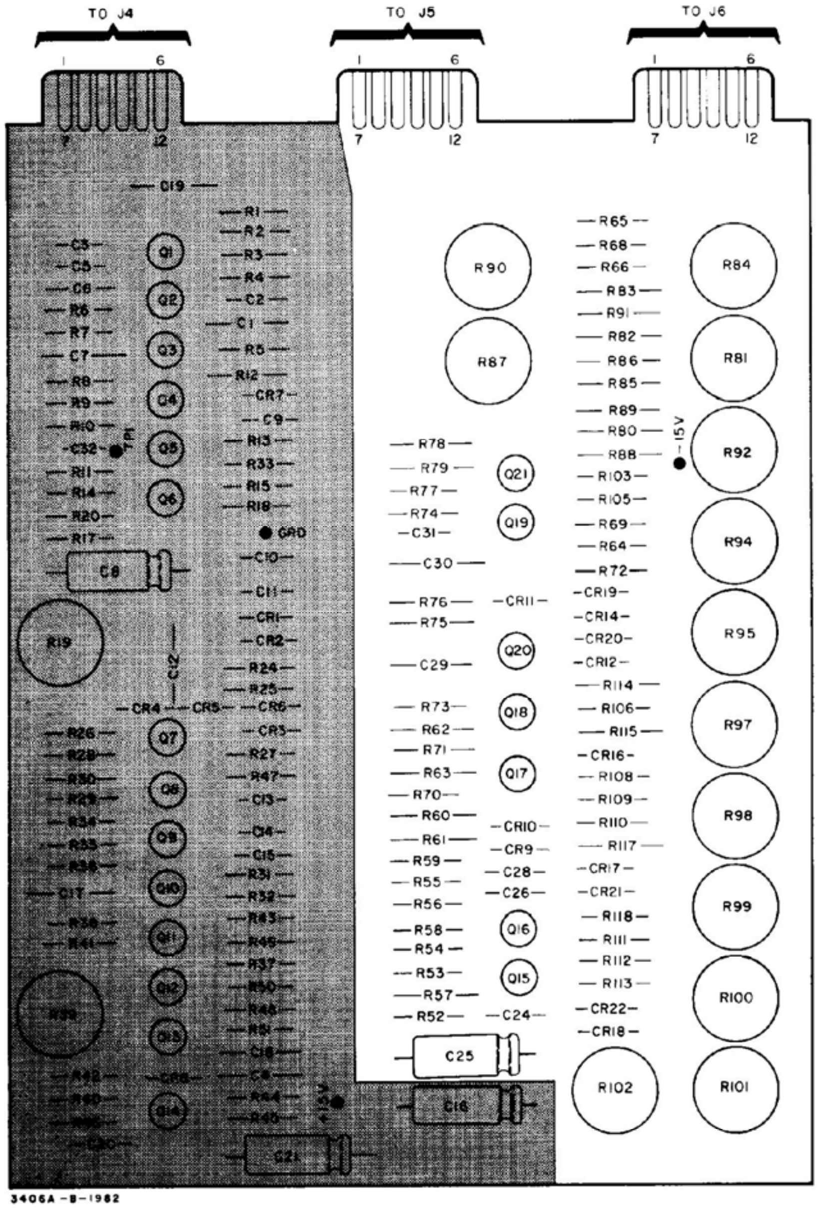


Figure 7-5. Input Circuits



A3
hp Part No. 03406-66502
Rev C

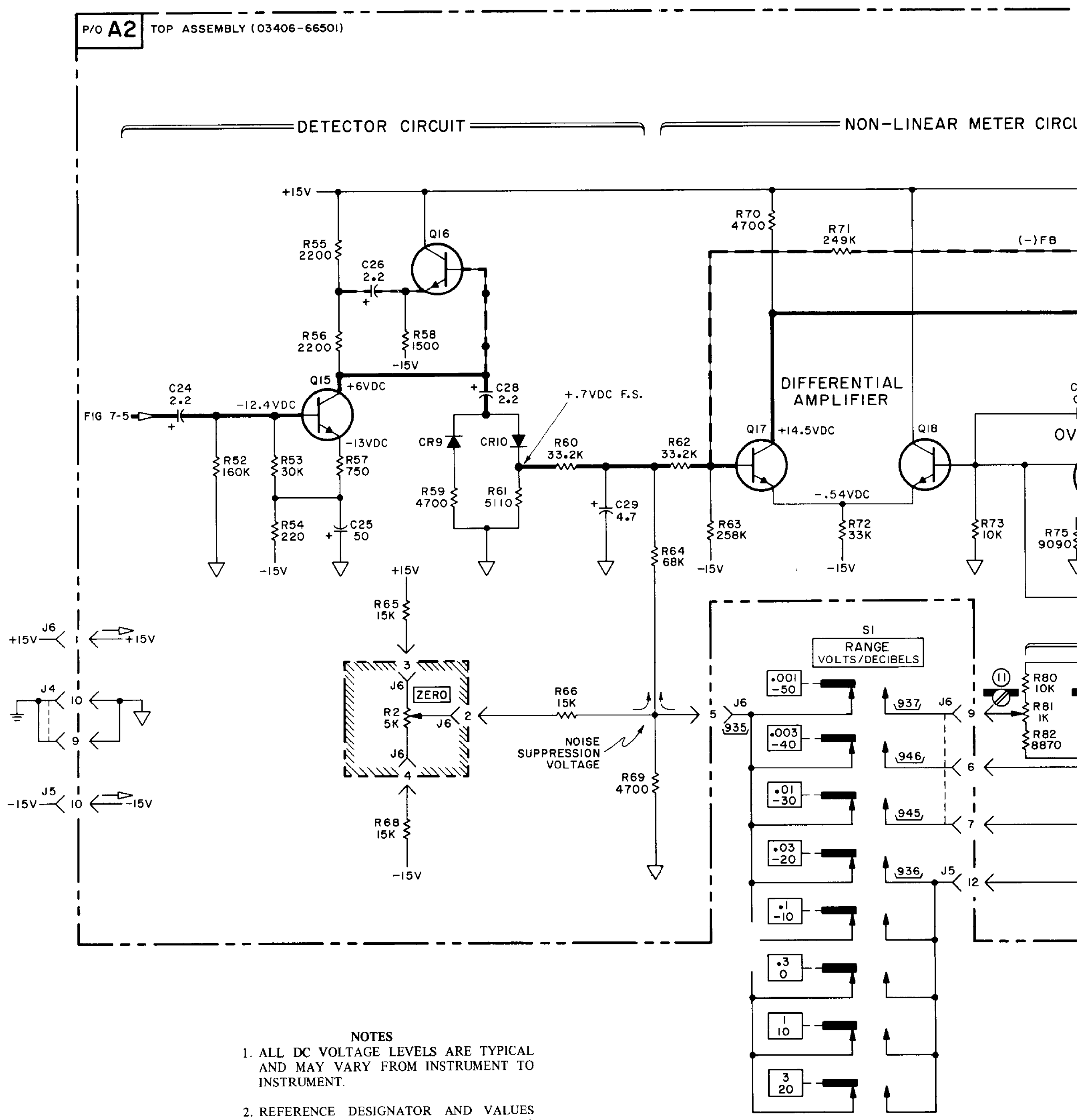
3406A-8-1986



A2

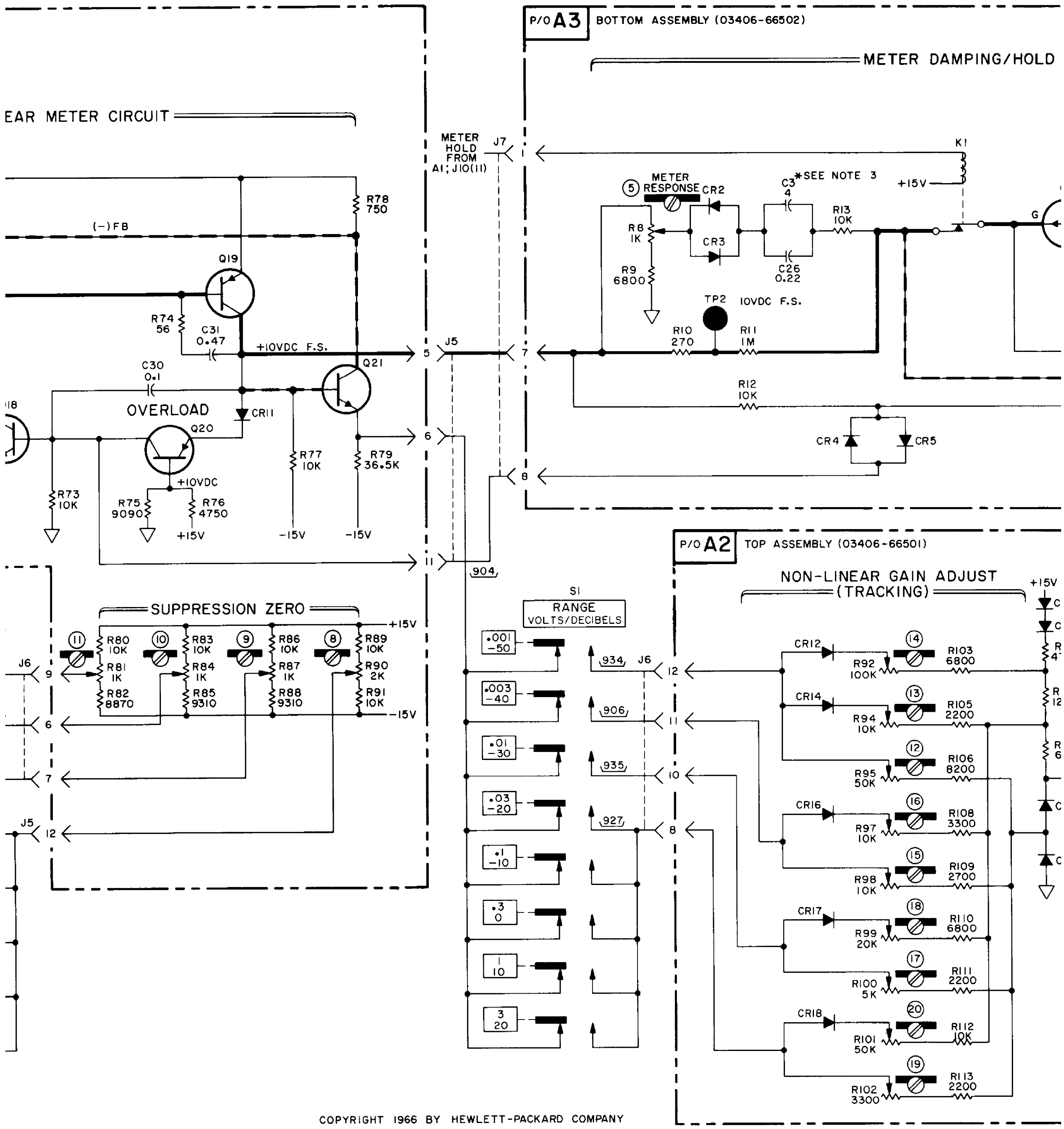
hp Part No. 03406-66501

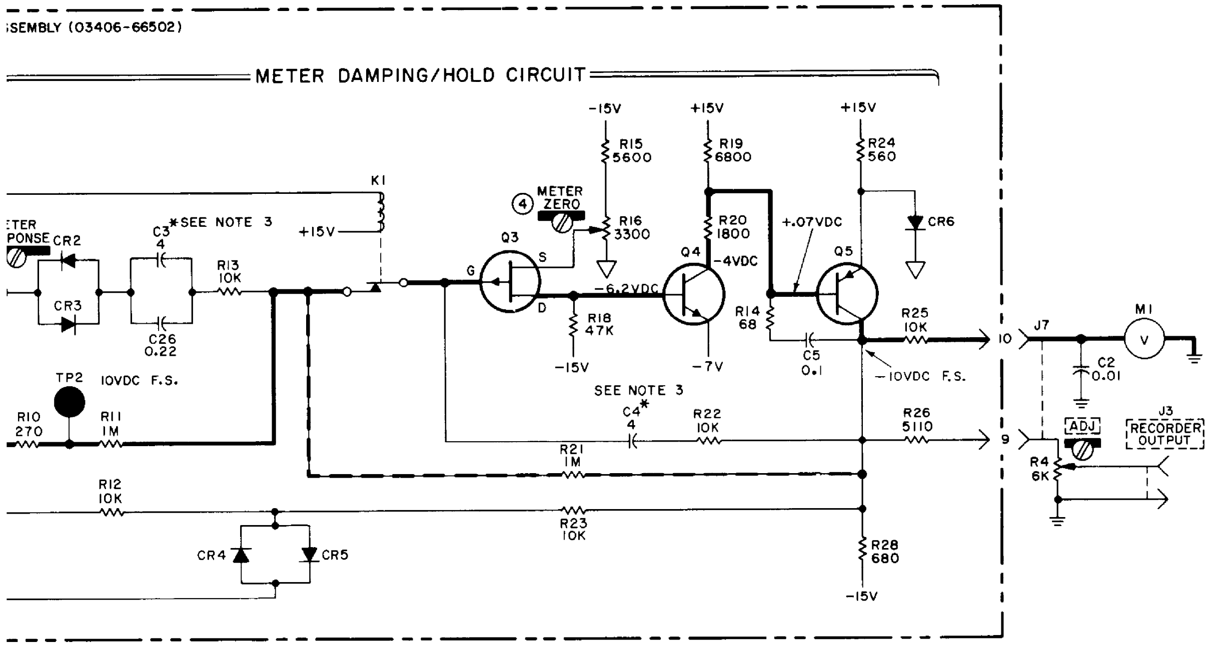
Rev A

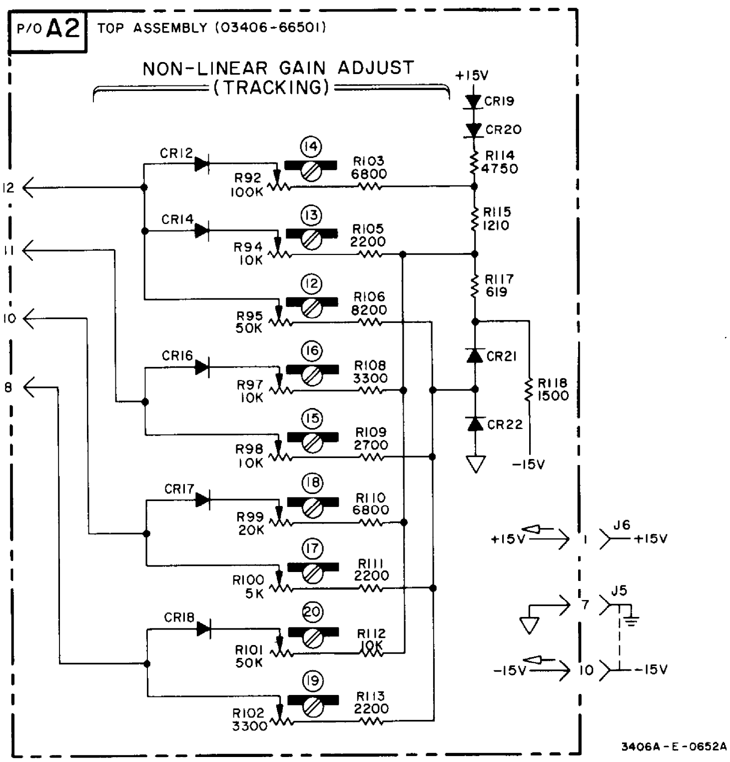


NOTES

1. ALL DC VOLTAGE LEVELS ARE TYPICAL AND MAY VARY FROM INSTRUMENT TO INSTRUMENT.
2. REFERENCE DESIGNATOR AND VALUES SHOWN IN RED DENOTES BACKDATING CHANGE. SEE APPENDIX C.
3. A3C3 AND A3C4 MATCHED PAIR TO WITHIN 1%.


P/O A3 BOTTOM ASSEMBLY (03406-66502)
METER DAMPING/HOLD
EAR METER CIRCUIT
METER HOLD FROM A1; J10(11)
J7
K1
METER RESPONSE
*SEE NOTE 3
+15V
R78 750
(-)FB
Q19
R74 56
C31 0.47
+10VDC F.S.
C30 0.1
Q21
OVERLOAD
Q20
CR11
R77 10K
R79 36.5K
+10VDC
R73 10K
R75 9090
R76 4750
+15V
-15V
-15V
R8 1K
R9 6800
CR2
CR3
C3 4
C26 0.22
R13 10K
TP2
10VDC F.S.
R10 270
R11 1M
R12 10K
CR4
CR5
J5
904
S1
RANGE VOLTS/DECIBELS
SUPPRESSION ZERO
+15V
-15V
R80 10K
R81 1K
R82 8870
R83 10K
R84 1K
R85 9310
R86 10K
R87 1K
R88 9310
R89 10K
R90 2K
R91 10K
J6
.001 -50
.003 -40
.01 -30
.03 -20
.1 -10
.3 0
1 10
3 20
934
906
935
927
P/O A2 TOP ASSEMBLY (03406-66501)
NON-LINEAR GAIN ADJUST (TRACKING)
CR12
R92 100K
R103 6800
CR14
R94 10K
R105 2200
R95 50K
R106 8200
CR16
R97 10K
R108 3300
R98 10K
R109 2700
CR17
R99 20K
R110 6800
R100 5K
R111 2200
CR18
R101 50K
R112 10K
R102 3300
R113 2200



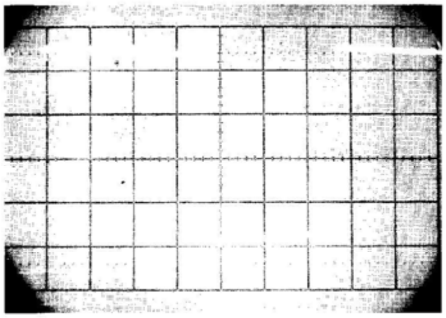


(11)	1MV		SUPPRESSION ZERO
(10)	3MV		
(9)	10MV		
(8)	30MV		
(14)	1MV	FS	TRACKING
(13)	1MV	.316FS	
(12)	1MV	.1FS	
(16)	3MV	FS	
(15)	3MV	.1FS	
(18)	10MV	FS	
(17)	10MV	.1FS	
(20)	30MV	FS	
(19)	30MV	.1FS	

Figure 7-6. Metering Circuits

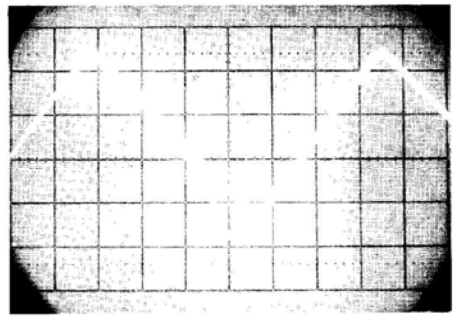
All waveforms should be observed with 10:1 Probe

(7) (Collector of A3Q7)



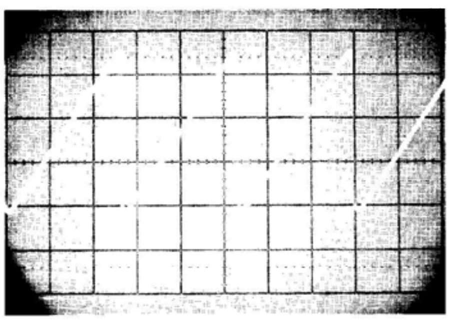
Vert = 2V/cm; sweep = 20ms/cm
AC coupled; Ext. trigger

(8) Triangle Generator Output
(Collector of A3Q8)



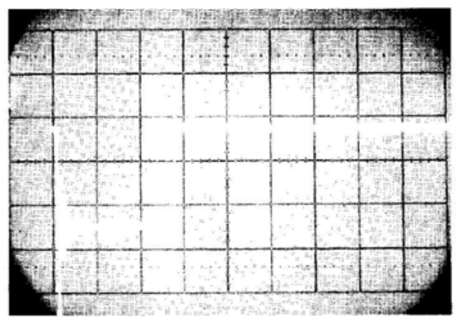
Vert = 2V/cm; sweep = 20ms/cm
AC coupled; Ext. trigger-collector A3Q7

(9) Sawtooth Generator Output
(Collector A3Q9)



Vert = 5V/cm; sweep = 20us/cm
AC coupled; sweeping stopped*

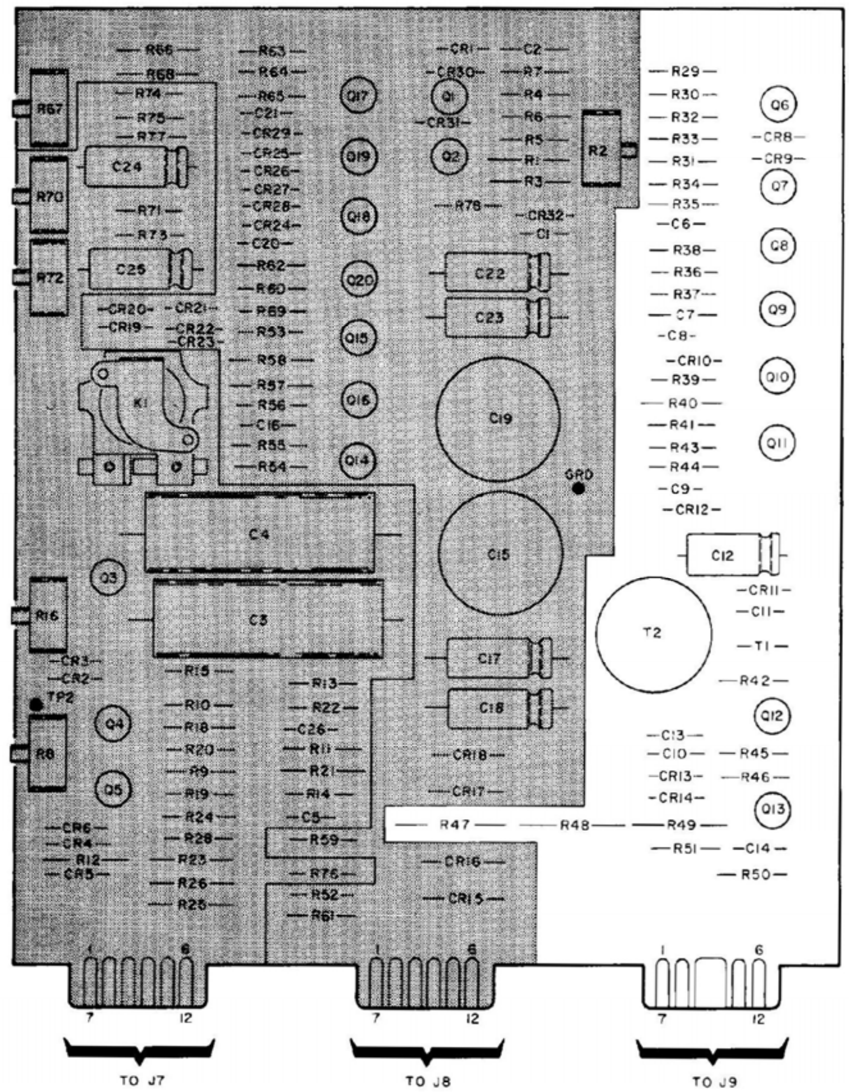
(10) Pulse Generator Output
(Anode of A3CR14)



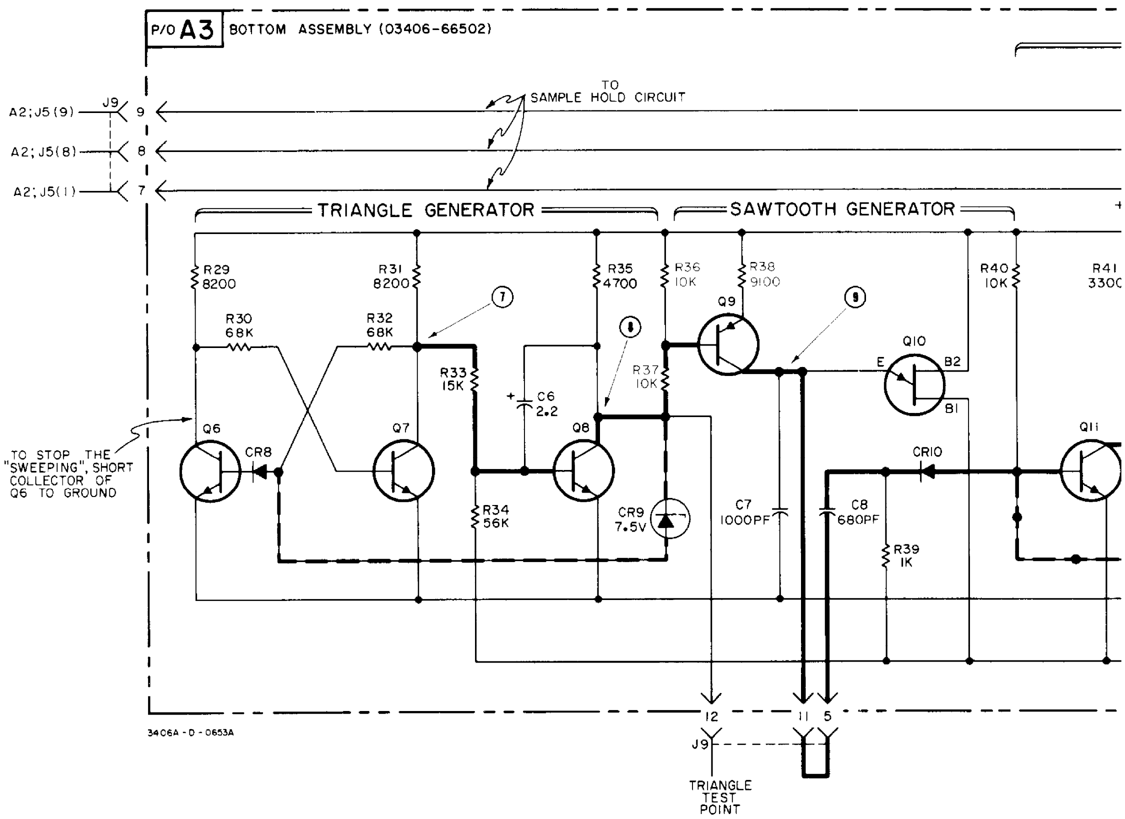
Vert = 5V/cm; sweep = 1us/cm
AC coupled; sweeping stopped*

*To stop sweeping, short collector of A3Q6 to ground (Figure 7-8).

Figure 7-7. Pulse Generator Circuit Waveforms



A3
hp Part No. 03406-66502
Rev C



NOTES

1. WAVEFORMS (8) AND (10) ARE OBSERVED WITH THE COLLECTOR OF A3Q6 SHORTED TO GROUND.
2. ALL DC VOLTAGE LEVELS ARE TYPICAL AND MAY VARY FROM INSTRUMENT TO INSTRUMENT.
3. REFERENCE DESIGNATOR AND VALUES SHOWN IN RED DENOTES BACKDATING CHANGE. SEE APPENDIX C.

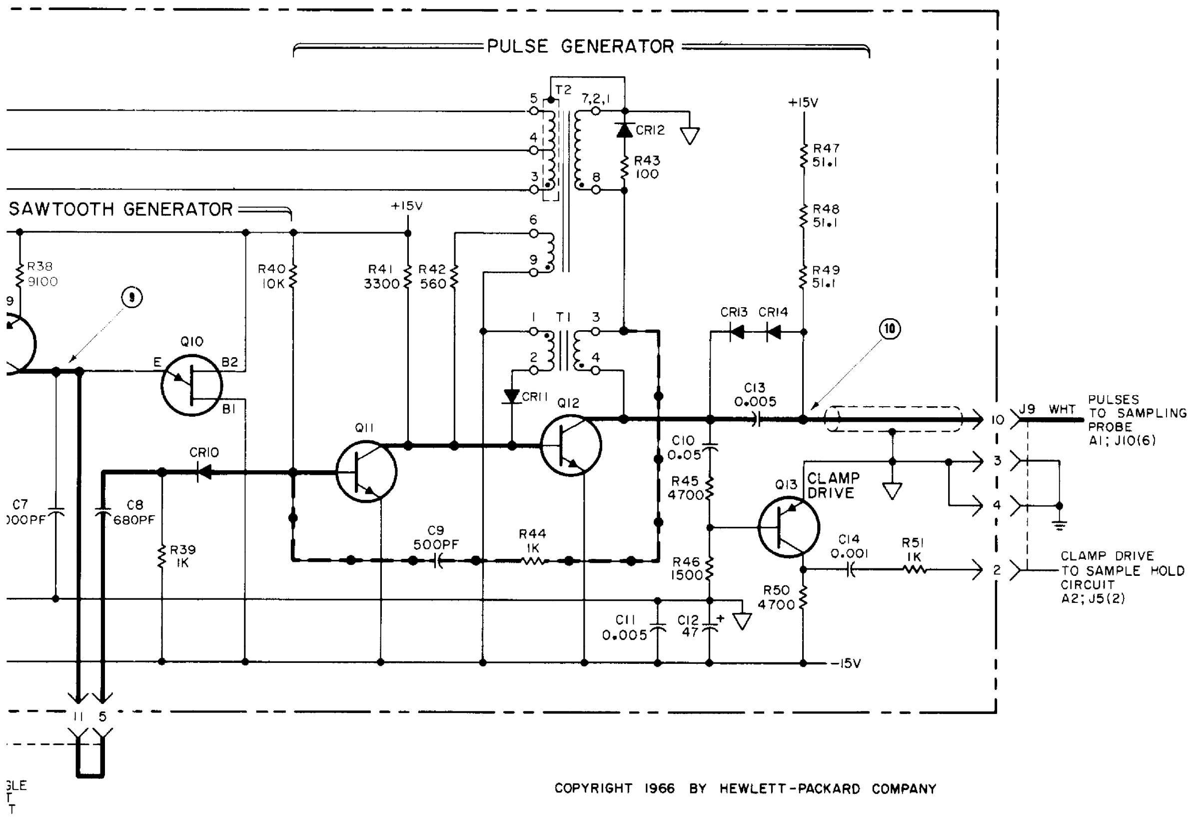


Figure 7-8. Pulse Generator Circuits

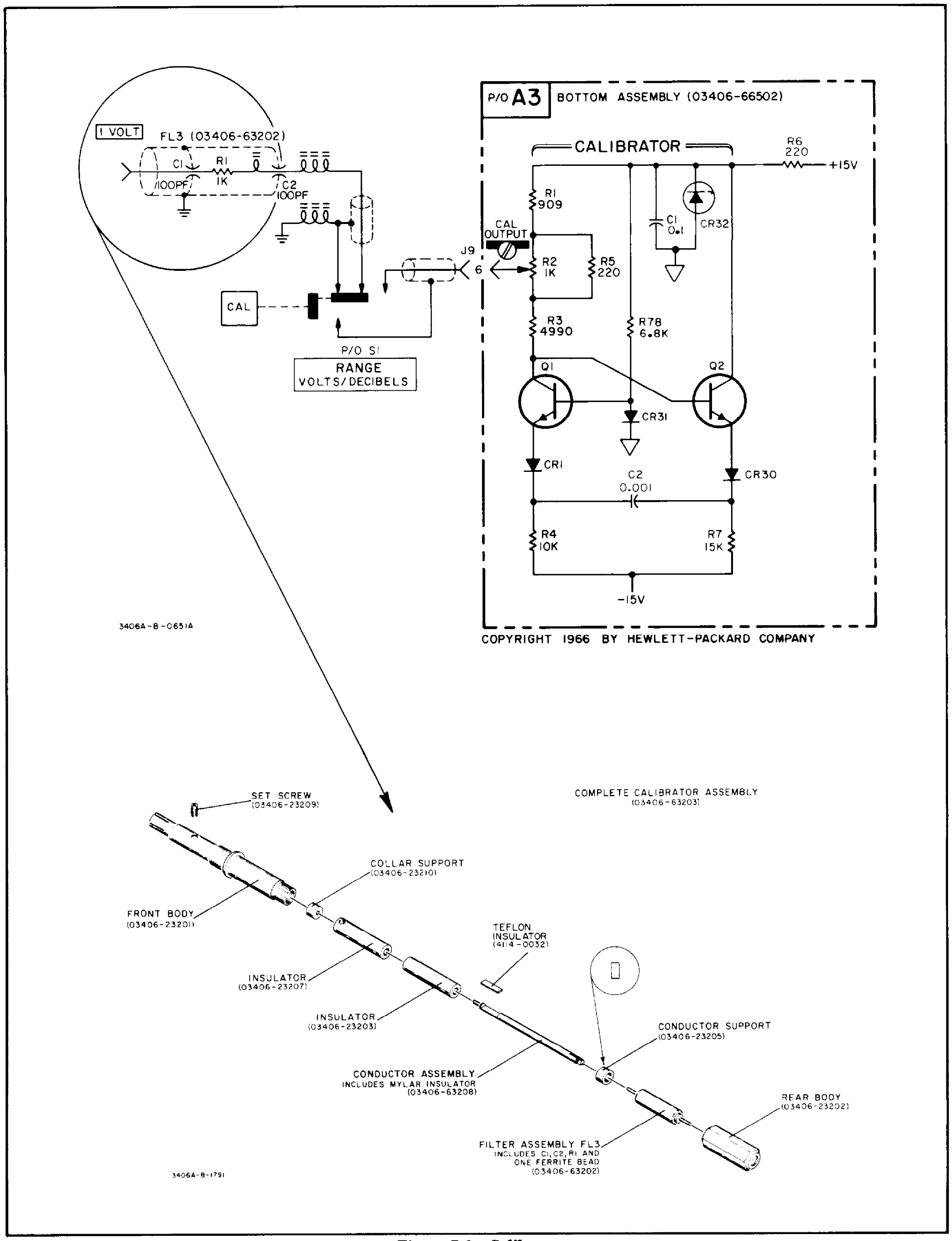
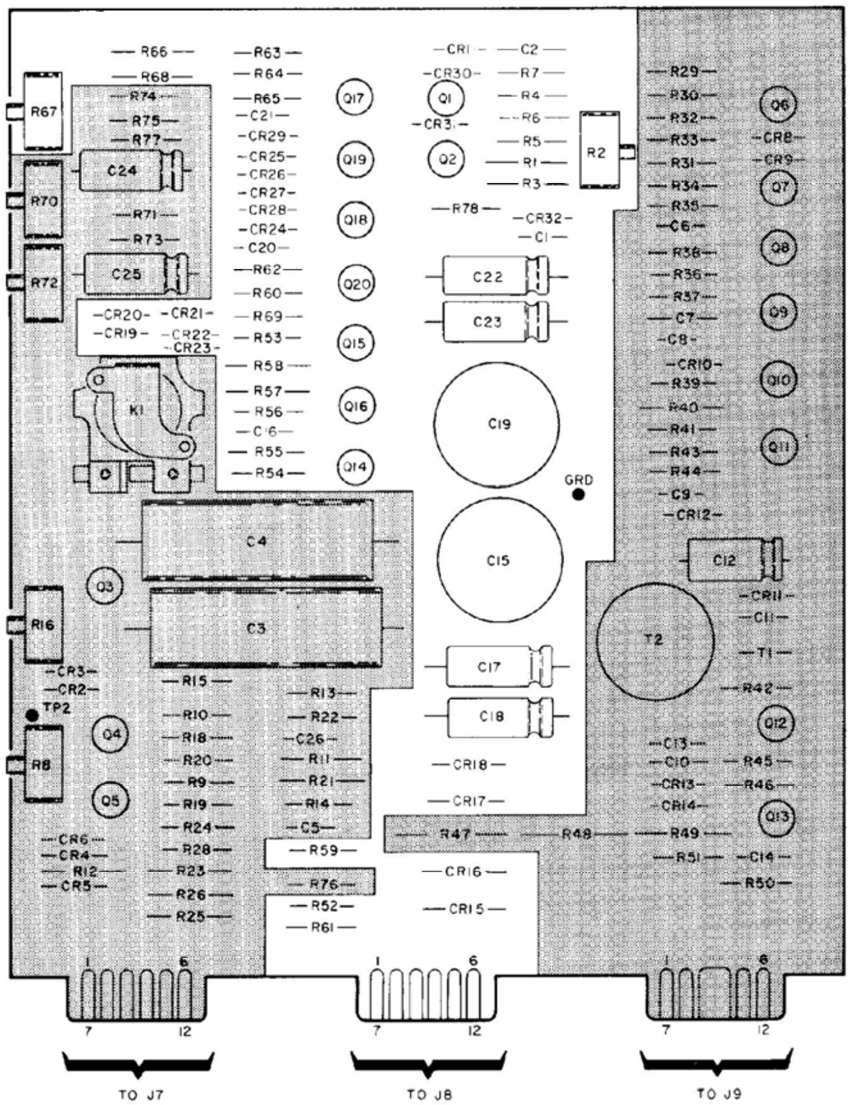


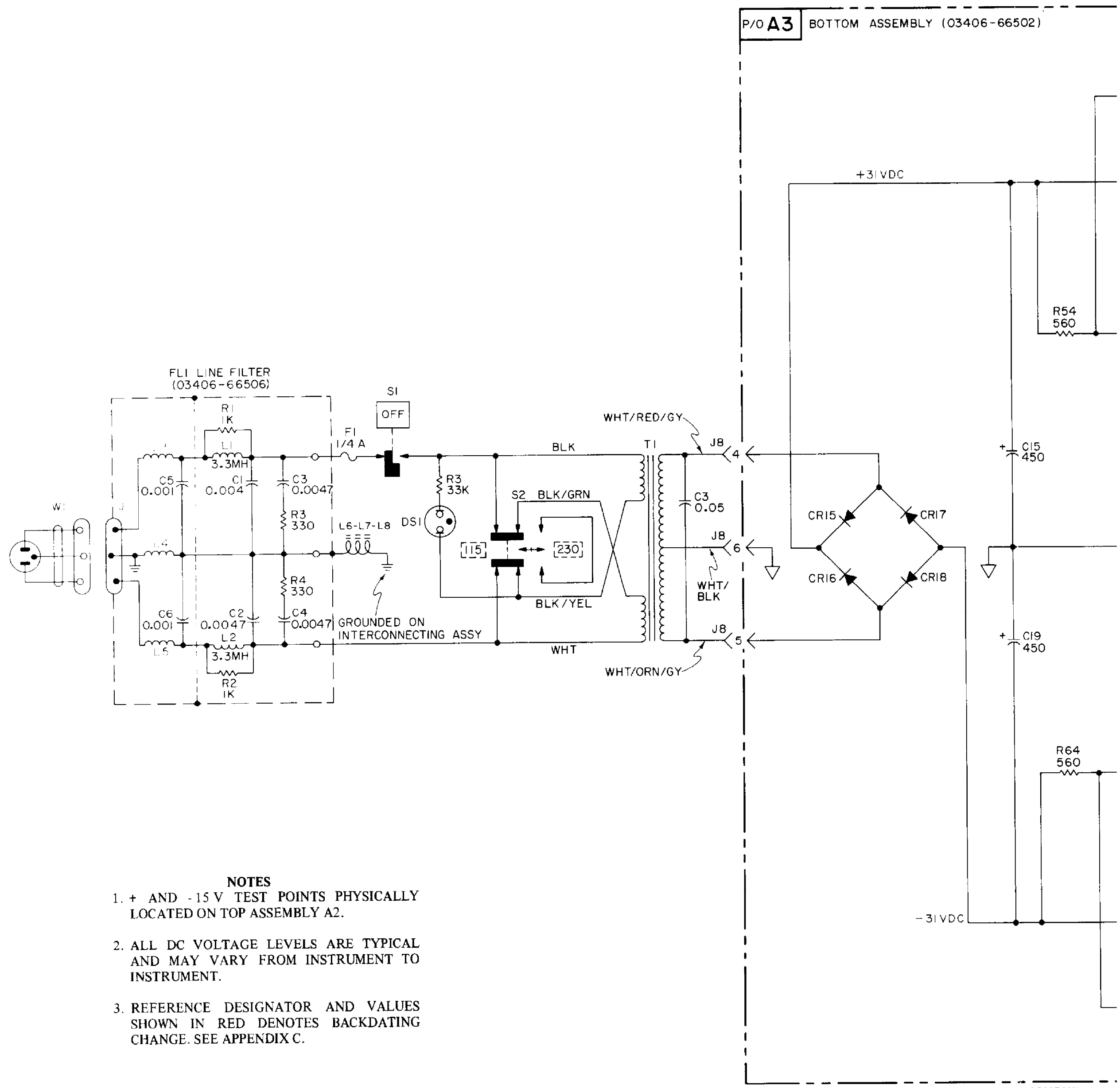
Figure 7-9. Calibrator



A3

hp Part No. 03406-66502

Rev C



NOTES

1. + AND -15 V TEST POINTS PHYSICALLY LOCATED ON TOP ASSEMBLY A2.
2. ALL DC VOLTAGE LEVELS ARE TYPICAL AND MAY VARY FROM INSTRUMENT TO INSTRUMENT.
3. REFERENCE DESIGNATOR AND VALUES SHOWN IN RED DENOTES BACKDATING CHANGE. SEE APPENDIX C.

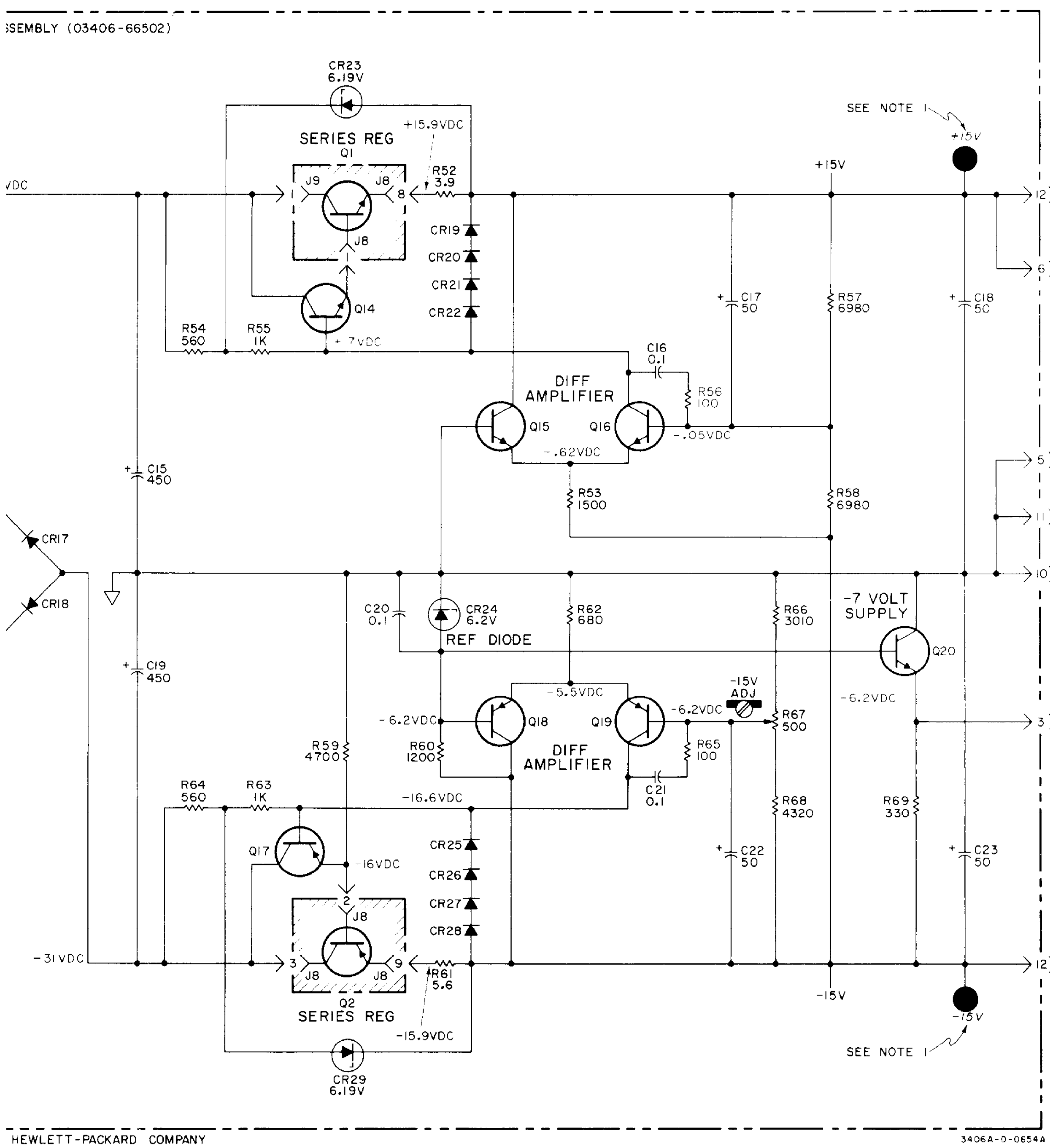


Figure 7-10. Power Supply Circuit

MANUAL BACKDATING CHANGES

MODEL 3406A

BROADBAND SAMPLING VOLTMETER

Manual backdating changes describe changes necessary to adapt this manual to earlier instruments. The backdating changes are listed in the following table according to instrument serial number or serial prefix.

Instrument component values that differ from those in the manual, yet are not listed in the Backdating Changes should be replaced with the part number given in Section VI of this manual.

Instrument Serial Numbers	Make Manual Changes
841-	1
830-	1 and 2
625-01001 to 01170	1 and 2
625-00751 to 01000	1 thru 3
625-00476 to 00750	1 thru 4
625-00475 and below	1 thru 5

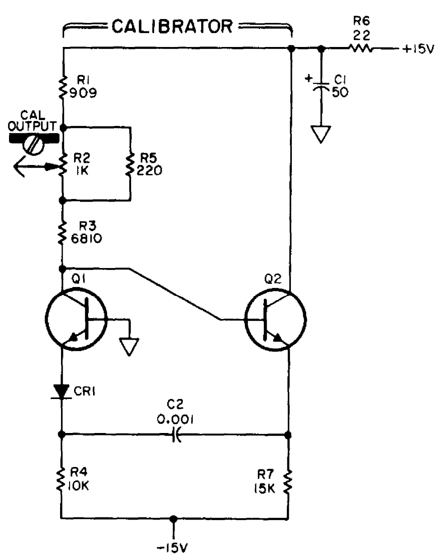
Instrument Serial Numbers	Make Manual Changes
606-00325 and below	1 thru 6
606-00250 and below	1 thru 7
606-00200 and below	1 thru 8
606-00150 and below	1 thru 9
606-00148 and below	1 thru 10

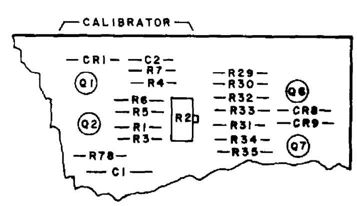
CHANGE NO. 1

1. Change FL1 Assembly Part No. to 03406-66506.
2. Change Rear panel to Part No. 03406-00202.
3. Change Rear sub-panel to Part No. 03406-00203.
4. Change Power Cord to Part No. 8120-0078.
5. Change each inductor FL1L3, FL1L4, FL1L5 to three ferrite beads Part No. 9170-0016.

CHANGE NO. 2

1. Change the calibrator schematic in Figure 7-9 and the A3 Assembly component location diagram in Figure 7-10 as shown below:





2. Change A3C1 to 50 uF, Part No. 0180-0058.
3. Change A3R3 to 6.81 kilohms 1% 1/4 W, Part No. 0757-0750.
4. Change A3R6 to 22 ohms 10% 1/2 W, Part No. 0687-2201.
5. Delete A3CR30, A3CR31, A3CR32.

CHANGE NO. 3

1. Change the Calibrator Assembly Part No. from 03406-63203 to 03406-63201.
2. Delete steps c thru e of Paragraph 5-39.

CHANGE NO. 4

1. In Figure 7-5 and Table 6-1. Change S1R13 to 189.7 ohms 1/4% 1/4 W Part No. 0698-3137, and delete R5.
2. Delete step q of Paragraph 5-37.

CHANGE NO. 5 (Ref. Service Note 3406A-5)

1. Change A3R56 and A3R65 in the Power Supply Circuit to 270 ohms ± 10% 1/2 W (Part No. 0687-2711).

NOTE

If any one of the above resistors fail, or if power supply oscillations are observed, replace both resistors with 100 ohm resistor identified in Table 6-1.

2. Change A3R28 to 680 ohms (Part No. 0687-6811) to prevent recorder output loading.

CHANGE NO. 6 (Ref. Service Note 3406A-2)

NOTE

Do not make Change No. 6 for the following instruments:

606-00201 thru 606-00203	606-00232	606-00240
606-00206 thru 606-00208	606-00234	606-00241
606-00231	606-00237	606-00243

1. Remove ferrite bead (Ō) in emitter circuit of A2Q2.

NOTE

If A2Q2 fails or starts to oscillate replace it with Part No. listed in Table 6-1 and add a ferrite bead (Part No. 9170-0016) around the emitter lead.

2. Change the following resistors in the Pulse Generator Circuit:
   A3R36 to 22 kilohms ± 10% 1/2 W (Part No. 0687-2231).
   A3R37 to 10 kilohms ± 10% 1/2 W (Part No. 0687-1031).
   A3R38 to 12 kilohms ± 10% 1/2 W (Part No. 0687-1231).

NOTE

If any one of the resistors above fail, or if a dip in frequency response occurs at 20 kHz, replace all three resistors with parts identified in Table 6-1.

CHANGE NO. 7 (Ref Service Note 3406A-3)

NOTE

Do not make Change No. 7 for the following instruments:

606-00203	606-00211	606-00215	606-00224	606-00235
606-00204	606-00212	606-00221	606-00228	606-00240
606-00208	606-00213	606-00222	606-00233	606-00246
606-00209	606-00214	606-00223	606-00234	606-00247
				606-00248

1. Change S1R14 and A2R34 in the Stabilizing Amplifier Circuit to 6.8 kilohms ± 10% 1/2 W (Part No. 0687-6821) and 22 kilohms ± 10% 1/2 W (Part No. 0687-2231) respectively.

NOTE

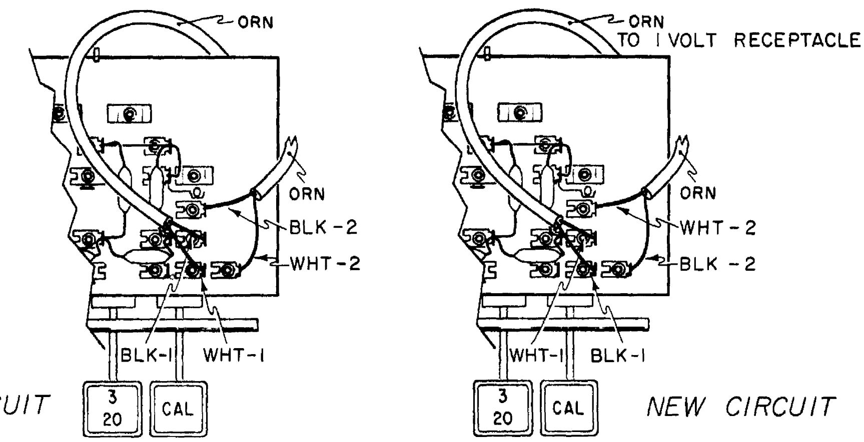
The value of S1R14 and A2R34 determines the Stabilizing Amplifier compensation and loop stability. If either resistor fails or if a ripple in the frequency response occur below 50 kHz replace both resistors with the parts identified in Table 6-1.

CHANGE NO. 8

1. All instruments 606-00200 and below should be changed to the "new" CAL pushbutton wiring shown below. This change will eliminate any ground loops when the CAL pushbutton is depressed, and make the wiring agree with the present schematic diagram.

NOTE

Some of the instruments may have been changed at the factory.



CHANGE NO. 9

1. Change A2R101 and A2R112 in the Non-linear Gain Adjust circuit to 20 kilohms ± 20% (Part No. 2100-0093) and 18 kilohms ± 10% 1/2 W (Part No. 0687-1831) respectively.

NOTE

To increase the range of the 30 MV FS TRACKING adjustment, replace both A2R101 and A2R112 with the parts identified in Table 6-1.

CHANGE NO. 10

NOTE

Do not make Change No. 10 for the following instruments:

606-00101 thru 606-00103	606-00122	606-00133
606-00107 thru 606-00111	606-00123	606-00137
606-00115	606-00127	606-00138
606-00116	606-00129	606-00141
606-00118	606-00131	606-00144

1. Change A3R7 and A3C2 in the Calibration circuit to 10 kilohms ± 10% 1/2 W (Part No. 0687-1031) and 0.0068 uF ± 10% 200 vdcw (Part No. 0160-0159).

NOTE

The front panel calibration accuracy may be improved by changing the resistor A3R7 and capacitor A3C7 to 15 kilohms ± 10% 1/2 W (Part No. 0687-1531) and 0.001 uF ± 10% (Part No. 0160-0153). If this change is made it will also be necessary to make Change No. 8.

MANUAL CHANGES

MODEL 3406 A

BROADBAND SAMPLING VOLTMETER

Manual Serial Prefixes: -942-
Manual Printed: OCT. 1969

To adapt this manual to instruments with other serial prefixes check for errata below, and make changes shown in tables.

Instrument Serial Prefixes	Make Manual Changes
G-902-	

ERRATA: In Section VI, Replaceable Parts and pertaining Schematics: -hp- Stock No.

Change:

On Page 7-7, Fig. 7-5, Add Ref. Des. R5 to
R:var., 5K, connected between S1R8/S1R13 to ground.

A1C7,C8	C:fxd., to 470 pF	
A2C5,9,20	to C:fxd., cer., 0.1 uF, +80-20%, 50 WVDC	0160-2914
A3C1,5,16,21	to C:fxd., cer., 0.1 uF, +80-20%, 50 WVDC	0160-2914
A3C10	to C:fxd., cer., 0.05 uF, +80-20%, 100 WVDC	0160-2917
A3CR15 thru CR18	Diode - Silicon 200 PIV to	1901-0026
A3Q12	Transistor NPN Si 2N3646 to	1854-0094
A3Q17	Transistor PNP Si to	1853-0090

For instruments equipped with German "Schuko" power cable change Assy Cable Power to 8120-0100

F1	Fuse 0.25 Amp. 250 V, slow-blow to	2110-0018
FL1	Assy - Line Filter to	03406-66506
FL1C5,6	to C:fxd., 0.001 uF, +100-20%, 500 WVDC	0160-2905
J1	Connector - AC Power	1251-0148

SERVICE NOTE

SUPERSEDES
P-03406-62103

-hp- Model 3406A Broadband Sampling Voltmeter

Serial Number 625-00550 and below

MODIFICATION REQUIRED FOR REPLACEMENT OF SAMPLING PROBE ASSEMBLY

This Serivce Note also applies to the deleted probe part no. 03406-62103

When the sampling probe assembly (A1) in the Hewlett-Packard Model 3406A is replaced, the instrument should be checked for the following modification. The modification will be done under warranty at your nearest -hp- Sales and Service office. The modification improves the probe/instrument compatibility.

With the new probe installed, the 3406A will have a new accuracy specification and maximum input specification as shown below.

10kHz – 20kHz	20kHz – 25kHz	25kHz – 100kHz	100kHz – 100MHz	100MHz – 700MHz	700MHz – 1GHz	1GHz – 1.2GHz
±13%	±8%	±5%	±3%	±5%	±8%	±13%

MAX PROBE INPUT: 30V p-p or 100VDC
MAX INPUT WITH 10:1 DIVIDER: 200VDC

All -hp- Model 3406A's with the serial numbers 625-00550 and below require the following modifications:

1. Replace A3C12 with -hp- Part No. 0180-0097 which is a 47 μF, 35 VDCW capacitor.
2. A3R40 is no longer a starred value. A3R40 should be a 10K resistor, -hp- Part No. 0757-0340.
3. A2R28 is now a starred value (1.5K to 3.3K). Its value should be chosen according to the following steps:
   a. ZERO the Model 3406A according the paragraph 3-8 in the manual.
   b. Insert the bare probe into 1 VOLT receptacle; depress 3 volt RANGE and CAL pushbuttons.
   c. Adjust front panel CALIBRATE control from full CW to full CCW and note meter indication at both ends. Range of CALIBRATE control should be at least .8 to 1.2V.
   d. Decrease value of A2R28 to obtain .8V reading; increase value of A2R28 to obtain 1.2V reading.
4. Replace Range Switch Assembly resistor S1R13 with a 200Ω, 1%, 1/4W resistor (-hp- Part No. 0698-4574). Also, add S1R15, 5K variable resistor (-hp- Part No. 2100-1760), according to the diagram on the following page.

   Set 1 volt level with probe in 1 VOLT receptacle and CAL button depressed. Switch to 3 volt range and adjust S1R15 for a 1 volt indication. Switch back to 1 volt range and if the indication is not 1 volt, split the difference with S1R15.

LAD/km/WA

6-73/9

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HEWLETT PACKARD

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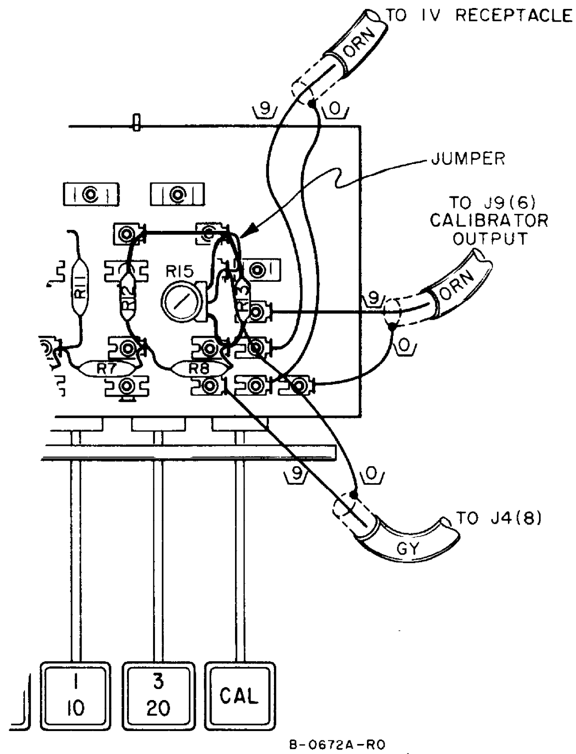
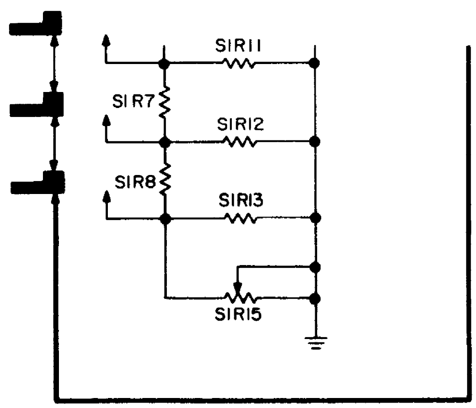


The schematic configuration is as follows:



5. On the A3 Board, interchange the leads of A3T1 as follows:

   Interchange leads 1 and 4
   Interchange leads 2 and 3

   The white doe on the transformer should now be on the transformer side nearest A3C11. Previously, the white dot was on the transformer side nearest A3R42.

   After these changes, the instrument should be completely recalibrated.