

OPERATING AND SERVICE MANUAL

MODEL 3580A SPECTRUM ANALYZER

Serial Number: 2030A6633

IMPORTANT NOTICE

If the Serial Number of your instrument is lower than the one on this title page, the manual contains revisions that do not apply to your instrument. Backdating information given in the manual adapts it to these earlier instruments.

Where practical, backdating changes are given on the schematic diagrams. These changes are indicated by a dagger sign (†) which refers to the corresponding backdating note on the schematic or apron page. Backdating changes not given on the schematics are flagged by a numbered delta (Δ_1) which refers to the corresponding numbered change in the Backdating Section (Section VIII).

WARNING

To help minimize the possibility of electrical fire or shock hazards, do not expose this instrument to rain or excess moisture.

Manual Part No. 03580-90003

Microfiche No. 03580-90093

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CERTIFICATION

Hewlett-Packard Company certifies that this product met its published specifications at the time of shipment from the factory. Hewlett-Packard further certifies that its calibration measurements are traceable to the United States National Bureau of Standards, to the extent allowed by the Bureau's calibration facility, and to the calibration facilities of other International Standards Organization members.

WARRANTY

This Hewlett-Packard product is warranted against defects in material and workmanship for a period of one year from date of shipment [,except that in the case of certain components listed in Section I of this manual, the warranty shall be for the specified period]. During the warranty period, Hewlett-Packard Company will, at its option, either repair or replace products which prove to be defective.

For warranty service or repair, this product must be returned to a service facility designated by -hp-. Buyer shall prepay shipping charges to -hp- and -hp- shall pay shipping charges to return the product to Buyer. However, Buyer shall pay all shipping charges, duties, and taxes for products returned to -hp- from another country.

Hewlett-Packard warrants that its software and firmware designated by -hp- for use with an instrument will execute its programming instructions when properly installed on that instrument. Hewlett-Packard does not warrant that the operation of the instrument, or software, or firmware will be uninterrupted or error free.

LIMITATION OF WARRANTY

The foregoing warranty shall not apply to defects resulting from improper or inadequate maintenance by Buyer, Buyer-supplied software or interfacing, unauthorized modification or misuse, operation outside of the environmental specifications for the product, or improper site preparation or maintenance.

NO OTHER WARRANTY IS EXPRESSED OR IMPLIED. HEWLETT-PACKARD SPECIFICALLY DISCLAIMS THE IMPLIED WARRANTIES OF MERCHANTABILITY AND FITNESS FOR A PARTICULAR PURPOSE.

EXCLUSIVE REMEDIES

THE REMEDIES PROVIDED HEREIN ARE BUYER'S SOLE AND EXCLUSIVE REMEDIES. HEWLETT-PACKARD SHALL NOT BE LIABLE FOR ANY DIRECT, INDIRECT, SPECIAL, INCIDENTAL, OR CONSEQUENTIAL DAMAGES, WHETHER BASED ON CONTRACT, TORT, OR ANY OTHER LEGAL THEORY.

ASSISTANCE

Product maintenance agreements and other customer assistance agreements are available for Hewlett-Packard products.

For any assistance, contact your nearest Hewlett-Packard Sales and Service Office. Addresses are provided at the back of this manual.



CATHODE-RAY TUBE WARRANTY AND INSTRUCTIONS

The cathode-ray tube (CRT) supplied in your Hewlett-Packard Instrument and replacement CRT's purchased from -hp- are warranted by the Hewlett-Packard Company against electrical failure for a period of one year from the date of shipment from Colorado Springs. Broken tubes and tubes with phosphor or mesh burns are not included under this warranty. No other warranty is expressed or implied.

INSTRUCTION TO CUSTOMERS

If the CRT is broken when received, a claim should be made with the responsible carrier. All warranty claims with Hewlett-Packard should be processed through your nearest Hewlett-Packard Sales/Service Office (listed at rear of instrument manual).

INSTRUCTIONS TO SALES/SERVICE OFFICE

Return defective CRT in the replacement CRT packaging material. If packaging material is not available, contact CRT Customer Service in Colorado Springs. The Colorado Springs Division must evaluate all CRT claims for customer warranty, Material Failure Report (MFR) credit, and Heart System credit. A CRT Failure Report form (see reverse side of this page) must be completely filled out and sent with the defective CRT to the following address:

HEWLETT-PACKARD COMPANY

1900 Garden of the Gods Road Colorado Springs, Colorado 80907

Parcel Post Address:

P.O. Box 2197 Colorado Springs, Colorado 80901

Attention: CRT Customer Service

Defective CRT's not covered by warranty may be returned to Colorado Springs for disposition. These CRT's, in some instances, will be inspected and evaluated for reliability information by our engineering staff to facilitate product improvements. The Colorado Springs Division is equipped to safely dispose of CRT's without the risks involved in disposal by customers or field offices. If the CRT is returned to Colorado Springs for disposal and no warranty claim is involved, write "Returned for Disposal Only" in item No. 5 on the form.

Do not use this form to accomplish CRT repairs. In order to have a CRT repaired, it must be accompanied by a customer service order (repair order) and the shipping container must be marked "Repair" on the exterior.

Printing History

New editions are complete revisions of the manual. Update packages, which are issued between editions, contain additional and replacement pages to be merged into the manual by the customer. The dates on the title page change only when a new edition or a new update is published. No information is incorporated into a reprinting unless it appears as a prior update; the edition does not change when an update is incorporated.

A software code may be printed before the date; this indicates the version level of the software product at the time the manual or update was issued. Many product updates and fixes do not require manual changes and, conversely, manual corrections may be done without accompanying product changes. Therefore, do not expect a one-to-one correspondence between product updates and manual updates.

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SAFETY SUMMARY

The following general safety precautions must be observed during all phases of operation, service, and repair of this instrument. Failure to comply with these precautions or with specific warnings elsewhere in this manual violates safety standards of design, manufacture, and intended use of the instrument. Hewlett-Packard Company assumes no liability for the customer's failure to comply with these requirements. This is a Safety Class 1 instrument.

GROUND THE INSTRUMENT

To minimize shock hazard, the instrument chassis and cabinet must be connected to an electrical ground. The instrument is equipped with a three-conductor ac power cable. The power cable must either be plugged into an approved three-contact electrical outlet or used with a three-contact to two-contact adapter with the grounding wire (green) firmly connected to an electrical ground (safety ground) at the power outlet. The power jack and mating plug of the power cable meet International Electrotechnical Commission (IEC) safety standards.

DO NOT OPERATE IN AN EXPLOSIVE ATMOSPHERE

Do not operate the instrument in the presence of flammable gases or fumes. Operation of any electrical instrument in such an environment constitutes a definite safety hazard.

KEEP AWAY FROM LIVE CIRCUITS

Operating personnel must not remove instrument covers. Component replacement and internal adjustments must be made by qualified maintenance personnel. Do not replace components with power cable connected. Under certain conditions, dangerous voltages may exist even with the power cable removed. To avoid injuries, always disconnect power and discharge circuits before touching them.

DO NOT SERVICE OR ADJUST ALONF

Do not attempt internal service or adjustment unless another person, capable of rendering first aid and resuscitation, is present.

USE CAUTION WHEN EXPOSING OR HANDLING THE CRT

Breakage of the Cathode-ray Tube (CRT) causes a high-velocity scattering of glass fragments (implosion). To prevent CRT implosion, avoid rough handling or jarring of the instrument. Handling of the CRT shall be done only by qualified maintenance personnel using approved safety mask and gloves.

DO NOT SUBSTITUTE PARTS OR MODIFY INSTRUMENT

Because of the danger of introducing additional hazards, do not install substitute parts or perform any unauthorized modification to the instrument. Return the instrument to a Hewlett-Packard Sales and Service Office for service and repair to ensure that safety features are maintained.

DANGEROUS PROCEDURE WARNINGS

Warnings, such as the example below, precede potentially dangerous procedures throughout this manual. Instructions contained in the warnings must be followed.

WARNING

Dangerous voltages, capable of causing death, are present in this instrument. Use extreme caution when handling, testing, and adjusting.

SAFETY SYMBOLS

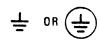
General Definitions of Safety Symbols Used On Equipment or In Manuals.



Instruction manual symbol: the product will be marked with this symbol when it is necessary for the user to refer to the instruction manual in order to protect against damage to the instrument.



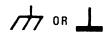
Indicates dangerous voltage (terminals fed from the interior by voltage exceeding 1000 volts must be so marked).



Protective conductor terminal. For protection against electrical shock in case of a fault. Used with field wiring terminals to indicate the terminal which must be connected to ground before operating equipment.



Low-noise or noiseless, clean ground (earth) terminal. Used for a signal common, as well as providing protection against electrical shock in case of a fault. A terminal marked with this symbol must be connected to ground in the manner described in the installation (operating) manual, and before operating the equipment.



Frame or chassis terminal. A connection to the frame (chassis) of the equipment which normally includes all exposed metal structures.

~

Alternating current (power line).

Direct current (power line).

 $\overline{}$

Alternating or direct current (power line).

DANGER

The DANGER sign denotes a hazard. It calls attention to an operating procedure, practice, condition or the like, which could result in injury or death to personnel even during normal operation.

WARNING

The WARNING sign denotes a hazard. It calls attention to a procedure, practice, condition or the like, which, if not correctly performed or adhered to, could result in injury or death to personnel.

ECAUTION 3

The CAUTION sign denotes a hazard. It calls attention to an operating procedure, practice, condition or the like, which, if not correctly performed or adhered to, could result in damage to or destruction of part or all of the product.

NOTE:

The NOTE sign denotes important information. It calls attention to procedure, practice, condition or the like, which is essential to highlight.

Table 1-1. Specifications.

FREQUENCY

Frequency Display Accuracy: ± 3.5 Hz;

0°C to 55°C, 1 Hz and 3 Hz Bandwidths only.

Display Accuracy: Frequency error between any two points is less than $\pm 2\%$ of their indicated separation.

| Bandwidths: | 1 Hz | 3 Hz | 10 Hz | 30 Hz | 100 Hz | 300 Hz |
|---------------------|-------|--------|-------|-------|--------|--------|
| (accuracy ± 15%) | (25°C | 2 ± 5° | cl | | | |

AMPLITUDE

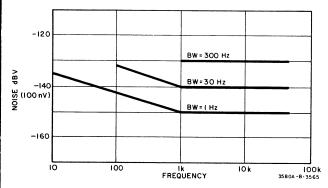
| Amplitude Accuracy: Frequency Response: * | Log | Linear |
|--|----------------------------------|---------------|
| 20 Hz—20 kHz 5 Hz—50 kHz | \pm 0.3 dB \pm 0.5 dB | ± 3% ± 5% |
| Switching Between Bandwidt 3 Hz—300 Hz 1 Hz—300 Hz | hs (25°C): ± 0.5 dB ± 1 dB | ± 5% ± 10% |
| Amplitude Display: | ± 2 dB | ± 2% |
| Input Attenuator: | $\pm0.3~dB$ | ± 3% |
| Amplitude Reference Level: (IF attenuator) | | |
| most sensitive range all other ranges | + 1 dB ± 1 dB | ± 10% ± 3% |

^{*}Standard 3580A and Option 002 unbalanced input.

Dynamic Range:

Display Range (Log 10 dB mode): > 80 dB

Noise Level: "Noise level is measured with 50 ohms placed across the input terminals. On the 30 to 300 Hz bandwidth use maximum display smoothing. The noise level as a function of frequency is:" (Refer to noise vs frequency graph).



Distortion (THD and 1M):

Std 3580A: > 80 dB below input reference level.

Option 002: > 80 dB below input reference level for signals below 0 dBm and above 100 Hz.

Spurious Responses: > 80 dB below input reference level Line Related Spurious: > 80 dB below input reference level or -135 dBV (0.18 μ V) Below -90 dBm for Option 002 Balanced-Terminated Input.

IF Feedthru:

Input Feedthru
> 10 V -60 dB or lower
< 10 V -70 dB or lower

Zero Response: > 30 dB below input reference level

Noise Sidebands (1 Hz Bandwidth): more than 70 dB below peak of CW signal ± 10 Hz away from center of response.

SWEEP

Sweep Times: 0.1 sec to 2,000 sec

Accuracy: ±5%

Log Sweep: 20 Hz to 43 kHz

Accuracy: 20% after 3 continuous sweeps

BALANCED INPUT (Option 002 only)

Frequency Response Δ_1 : \pm 0.5 dB, 40 Hz to 20 kHz for signals bebelow \pm 20 dBm.

Common Mode Rejection: > 70 dB at 60 Hz

OUTPUTS

Recorder Outputs:

X-Axis: 0 V to + 5 V \pm 2.5% Y-Axis: 0 V to + 5 V \pm 2.5%

Tracking Oscillator Output:

Frequency Response:

Std 3580A: ± 3%, 5 Hz to 50 kHz

Opt. 002: $\pm\,0.5$ dB, 100 Hz to 20 kHz, 10 kHz Reference, 600 Ω load.

L.0. Output: Frequency of output signal varies from 1.0 MHz to 1.5 MHz as analyzer frequency is tuned from 0 Hz to 50 kHz

Frequency Accuracy: The tuned frequency can be read to an accuracy of ± 5 Hz using an external counter.

 Δ_1 Serial No. 1312A-00465 and below: Change Frequency Response Specification to $\pm\,0.5$ dB, 300 Hz to 20 kHz.

 Δ_{16} Serial No. 1415A04280 and below: Change Frequency Dial Accuracy to \pm 100 Hz, 20° to 30°C; \pm 300 Hz 0° to 55°C.

SECTION I GENERAL INFORMATION

1-1. DESCRIPTION.

- 1-2. The Hewlett-Packard Model 3580A Spectrum Analyzer is a low frequency instrument that has been optimized for use in the 5 Hz to 50 kHz range. The 3580A functions as a signal analyzer or as a network analyzer. When used as a signal analyzer, the 3580A provides a graphical display of the spectral components of an input signal. When used as a network analyzer, the 3580A plots the amplitude vs. frequency characteristics of 2-port networks such as amplifiers, attenuators and filters.
- 1-3. The major features of the 3580A include a digitally stored display, adaptive sweep, six selectable bandwidths (1 Hz 300 Hz), 30 nV sensitivity and 80 dB dynamic range. These standard features, along with optional balanced inputs and an internal rechargeable battery pack, make the 3580A ideally suited for communications, geophyical, oceanography and metrology applications.

1-4. SPECIFICATIONS.

- 1-5. Table 1-1 is a complete list of the Model 3580A critical specifications that are controlled by tolerances. Table 1-2 contains general information describing the operating characteristics of the 3580A.
- 1-6. Any changes in specifications due to manufacturing, design, or traceability to the U.S. National Bureau of Standards are included in Table 1-1 in this manual. Specifications listed in this manual supersede all previous specifications for the Model 3580A.

1.7. OPTIONS.

1-8. There are three options available for the 3580A. Option 001 and Option 002 are listed in the following table. For further information concerning those options, refer to Table 1-2 or Section III in this manual or contact the nearest -hp- Sales and Service Office. Option 910 is an additional Operating and Service Manual.

| 3580A Option (Factory Installed) | Description |
|-------------------------------------|---|
| 001* | Internal rechargeable battery pack and front panel cover for complete portability |
| 002 | Balanced inputs; balanced tracking oscil- lator output |

^{*}Field Installation Kit Battery Pack only 03580-69508 Front Panel Cover Accessory -hp- 10101B.

1-9. Warranty Exceptions.

1-10. Batteries in Option 001 instruments are warranted for 90 days.

1-11. ACCESSORIES SUPPLIED.

1-12. The following is a list of accessories supplied with the 3580A.

| ltem | Qty. | -hp- Part No. |
|---|-------|---------------|
| Accessory Kit Includes the following: | 1 ea. | 03580-84401 |
| PC Board Extender (15 pin) | 2 ea. | 5060-0049 |
| PC Board Extender (10 pin) | 2 ea. | 5060-5917 |
| Fuse: 0.25 A, 250 V Normal Blo (for 220 V/240 V operation) | 1 ea. | 2110-0004 |

1-13. ACCESSORIES AVAILABLE.

1-14. The following is a list of Hewlett-Packard accessories available for use with the Model 3580A:

| -hp- Model | Description |
|----------------|----------------------------|
| 10004B | Voltage Divider Probe |
| 10101B | Front Panel Cover Assembly |
| 7035B Opt. 020 | X/Y Recorder |
| 197A or 198A | Oscilloscope Camera |

1-15. INSTRUMENT AND MANUAL IDENTIFICATION.

- 1-16. The instrument serial number is located on the rear panel. Hewlett-Packard uses a two-section serial number consisting of a four-digit prefix and a five-digit suffix. A letter between the suffix and prefix identifies the country in which the instrument was manufactured (A = USA, G = West Germany, J = Japan, U = United Kingdom). All correspondence with Hewlett-Packard should include the complete serial number.
- 1-17. If the serial number of your instrument is lower than the one on the title page of this manual, refer to Section VIII for backdating information that will adapt this manual to your instrument.

Table 1-2. General Information.

INPUT CHARACTERISTICS (Standard 3580A)

Connector: female banana plug
Impedance: 1 megohm, 30 pF

Maximum (ac) Input Level:

| Input Sensitivity | Maximum Input |
|----------------------------------|---------------|
| + 30 dB (20 V) to -10 dB (0.2 V) | 100 V rms |
| -20 dB (0.1 V to -70 dB (0.2 mV) | 50 V rms |

Maximum (dc) Input Voltage: ± 100 Vdc

Coupling: capacitive

DC Isolation: none (input common referenced to frame ground)

INPUT CHARACTERISTICS (Option 002)

Selectable Input Configurations:

Unbalanced Balanced Bridged Balanced Terminated

Connector: female banana plug

Impedance:

Unbalanced: 1 megohm, 40 pF

Greater than 12 K (typically 14 K at 1 kHz) Terminated: 600 ohms or 900 ohms

Maximum Input Levels:

Unbalanced: same as Standard 3580A Bridge: 100 V dc max, 35 V rms ac max

Terminated: +27 dBm at 0 V dc. (see Paragraph 3-187)

DC Isolation:

Unbalanced: none (input common referenced to frame

ground)

Bridged and Terminated: floating input

AMPLITUDE CHARACTERISTICS:

Amplitude Modes:

Linear: Absolute measurements in rms volts (average responding): relative measurements in percent of full scale.

Log 10dB/div.: Absolute measurements in dBV (1 V rms = 0 dBV) or dBm/600 ohms; relative measurements in dB. Display sensitivity is 10 dB per division; display range is > 80 dB.

Log 1 dB/div.: Display sensitivity is 1 dB per division; display range is 10 dB. Any 10 dB portion of 80 dB range can be displayed by changing the AMPLITUDE REF LEVEL control setting.

Full-Scale Sensitivity:

Linear Mode:

Calibrated: 20 V rms to 0.1 μ V rms (18 ranges) Uncalibrated: 100 V rms to 0.2 μ V rms

Log 10 dB Mode:

Calibrated: +30 dBV/dBm to -70 dBV/dBm (11 ranges) Uncalibrated: +40 dBV/dBm to -60 dBV/dBm

Overload Indicator: An LED Overload indicator on the front panel lights to indicate that the input signal exceeds the maximum (full scale) input level set by the INPUT SENSITIVITY switch and amplitude VERNIER.

Internal Calibration Signal: An internally generated calibration signal can be used to calibrate the amplitude section (following input attenuator) to an accuracy of $\pm 1.5\%$ at 10 kHz. The calibration signal can also be used to verify the frequency accuracy of the instrument.

FREQUENCY CHARACTERISTICS:

Frequency Range: 5 Hz to 50 kHz

Frequency Control: The front panel FREQUENCY control tunes the frequency of the analyzer over the 0 Hz to 50 kHz range. The control can be used to set either the start or center frequency of linear sweeps.

Δ16 Course and Fine Tuning: Course and fine tuning is performed by using the concentric knobs in the upper right corner of the front panel. The knob closest to the front panel controls the course tuning. The knob furthest from the front panel controls the fine tuning. The fine tuning knob is also used to set the displayed frequency to 20 Hz in the LOG ZERO sweep mode.

Frequency Display: Indicates start or center frequency in Hz. In the Manual Mode, the Frequency Display indicates the marker frequency.

Range: 00.0 kHz to approximately 50.8 kHz.

Resolution: 20 Hz (one minor division)

Typical Frequency Stability: \pm 10 Hz/hr. after 1 hour; \pm 5 Hz/° C

Bandwidth Settings: 1 Hz, 3 Hz, 10 Hz, 30 Hz, 100 Hz, 300 Hz

Bandpass Characteristic: closely approximates a gaussian response.

Shape Factor: 10:1 on 1 Hz thru 100 Hz bandwidths; 8:1 on 300 Hz bandwidth

Equivalent Noise Bandwidth: Typically 12% wider than absolute 3 dB bandwidth.

Display Smoothing (noise filtering):

3 Settings: min, med, max

Response: determined by Bandwidth setting.

SWEEP CHARACTERISTICS:

Sweep Modes:

Repetitive: The instrument sweeps continuously over the selected frequency range.

Single: The instrument sweeps one time over the selected frequency range and stops at the end frequency

Reset: Sweep is reset to left-hand side of screen; instrument remains at start frequency of sweep.

Manual: The electronic sweep is disabled and a front panel potentiometer is used to manually sweep the frequency and the refresh trace on the CRT. The manual sweep fully duplicates the span of the electronic sweep.

Log Zero: Used to set the correct starting point for log sweep.

Log: Front panel frequency and sweep controls are disabled. The instrument sweeps logarithmically from 20 Hz

Table 1-2. General Information (Cont'd).

to 43 kHz. The log sweep is repetitive; sweep time is approximately 5 seconds.

Typical Sweep Linearity: ± 1%

Frequency Span Settings: O Hz, 5 Hz/div to 5 kHz/div.

When the 0 Hz span setting is selected, the frequency sweep is disabled and the instrument remains at the frequency indicated on the frequency display. The display continues to sweep at the panel-selected rate. This provides a graphical display of amplitude vs. time.

Overall Span: 50 Hz to 50 kHz (10 span settings)

Sweep Time Settings: 0.01 sec/div. to 200 sec/div. (14 settings)

Overall Sweep Time: 0.1 sec to 2,000 sec

Sweep Error Light: A front panel LED indicator lights when sweep rate is too fast.

Out of Range Indication: The CRT display is cleared in areas where the sweep goes below 0 Hz or above 50 kHz.

Adaptive Sweep: The front panel Adaptive Sweep control is used to set a baseline threshold on the CRT. In areas where responses are below the baseline threshold, the instrument sweeps 20 to 25 times faster than the panel-selected rate. When the sweep reaches a response that rises above the baseline threshold, it backs up slightly, pauses to allow the IF Filter to settle and then sweeps slowly over the response at the panel-selected rate. By sweeping rapidly through unused portions of the spectrum, the Adaptive Sweep greatly reduces the measurement time for certain applications.

External Triggering: A rear panel External Trigger Input connector is provided to allow the frequency sweep to be remotely triggered by a contact closure or TTL logic levels. External triggering can be used in the Repetitive, Single or Log sweep mode.

OUTPUTS:

Recorder Outputs:

X-Axis: Supplies dc voltage corresponding to position of frequency sweep on CRT.

Output Voltage: 0 V (left-hand edge) to +5 V (right-hand edge)

Output Resistance: 1 kilohm

Y-Axis: Supplies dc voltage proportional to amplitude.

Output Voltage: 0 V (bottom of screen) to + 5 V (top of screen).

Output Resistance: 1 kilohm

Pen Lift: Provides a contact closure during single sweeps. If Adaptive Sweep is used, closure is present only when instrument is sweeping slowly over a response.

Tracking Oscillator Output:

Frequency: 5 Hz to 50 kHz; tracks turned or swept frequency of instrument.

Output Level: 0 V to > 1 V rms into 600 Ω (adjustable)

Output Impedance: 600 ohms

Tracking Oscillator Input: Tracking oscillator output signal can be offset or frequency modulated by applying an external reference signal (about 100 kHz) to the rear panel Tracking Oscillator Input connector.

L.O. Output:

Frequency: Varies from 1.0 MHz to 1.5 MHz as 3580A frequency is tuned from 0 Hz to 50 kHz.

Output Level: Varies from about 300 mV p-p to 600 mV p-p depending on frequency.

Output Impedance: 1 kilohm

GENERAL:

Operating Temperature Range:

Standard 3580A: 0°C to 55°C Option 001: 0°C to +40°C

Storage Temperature Range:

Standard 3580A: -40° C to $+75^{\circ}$ C Option 001: -40° C to $+50^{\circ}$ C

Charge Temperature Range (Option 001): O°C to +40°C

Power Requirements: 100 V, 120 V, 220 V or 240 V +5% -10%, 48 Hz to 440 Hz, 35 watts maximum

Battery Characteristics (Option 001):

Operating Time: 5 hours from full charge

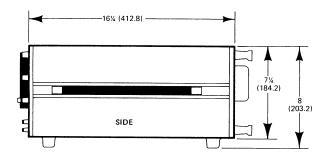
Charge Time: 14 hours to recharge fully discharged bat-

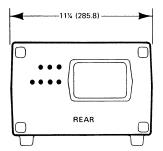
tery pack

Battery Life: more than 100 charge/discharge cycles Protection: The batteries are protected from excessive

discharge by an automatic cut out.

Dimensions:





DIMENSIONS SHOWN IN INCHES AND (MILLIMETERS)

Weight:

Standard 3580A: Net 27 lbs. Option 001: Net 35 lbs.

SECTION II INSTALLATION

2-1. INTRODUCTION.

2-2. This section contains information and instructions necessary for installing and shipping the Model 3580A Spectrum Analyzer. Included are initial inspection procedures, power and grounding requirements, environmental information, installation instructions and instructions for repackaging for shipment.

2-3. INITIAL INSPECTION.

2-4. This instrument was carefully inspected both mechanically and electrically before shipment. It should be free of mars or scratches and in perfect electrical order upon receipt. To confirm this, the instrument should be inspected for physical damaged in transit. If the instrument was damaged in transit, file a claim with the carrier. Check for supplied accessories (Paragraph 1-11) and test the electrical performance of the instrument using the performance test procedures outlined in Section V. If there is damage or deficiency, see the warranty in the front of this manual.

2-5 POWER REQUIREMENTS.

2-6. The Model 3580A can be operated from any power source supplying 100 V, 120 V, 220 V or 240 V (+5% -10%), 48 Hz to 440 Hz. Power dissipation is 35 watts, maximum. Refer to Paragraph 3-192 (Section III) for the Instrument Turn On Procedure.

2-7. Power Cords And Receptacles.

2-8. Figure 2-1 illustrates the standard power receptacle (wall outlet) configurations that are used throughout the United States and in other countries. The -hp- part number shown directly below each receptacle drawing is the part number for a 3580A power cord equipped with the appropriate mating plug for that receptacle. If the appropriate power cord is not included with the instrument, notify the nearest -hp- Sales and Service Office and a replacement cord will be provided.

2-9. GROUNDING REQUIREMENTS.

2-10. To protect operating personnel, the National

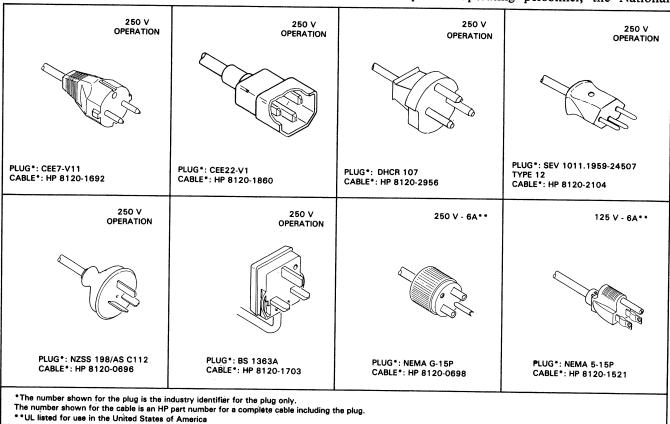


Figure 2-1. Power Receptacles.

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Electrical Manufacturer's Association (NEMA) recommends that the instrument panel and cabinet be grounded. The Model 3580A is equipped with a three conductor power cable which, when plugged into an appropriate receptacle, grounds the instrument. The offset pin on the power plug is the ground connection.

2-11. For battery powered instruments (Option 001), the common binding post of the INPUT connector (Case Ground \checkmark) should be connected to earth ground or to an appropriate system ground. If a system ground is used, extra care should be taken to ensure that it is actually at ground potential and is not a voltage source.

2-12. ENVIRONMENTAL REQUIREMENTS.

2-13. Operating and Storage Temperature (Standard 3580A).

Operating Temperature Range: 0°C to 55°C Storage Temperature Range: -40°C to +75°C

2-14. Operating and Storage Temperature (Option 001).

Operating Temperature Range: 0°C to +40°C Storage Temperature Range: -40°C to +50°C Charge Temperature Range: 0°C to +40°C

2-15. INSTALLATION.

2-16. The Model 3580A is a portable instrument and does not require installation. The instrument is shipped with rubber feet and tilt stand in place, ready for use as a bench instrument.

2-17. REPACKAGING FOR SHIPMENT.

2-18. The following paragraphs contain a general guide for repackaging the instrument for shipment. Refer to Paragraph 2-19 if the original container is to be used; 2-20 if it is not. If you have any questions, contact the nearest -hp- Sales and Service Office (see Appendix B for office locations).

NOTE

If the instrument is to be shipped to Hewlett-Packard for service, or repair, attach a tag to the instrument identifying the owner and indicating the service or repair to be accomplished. Include the model number and full serial number of the instrument. In any correspondence, identify the instrument by model number and full serial number.

- 2-19. Place instrument in original container with appropriate packing material and seal well with strong tape or metal bands. If original container is not available, one can be purchased from your nearest -hp-Sales and Service Office.
- 2-20. If original container is not to be used, proceed as follows:
- a. Wrap instrument in heavy paper, or plastic before placing in an inner container.
- b. Place packing material around all sides of instrument and protect panel face with cardboard strips.
- c. Place instrument and inner container in a heavy carton or wooden box and seal with strong tape or metal bands.
- d. Mark shipping container "DELICATE INSTRU-MENT," "FRAGILE," etc.
- 2-21. Option: Option 910 is an additional Operating and Service Manual -hp- Part Number 03580-90003.
- 2-22. Option 001 for internal rechargeable battery pack may be installed in the field by the following procedure:
 - a. Turn power OFF and disconnect the power cord.
 - b. Remove the bottom cover.
 - c. Insert the battery pack in the space provided.
- d. Attach 2 each of screws provided through each of the side panels to the battery pack.
- e. Connect battery plug to switch assembly A10 P1, located near front.
 - f. Replace the bottom cover.

SECTION III OPERATING INSTRUCTIONS

3-1. INTRODUCTION.

3-2. This section contains complete operating instructions for the Model 3580A Spectrum Analyzer. Included is a brief description of the instrument, a description of controls, general operating information and basic operating procedures.

3-3. ABOUT THE SPECTRUM ANALYZER.

- 3-4. The first spectrum analyzers were introduced during World War II for the use in the development of pulse radar systems. Early spectrum analyzers were difficult to operate and interpret since they lacked such refinements as calibrated controls. They were, however, adequate tools which enabled scientists to observe the spectra of radar pulses and subsequently optimize the gain and bandwidth of radar receivers. Since that time, spectrum analyzers have evolved into general purpose instruments with unlimited applications in the RF and audio frequency ranges.
- 3-5. The 3580A is a low frequency spectrum analyzer designed specifically for use in the audio frequency range. It can be used as a signal analyzer or as a network analyzer. When used as a signal analyzer, the 3580A measures the amplitudes and frequencies of the spectral components of an input signal. When used as a network analyzer, the 3580A plots the amplitude vs. frequency characteristics of 2-port networks such as amplifiers, attenuators and filters.

3-6. Operating Features.

3-7. The 3580A has many unique operating features that make it versatile, easy to use and ideally suited for low-frequency work. The three most significant features are its digitally stored display, Adaptive Sweep and 1 Hz bandwidth. Details of these and other features outlined in Table 3-1 are given in the General Operating Section (Paragraph 3-10).

3-8. CONTROLS, CONNECTORS AND INDICATORS.

3-9. Figures 3-1 and 3-2 illustrate and describe the function of all front and rear panel controls, connectors and indicators. The description of each item keyed to the drawing within the figure.

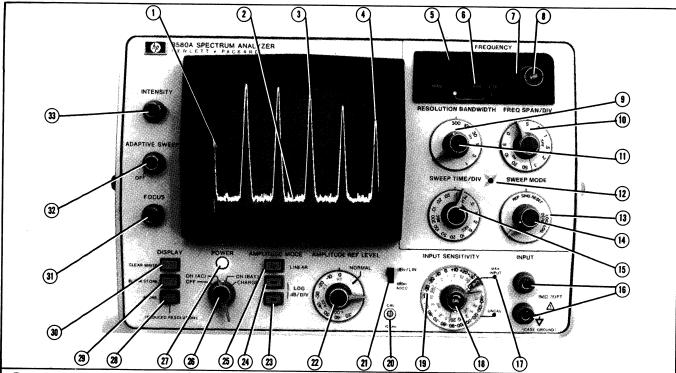
3-10. GENERAL OPERATING INFORMATION.

3-11. Input Cable Requirements.

3-12. The input signal can be applied to the 3580A through a twisted pair, a shielded cable equipped with banana-plug connectors (-hp- 11000A Cable Assy.) or a 10:1 Voltage Divider Probe (-hp- 10004B). Input leads should be kept as short as possible to minimize extraneous pickup. When using a 10:1 Voltage Divider Probe, the probe must be compensated as outlined in Paragraph 3-203.

Table 3-1. Operating Features

| Feature | Paragraph | Feature | Paragraph |
|--|--|---|--|
| High Input Impedance: 1 MΩ, 30 pF Frequency Range: 5 Hz to 50 kHz Six Selectable Bandwidths: 1 Hz – 300 Hz | 3-13 3-80 | Log 10 dB: scale 10 dB/div; absolute measurements in dBV or dBm/600 ohms; relative measurements in dB; 80 dB dynamic range. | 3-66 |
| Calibrated Frequency Display: 1. Selects start or center frequency of sweep 2. Coarse or fine tuning | 3-96 | 3. Log 1 dB: scale 1 dB/div; 10 dB display range Measurement Range: 1. Calibrated: 0.1 μ V rms (-140 dBV/dBm) full-scale to 20 V rms (+ 30 dBV/dBm) full-scale. | |
| Eleven Frequency Span Settings: 0 Hz, 50 Hz $-$ 50 kHz | 3-103 | 2. Uncalibrabed: $0.1~\mu V$ rms (-140 dBV/dBm) full-scale to 100 V rms (+40 dBV/dBm) full-scale. | |
| Sweep Modes: 1. Single or repetitive linear sweep | 3-13 | 80 dB Dynamic Range Digitally Stored Display | 3-49 3-158 |
| Manual Sweep Log sweep | | Internal Calibration Signal Recorder Outputs: | 3-77 |
| Fourteen Sweep Time Settings: 0.1 sec – 2,000 sec. Optimum Sweep Rate Indicator Frequency Out-Of-Range Indication On CRT Adaptive Sweep Three Amplitude Modes: 1. Linear: absolute measurements in rms volts; relative measurements in percent of full-scale. | 3-133 3-137 3-108 3-147 3-32 3-51 | 1. X-AXIS 2. Y-AXIS 3. PEN LIFT Tracking Oscillator Output Tracking Oscillator Input L.O. Output Portability, Battery Operation (Option 001) | 3-165 3-168 3-170 3-171 3-175 3-178 |
| | | Balanced Inputs, Balanced Tracking Oscillator Output (Option 002) | 3-182 3-187 |

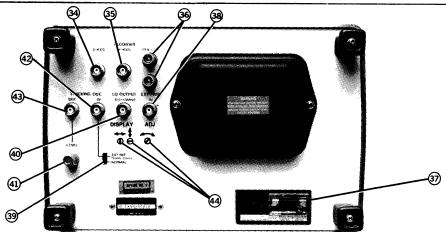


- LOG Marklings: In the LOG 10 dB mode, these markings indicate signal amplitude in dB below full scale.
- Frequency Markings: These markings indicate 20 Hz, 200 Hz, 2 kHz and 20 kHz decade frequencies of log sweep. (Paragraph 3-125)
- CRT Display: (Paragraph 3-158).
- (4) LIN Markings: In the LIN mode, these markings indicate signal amplitude in percent of full scale (1.0 = 100%, 0.4 = 40%, etc.).
- (5) FREQUENCY Display: Indicates start or center frequency of linear sweep. The Frequency Display also indicates the frequency at which the manual sweep is set. In the LOG mode, the Frequency Display blanks (Paragraph 3-99).
- 6 START/CTR Switch: When set to START position, FREQUENCY display indicates start frequency of linear sweep; when set to CTR position, FREQUENCY display indicates center frequency of linear sweeps. When the MANUAL mode is selected, the START/CTR switch is no longer applicable because the displayed frequency corresponds to the marker position instead of the start or center frequency. An amber "MAN" light is provided to indicate the instrument is in the MANUAL mode. The "START/CTR" light indicates the instrument is in either the REPETATIVE, SINGLE, or RESET mode. (Paragraph 3-100)
- COARSE FREQUENCY Control: Tunes frequency of instrument over 0 Hz to 50 kHz range. Is used to set start or center frequency of linear sweeps.
- FINE FREQUENCY Control: Used for fine tuning of the instruments frequency. Also used to set the start frequency to 20 Hz in the LOG ZERO sweep mode.
- BANDWIDTH Control: Controls 3 dB bandwidth of IF Filter. Is used to select the desired frequency resolution. The six BANDWIDTH settings are: 300 Hz, 100 Hz, 30 Hz, 10 Hz, 3 Hz and 1 Hz. (Paragraph 3-80)
- (10) FREQ SPAN Control: Determines width of spectrum to be observed. Span settings range from 5 Hz per division (50 Hz) to 5 kHz per division (50 kHz). (Paragraph 3-103)
- (1) DISPLAY SMOOTHING Switch: Provides three levels of noise filtering for video presentation.

- ADJUST Indicator: Lights to indicate that sweep rate is too fast. Will go out when SWEEP TIME is increased, BANDWIDTH is widened or when FREQUENCY SPAN is narrowed. (Paragraph 3-137)
- 3 SWEEP MODE Switch: Permits selection of six sweep modes: REP (Repetitive), SING (Single), RESET, MAN (Manual), LOG ZERO and LOG. (Paragraph 3-113)
- (A) MANUAL VERNIER: Tunes analyzer frequency and positions horizontal trace when SWEEP MODE switch is set to MAN position. (Paragraph 3-121)
- (15) SWEEP TIME Control: Sets duration of single and repetitive sweeps. Settings range from 0.01 second per division (0.1 sec.) to 200 seconds per division (2,000 sec.). (Paragraph 3-133)
- (16) INPUT Connector: Accepts male, banana-plug connector; input impedance is 1 megohm, 30 pF. (Paragraph 3-13)
- OVERLOAD Indicator: Lights to indicate that input signal exceeds maximum input level set by INPUT SENSITIVITY and amplitude VERNIER controls. (Paragraph 3-37)
- (8) Amplitude VERNIER: For absolute measurements VERNIER must be set to CAL (fully CW) position. For relative measurements, VERNIER adjusts gain of analyzer to establish a full-scale reference. As the VERNIER is rotated counterclockwise, the gain decreases and the full-scale input level increases. (Paragraph 3-36, 3-39).
- [9] INPUT SENSITIVITY Switch: Selects maximum (full scale) input level and measurement range. For absolute measurements, full-scale settings range from + 30 dBV/dBm to -70 dBV/dBM in Log 10 dB mode or from 20 V rms to 0.2 mV rms in the Linear mode. In the Linear mode, seven additional ranges (0.1 mV to 0.1 μV) can be selected by the AMPLITUDE REF LEVEL switch (Paragraph 3-39, 3-53 and 3-68). With the switch in the CAL position, the INPUT terminals are disconnected and an internally generated calibration signal is applied to the input circuits (Paragraph 3-77).
- (20) CAL 10 kHz Potentiometer: Adjusts gain of amplitude circuits to compensate for slight variations in amplitude accuracy caused by temperature changes or changes in bandwidth. (Paragraph 3-199)
- (1) dBV/LIN dBm Switch: Set to dBV/LIN position for measurements in dBV or rms volts; set to dBM 600 OHM position for measurements in dBm 600 ohms. For measurements in dBm/600 ohms, an external termination is required.

- AMPLITUDE REF LEVEL Switch: Operates in conjunction with IN-PUT SENSITIVITY switch to establish full-scale sensitivity and measurement range. In linear mode it controls the IF attenuation. When rotated in a clockwise direction, full-scale sensitivity increases in a 20 V, 10 V, 2 V, 1 V sequence (Paragraph 3-55). In the Log 10 dB mode, changing the Amplitude Ref Level setting offsets the entire display in 10 dB increments (Paragraph 3-69). In Log 1 dB mode, the Amplitude Ref Level control offsets the display to select any 10 dB portion of the 80 dB range (Paragraph 3-71).
- (23) LOG 1 dB Button: (push to set; push LIN or LOG 10 dB to release) Selects Log 1 dB amplitude mode. Display sensitivity is 1 dB per division; display range is 10 dB. Any 10 dB portion of the 80 dB range can be displayed by changing the AMPLITUDE REF LEVEL setting. (Paragraph 3-71)
- 24 LOG 10 dB Button: (push to set; push LIN or LOG 1 dB to release) Selects Log 10 dB amplitude mode for absolute measurements in dBV or dBm/600 ohms or relative measurements in dB. Display sensitivity is 10 dB per division; display range is 80 dB. (Paragraph 3-66)
- (25) LINEAR Button: (push to set; push LOG 1 dB or LOG 10 dB to release) Selects Linear amplitude mode for absolute measurements in rms volts or relative measurements in percent of full scale. (Paragraph 3-51)
- (26) POWER Switch: Applies line voltage to instrument when set to ON (AC) position; applies battery power to Option 001 instruments

- when set to ON (BAT) position; applies line voltage to Option 001 instruments to recharge batteries when set to CHARGE position. (Paragraph 3-192)
- POWER Light: Lights when POWER switch is set to ON (AC), ON (BAT) or CHARGE.
- STORE Button: (push to set; push to release) When initially pressed, trace currently being displayed is permanently stored in memory. When released, permanently stored trace is cleared from memory. (Paragraph 3-160)
- 29 BLANK STORE Button: (push to set; push to release) When pressed, permanently stored trace is blanked from the display. When released, stored trace returns to display. (Paragraph 3-160)
- (30) CLEAR WRITE Button: (momentary pushbutton) Clears display and resets sweep.
- 30 FOCUS Control: Focuses CRT trace. (Paragraph 3-158)
- (32) ADAPTIVE SWEEP Control: Turns Adaptive Sweep on or off; is used to set baseline threshold on CRT display. (Paragraph 3-147)
- (33) INTENSITY Control: Adjusts brightness of CRT trace. Intensity can be set to any level without danger of burning the CRT face. (Paragraph 3-158)



- X-AXIS Output: Female BNC connector supplies dc voltage corresponding to position of frequency sweep on CRT. Output voltage ranges from 0 V (left-hand edge) to + 5 V (right-hand edge). Output resistance is 1 kilohm, nominal. (Paragraph 3-165)
 - Y-AXIS Output: Female BNC connector supplies dc voltage proportional to amplitude. Output voltage ranges from 0 V (bottom of screen) to \pm 5 V (top of screen). Output resistance is 1 kilohm, nominal. (Paragraph 3-168)
- PEN LIFT Output: A closure is present across these terminals during single sweeps. If Adaptive Sweep is used, the closure is present only when the instrument is sweeping slowly over a response. (Paragraph 3-170)
- Power Input Module: Accepts power cord supplied with instrument. Contains line fuse and PC board for selecting line voltage. (Paragraph 3-193)
- 33 EXT TRG IN Connector: Female BNC connector accepts contact closure or TTL logic levels to remotely trigger the frequency sweep. (Paragraph 3-143)
- 39 EXT REF/NORMAL Switch: In the NORMAL position, the tracking oscillator receives its reference from an internal 100 kHz crystal oscillator. In the EXT REF position, the tracking oscillator reference is an external signal applied to the TRACKING OSC IN connector. With the switch in the EXT REF position, the tracking oscillator will be inoperative unless an external reference signal is applied. (Paragraph 3-176)

- 40 L.O. OUTPUT: Female BNC connector supplies a 100 mV rms signal whose frequency varies from 1 MHz to 1.5 MHz as the analyzer frequency is tuned from 0 Hz to 50 kHz. Output impedance is approximately 1 kilohm. (Paragraph 3-178)
- (1) LEVEL Control: Sets the amplitude of the Tracking Oscillator Output signal (0 V to 2 V rms)
- (2) TRACKING OSC IN: Female BNC connector. An external reference signal can be applied to this connector to offset or frequency-modulate the Tracking Oscillator Output signal. (Paragraph 3-175)
- (43) TRACKING OSC OUT: Female BNC connector supplies 0 Hz to 50 kHz signal that tracks the tuned or swept frequency of the instrument. Output level can be adjusted from 0 V to 2 V rms using the rear panel LEVEL control Output impedance is 600 ohms, nominal. (Paragraph 3-171)
- (44) DISPLAY ADJUSTMENTS: Positions the vertical, horizontal, and rotational axis of the display. These controls can be adjusted with an alignment tool to allow precise alignment of the display. $\Delta 21$

Figure 3-2. Rear Panel.

Revised: September 1987

3-13. Input Impedance.

- 3-14. The input impedance of the 3580A is 1 megohm shunted by 30 pF (28 pF nominal). This high input impedance has a minimum loading effect on the input signal and further permits the use of a 10 megohm, 10 pF Voltage Divider Probe (-hp- 10004B).
- 3-15. Figure 3-3 shows the equivalent circuit for the 3580A Input. The resistor, $R_{\rm in}$, represents the 1 megohm input resistance and the capacitor, $C_{\rm s}$, represents the 28 pF shunt capacitance. Figure 3-4 shows the input impedance, $Z_{\rm t}$, as a function of frequency. At low frequencies the reactance of $C_{\rm s}$ is very high, making $Z_{\rm t}$, nearly equal to $R_{\rm in}$. As frequency increases, the decreasing reactance of $C_{\rm s}$ becomes more and more significant, causing $Z_{\rm t}$ to decrease. At 50 kHz, $Z_{\rm t}$ is approximately 100 kilohms.

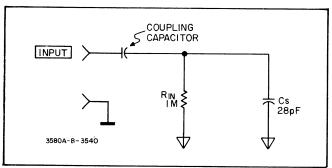


Figure 3-3. Equivalent Input Circuit.

3-16. Input Constraints.

3-17. The maximum ac voltage that can be safely applied to the 3580A INPUT is determined by the INPUT SENSITIVITY switch setting (Paragraph 3-39). Maximum input levels are listed in Table 3-2. The 3580A input circuits are well protected and can withstand momentary (<5 second) overloads up to 100 V rms on all input ranges. The instrument can withstand continuous overloads up to 100 V rms on the +30 dB through -10 dB ranges and overloads up to 50 V rms on

the -20 dB through -70 dB ranges. Overloads greater than this may damage the instrument.

ECAUTION

3580A STD Input Levels exceeding 100V rms on the +30 dB through -10 dB ranges, 50 V rms on the -20 dB through -70 dB ranges or ± 100 V dc may damage the instrument. See Paragraph 3-187 for Option 002.

- **3-18. DC** Isolation. The STD 3580A INPUT is capacitively coupled to provide dc isolation. The maximum dc voltage that can be safely applied to the INPUT is ± 100 V dc. Exceeding this limit can cause breakdown of the input capacitor resulting in damage to the input amplifier circuitry.
- 3-19. The 3580A cannot be operated in a floating condition. All input and output commons are connected directly to outer-chassis (frame) ground which connects to earth ground through the offset pin of the power cord connector or the common side of the INPUT connector. The 3580A Option 992, when operated in the unbalanced mode, has the same input restrictions as the 3580A standard. However, when the 3580A Option 002 is used in the bridged mode or the terminated mode, there is no input connection to chassis ground.

3-20. Grounding.

- 3-21. To protect operating personnel, the 3580A chassis must be grounded. The 3580A is equipped with a three conductor power cable which, when plugged into an appropriate receptacle, grounds the instrument. The offset pin on the power plug is the ground connection.
- 3-22. To preserve the protection feature when operating the instrument from a two contact outlet, use a three-prong to two-prong adapter and connect the lead on the adapter to earth ground.

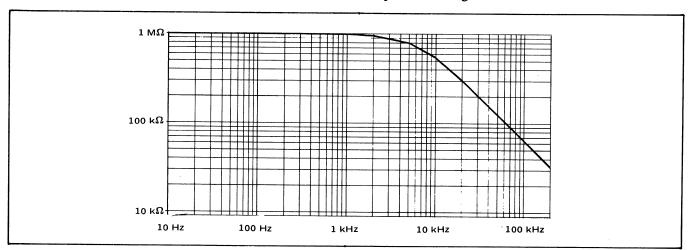


Figure 3-4. Graph Z_t vs. Frequency.

3-23. For battery powered instruments (Option 001), the common binding post of the INPUT connector (Case Ground \Diamond) should be connected to earth ground or to an appropriate system ground. If a system ground is used, extra care should be taken to ensure that it is actually at ground potential and is not a voltage source.

3-24. Ground Loops.

- 3-25. In the design of the 3580A, extra care has been taken to control internal ground currents that could produce undesirable responses or degrade the accuracy of low level measurements. Due to its wide dynamic range and high sensitivity, however, the 3580A can be affected by external ground currents or "ground loops" which are normally caused by poor grounding. The following paragraphs briefly describe the common power-line ground loop and outline the steps that can be taken to minimize ground loop problems.
- 3-26. Figure 3-5A shows the input arrangement for a simple grounded measurement. Ein represents the source being measured along with any noise associated with it and is generally called the "normal-mode source". R_s represents the source resistance and the resistance of the high lead; R_g represents the resistance of the ground lead. Current from E_{in} (normal-mode current) flows through R_s, Z_l and R_g and the instrument responds to the drop across Z_1 . As long as the grounds on both sides of R_g are identical, extraneous currents cannot circulate between the source ground and the instrument ground. If, however, the grounds are different due to voltage drops in the ground lead or currents induced into it, a new source is developed and the measurement appears as shown in Figure 3-5B. The new source E_{cm} (the difference between grounds), is called the "common-mode source" because it is common to both the high and ground lines. (Common-mode current

can flow through R_g or through R_s and Z_l . Since Z_l is usually much larger than R_s and since they are both in parallel with R_g , most of the voltage across R_g will appear across Z_l causing an error in the amplitude reading.

- 3-27. To minimize power-line ground loops, the following guidelines should be observed:
 - a. Keep input leads as short as possible.
 - b. Provide good ground connections to minimize R_g .
- c. Connect the signal source and the 3580A to the same power bus.
- d. If a removable ground strap is provided on the signal source, float the source to break the common-mode current path.
- e. Option 001: Battery operate the 3580A; connect a separate ground lead between the common terminal of the 3580A INPUT connector and the ground terminal of the signal source.

3-28. Measurement Configurations.

- 3-29. The 3580A can be used in either of two measurement configurations: open loop or closed loop. These configurations are illustrated in Figure 3-6.
- **3-30. Open Loop.** In the open-loop configuration, the 3580A functions as a *signal analyzer* which divides the input signal into its various frequency components. The amplitudes of these components are displayed as a function of frequency on the CRT. The amplitude vs. frequency display shows how energy is distributed as a function of frequency and, in effect, is the Fourier spectrum of the input signal. Some of the more common

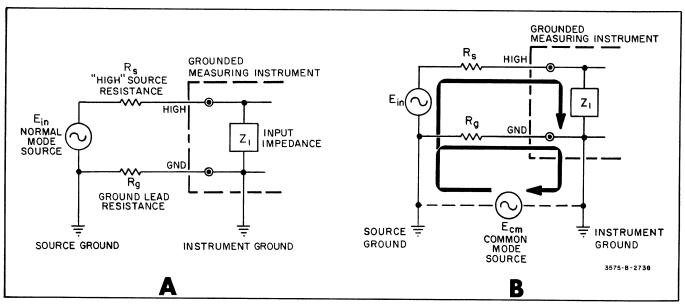


Figure 3-5. Power Line Ground Loop.

measurements that can be made using the open-loop configuration include harmonic distortion, intermodulation distortion, spurious, square-wave symmetry and noise.

3-31. Closed Loop. In the closed-loop configuration, the 3580A functions as a network analyzer for characterizing two-port devices such as amplifiers, attenuators and filters. For closed-loop measurements the network to be tested is inserted between the rear panel TRACK-ING OSC OUT and the front panel INPUT. The tracking oscillator supplies the stimulus to the network and the 3580A measures the response. As the frequency is swept over the band of interest, the instrument responds to the amplitude variations introduced by the network. The resulting display is an amplitude vs. frequency plot of the network.

3-32. Amplitude Modes.

3-33. The front panel AMPLITUDE MODE switch permits selection of three amplitude modes: Linear (LIN), Log 10 dB and Log 1 dB. When the Linear mode is selected and the amplitude VERNIER is in the CAL position, the vertical axis of the display is calibrated in rms volts (average responding). The bottom line of the display graticule represents 0 volts while the top line represents the full scale input voltage determined by the INPUT SENSITIVITY and AMPLITUDE REF LEVEL control settings (Paragraph 3-53). When either of the Log modes is selected, the verical axis of the display is calibrated in dBV (1 V rms = 0 dBV) or dBm/600 ohms, depending on the position of the dBV/LIN-dBm slide switch. In the Log 10 dB mode, the vertical scale is 10 dB per division and the maximum display range is greater than 80 dB (Paragraph 3-67). In the Log 1 dB mode, the vertical scale is expanded to 1 dB per division with a maximum display range of 10 dB. Any 10 dB portion of the 80 dB display range can be displayed by changing the AMPLITUDE REF LEVEL setting (Paragraph 3-71).

3-34. Absolute/Relative Measurements.

3-35. Absolute Measurements. Absolute measurements reveal the actual amplitude of responses appearing on the CRT display. The 3580A can be calibrated for absolute measurements in rms volts, dBV (1 V rms = 0 dBV) or dBm/600 ohms. For absolute measurements with the 3580A, the front panel amplitude VERNIER control must be set to the CAL (full clockwise) position and the instrument must be calibrated as outlined in Paragraph 3-199.

3-36. Relative Measurements. In signal analysis, relative measurements are used for comparing the amplitudes of two or more frequency components of a signal. In network analysis, relative measurements are used to compare the amplitude variations of a response curve at two or more frequencies. Relative measurements do not require a calibrated scale. That is, using the amplitude VERNIER and other amplitude controls, the gain of the analyzer can be adjusted so that any input level within the range of 100 V rms to 0.1 μ V rms will produce full scale deflection on the CRT display. This arbitrary full scale input level then serves as a reference for measuring signals that are lower in amplitude. In the Linear mode with the VERNIER not in the CAL position, the vertical scale on the CRT is no longer calibrated in volts per division. Thus, the unit of measure becomes "percent of full scale" where the reference is 100% and one vertical division is 10%. In the Log modes the vertical scale is always 10 dB per division or 1 dB per division even though the full scale reference is arbitrary. For relative measurements in the Log 10 dB mode, the top line of the display graticule (full scale) represents 0 dB and signals are measured in dB below the 0 dB reference level.

3-37. Overload Indicator.

3-38. Figure 3-7. is a simplified block diagram showing the 3580A Input Section. The INPUT SENSITIVITY

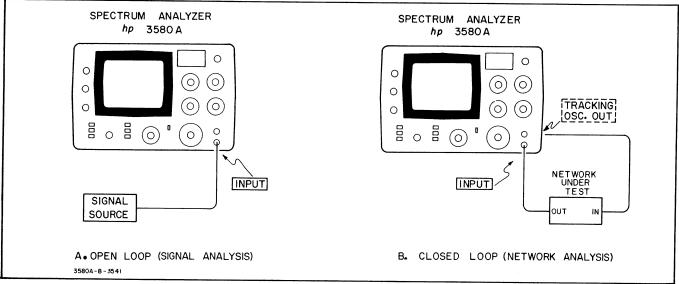


Figure 3-6. Measurement Configurations.

switch and its associated VERNIER potentiometer control the input attenuation and gain of the Input Circuits to maintain the proper signal level at the input of the Mixer. This is an important function since signals that overdrive the Mixer can produce harmonic and spurious mixing products which ultimately appear on the display. The Overload Detector at the input of the Mixer senses when the signal level exceeds the design limits and, in turn, lights the front panel OVERLOAD indicator. As indicated in Paragraph 3-17, the 3580A Input Circuits are well protected and continuous overloads up to 100 V rms on the + 30 dB through -10 dB ranges or up to 50 V rms on the -20 dB through -70 dB ranges will not damage the instrument. In most cases, an OVERLOAD indication simply means that the input signal is overdriving the Mixer and unwanted responses may appear on the display. Generally, any time the OVERLOAD light is off, instrument-inducted distortion and spurious is more than 80 dB below the input reference level.

3-39. Maximum Input Level.

3-40. The maximum input level is the maximum level that can be applied to the INPUT without overloading the instrument. The maximum input level is determined only by the INPUT SENSITIVITY and amplitude VER-NIER settings and is *not* affected by the AMPLITUDE REF LEVEL setting. With the amplitude VERNIER control in the CAL (fully CW) position, the maximum input level is indicated by a black panel index adjoining the INPUT SENSITIVITY switch dial and the OVER-LOAD indicator (Figure 3-8). In both Linear and Log modes, the maximum input level is determined by the black (dB) markings on the INPUT SENSITIVITY switch dial. These markings represent either dBV or dBm/600 ohms, depending on the position of the dBv/LIN-dBm slide switch. When the amplitude VER-NIER control is rotated counterclockwise away from the CAL position, the gain of the input circuit decreases, the maximum input level increases and the markings on the INPUT SENSITIVITY switch dial no longer apply. Table 3-2 lists the maximum input levels for each INPUT SENSITIVITY setting with the

amplitude VERNIER in the CAL and fully counterclockwise positions. The maximum levels listed in the table are, in some cases, considerably lower than the absolute maximum levels that will produce an OVER-LOAD indication. Observing these maximum levels will ensure optimum performance on all ranges.

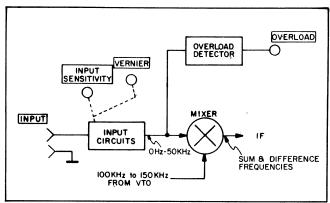


Figure 3-7. Input Section.

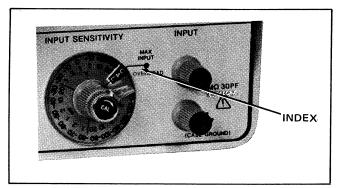


Figure 3-8. Maximum Input Index.

3-41. Sensitivity.

3-42. Sensitivity is a figure of merit that defines the analyzer's ability to detect or respond to a given input level. There are three types of sensitivity that are of interest when operating the 3580A:

| Input Sensitivity Setting | (VERN Linear Mode | IER in CAL Log Mode | (VERNIE Linear Mode | R fully CCW) Log Mode | Potential Damage Level (Continuous Overload |
|------------------------------|----------------------|------------------------|------------------------|----------------------------|---|
| + 30 dB/20 V | 31.6 V | + 30 dBV/dBm | 100 V* | + 40 dBV/dBm | 100 V* |
| + 20 dB/10 V | 10 V | + 20 dBV/dBm | 20 V | + 30 dBV/dBm | ''' |
| + 10 dB/2 V | 3.16 V | + 10 dBV/dBm | 10 V | + 20 dBV/dBm | |
| 0 dB/1 V | 1 V | 0 dBV/dBm | 2 V | + 10 dBV/dBm | |
| -10 dB/0.2 V | 0.32 V | -10 dBV/dBm | 1 V | 0 dBV/dBm | 100 V* |
| -20 dB/0.1 V | 0.1 V | -20 dBV/dBm | 0.2 V | -10 dBV/dBm | 50 V |
| -30 dB/20 mV | 32 mV | -30 dBV/dBm | 0.1 V | -20 dBV/dBm | |
| -40 dB/10 mV | 10 mV | -40 dBV/dBm | 20 mV | -30 dBV/dBm | |
| -50 dB/2 mV | 3.2 mV | -50 dBV/dBm | 10 mV | -40 dBV/dBm | |
| -60 dB/1 mV | 1 mV | -60 dBV/dBm | 2 mV | -50 dBV/dBm | |
| -70 dB/0.2 mV | 0.32 mV | -70 dBV/dBm | 1 mV | -60 dBV/dBm | 50 V |

Table 3-2. Maximum Input Levels.

^{*}Absolute maximum input voltage.

- a. Maximum Sensitivity.
- b. Full Scale Sensitivity.
- c. Display Sensitivity.
- 3-43. Maximum Sensitivity. Maximum Sensitivity refers to the smallest signal that can be detected by the analyzer. The maximum sensitivity of the analyzer is limited by its own internally generated noise and is commonly defined as the point where the signal level is equal to the noise level. This is sometimes called "tangential sensitivity".
- 3-44. Nyquist's Noise Equation¹ reveals two important things about noise that apply to the 3580A:
- a. Noise is proportional to the square root of bandwidth... Noise level decreases and sensitivity increases as the BANDWIDTH setting is narrowed.
- b. Noise is proportional to the square root of input resistance. The 3580A has a high (1 Megohm) input resistance. This means that noise is largely dependent on the source resistance placed at the INPUT terminals. Signal sources having low output resistances will produce a lower noise level than those having high output resistances.
- 3-45. Noise level is also dependent on the tuned frequency of the instrument. Semiconductors in the input stages of the instrument exhibit surface noise which has a 1/f frequency spectrum. This surface noise is predominate at frequencies below 1 kHz. When the 3580A is tuned below 1 kHz, the noise level increases and sensitivity decreases.
- 3-46. Figure 3-9 is a family of curves showing the specified noise levels vs. frequency for the 300 Hz, 30 Hz and 1 Hz BANDWIDTH settings. Typically, if the source resistance is less than 10 kilohms, the noise levels will be below those indicated by the curves.

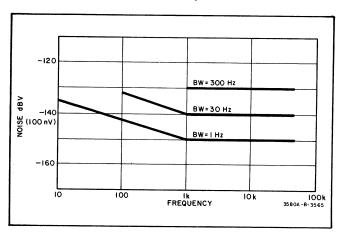


Figure 3.9. Noise vs. Frequency.

- 3-47. Full Scale Sensitivity. Full scale sensitivity defines the input level that will produce full scale deflection on any given range. For absolute measurements, full scale sensitivity ranges from 20 V rms to 0.1 μ V rms in the Linear mode and from +30 dBV/dBm to -140 dBV/dBm in the Log (10 dB) mode. With the amplitude VERNIER control set fully counterclockwise, full scale sensitivity ranges from approximately 100 V rms to 0.2 μ V rms in the Linear mode and from +40 dBV/dBm to -130 dBV/dBm in the Log mode.
- 3-48. Display Sensitivity. Display Sensitivity or "scale calibration" expresses the analyzer's response in units per vertical division. For absolute measurements in the Linear mode, display sensitivity ranges from 2 V per division to 10 nV per division. For absolute or relative logarithmic measurements, display sensitivity is 10 dB per division in the Log 10 dB mode and 1 dB per division in the Log 1 dB mode.

3-49. Dynamic Range.

3-50. The dynamic range of a spectrum analyzer defines its ability to detect large and small signals and display them simultaneously. For operating purposes, dynamic range can be expressed as the ratio of the largest to smallest signals that can be simultaneously displayed on the CRT. In both the Linear and Log modes, the largest signal that can be displayed (full scale sensitivity) is determined by the INPUT SENSITIVITY, amplitude VERNIER and AMPLITUDE REF LEVEL control settings. The smallest signal that can be displayed is determined by the display range or by the internal noise floor (Maximum sensitivity). In the Linear mode the smallest signal that can be displayed is approximately 1% of full scale. Thus, the dynamic range is approximately 40 dB as long as the internal noise floor is more than 40 dB below full scale. With the AMPLITUDE REF LEVEL switch in the NORMAL position, the display range is the Log 10 dB mode is greater than 80 dB. The dynamic range is, therefore, at least 80 dB as long as the noise floor is more than 80 dB below full scale. In the Log 1 dB mode, the display sensitivity is increased to 1 dB per division and the dynamic range, determined by the display range, is 10 dB.

3-51. Amplitude Measurements (Linear Mode).

3-52. Figure 3-10 is a simplified block diagram showing a portion of the 3580A amplitude section in the Linear mode. The INPUT SENSITIVITY switch and amplitude VERNIER potentiometer control the input attenuation and gain of the Input Circuits and establish the maximum input level as outlined in Paragraph 3-40. In addition, the INPUT SENSITIVITY switch operates in conjunction with the AMPLITUDE REF LEVEL switch to establish the full-scale sensitivity and measurement range.

 $^{^{1}}E_{n} = (4 \text{ kTBR})^{\frac{1}{2}}$ Where $E_{n} = \text{noise level}$; k = Boltzmann's constant; T = temperature (°K); B = bandwidth (Hz); R = input resistance.

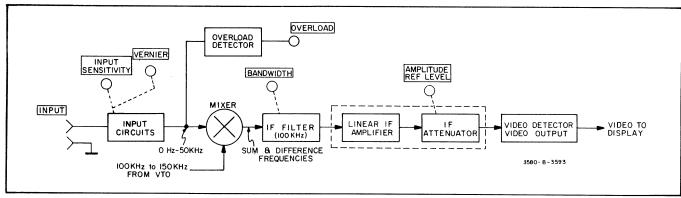


Figure 3-10. Amplitude Section (Linear Mode).

- 3-53. The INPUT SENSITIVITY switch has 12 positions: a CAL position and 11 voltage range settings. With the amplitude VERNIER in the CAL position and the AMPLITUDE REF LEVEL switch in the NORMAL (X1) position, the full-scale sensitivity, as determined by INPUT SENSITIVITY switch setting, ranges from 20 V rms to 0.2 mV rms.
- 3-54. For any given INPUT SENSITIVITY setting, the dynamic range of the Input Circuits, Mixer and IF Filter is at least 80 dB as long as the noise floor is more than 80 dB below full scale. Thus, with the INPUT SENSITIVITY switch in the 0.2 mV position, an input signal as low as 0.1 μ V rms could be detected at the output of the IF Filter. In the Linear mode, however, the dynamic range of the display is limited to approximately 40 dB. This means that on the 0.2 mV range the smallest signal that can be displayed is approximately 2μ V or 1% of full scale. Moreover, the 2 μ V signal might be visible on the display but it would be too small to be measured accurately. For all practical purposes, then, the dynamic display range is limited to approximately 20 dB.
- 3-55. To utilize the full measurement range of the instrument in the Linear mode, it is necessary to increase the display sensitivity. To accomplish this a variable IF Attenuator, controlled by the AMPLITUDE REF LEVEL switch, is inserted between the Linear IF Amplitude and Video Detector. With the AMPLITUDE REF LEVEL switch set to the NORMAL (X1) position, the IF attenuation is *maximum*. As the AMPLITUDE REF LEVEL switch is rotated in a clockwise direction, the IF attenuation decreases, the effective IF gain increases and the display sensitivity increases. The IF Attenuator provides seven additional ranges which allow the full-scale sensitivity to be varied from 0.1 mV rms to 0.1 μ V rms.
- 3-56. By observing the INPUT SENSITIVITY and AMPLITUDE REF LEVEL controls, it can be noted that the full-scale (blue) markings on the INPUT SENSITIVITY switch dial are indicated by a white window that is mechanically linked to the AMPLITUDE REF LEVEL switch. Changing the position of either switch changes the full-scale sensitivity in a 20 V, 10 V 2 V, 1 V

- sequence. Changing the AMPLITUDE REF LEVEL setting, however, does not change the maximum input level. For example, with the INPUT SENSITIVITY switch set for a maximum input of 1 V rms and the AMPLITUDE REF LEVEL switch set to the X0.1 position, the full-scale sensitivity is 0.1 V rms, the display sensitivity is 10 mV per division but the maximum input level is still 1 V rms. Input signals greater than 0.1 V rms but less than or equal to 1 V rms will not overdrive the mixer or produce an OVERLOAD indication. They will, however, peak the display when the analyzer is tuned to their specific frequency. This does not damage the instrument nor hinder its ability to measure signals within the display range.
- 3-57. Using the AMPLITUDE REF LEVEL Control. Whenever possible, the AMPLITUDE REF LEVEL switch should be left in the NORMAL (X1) position and the INPUT SENSITIVITY switch should be used to set the full-scale sensitivity. This is because the Amplitude Calibration Procedure (Paragraph 3-199) is performed with the AMPLITUDE REF LEVEL switch in the NORMAL position and any error introduced by the IF Attenuator is adjusted out. When the AMPLITUDE REF LEVEL setting is changed from the NORMAL position, the accuracy of the IF Attenuator must be considered. This means that a possible worst-case error ± 3% of full scale must be added to the amplitude accuracy specification. Amplitude accuracy is discussed in Paragraph 3-72.
- 3-58. There are commonly two occasions when it is necessary to change the AMPLITUDE REF LEVEL setting:
- a. When the required full-scale is within the range of 0.1 mV rms to 0.1 μ V rms and the amplitude of the input signal is less than or equal to 0.1 mV rms. In this case, the INPUT SENSITIVITY switch is set to the 0.2 mV range (fully clockwise) and the appropriate range is selected using the AMPLITUDE REF LEVEL switch.
- b. For expanded-scale measurements where the amplitude of the input signal is 0.2 mV rms or greater and the signal or signals of interest are less than 10% of

full scale with the INPUT SENSITIVITY switch set to the lowest range that does not produce an OVERLOAD indication. In this case, the AMPLITUDE REF LEVEL switch is initially set to the X1 position and the INPUT SENSITIVITY switch is set to the lowest range that does not produce an OVERLOAD indication. The AMPLITUDE REF LEVEL switch is then set so that the low-level signals of interest can be measured. Signals greater than the full-scale level indicated by the white window on the INPUT SENSITIVITY switch dial will peak the display but will not damage the instrument nor introduce harmonic or spurious responses.

- **3-59.** Scale Factor. The blue markings on the AMPLITUDE REF LEVEL switch dial indicate the scale factor which, for *absolute* measurements is the factor by which the INPUT SENSITIVITY (Max. Input) setting must be multiplied to determine the full-scale sensitivity. For example, if the INPUT SENSITIVITY switch is set to the 2 V range and the AMPLITUDE REF LEVEL switch is set to the X0.01 position, the full-scale sensitivity is: 2 V X0.01 = 0.02 V or 20 mV.
- 3-60. For absolute measurements the full-scale sensitivity is conveniently indicated by the white window on the INPUT SENSITIVITY switch dial and the scale factor can generally be ignored. If, for some reason, the scale factor is to be used, note that the even numbered positions on the AMPLITUDE REF LEVEL dial are not marked. This is because the scale factor in these positions depends on the INPUT SENSITIVITY switch setting. If the INPUT SENSITIVITY switch is set to the 20 V, 2 V, 0.2 V, etc. position, the unmarked positions on the AMPLITUDE REF LEVEL switch dial represent X0.5, X0.05, X0.005 and X0.0005. If the INPUT SEN-SITIVITY switch is set to 10 V, 1 V, 0.1 V, etc., the unmarked positions represent X0.2, X0.02, X0.002 and X0.0002. This applies only when the amplitude VER-NIER is in the CAL position.
- 3-61. For relative measurements where the amplitude VERNIER is not in the CAL position, the full-scale markings on the INPUT SENSITIVITY switch dial do not apply and, for expanded-scale measurements, a scale factor must be used. In relative measurements the scale factor is the factor by which a relative amplitude reading must be multiplied to obtain the correct reading in percent of full scale.
- 3-62. When making relative measurements it is important to remember that any time the VERNIER is not in the CAL position, the relationship between the marked and unmarked positions of the AMPLITUDE REF LEVEL Switch varies as a function of both the INPUT SENSITIVITY and amplitude VERNIER settings. There is always a X1, X0.1, X0.01, X0.001 relationship between the marked positions and this same relationship exists between the unmarked positions. However, there is no longer a X1, X0.5, X0.1 or X1, X0.2, X0.1 relationship between the marked and unmarked posi-

- tions. To obtain the correct scale factor the following guidelines must be observed:
- a. If the full-scale reference is set with the AMPLITUDE REF LEVEL switch in a marked position, all measurements must be made using marked positions.
- b. If the full-scale reference is set with the AMPLITUDE REF LEVEL switch in an unmarked position, all measurements must be using unmarked postions.
- c. The AMPLITUDE REF LEVEL setting on which the full-scale reference level is established becomes the X1 setting. If the X1 setting is a marked position, the scale factors for the remaining marking positions become X0.1, X0.01, etc. Similarly, if the X1 setting is an unmarked position the remaining unmarked positions become X0.1, X0.01, etc.
- 3-63. Examples. Consider the case where the fundamental frequency component of an input signal is 0.75 V and it is necessary to measure the second harmonic component whose relative amplitude is 1%. With the AMPLITUDE REF LEVEL control initially set to the NORMAL (X1) position and the amplitude VER-NIER fully counterclockwise, the INPUT SENSITIVI-TY switch can be set to the 0.2 V position without overloading the instrument. The amplitude VERNIER can then be adjusted so that the amplitude of the fundamental frequency component is 100% of full scale. The % second harmonic will perhaps be visible on the display but an expanded scale will be required to measure it accurately. In this case, the full-scale reference was established with the AMPLITUDE REF LEVEL switch in the X1 position. Thus, the unmarked positions cannot be used and the scale factors of the marked positions are as indicated on the switch dial. By setting the AMPLITUDE REF LEVEL control to the X0.01 position, the 1% second harmonic can be expanded to 100% of full scale. It will be necessary to multiply the 100% reading by the X0.01 scale factor to obtain the correct reading: $100 \times 0.01 = 1\%$.
- 3-64. Next, consider the case where the amplitude of the fundamental frequency component is 1.8 mV and it is necessary to measure a harmonic component whose relative amplitude is 4%. With the AMPLITUDE REF LEVEL switch in the NORMAL (X1) position and the amplitude VERNIER fully counterclockwise, the IN-PUT SENSITIVITY switch can be set to the 0.2 mV (lowest) range. With a fundamental frequency component of less than 0.2 mV, a full-scale reference cannot be obtained on the 0.2 mV range. It is, therefore, necessary to go to the 0.1 mV range using the AMPLITUDE REF LEVEL switch. In this case, the full-scale reference will be established with the AMPLITUDE REF LEVEL switch in an unmarked position. This unmarked position becomes the X1 position. To expand the harmonic to a measureable level, it will be necessary to rotate the

AMPLITUDE REF LEVEL control clockwise to the next unmarked position. This unmarked position has a scale factor of X0.1 and will expand the 4% harmonic to 40% of full scale. The correct reading can then be obtained by multiplying the 40% reading by the X0.1 scale factor: $40 \times 0.1 = 4\%$.

3-65. Alternative Method. An alternative method for determining the relative amplitude of two signals is to first measure the absolute voltage levels and then calculate their relative amplitude using the following formula:

$$A = \frac{V2}{V1} X100$$

Where: A = relative amplitude in percent

V1 = reference level in rms volts

V2 = signal level in rms volts

3-66. Amplitude Measurements (Log Mode).

3-67. Figure 3-11 is a simplified block diagram showing a portion of the 3580A amplitude section in the Log mode. By comparing Figures 3-10 and 3-11, it can be noted that in the Log mode, the IF Amplifier/Attenuator is replaced by a Log Amplifier. The Log Amplifier provides an 80 dB display range.

3-68. With a dynamic display range of 80 dB, only eleven full-scale ranges are needed to utilize the full measurement range of the instrument. These eleven ranges are selected by the INPUT SENSITIVITY switch. With the amplitude VERNIER in the CAL position and the AMPLITUDE REF LEVEL control in the NORMAL (0 dB) position, the full-scale sensitivity, as determined by the INPUT SENSITIVITY switch setting, ranges from +30 dBV/dBm to -70 dB/dBm.

3-69. As in the Linear mode, the maximum input level is determined by the INPUT SENSITIVITY and amplitude VERNIER settings. Likewise, the full-scale sensitivity is indicated on the INPUT SENSITIVITY switch dial by the white window that is linked to the AMPLITUDE REF LEVEL switch. In the Log mode, however, the AMPLITUDE REF LEVEL switch con-

trols the dc operating point of the Video Output circuits and cannot be used to extend the measurement range. In th Log 10 dB mode, rotating the AMPLITUDE REF LEVEL switch in a clockwise direction offsets the entire display in 10 dB increments. Each time the display is offset the value of the top line of the display graticule (full scale) becomes 10 dB lower as indicated by the white window. At the same time, however, the dynamic range of the display decreases by 10 dB. With the AMPLITUDE REF LEVEL switch set to the -70 dB position, the full-scale sensitivity is 70 dB below its original value but the dynamic display range is only about 10 dB.

3-70. The ability to offset the display in the Log 10 dB mode is useful for some measurement applications. In most cases, however, all measurements can be made with the AMPLITUDE REF LEVEL switch set to the NORMAL position. Any time the AMPLITUDE REF LEVEL setting is changed from the NORMAL position, the dynamic display range decreases and a possible worst-case error ± 1 dB must be added to the overall amplitude accuracy specification.

3-71. Expanded-Scale Measurements. When the Log 1 dB mode is selected, the display sensitivity is increased to 1 dB per division and, with 10 vertical divisions, the maximum display range is 10 dB. The display in the Log 1 dB mode corresponds to the top 10 dB of the display in the Log 10 dB mode. Thus, by offsetting the display using the AMPLITUDE REF LEVEL control, any 10 dB portion of the 80 dB range can be displayed. In the Log 1 dB mode, the black (dB) markings on the AMPLITUDE REF LEVEL switch dial indicate the value of the top line of the display graticule with respect to the 0 dB (full scale) reference. For example, with the switch in the -10 dB position the top line of display graticule represents -10 dB and the display ranges from -10 dB to -20 dB. Similarly, with the switch in the -60 dB position the top line of the display graticule represents -60 dB and the display ranges from -60 dB to -70 dB.

3-72. Amplitude Accuracy.

3-73. The Amplitude Accuracy Specification listed in Table 1-1 is as follows:

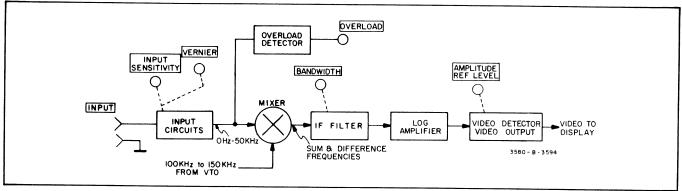


Figure 3-11. Amplitude Section (Log Mode).

| Amplitude: | Log | Linear | |
|--|-------------|--------|--|
| Frequency Response: | | | |
| 20 Hz-20 kHz | ± .3 dB | ±3% | |
| 5 Hz-50 kHz | ± .5 dB | ± 5% | |
| Switching between bandwid | ths (25°C): | | |
| 3 Hz-300 Hz | ± .5 dB | ± 5% | |
| 1 Hz-300 Hz | ± 1 dB | ± 10% | |
| Amplitude Display: | ± 2 dB | ± 2% | |
| Input Attenuator: | ± .3 dB | ± 3% | |
| Amplitude Reference Level: (IF Attenuator) | | | |
| most sensitive range | ± 1 dB | ± 10% | |
| all other ranges | ± 1 dB | ± 3% | |

3-74. The Amplitude Accuracy Specification is broken down so that portions of the specification that do not apply to a particular measurement can be eliminated. All applicable portions of the specification must be added together to obtain the overall accuracy specification. It should be noted that the overall accuracy specification reflects the absolute worst-case error that could possibly be encountered. Typically, all parameters are well within their specified tolerances and the probability of having a worst-case condition is very slight. As more parameters are added to the specification, the magnitude of the possible worst-case error increases but the probability of having a worst-case condition greatly decreases.

3-75. The Frequency Response, Amplitude Display and Input Attenuator Specifications must always be taken into account when calculating the overall accuracy specification. Excluding the Switching Between Bandwidths and Amplitude Ref Level specifications, the worst case error is ± 2.8 dB in the Log mode or $\pm 10\%$ of reading in the Linear mode.

3-76. The Switching Between Bandwidths specification can be disregarded as long as the Amplitude Calibration Procedure is performed on the BANDWIDTH setting that is used for measurements. If the BANDWIDTH setting is changed, the Switching Between Bandwidths specification must be added to the overall accuracy specification. Similarly, the Amplitude Ref Level specification can be disregarded as long as the AMPLITUDE REF LEVEL control is in the NORMAL position. If the AMPLITUDE REF LEVEL setting is changed, the Amplitude Ref Level specification must also be added to the overall accuracy specification.

3-77. Internal Cal Signal.

3-78. With the INPUT SENSITIVITY switch set to the CAL position, the high INPUT terminal on the front panel is disconnected and an internally generated calibration signal is applied to the Input Amplifier. The calibration signal is a highly accurate 15/85 duty cycle

pulse train which provides a 10 kHz fundamental frequency component along with odd and even harmonic components spaced at 10 kHz intervals (Figure 3-12). The magnitude of the pulse is such that the fundamental frequency component produces full scale deflection when the instrument is properly calibrated. The amplitudes of the harmonic components are not meaningful. The calibration signal can be used for amplitude calibration or to verify the frequency accuracy of the instrument.

3-79. In the Amplitude Calibration Procedure (Paragraph 3-199), the front panel 10 kHz CAL potentiometer is adjusted so that the 10 kHz fundamental frequency component of the cal signal produces full scale deflection. This calibrates all circuitry following the input attenuator to a full scale accuracy of $\pm 1.5\%$ at 10 kHz.

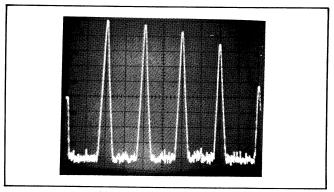


Figure 3-12. Cal Signal.

3-80. Bandwidth Setting.

3-81. Refer to Figure 3-13 for the following discussion. The 3580A uses a hetrodyne technique where the 0 Hz to 50 kHz input signal is mixed with a 100 kHz to 150 kHz signal from a Voltage-Tuned Local Oscillator (VTO). To select a given frequency present at the input of the Mixer, the VTO frequency is tuned so that the difference between it and the frequency of interest is 100 kHz intermediate frequency (IF) is fed through the IF Filter, detected and applied to the vertical axis of the CRT display. Signals outside the pass band of the IF Filter are rejected. The BANDWIDTH setting determines the bandwidth of the IF Filter and thus, the selectivity of the instrument.

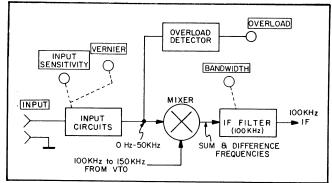


Figure 3-13. Frequency Tuning.

3-82. For operating purposes, the 3580A input channel can be pictured as a bandpass filter than can be manually tuned or swept over the 0 Hz to 50 khz frequency range. The instrument responds only to signals passing through the filter and thereby sorts out the various frequency components present at the input. The BAND-WIDTH setting determines the width of the filter skirts at the -3 dB points above and below the tuned frequency:

Lower 3 dB Point =
$$f_0 - \frac{BW}{2}$$

Upper 3 dB Point = $f_0 + \frac{BW}{2}$

Where:

f₀ = Tuned Frequency (0 Hz to 50 kHz) BW = BANDWIDTH Setting (1 Hz—300 Hz)

3-83. IF Bandpass Characteristic. Many signal analyzers use active filters that have very steep skirts and a square-shaped bandpass characteristic that approaches the ideal "window filter". This type of filtering provides a high degree of selectivity, but because of its long transient response time, is not well suited for swept frequency applications. The 3580A IF Filter consists of 5 synchronously-tuned crystal filter stages. The bandpass characteristic of the synchronously-tuned filter (Figure 3-14) closely approximates a gaussian response. The gaussian filter provides good selectivity and, because of its relatively short transient response time, is considered optimum for sweeping.

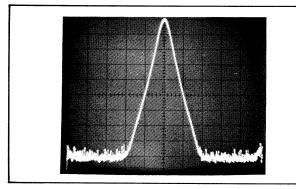


Figure 3-14. IF Filter Response.

3-84. Shape Factor. The shape factor of the 3580A IF Filter is approximately 10:1 on the 1 Hz through 100 Hz bandwidths and 8:1 on the 300 Hz bandwidth. A shape factor of 10:1 means that the filter skirts are 10 times wider at the -60 dB points than at the -3 dB points. Similarly, a shape factor of 8:1 means that the skirts are 8 times wider at the -60 dB points than at the -3 dB points. On the 10 Hz bandwidth, for example, the -3 dB points are 10 Hz apart and the -60 dB points are 10 x 10 or 100 Hz apart. The filter is, in effect, centered on the tuned frequency, f_0 , and exhibits 3 dB of rejection to signals that are \pm 5 Hz away from f_0 and 60 dB of rejection to signals that are \pm 5 Hz away from f_0 .

3-85. Equivalent Noise Bandwidth. When making noise measurements with the 3580A, it is necessary to use the "equivalent noise bandwidth" rather than the 3 dB bandwidth indicated by the BANDWIDTH setting. In the 3580A, the equivalent noise bandwidth is 12% wider than the absolute 3 dB bandwidth. Note that the specified bandwidth tolerance is $\pm 15\%$. This means that the absolute 3 dB bandwidth can be 15% wider or narrower than the BANDWIDTH setting. For optimum accuracy, measure the absolute 3 dB bandwidth of your instrument and use that figure to calculate the equivalent noise bandwidth.

3-86. Bandwidth Selection. There are 4 things to consider when selecting a BANDWIDTH setting:

- 1) Resolution
- 2) Low Frequency Limit
- 3) Response Time
- 4) Noise Rejection

3-87. Resolution. Resolution is the ability of the analyzer to separate signals that are closely spaced in frequency. An important point here is that the response of the analyzer to a CW signal is an amplitude vs. frequency plot of the IF Filter (Figure 3-15). The width and shape of the filter skirts are, therefore, the major limitations of resolution. If two CW signals appear in the passband (±3 dB points) simultaneously, they cannot be separated (Figure 3-16). If two signals differing widely in amplitude are both inside the filter skirts, the response of the larger signal can hide or obscure that of the smaller signal (Figure 3-17). If the amplitude of the smaller signal is greater than that of the skirt produced by the larger signal the peak of the smaller signal can be resolved (Figure 3-18). For optimum resolution, the bandwidth should be narrowed to the point where only one signal is inside the filter skirts at any given time. Generally, the width of the filter skirts at the -80 dB bandwidth. Thus, optimum resolution can always be obtained when the frequency separation between signals is at least 15 times the BANDWIDTH setting.

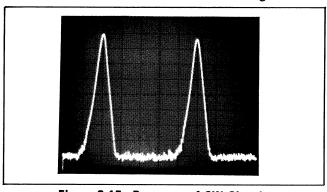


Figure 3-15. Response of CW Signals.

3-88. Table 3-3 lists the approximate maximum resolution for two signals whose relative amplitude is within the range of 0 dB to 70 dB. For example, on the 100 Hz Bandwidth, it is possible to resolve two signals that are

equal in amplitude and 2 X BW or 200 Hz apart. Similarly, it is possible to resolve two signals that differ in amplitude by 40 dB and are 5 X BW or 500 Hz apart.

3-89. In some analyzers, resolution is further limited by noise sidebands caused by residual FM in the local oscillator. In the 3580A, however, the 1 Hz bandwidth is the only bandwidth on which the noise sidebands can be resolved. On the 1 Hz bandwidth the noise sidebands are more than 70 dB below the peak of a CW response \pm 10 Hz away from the center frequency, f_0 (Figure 3-19). In some isolated cases, the noise sidebands may slightly degrade the resolution on the 1 Hz bandwidth. For the most part, however, noise sidebands can be ignored.

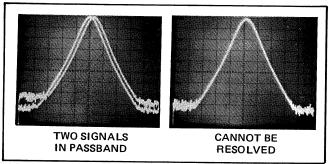


Figure 3-16. Two Signals In Passband.

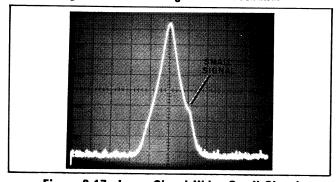


Figure 3-17. Large Signal Hides Small Signal.

Table 3-3. Frequency Resolution.

| AMPL | Max. |
|------------|------------|
| Difference | Resolution |
| 0 dB | 2 X BW |
| 10 dB | 2 X BW |
| 20 dB | 5 X BW |
| 30 dB | 5 X BW |
| 40 dB | 5 X BW |
| 50 dB | 10 X BW |
| 60 dB | 10 X BW |

BW = BANDWIDTH setting

3-90. Low Frequency Limit. To utilize the full dynamic range of the instrument at low frequencies, the lowest frequency to be resolved must be at least 5 times the selected BANDWIDTH. This low frequency limit is due to the zero response described in the following paragraphs.

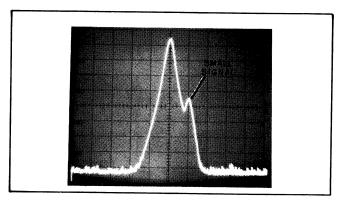


Figure 3-18. Small Signal Resolved.

3-91. As the 3580A frequency is tuned toward 0 Hz, the VTO frequency approaches the 100 kHz IF. Although the VTO signal is suppressed by the use of a double balanced mixer, part of the VTO signal feeds through the 100 kHz IF Filter and appears on the display. The response produced by the VTO signal peaks at 0 Hz and is appropriately called the "zero response". As with any other CW signal, the zero response on the display is an amplitude vs. frequency plot of the IF Filter (Figure 3-20). The wider the bandwidth, the wider the zero response.

3-92. The amplitude and bandwidth of the zero response determines the lowest frequency that can be resolved. On any BANDWIDTH setting, the peak amplitude of the zero response is more than 30 dB below the full scale reference set by the INPUT SENSITIVITY and amplitude VERNIER controls (AMPLITUDE REF LEVEL switch in NORMAL position). With the zero response more than 30 dB below full scale and a dynamic display range of 80 dB, the maximum difference between the peak of the zero response and any measureable input signal is between 40 dB and 50 dB. Table 3-3 indicates that the maximum resolution between two signals whose relative amplitude is between 40 dB and 50 dB is 5 times the BANDWIDTH setting.

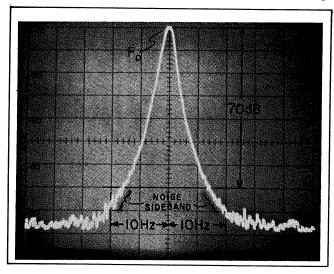


Figure 3-19. Noise Sidebands (1 Hz BW)

3-93. Response Time. Generally, when making swept frequency measurements, it is desirable to have good resolution and, at the same time, sweep as rapidly as possible. This involves a definite trade off since the narrow bandwidths provide the greatest resolution but require slower sweep rates. As the bandwidth is narrow, the IF Filter takes longer to respond to electrical changes taking place at its input. Consequently, the sweep rate must be slow so that the signal remains in the passband long enough for the filter to fully respond. Optimum sweep rate is discussed in Paragraph 3-135.

3-94. For applications where narrow bandwidths and slow sweep rates are required, the 3580A Adaptive Sweep feature can often be used to substantially reduce the measurement time. Adaptive Sweep is discussed in Paragraph 3-147.

3-95. Noise Rejection. The maximum sensitivity of the analyzer is limited by its own internally generated noise. As outlined in Paragraph 3-44, internal noise is a function of bandwidth, input resistance and tuned frequency. The narrower bandwidths provide the greatest noise rejection.

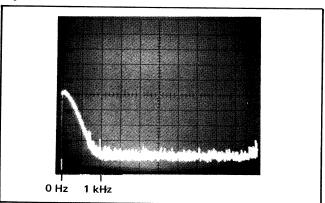


Figure 3-20. Zero Response (300 Hz BW).

3-96. Frequency Setting. Δ 16

3-97. The front panel FREQUENCY controls tune the frequency of the analyzer over the 0 Hz to 50 kHz range. The controls can be used to set either the start or center frequency of a linear sweep. The start or center frequency, selected by the FREQUENCY controls, is indicated on the FREQUENCY display.

3-98. The FREQUENCY controls consist of a coarse FREQUENCY control (located closest to the front panel) and a fine FREQUENCY control. Turning the coarse control will change the frequency in the display at a rate of about 5.7 kHz per resolution. The fine control will change the displayed frequency at a rate of about 75 Hz per revolution.

3-99. Frequency Display. The FREQUENCY display indicates the start or center frequency in Hz. When the instrument is in the Manual Sweep mode, the FREQUENCY display shows the frequency at which the

Manual Sweep is set. In the LOG mode the FREQUEN-CY display blanks. The FREQUENCY display resolution is 1 Hz represented by the least significant digit. The range of frequencies that may be displayed is 0 Hz to above 50 kHz. When the instrument is properly calibrated (Paragraph 3-195), the FREQUENCY display accuracy is: 3.5 Hz in the 1 Hz or 3 Hz Bandwidths when the ambient temperature is within the range of 0°C (32°F) to 55°C (131°F).

NOTE

Whenever the frequency display attempts to go below 0 Hz, the numeric readout is replaced with five dots. Thus, a display of five dots indicates that the frequency of the display is less than zero hertz.

3-100. Start/Center. With the START/CENTER slide switch in the START position, the FREQUENCY display indicates the frequency represented by the first vertical line on the left-hand side of the display graticule. This is the "start frequency" or frequency at which the sweep begins. With the switch in the CENTER position, the FREQUENCY display indicates the frequency represented by the center vertical line on the display graticule. This is the "center frequency" of the sweep. The START/CENTER switch is useful only in the REPetative, SINGle, or RESET mode. To indicate this, an amber light is provided to indicate when the instrument is in one of these sweep modes. When the instrument is in the MANual mode, the "MAN" indicator lights. If the instrument is in the LOG ZERO or LOG mode, both lights go out.

3-101. When surveying a spectrum containing two or more signals, it is generally convenient to leave the START/CENTER switch in the START position. The FREQUENCY controls can then be used to set the start frequency and the FREQUENCY SPAN control can be used to set the spectrum width or "end frequency". To observe one frequency component in a spectrum, set the START/CENTER switch to the CENTER position and set the FREQUENCY display to the frequency of interest. The frequency of interest will appear in the center of the display. The width of the center frequency response can be adjusted by changing the FREQUENCY SPAN or BANDWIDTH setting.

3-102. Fine FREQUENCY control used as Log Zero control. In the LOG ZERO Sweep mode, the fine FREQUENCY control is used to calibrate the Logarithmic frequency scale used in the LOG SWEEP mode. This is accomplished by adjusting the fine FREQUENCY control (in the LOG ZERO mode) such that the FREQUENCY display shows 20 Hz. In this way, the Logarithmic scale of the LOG SWEEP mode remains calibrated until the fine FREQUENCY control is readjusted.

Operating Instructions Model 3580A

3-103. Frequency Span Setting.

3-104. The FREQUENCY SPAN control sets the width of the spectrum to be observed during linear or manual sweeps. Excluding the 0 Hz position, there are ten FREQUENCY SPAN settings ranging from 5 Hz per division to 5 kHz per division. With ten horizontal divisions on the display, the overall spectrum width can be adjusted from 50 Hz to 50 kHz.

3-105. 0 Hz Span. With the FREQUENCY SPAN switch set to the 0 Hz position, the instrument remains at the start or center frequency indicated on the FREQUENCY display. The display, however, continues to sweep at the rate selected by the SWEEP TIME setting. The result is a graphical display of amplitude vs. time.

3-106. The amplitude vs. time feature is useful for observing the amplitude variations of a signal that occur over relatively long periods of time. For example, the amplitude of the 10 kHz sine wave shown in Figure 3-21A appears stable on a conventional oscilloscope but is actually varying at a very slow rate. In Figure 3-21B, the 3580A was used to monitor the amplitude of the 10 kHz signal over a 2, 000 second period. The 3580A amplitude vs. time display shows that the 10 kHz signal is amplitude modulated by a triangular-shaped signal whose frequency is 0.00166 Hz.

3-107. Because of its narrow bandwidth, the 3580A cannot respond to rapid changes in amplitude. The maximum modulating frequency that can be observed and measured with any accuracy is approximately 100 Hz on the 300 Hz BANDWIDTH setting.

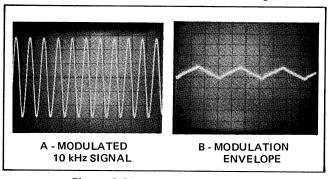


Figure 3-21. Amplitude vs. Time.

3-108. Frequency Out of Range.

3-109. There are a number of cases where the FRE-QUENCY and FREQUENCY SPAN settings are such that the frequency sweep attempts to go below 0 Hz or above 50 kHz. For example, if the start frequency is set to 10 kHz and the FREQUENCY SPAN setting is 5 kHz/div (50 kHz), the end frequency of the sweep is 60 kHz which is 10 kHz above the 50 kHz limit. If the instrument is set for a center frequency of 0 Hz, the start frequency is a negative value and the area between the start frequency and the center frequency is not meaningful.

3-110. To minimize erroneous indications, an internal detector senses when the frequency sweep tries to go below 0 Hz or above 50 kHz and, in turn, clears the display. The result is a clean baseline in areas where the frequency limits are exceeded (Figure 3-22).

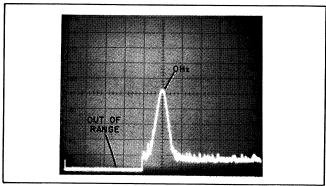


Figure 3-22. Frequency Out of Range.

3-111. The frequency out-of-range detector is not exact. Consequently, there are margin areas below 0 Hz and above 50 kHz where signals can be displayed. Typically, the margin below 0 Hz is about 500 Hz wide. Signals displayed in this negative margin are the images of the 0 Hz to 500 Hz signals displayed on the positive side of 0 Hz (Figure 3-23). The margin above 50 kHz is about 800 Hz wide and signals up to 50.8 kHz can generally be displayed.

3-112. The frequency sweep will go out of range under any of the following conditions:

```
a. When: Fstart + 10 Fspan = > 50 kHz
b. When: Fcenter + 5 Fspan = > 50 kHz
c. When: Fcenter - 5 Fspan = < 0 Hz</li>
Where: Fstart = start frequency of sweep
Fspan = FREQUENCY SPAN setting
Fcenter = center frequency of sweep
```

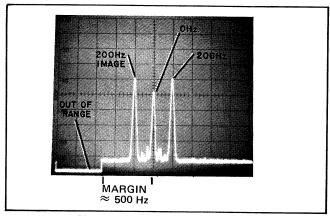


Figure 3-23. Margin Below 0 Hz.

3-113. Sweep Modes Δ 16.

3-114. The front panel SWEEP MODE switch permits selection of six sweep modes:

- 1) REP (Repetitive)
- 2) SING (Single)
- 3) RESET
- 4) MAN (Manual)
- 5) LOG ZERO
- 6) LOG
- 3-115. Repetitive Mode. In the Repetitive sweep mode the instrument sweeps continuously over the selected frequency range (the STR/CTR indicator is on in this mode). The duration of each sweep is determined by the SWEEP TIME setting. If the FREQUENCY controls are varied during a sweep, there will be no change in the FREQUENCY display until the beginning of another sweep. This may seem inconvenient during long sweeps. To overcome this, simply press CLEAR WRITE while varying the FREQUENCY controls.
- 3-116. Single Mode. When the Single sweep mode is selected, the instrument sweeps one time over the selected frequency range and stops at the end frequency (the STR/CRT indicator is on in this mode). The instrument remains at the end frequency until another sweep mode is selected or until a new sweep is initiated. A new sweep can be initiated by:
- a. Setting the SWEEP MODE switch to RESET and back to SING.
- b. Pressing the CLEAR WRITE button. This clears the display and simultaneously resets the sweep. Do not use clear-write when making x y recordings.
 - c. External triggering as outlined in Paragraph 3-143.
- 3-117. The Single sweep mode is particularly useful for making X-Y recordings using an external plotter connected to the rear panel RECORDER outputs. The operator can start the sweep, go about his business and return later to retrieve the completed recording.
- 3-118. It should be noted that the rear panel PEN LIFT output is operative *only* in the Single sweep mode. The PEN LIFT output is provided for use with X-Y recorders that have an electrically operated pen lift circuit enabling the pen to be remotely actuated by a contact closure (Paragraph 3-170).
- **3-119.** Reset Mode. When the Reset mode is selected, the sweep is reset to the left-hand side of the screen and the instrument remains at the *start* frequency determined by the FREQUENCY display setting.
- 3-120. The Reset mode is used primarily to facilitate the quick reseting of the start or center frequency. Since the FREQUENCY display updates only between sweeps (in

the Repetative mode), difficulty may arise in attempting to adjust the start or center frequency during slow sweeps. This difficulty is easily overcome by switching to the Reset mode, adjusting the start or center frequency (the STR/CRT indicator is on in this mode) and switching back to the Repetative mode. In general, for facile and expeditious tuning, switch to the Reset mode when adjusting the start or center frequency.

- 3-121. Manual Mode. In the Manual Sweep mode, the electronic frequency sweep is disabled and frequency control is transferred to the MANUAL VERNIER potentiometer. By adjusting the MANUAL VERNIER, the frequency can be set anywhere within the selected spectrum. With the MANUAL VERNIER set fully counterclockwise, the CRT sweep is at the left-hand side of the screen and the instrument is tuned to the start frequency determined by the FREQUENCY setting. As the vernier is rotated in a clockwise direction, the frequency increases and the video information is written (and retained) on the CRT just as it is when using the electronic sweep. In addition, the frequency at which the manual sweep is set, is shown on the FREQUENCY display. The "MAN" indicator lights to show the instrument is in the MANual mode.
- 3-122. The Manual sweep is useful for applications where it is necessary to precisely measure the frequency of a signal within the spectrum. For precise frequency measurements, simply manually tune to the desired signal and read the frequency directly from the FRE-QUENCY display. For an alternate method of frequency measurement, an electronic counter is connected to the rear panel TRACKING OSC OUT or LO OUTPUT to monitor the frequency. Using a narrow bandwidth such as 10 Hz or 30 Hz, the MANUAL VERNIER is adjusted so that the CRT sweep is at the peak of the signal to be measured. If the TRACKING OSC OUT is used, the frequency of the signal can then be read directty from the counter. If the LO OUTPUT is used, the frequency must be calculated by dividing the counter reading by ten and substracting 100 kHz (Paragraph 3-178).

NOTE

When the SWEEP MODE setting is changed from LOG ZERO to MAN or from RESET to MAN, the frequency sweep jumps from the start frequency to the frequency set by the MANUAL VERNIER. Conversely, when the SWEEP MODE is changed from MAN to LOG ZERO or from MAN to RESET, the frequency sweep jumps from the frequency set by the MANUAL VER-NIER to 0 Hz or to the start frequency. In either case, the rapid change in frequency will distort the trace being displayed on the CRT. If it is desirable to retain a specific trace when switching to or from the Manual mode, set the MANUAL VERNIER fully counterclockwise before changing the SWEEP MODE setting.

3-123. Log Zero Mode. The Log Zero mode is used to establish the correct starting frequency for the LOG sweep. When the Log Zero mode is selected, the sweep is reset to the left-hand side of the screen, the coarse FREQUENCY and FREQUENCY SPAN controls are disabled. To calibrate the log sweep, the front panel FINE FREQUENCY potentiometer is adjusted such that the FREQUENCY display reads 20 Hz. This ensures that the log sweep will start at 20 Hz.

3-124. LOG Sweep. When the LOG sweep mode is selected, the following things take place:

- a. The coarse FREQUENCY, FREQUENCY SPAN and SWEEP TIME controls are disabled and their settings do not effect the log sweep. The FINE FREQUENCY potentiometer remains operative and, to ensure the proper starting point for the log sweep, it must be adjusted so that the FREQUENCY display reads 20 Hz in the Log Zero mode.
- b. The instrument sweeps logarithmically over the 20 Hz to 43 kHz frequency range. The log sweep is repetitive and the duration of each sweep is approximately 5 seconds.

NOTE

When the LOG sweep mode is first selected or when the LOG sweep is initiated by external triggering, optimum frequency accuracy will not be obtained until 3 or 4 continuous sweeps have been made. This peculiarity of the LOG sweep is caused by dielectric absorption (soak effect) in the integrating capacitor of the LOG sweep generator.

- 3-125. By observing the CRT display it can be noted that each decade frequency of the LOG sweep is marked at the bottom of the graticule. The first vertical line on the left-hand side of the graticule represents 20 Hz, the second line represents 43 Hz and the third line 98.2 Hz. This sequence is repeated for each decade of frequency.
- 3-126. Figure 3-24 is a plot of frequency vs. time during a LOG sweep. At the beginning of the sweep the slope of the curve is gradual. A gradual slope indicates a small change in frequency for a given unit of time and thus, a slow sweep rate. As the sweep progresses the slope becomes steeper and the seep rate increases exponentially.
- 3-127. Because the 3580A is a narrow band instrument, the continuously increasing sweep rate presents a problem. At low frequencies narrow bandwidths are required to obtain good resolution. Narrow bandwidths can be used at low frequencies because the sweep rate is slow. As the frequency and sweep rate increases, however, the bandwidth must be widened so that the instrument can respond properly.

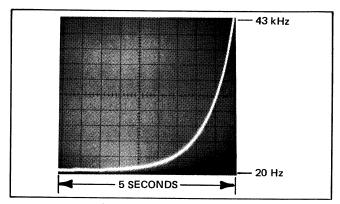


Figure 3-24. Frequency vs. Time (LOG Sweep).

3-128. The 300 Hz BANDWIDTH is the only bandwidth that allows the instrument to respond properly over the entire range of the LOG sweep. For this reason, the ADJUST light comes on when any bandwidth other than 300 Hz is selected. On the 300 Hz bandwidth, however, low frequency measurements are not possible because the resolution is poor and the skirt produced by the zero response covers nearly half of the display (Figure 3-25). For measurements at low frequencies a narrow bandwidth must be used. Table 3-4 lists the recommended bandwidths for measurements in given portions of the spectrum.

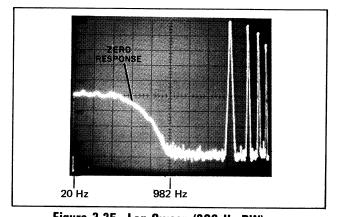


Figure 3-25. Log Sweep (300 Hz BW).

Table 3-4. Recommended Bandwidths (LOG Sweep).

| Frequency | Recommended | |
|-----------------|-------------|--|
| Range | Bandwidth | |
| 20 Hz-200 Hz | 10 Hz | |
| 200 Hz-982 Hz | 30 Hz | |
| 982 Hz-9.82 kHz | 100 Hz | |
| 9.82 kHz-43 kHz | 300 Hz | |

3-129. The LOG sweep is intended primarily for making log amplitude vs. log frequency plots of 2-port devices. For this application, the network to be tested is connected in the closed-loop configuration where the rear panel Tracking Oscillator Output supplies the stimulus and the 3580A measures the response.

NOTE

Because of the relatively fast sweep rates used in the Log sweep mode, conventional X-Y recorders connected to the rear panel RECORDER outputs cannot respond properly during LOG sweeps (see Paragraph 3-163).

3-130. During closed loop measurements the bandwidth limitations are not quite as stringent as those previously described. This is because the input frequency, derived from the Tracking Oscillator Output, is always in or near the center of the passband. The only requirement is that the bandwidth be wide enough to permit the instrument to fully respond to amplitude variations introduced by the network under test. If the network under test does not have extremely steep skirts, a relatively narrow bandwidth can be used. For example, Figure 3-26 is a log amplitude vs. log frequency plot of a 20 kHz notch filter. The plot was made using a 30 Hz bandwidth.

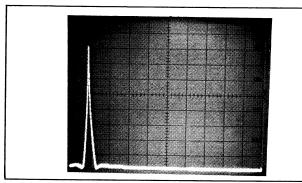


Figure 3-26. Log Amplitude vs. Log Freq. Plot of 20 kHz Notch Filter (30 Hz BW).

3-131. The easiest way to select the proper bandwidth for the Log sweep is to start with a wide bandwidth such as 100 Hz and then narrow the bandwidth until the amplitude or shipe of the response curve begins to change. When the response curve starts to change, the bandwidth is too narrow.

3-132. Sweep Time and Sweep Rate.

3-133. Sweep Time Control. The front panel SWEEP TIME control provides 14 sweep time settings ranging from 0.01 second per division to 200 seconds per division. With 10 horizontal divisions, total sweep time ranges from 0.1 second to 2,000 seconds.

3-134. Sweep Rate. The sweep rate in Hz per second is determined by the FREQ SPAN and SWEEP TIME settings:

$$R = \frac{F_s}{T}$$

Where: R = sweep rate in Hz/sec

F_s = FREQ SPAN setting T = SWEEP TIME setting Increasing the frequency span or decreasing the sweep time increases the sweep rate.

3-135. Optimum Sweep Rate. The optimum sweep rate is the maximum rate at which the frequency can be swept without excessively compressing or skewing the amplitude response. When the 3580A is sweeping at what is considered to be the optimum rate, the amplitude compression is about 2%.

3-136. The optimum sweep rate is determined by the response time of the instrument. If the response time is long, the sweep rate must be slow so that the instrument can respond properly. The response time of the 3580A is determined by the BANDWIDTH and DISPLAY SMOOTHING settings. Narrowing the bandwidth or increasing the display smoothing increases the response time and, therefore, decreases the optimum sweep rate.

3-137. Optimum Sweep Indicator. The 3580A is equipped with an internal detector that monitors the BAND-WIDTH, DISPLAY SMOOTHING, FREQUENCY SPAN and SWEEP TIME control settings. When these control settings are such that the sweep rate exceeds the optimum sweep rate, the front panel ADJUST indicator illuminates.

3-138. To sweep at the optimum rate, first set the FRE-QUENCY, FREQUENCY SPAN, BANDWIDTH and DISPLAY SMOOTHING controls to obtain the desired measurement parameters. Then starting with a slow SWEEP TIME setting, increase the sweep rate until the ADJUST light first comes on. When the ADJUST light comes on, rotate the SWEEP TIME control one position counterclockwise. The ADJUST light will go out and the instrument will sweep at the optimum rate.

3-139. Table 3-5 lists the optimum SWEEP TIME settings for various FREQ SPAN, BANDWIDTH and DISPLAY SMOOTHING settings.

3-140. For closed-loop measurements where the 3580A is used as a network analyzer, the optimum sweep rate is determined by the 3580A BANDWIDTH and DISPLAY SMOOTHING control settings and by the bandwidth of the network under test. During closed-loop measurements, the input frequency is always near the center of the passband and the IF Filter is required to respond only to amplitude variations introduced by the network. For this reason, the optimum sweep rate for closed-loop measurements is generally much faster than it is for open-loop measurements. In many closed-loop measurement applications the sweep rate can be set 20 to 25 times faster than the optimum rate indicated by the ADJUST light.

3-141. If the optimum sweep rate is not limited by the bandwidth of the 3580A, it may be limited by the bandwidth of the network under test. For bandpass and low pass filters, a rough approximation of optimum sweep rate can be made using the following formula:

$$R = \frac{BW^2}{2}$$

Where: R = optimum sweep rate in Hz/sec BW = bandwidth of network under test

3-142. In practice it is often difficult to predict the optimum sweep rate. For this reason, the simplest approach is to start with the optimum rate set using the ADJUST light. Then, while observing the response curve, gradually increase the sweep rate until the amplitude or shape of the curve begins to change. When

the curve begins to change the sweep rate is too fast.

3-143. External Triggering.

3-144. The EXT TRIG IN connector enables the frequency sweep to be remotely inhibited using a contact closure or TTL Logic Levels. This signal may be used to inhibit the sweep in the single, repetitive or Log Sweep mode.

Table 3-5. Optimum Sweep Time Settings.

| Bandwidth Setting | Freq Span/Div | Spectrum Width | Optimum SWP Time (Smoothing Min.) | Optimum SWP Time (Smoothing Med.) | Optimum SWP Time (Smoothing Max.) |
|----------------------|------------------|-------------------|--------------------------------------|--------------------------------------|--------------------------------------|
| 1 Hz | 5 Hz | 50 Hz | 10 sec. | 100 sec. | _ |
| 1 Hz | 10 Hz | 100 Hz | 20 sec. | 200 sec. | _ |
| 1 Hz | 20 Hz | 200 Hz | 50 sec. | _ | - |
| 1 Hz | 50 Hz | 500 Hz | 100 sec. | - | _ |
| 1 Hz | 0.1 kHz | 1 kHz | 200 sec.* | _ | - |
| 3 Hz | 5 Hz | 50 Hz | 1 sec. | 10 sec. | 100 sec. |
| 3 Hz | 10 Hz | 100 Hz | 2 sec. | 20 sec. | 200 sec. |
| 3 Hz | 20 Hz | 200 Hz | 5 sec. | 50 sec. | |
| 3 Hz | 50 Hz | 500 Hz | 10 sec. | 100 sec. | _ |
| 3 Hz | 0.1 kHz | 1 kHz | 20 sec. | 200 sec.* | _ |
| 3 Hz | 0.2 kHz | 2 kHz | 50 sec. | | - |
| 3 Hz | 0.5 kHz | 5 kHz | 100 sec. | - | |
| 3 Hz | 1 kHz | 10 kHz | 200 sec.* | _ | _ |
| 10 Hz | 5 Hz | 50 Hz | 0.1 sec. | 1 sec. | 10 sec. |
| 10 Hz | 10 Hz | 100 Hz | 0.2 sec. | 2 sec. | 20 sec. |
| 10 Hz | 20 Hz | 200 Hz | 0.5 sec. | 5 sec. | 50 sec. |
| 10 Hz | 50 Hz | 500 Hz | 1 sec. | 10 sec. | 100 sec. |
| 10 Hz | 0.1 kHz | 1 kHz | 2 sec. | 20 sec. | 200 sec.* |
| 10 Hz | 0.2 kHz | 2 kHz | 5 sec. | 50 sec. | 200 360. |
| 10 Hz | 0.5 kHz | 5 kHz | 10 sec. | 100 sec. | |
| 10 Hz | 1 kHz | 10 kHz | 20 sec. | 200 sec.* | _ |
| 10 Hz | 2 kHz | 20 kHz | 50 sec. | _ | _ |
| 10 Hz | 5 kHz | 50 kHz | 100 sec. | _ | _ |
| 30 Hz | 5 Hz | 50 Hz | 0.01 sec.** | 0.1 sec. | 1 |
| 30 Hz | 10 Hz | 100 Hz | 0.02 sec. | 0.1 sec. | 1 sec. |
| 30 Hz | 20 Hz | 200 Hz | 0.05 sec. | 0.5 sec. | 2 sec. 5 sec. |
| 30 Hz | 50 Hz | 500 Hz | 0.1 sec. | 1 sec. | 10 sec. |
| 30 Hz | 0.1 kHz | 1 kHz | 0.2 sec. | 2 sec. | 20 sec. |
| 30 Hz | 0.2 kHz | 2 kHz | 0.5 sec. | 5 sec. | 50 sec. |
| 30 Hz | 0.5 kHz | 5 kHz | 1 sec. | 10 sec. | 100 sec. |
| 30 Hz | 1 kHz | 10 kHz | 2 sec. | 20 sec. | 200 sec.* |
| 30 Hz | 2 kHz | 20 kHz | 5 sec. | 50 sec. | _ |
| 30 Hz | 5 kHz | 50 kHz | 10 sec. | 100 sec. | |
| 100 Hz | 5 Hz | 50 Hz | 0.01 sec.** | 0.01 sec.** | 0.1 |
| 100 Hz | 10 Hz | 100 Hz | 0.01 sec.** | 0.01 sec. | 0.1 sec. |
| 100 Hz | 20 Hz | 200 Hz | 0.01 sec.** | 0.02 sec. | 0.2 sec. 0.5 sec. |
| 100 Hz | 50 Hz | 500 Hz | 0.01 sec. * * | 0.1 sec. | 0.5 sec. 0.5 sec. |
| 100 Hz | 0.1 kHz | 1 kHz | 0.02 sec. | 0.1 sec. | 1 sec. |
| 100 Hz | 0.2 kHz | 2 kHz | 0.05 sec. | 0.5 sec. | 2 sec. |
| 100 Hz | 0.5 kHz | 5 kHz | 0.1 sec. | 1 sec. | 10 sec. |
| 100 Hz | 1 kHz | 10 kHz | 0.2 sec. | 2 sec. | 20 sec. |
| 100 Hz | 2 kHz | 20 kHz | 0.5 sec. | 5 sec. | 50 sec. |
| 100 Hz | 5 kHz | 50 kHz | 1 sec. | 10 sec. | 100 sec. |
| 300 Hz | 5 Hz | 50 Hz | 0.01 sec.** | 0.01 sec.** | |
| 300 Hz | 10 Hz | 100 Hz | 0.01 sec.** | 0.01 sec. * * | 0.01 sec.** 0.02 sec. |
| 300 Hz | 20 Hz | 200 Hz | 0.01 sec.** | 0.01 sec.** | 0.02 sec. 0.05 sec. |
| 300 Hz | 50 Hz | 500 Hz | 0.01 sec.** | 0.01 sec.** | 0.05 sec. 0.1 sec. |
| 300 Hz | 0.1 kHz | 1 kHz | 0.01 sec.** | 0.02 sec. | 0.1 sec. 0.2 sec. |
| 300 Hz | 0.2 kHz | 2 kHz | 0.01 sec.** | 0.05 sec. | 0.2 sec. |
| 300 Hz | 0.5 kHz | 5 kHz | 0.01 sec.** | 0.1 sec. | 1 sec. |
| 300 Hz | 1 kHz | 10 kHz | 0.02 sec. | 0.2 sec. | 2 sec. |
| 300 Hz | 2 kHz | 20 kHz | 0.05 sec. | 0.5 sec. | 5 sec. |
| 300 Hz | 5 kHz | 50 kHz | 0.1 sec. | 1 sec. | 10 sec. |

3-145. In order to inhibit the sweep, the externally applied signal into the EXT TRIG IN connector is kept low. To allow the 3580A to perform a single sweep the inhibit signal is allowed to go high for greater than 1 msec, but for less than the total sweep time. If the inhibit signal is not returned to low within the specified time, additional sweeps may be initiated.

3-146. To remotely inhibit the frequency sweep apply the following levels to the center terminals of the EXT TRIG IN connector:

Sweep Inhibit: Ground (through < 10 K) or -0.5 V

dc to 0.5 V dc.

Sweep Enable: Open or +2.5 V dc to +5 V dc.

NOTE

The outer shield of the EXT TRIG IN connector is connected to case ground. The center terminal of the connector is the inhibit line.

3-147. Adaptive Sweep.

3-148. One of the inconveniences associated with low frequency spectrum analyzers is the extremely slow sweep rates required when using narrow bandwidths. For example, to sweep over a 200 Hz spectrum using a 1 Hz bandwidth, the optimum sweep time setting is 50 seconds per division. This makes the overall measurement time 500 seconds or about 8 minutes. If a sweep time setting of 200 seconds per division is used, the total measurement time is 2,000 seconds or 33 minutes.

3-149. In many applications relatively wide portions of the spectrum being swept do not contain useful information. The plot shown in Figure 3-27, for example, has a number of narrow spectral components but more than 98% of the display is nothing but noise floor. Using a conventional sweep at the optimum sweep rate, it took more than 15 minutes to trace the plot shown in Figure 3-27. Using the 3580A Adaptive Sweep feature, however, the same plot (minus the noise floor) was traced in about 1.5 minutes (Figure 3-28).

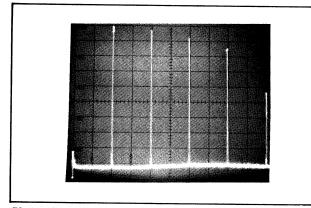


Figure 3-27. Plot Using Conventional Sweep (15 minutes).

3-150. To use the Adaptive Sweep feature, the operator sets a baseline threshold using the front panel ADAPTIVE SWEEP control. The baseline threshold can be adjusted anywhere from the bottom of the screen to approximately 70% of full scale. For the plot shown in Figure 3-28, the baseline threshold was set about 10 dB above the noise floor.

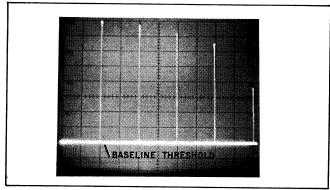


Figure 3-28. Plot Using Adaptive Sweep.

3-151. At the beginning of the Adaptive Sweep, the instrument sweeps at the rate selected by the SWEEP TIME setting. This ensures that the zero response or any other signal on or near the start frequency will be properly detected. After the sweep passes through any initial responses, the sweep rate is automatically increased to 20 or 25 times the selected rate. When the sweep reaches a response that rises above the baseline threshold, it backs up slightly, pauses to allow the IF Filter to settle and then sweeps slowly over the response at the panel-selected rate. When the response has been completely traced, the sweep is again speeded up until another response is encountered. As a result, the portions of the spectrum below the threshold level are not displayed, but the spectral responses above the threshold level are displayed just as they are using a conventional sweep. By sweeping rapidly through unused portions of the spectrum, the Adaptive Sweep greatly reduces the overall measurement time.

3-152. Setting the Baseline Threshold. When setting the baseline threshold for the Adaptive Sweep, the following guidelines must be observed:

a. In the Linear amplitude mode the threshold must be at least 60% below the peak of the smallest signal to be displayed. For example, if the peak of the smallest signal to be displayed is 4 vertical divisions, the threshold must be at least 2.4 divisions (0.6 X 4) below it. Similarly, if the peak of the smallest signal to be displayed is 1 vertical division, the threshold must be at least 0.6 of a division below it.

b. In the Log amplitude mode, the threshold must be at least 8 dB below the peak of the smallest signal to be displayed.

3-153. The reason for setting the baseline threshold below the peak of the smallest signal to be displayed is that the responses are detected when the instrument is sweeping 20 to 25 times faster than the panel-selected rate. During these fast sweeps the IF Filter does not have time to fully respond. As a result, the signals applied to the internal threshold detector are about 6 dB (50%) lower in amplitude than they are when sweeping at the optimum rate. If the threshold is not more than 6 dB below the peak of a signal, that signal will not be detected and consequently, will not be displayed.

NOTE

Adaptive Sweep cannot be used on the 0.05 sec., 0.02 sec. and 0.01 sec. SWEEP TIME settings.

3-154. With the SWEEP TIME control set to one position slower than the optimum rate, the signal compression during fast sweeps is approximately 3 dB or 30%. This allows the baseline threshold to be set 4 dB or 45% below the peak of the smallest signal to be displayed. The trade off here is that the measurement time is considerably longer than it is when sweeping at the optimum rate.

3-155. Adaptive Sweep, Log 1 dB Mode. The Adaptive Sweep is difficult to use in the Log 1 dB amplitude mode. This is because the display range is only 10 dB and, when sweeping at the optimum rate, the baseline threshold must be at least 8 dB below the peak of the smallest signal to be displayed. With the baseline threshold at the bottom of the screen, signals more than 2 dB below full scale will not be displayed. If the Adaptive Sweep is to be used in the Log 1 dB mode, set the SWEEP TIME control one or two positions slower than the optimum sweep rate. This will reduce the amplitude compression during fast sweeps and allow at least 50% of the display range to be used. If the Adaptive Sweep is not to be used in the Log mode, be sure the ADAPTIVE SWEEP control is in the OFF position.

3-156. Adaptive Sweep Marker. When the ADAPTIVE SWEEP control is set to the ON position, a sweep marker appears on the display. The sweep marker is a blank spot or gap in the trace that indicates the position of the frequency sweep. The sweep marker is provided because the digital memory that generates the display does not track the fast-forward and fast-backward excursions of the Adaptive Sweep. The sweep marker enables the operator to observe these excursions, making it easy to verify that the Adaptive Sweep is operating properly.

3-157. In some cases it may be desirable to display the sweep marker without using the Adaptive Sweep. This can be done in the Linear and Log 10 dB modes by setting the ADAPTIVE SWEEP control to the ON position and leaving the baseline threshold at the bottom of the display. With the baseline threshold at the bottom of

the display, the video level exceeds the threshold level causing the instrument to continually sweep at the panel-selected rate.

3-158. Digitally-Stored Display.

3-159. A unique feature of the 3580A is its digitally-stored display. The digitally-stored display provides a number of unusual operating conveniences. For example, display adjustments are not required when the sweep parameters are changed. The digitally-stored trace is automatically cleared and updated at the correct rate. The INTENSITY and FOCUS controls have the same effect as those of a regular oscilloscope. Once they are set, they do not need to be readjusted. Moreover, the INTENSITY control can be set to any level without danger of burning the CRT face. Digital storage provides a bright, crips, flicker-free presentation. There is no blooming of display ambiguity.

3-160. One of the major advantages of digital storage is its ability to retain a trace indefinitely, i.e., as long as power is applied to the instrument. When a signal sweep is made, the trace that is generated will continue to be displayed until the CLEAR WRITE button is pressed or until it is replaced by a new sweep. If a trace is needed for furture reference, it can be permanently stored in memory by simply pressing the STORE pushbutton. The "stored trace" and a current or "refresh trace" can then be displayed simultaneously (Figure 3-29). If desired, the stored trace can be blanked from the display by pressing the BLANK STORE button. Releasing the BLANK STORE button returns the stored trace to the display.

3-161. A permanently stored trace is not effected by changing the control settings or by pressing the CLEAR WRITE button. The only way the stored trace can be cleared from memory is by releasing the STORE button or turning the power off. When the STORE button is initially released, the stored trace disappears and a series of dots appear on the display (Figure 3-30). The dots are automatically cleared when the display is updated by a new sweep.

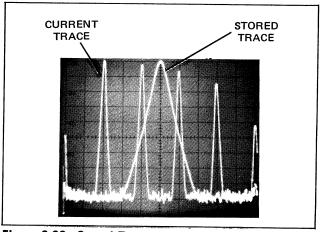


Figure 3-29. Stored Trace and Current Trace Displayed Simultaneously.

3-162. Reduced Resolution. The digital memory in the 3580A has 1024 addresses when the Y-axis amplitude information is stored. When the STORE button is not presses, each address corresponds to a given position of the frequency sweep and the X-axis of the display is divided into 1024 discreet segments. When the STORE button is pressed, the memory is split in half. One half (512 addresses) is used for the stored trace and the other half is for the refresh trace. Since only 512 addresses are used for each trace, the display resoultion is decreased. This means that the display is not quite as detailed as it is with a single trace stored in 1024 addresses. The techniques used for storing information and splitting the memory are such that the peaks of the responses are always retained. Thus, the reduced resolution does not normally obsecure any useful information.

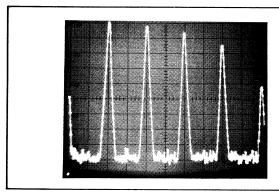


Figure 3-30. Store Button Released.

3-163. Recorder Outputs.

3-164. Recorder outputs are provided on the rear panel of the 3580A to permit the use of an external X-Y recorder/plotter. The -hp- Model 7035B Option 020 X-Y Recorder is recommended. Although the standard Model 7035B and other X-Y recorders can be used, the 7035B Option 020 is preferable because it has some special features that simplify scale calibration. In addition, the Model 7035B Option 020 is equipped with an X-axis log converter which can be used to scale the 3580A linear sweep to obtain a full log sweep over a 3-decade (10 Hz to 10 kHz) range.

3-165. X-Axis Output. The X-AXIS output supplies a dc voltage proportional to the position of the frequency sweep on the CRT display. When the sweep is at the start frequency, the X-Axis output is 0 V dc; when the sweep is at the end frequency, the output is + 5 V dc. The output resistance is 1 Kilohm, nominal.

3-166. In the Repetitive and Single sweep modes when Adaptive Sweep is not used, the X-Axis output is a 0 V to +5 V linear ramp. When Adaptive Sweep is used, the output voltage tracks the forward and reverse excursions of the sweep. In the Manual Sweep mode, the X-Axis output voltage corresponds to the sweep position set by the MANUAL VERNIER control. When the Reset or Log Zero mode is selected, the X-Axis output remains at 0 V dc.

3-167. In the Log Sweep mode, the frequency is swept logarithmically but the X-Axis output is still a 0 V to +5 V linear ramp. An output of 0 V dc corresponds to the 20 Hz start frequency, an output of +2.5 V dc corresponds to 982 Hz at the center of the display and an output of +5 V dc corresponds to the 43 kHz end frequency.

NOTE

Because of the relatively fast sweep rates used in the Log Sweep mode, conventional X-Y recorders connected to the X-AXIS output cannot respond properly. To make amplitude vs. log-frequency recordings, use an X-Y recorder that has a built-in log converter for the X-axis input (-hp-7035B Opt. 020). Connect the 3580A X-AXIS output to the X-axis input of the recorder. With the recorder set to the Log mode, sweep the 3580A at a slow linear rate using the Single or Repetitive Sweep mode.

3-168. Y-Axis Output. The Y-AXIS output supplies a dc voltage proportional to the amplitude of the responses appearing on the display. An output of 0 V dc corresponds to the bottom of the screen; The Y-Axis output voltage is 0.5 V per division in the Linear Amplitude mode, 0.05 V per dB in the Log 10 dB mode and 0.5 V per dB in the Log 1 dB mode. Output resistance is 1 kilohm nominal

3-169. There are several things about the Y-AXIS output that should be noted:

a. In the Log 10 dB mode, rotating the AMPLITUDE REF LEVEL control in a clockwise direction offsets the display in steps of 10 dB. This also offsets the Y-Axis output in steps of + 0.5 V.

b. In the Log 1 dB mode, the display ranges from 0 dB (+5 V) to -10 dB (0 V). The Y-Axis output, however, extends from approximately +1 dB (+5.5 V) to -13 dB (-1.5 V).

c. Changing the baseline threshold using the ADAP-TIVE SWEEP control does not effect the Y-Axis output voltage.

3-170. Pen Lift Output. The PEN LIFT output is provided for use with X-Y recorders having electrically operated pen lift circuits that allow the pen to be remotely actuated by a contact closure. The PEN LIFT output is operative only in the Single sweep mode. If Adaptive Sweep is not used, a contact closure is present between the PEN LIFT output terminals for the duration of the single sweep. If Adaptive Sweep is used, the contact closure is present only when the instrument is sweeping slowly over a response. This prevents the fastforward and fast-backward excursions of the sweep from being recorded. The PEN LIFT output terminals are isolated from case ground. Do not use clear-write to reset sweep.

3-171. Tracking Oscillator Output.

3-172. The rear panel TRACKING OSC OUT connector supplies a 5 Hz to 50 kHz sinusodial output signal that tracks the tuned or swept frequency of the instrument. The specified frequency response of the tracking oscillator output signal is $\pm 3\%$ over the 5 Hz to 50 kHz frequency range. Total harmonic distortion and spurious is more than 40 dB below a 1 V rms signal level. The output impedance is 600 ohms, nominal. When the output is terminated in 600 ohms, the LEVEL control may be used to adjust the output from 0 V to 1 V rms.

3-173. The frequency accuracy of the tracking oscillator output signal is specified at ± 2.5 Hz relative to the center of the instrument's passband. On the 1 Hz and 3 Hz bandwidths, the passband is less than 2.5 Hz above and below the center frequency. Thus, the tracking oscillator output frequency may be slightly outside of the passband. This is of little consequence except during closed-loop measurements where the tracking oscillator signal is fed into the INPUT through a network under test. If the tracking oscillator frequency is outside the passband, insertion loss will be encountered. Under worst case conditions, maximum insertion loss is approximately 30 dB on the 1 Hz bandwidth and 8 dB on the 3 Hz bandwidth. Typically, the insertion loss is about 5 dB on the 1 Hz bandwidth and 2 dB on the 3 Hz bandwidth.

WARNING

Dangerous voltages, capable of causing death, are present in this instrument. Refer the adjustment of A2C4 (discussed in Paragraph 3-174) to Service Trained Personnel only.

3-174. For most closed-loop measurements optimum results will be obtained using the 10 Hz or 30 Hz bandwidth. If, for some reason, the 1 Hz or 3 Hz bandwidth is used insertion loss can be minimized by removing the top cover and adjusting A2C4 (100 kHz ADJ) so that the tracking oscillator frequency is in the center of the passband. An alternative approach is to apply an external reference signal to the TRACKING OSC IN connector and adjust the frequency of the reference so that the tracking oscillator frequency is in the center of the passband (see Paragraph 3-176).

3-175. Tracking Oscillator Input.

3-176. Figure 3-31 is a simplified block diagram of the Tracking Oscillator circuit. With the rear panel slide switch in the NORMAL position, the 100 kHz to 150 kHz signal from the VTO is mixed with a 100 kHz signal from a Crystal Oscillator. The 0 Hz to 50 kHz difference frequency is fed through a 50 kHz Low-Pass Filter and applied to the TRACKING OSC OUT connector. With the slide switch in the EXT REF position,

the 100 kHz Crystal Oscillator is disconnected and an external reference signal can be applied to the Mixer through the TRACKING OSC IN connector. The frequency of the external reference signal can be varied about 100 kHz to offset or frequency modulate the tracking oscillator output signal. Increasing the frequency of the external reference signal decreases the tracking oscillator output frequency; decreasing the external reference frequency increases the tracking oscillator output frequency.

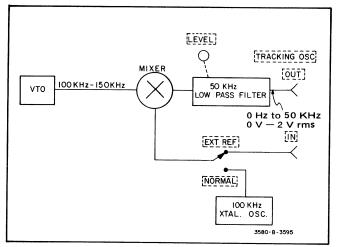


Figure 3-31. Tracking Oscillator.

3-177. The signal level applied to the TRACKING OSC IN connector should be 100 mV rms \pm 10%. Use a highly stable signal source such as an -hp- Model 3320A/B or 3330A/B Frequency Snythesizer. The impedance of the tracking oscillator input is approximately 3.6 kilohms.

3-178. L.O. Output.

3-179. The VTO in the 3580A generates 1 MHz signal which is divided in frequency to obtain the 100 kHz to 150 kHz VTO signal that is applied to the Input Mixer and Tracking Oscillator. The 1 MHz to 1.5 MHz signal from the VTO is available at the rear panel LO OUT-PUT connector. The signal level at the LO OUTPUT is 10 mV rms; output impedance is 1 kilohm, nominal.

3-180. The tuned frequency of the instrument can be measured to an accuracy of ± 5 Hz with an electronic frequency counter connected to the LO OUTPUT. The following formula can be used to calculate the tuned frequency from the counter reading:

$$F_t = \frac{F_c}{10} - 100 \text{ kHz}$$

Where: F_t = tuned frequency F_c = counter reading

3-181. The tuned frequency of the instrument can be measured using either the L.O. Output or the Tracking Oscillator Output. It is generally preferable to use the

L.O. Output because it provides greater frequency resolution. Also the L.O. Output frequency can be measured using a 0.1 second gate time for fast response.

3-182. Option 001.

3-183. The 3580A Option 001 is equipped with an internal rechargeable battery pack and a protective front panel cover for complete portability.

WARNING

To protect operating personnel, the 3580A Option 001 chassis must be grounded. For power line operation connect the power cord to a three-prong grounded receptacle. For battery operation connect the common (black) input terminal to earth ground or to an appropriate system ground. If a system ground is used be sure it is actually at ground potential and is not a voltage source.

3-184. The 3580A Option 001 can be operated from the ac power line or from its own internal battery pack. With POWER switch set to the ON (AC) position, the instrument receives its power from the ac power line and a trickle charge is applied to the batteries. The trickle charge prevents the batteries from discharging, but is not sufficient to recharge the batteries in a reasonable time. With the POWER switch in the ON (BAT) position, the ac power is tuned off and the instrument receives it power solely from the internal battery pack. A fully charged battery pack will operate the instrument for more than 5 hours. When the batteries are discharged to the point where they cannot operate the instrument properly, the power is automatically shut off. This eliminates erroneous measurements caused by weak batteries and further prevents the batteries from being damaged due to excessive discharge.

3-185. To recharge the batteries, connect the instrument to an appropriate ac power source and set the POWER switch to the CHARGE position. The POWER light will illuminate. The instrument cannot be operated while the batteries are being charged. Recharge time for completely discharged batteries is 14 hours. The useful life of the batteries is more than 100 charge/discharge cycles.

ECAUTION?

The instrument should not be left in the CHARGE mode for prolonged periods. A charge period of 14 hours is sufficient to recharge a fully discharged battery pack. Extended periods of overcharge in ambient temperatures exceeding 30°C (86°F) will severely degrade battery life and capacity by causing the cells to overheat.

3-186. Temperature Limits. To prevent battery damage, the following temperature limits must be observed:

a. Operating Temperature: $0^{\circ}C (+32^{\circ}F) t_0 + 40^{\circ}C (+140^{\circ}F)$

b. Charge Temperature Range: 0° C (+32°F) to +40°C (+104°F)

c. Storage Temperature Range: -40°C (-40°F) to +50°C (+122°F)

3-187. Option 002.

3-188. The 3580A Option 002 is equipped with a front panel slide switch which permits selection of three input configurations: Unbalanced, Balanced Bridge, and Balanced Terminated. These input configurations are illustrated in Figure 3-32. In addition, the 3580A Option 002 TRACKING OSC OUT is transformer coupled to provide a 600-ohm balanced output configuration, with an output level of 0 V to > 1 V rms into 600-ohms (Adjustable).

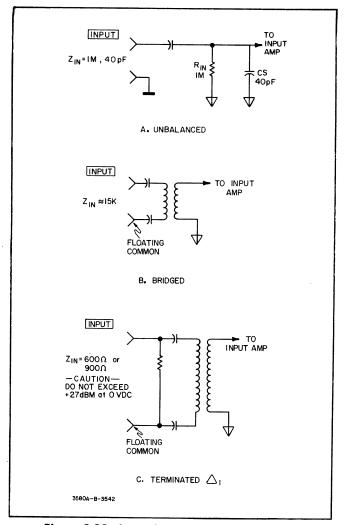


Figure 3-32. Input Configurations (Option 002).

ECAUTION 3

Δ

The differential signal level applied to the Option 002 Balanced Terminated input must not exceed + 27 dBm at 0 V dc. The combined ac and dc levels must be such that the power dissipated by the terminating resistor is less than 0.5 watt.

3-189. The 3580A Option 002 can be calibrated for absolute measurements in rms volts, dBm/600 ohms or

dBm/900 ohms. The selection is made using the front panel dBm 900 ohm/LIN - - dBm 600 ohm slide switch. Relative measurements can be made in dB or percent of full scale.

3-190. It should be noted that in the unbalanced input configuration, the input shunt capacitance is 40 pF, nominal. This differs from the 30 pF shunt capacitance of the standard Model 3580A. If a 10:1 divider probe is used, it must have sufficient adjustment range to compensate for the 40 pF shunt capacitance. An -hp- Model 10003A Voltage Divider Probe is recommended.

Δ₁ Refer to Section VIII Backdating.

- 3-191. BASIC OPERATING PROCEDURES.
- 3-192. Instrument Turn On.
- 3-193. Power Line Operation.
 - a. Check the line voltage at the point of installation.
- b. Refer to Figure 3-33 and set the 3580A for the line voltage to be used (100 V, 120 V, 220 or 240 V). Line voltage must be within +5% to -10% of voltage setting.
- c. Verify that the proper fuse is installed in the fuse holder:

| Line Setting | Fuse Type | -hp- Part No. |
|--------------|------------------------------|---------------|
| 100 V/120 V | 0.5 A, 250 V Normal Blow | 2110-0012 |
| 220 V/240 V | 0.25 A, 250 V Normal Blow | 2110-0004 |

- d. Connect the detachable ac power cord to the rear panel power receptacle and to the power source.
- e. Set the POWER switch to the ON (AC) position. The POWER light will illuminate.
- f. Allow approximately 15 seconds for the CRT to warm up. Adjust the INTENSITY and FOCUS controls for a bright, clear presentation on the CRT. When the instrument is initially turned on, the display may be similar to the one shown in Figure 3-34. This display reflects the preferred states of the storage elements in the digital memory and is not meaningful. To clear the display, press the CLEAR WRITE button.
- g. Allow a warm-up period of at least 1 hour before using the 3580A in a critical measurement application.

3-194. Battery Operation (Option 001).

- a. Connect the low (black) terminal of the front panel INPUT connector to earth ground or to an appropriate system ground.
- b. Set the POWER switch to the ON (BAT) position. The POWER light will illuminate.
- c. Allow approximately 15 seconds for the CRT to warm up. Adjust the INTENSITY and FOCUS controls for a bright, clear presentation on the CRT. When the instrument is initially turned on, the display may be similar to the one shown in Figure 3-34. This display reflects the preferred states of the storage elements in the digital memory and is not meaningful. To clear the display, press the CLEAR WRITE button.
- d. Allow a warm-up period of at least 1 hour before using the 3580A in a critical measurement application.
- e. To recharge the batteries, perform Steps a through d of the power-line turn on procedure (Paragraph 3-193). Set the POWER switch to the CHARGE position. The POWER light will illuminate. The instrument cannot be used while the batteries are being charged.

ECAUTION 3

The instrument should not be left in the CHARGE mode for prolonged periods. A charge period of 14 hours is sufficient to recharge a fully discharged battery pack. Extended periods of overcharge in ambient temperatures exceeding 30°C (86°F) will severely degrade battery life and capacity by causing the cells to overheat.

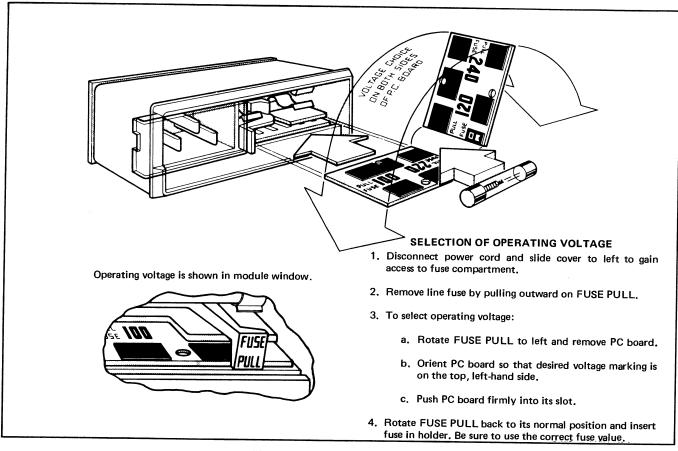


Figure 3-33. Voltage Selection.

3-195. Frequency Calibration Procedure (Log Sweep only) $\Delta \text{16}.$

3-196. This procedure should be performed before each use of the Log Sweep mode.

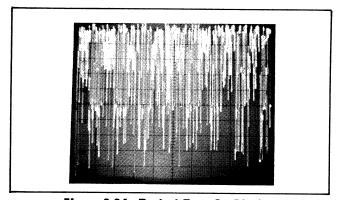


Figure 3-34. Typical Turn On Display.

3-197. Set the -hp- 3580A controls as follows:

| ADAPTIVE SWEEP DISPLAYSTORE and | |
|---------------------------------|----------|
| | Released |
| AMPLITUDE MODE | |
| AMPLITUDE REF LEVEL | NORMAI |

| dBV/LIN - dBm Switch | dRV/I IN |
|----------------------|-----------------|
| INPUT SENSITIVITY | CAI |
| VERNIER | CAL |
| | |
| FREQUENCY | N/A |
| START CTR | N/A |
| BANDWIDTH | 30 Hz |
| DISPLAY SMOOTHING | MIN |
| FREQ. SPAN/DIV | N/A |
| SWEEP TIME/DIV | N/A |
| SWEEP MODE | LOG ZERO |
| PRESS | EAR WRITE |
| | 27 717 17 17 17 |

3-198. Adjust the fine FREQUENCY control until the FREQUENCY display reads 20 Hz.

3-199. Amplitude Calibration Procedure Δ 16.

3-200. The Amplitude Calibration Procedure should be performed initially after warm-up and each time the BANDWIDTH setting is changed.

3-201. For operation on the 1 Hz or 3 Hz BAND-WIDTH proceed as follows:

- a. Turn the instrument on (Paragraph 3-192).
- b. Set the 3580A controls as follows:

| DISPLAYSTOR | E and BLANK STORE |
|----------------------|--------------------------|
| | Released |
| AMPLITUDE MODE | LOG 10 dB/DIV |
| AMPLITUDE REF LEV | ELNORMAL |
| dBV/LIN - dBm Switch | dBV/LIN |
| INPUT SENSITIVITY. | |
| VERNIER | |
| | (Fully CW) |
| FREQUENCY | |
| START CTR | |
| BANDWIDTH | |
| | whichever is to be used) |
| DISPLAY SMOOTHING | |
| FREQ. SPAN/DIV | |
| SWEEP TIME/DIV | N/A |
| SWEEP MODE | |
| | |

- c. Turn the ADAPTIVE SWEEP control to the on position so the sweep marker (gap) appears on the horizontal trace. Leave the baseline threshold at the bottom of the screen.
- d. While pressing the CLEAR WRITE button, adjust the MANUAL VERNIER so that the sweep marker is in the center of the display. Release the CLEAR WRITE button and set the ADAPTIVE SWEEP control to the OFF position.
- e. Carefully adjust the fine FREQUENCY control for a peak 10 kHz response in the center of the display.
- f. Using a small screwdriver, adjust the front panel CAL 10 kHz potentiometer so that the peak of the 10 kHz response is exactly full scale.
- g. Set the AMPLITUDE MODE to LOG 1 dB/DIV. Repeat Step f.
- 3-202. For operation on the 10 Hz, 30 Hz, 100 Hz or 300 Hz BANDWIDTH proceed as follows:
 - a. Turn the instrument on (Paragraph 3-192).
 - b. Set the 3580A controls as follows:

| DISPLAYSTORE and BLANK STORE |
|------------------------------|
| Released |
| AMPLITUDE MODELOG 10 dB/DIV |
| AMPLITUDE REF LEVELNORMAL |
| dBV/LIN - dBm SwitchdBV/LIN |
| INPUT SENSITIVITYCAL |
| VERNIER CAL |
| (Fully CW) |
| FREQUENCY10000 Hz |
| START CTRCTR |
| BANDWIDTH10 Hz - 300 Hz |
| (whichever is to be used) |
| DISPLAY SMOOTHINGMIN |
| FREQ. SPAN/DIVSee Table 3-6 |
| SWEEP TIME/DIVSee Table 3-6 |
| SWEEP MODEREP |
| |

- c. Using the ADAPTIVE SWEEP control, set the baseline threshold to -60 dB on the display.
- d. Using a small screwdriver, adjust the front panel CAL 10 kHz potentiometer so that the peak of the 10 kHz response is exactly full scale.
- e. Set the AMPLITUDE MODE to LOG 1 dB/DIV. Using the ADAPTIVE SWEEP control, set the baseline threshold to the bottom of the display. Repeate Step d.

Table 3-6. Control Settings (Amplitude Calibration).

| Bandwidth Setting | Freq. Span/Div | Sweep Time/Div |
|----------------------|----------------|----------------|
| 10 Hz | 20 Hz | 0.5 sec |
| 30 Hz | 0.1 kHz | 0.2 sec |
| 100 Hz | 0.5 kHz | 0.1 sec |
| 300 Hz | 1 kHz | 0.02 sec |

3-203. Input Probe Compensation.

- 3-204. Before using a 10:1 voltage divider probe it is necessary to adjust the probe for optimum frequency response. Once the probe is properly adjusted, it should not require further attention. It is good practice, however, to perform periodic verification tests to ensure that optimum adjustment is maintained.
- a. Turn the instrument on as outlined in Paragraph 3-192.
- b. Connect the probe to the 3580A INPUT using a BNC to banana-plug (-hp- Part Number 1251-2277).
 - c. Set the 3580A controls as follows:

| ADAPTIVE SWEEPOF | F |
|-----------------------------|----|
| DISPLAYSTORE and BLANK STOR | E |
| Release | d |
| AMPLITUDE MODELOG 10 dB/DI | V |
| AMPLITUDE REF LEVELNORMA | L |
| INPUT SENSITIVITY10 d | |
| FREQUENCY000000 H | Z |
| START CTRSTAR | |
| BANDWIDTH300 H | ĺΖ |
| DISPLAY SMOOTHINGMI | N |
| FREQ. SPAN/DIV2 KH | z |
| SWEEP TIME/DIV0.05 SEG | |
| SWEEP MODERE | |

- d. Set the rear panel LEVEL control fully clockwise (facing rear panel).
- e. Connect the probe tip to the rear panel TRACK-ING OSC OUT connector. Connect the ground lead of the probe to case ground.

- f. Adjust the front panel amplitude VERNIER so that the horizontal trace is between 0 dB and -10 dB on the display.
 - g. Set the AMPLITUDE MODE to LOG 1 dB/DIV.
- h. Adjust the probe so that its response is flat over the entire frequency range (Figure 3-35).

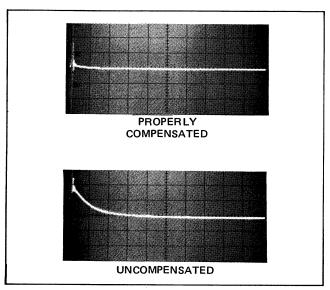


Figure 3-35. Probe Compensation.

3-205. Familiarization Exercise.

3-206. The following procedure demonstrates the Digital Storage, Adaptive Sweep and other operating features of the 3580A.

- a. Turn the instrument on as outlined in Paragraph 3-92. Perform the Amplitude Calibration Procedure (Paragraph 3-199). Perform the amplitude calibration using the 100 Hz BANDWIDTH and the LOG 10 dB AMPLITUDE MODE.
 - b. Set the 3580A controls as follows:

| ADAPTIVE SWEEPOFF |
|------------------------------|
| DISPLAYSTORE and BLANK STORE |
| Released |
| |
| AMPLITUDE MODELOG 10 dB/DIV |
| AMPLITUDE REF LEVELNORMAL |
| dBV/LIN - dBm SwitchdBV/LIN |
| INPUT SENSITIVITYCAL |
| VERNIER |
| (Fully CW) |
| FREQUENCY000000 Hz |
| START CTRSTART |
| BANDWIDTH100 Hz |
| DISPLAY SMOOTHINGMIN |
| |
| FREQ. SPAN/DIV |
| SWEEP TIME/DIV1 SEC |
| SWEEP MODEREP |

- c. The spectral components of the 10 kHz calibration signal will now appear on the display. If the instrument is properly calibrated, the peak of the 10 kHz fundamental frequency component will be at full scale and the zero response will coincide with the first line on the left-hand side of the display graticule.
- d. Set the BANDWIDTH switch to the 30 Hz position. The ADJUST light will illuminate to indicate that the sweep rate is too fast. As the trace is updated by a new sweep, the amplitudes of the various frequency components will be compressed because the IF Filter does not have time to fully respond.
- e. Rotate the SWEEP TIME control counterclockwise until the ADJUST light goes out (10 SEC). When the ADJUST light goes out, the instrument is sweeping at the optimum rate.
- f. Set the SWEEP MODE switch to the SING (Single) position. Press and release the CLEAR WRITE button. This will clear the display and initiate a new sweep. Allow 100 seconds for the display to be updated. The trace generated by the single sweep will continue to be displayed until it is cleared or replaced by a new sweep.
- g. Press the STORE button and then press the BLANK STORE button. The trace currently being displayed is now permanently stored in memory and can be recalled at any time by releasing the BLANK STORE button.
- h. Using the ADAPTIVE SWEEP control, set the baseline threshold about 10 dB above the noise floor.
- i. Press and release the CLEAR WRITE button to initiate a new sweep. Observe the fast and slow excursions of the Adaptive Sweep. Note that the pen lift relay clicks each time the instrument begins to sweep slowly over a response. The Adaptive Sweep takes only about 15 seconds to trace the plot that previously took 100 seconds.
- j. Set the ADAPTIVE SWEEP control to the OFF position. Release the BLANK STORE button to compare the 15 second trace and the 100 second trace. The two traces will be identical except the 15 second trace obtained using the Adaptive Sweep will not have a noise floor. Again press the BLANK STORE button. The permanently stored trace will disappear.
- k. Set the SWEEP MODE switch to the REP (Repetitive) position.
- 1. To examine the 20 kHz frequency component in greater detail, set the START/CTR switch to CTR, set the FREQ SPAN/DIV to 0.5 KHz and set the SWEEP TIME/DIV to 1 SEC. At this point, the center of the display is 0 Hz and the negative frequencies on the left-hand side of 0 Hz are blanked. Set the FREQUENCY

display to 20000 Hz. When the trace is updated by a new sweep, the 20 kHz frequency component will appear in the center of the display.

- m. Set the BANDWIDTH switch to 300 Hz. This will make the 20 kHz component wider because the analyzer's response to a CW signal is an amplitude vs. frequency plot of the IF Filter.
- n. Release the BLANK STORE button. The permanently stored trace will reappear on the display. Even though the sweep parameters have been changed, the stored trace appears exactly as it did when the STORE button was initially pressed.
- o. Set the FREQ SPAN/DIV to 5 KHz and allow 10 seconds for the display to be updated.
- p. Release the STORE button. The previously stored trace will disappear and a series of dots will appear on the current trace. The dots will be cleared when the display is updated by a new sweep.

3-207. Technique For Measuring Noise.

3-208. The 3580A uses peak detection on the sweep spectrum. Therefore, the noise displayed is peak noise and can be several dB higher than average noise. Average noise measurements can be made if the following technique is used:

- a. Use display smoothing.
- b. Ignoring the adjust warning light, decrease Sweep Time/Div until the display noise level no longer decreases. The spectrum shape of the noise should be gradually changing, not abruptly, allowing the spectrum analyzer to follow it well.
- 3-209. Average Detection Error. The video detector is an average responding full wave detector. This type of detector has an inherent error when detecting noise. In the 3580A, the error occurs in both the linear and log modes of operation. To correct for this error, multiply the displayed reading by 1.128 to get the rms value.
- **3-210.** Log Conversion Error. In the Log mode of operation, an additional correction must be made to compensate for log conversion error. Add 1.5 dB to the corrected display reading.

NOTES

1. Only "Gently" varying noise spectra can be accurately measured using this technique. Accurate measurement of both discrete lines and noise levels in the same spectrum is generally not possible.

- 2. To calculate the equivalent noise bandwidth, multiply the 3 dB bandwidth by 1.12. Remember that the 3 dB bandwidth has a tolerance of \pm 15% and therefore should be measured if accurate results are desired.
- 3. The recorder Y Axis output is linear and continuous. Noise measurements can be made by connecting a true rms reading voltmeter to this output. See Paragraph 3-168 for operating information concerning the Y Axis output. The use of an X-Y recorder may also prove beneficial in making noise measurements.

3-211. Rear Panel Display Adjustment Controls. △21

Instruments with serial prefix 2030A06030 or greater have these controls which allow for the precise adjustment of the horizonal, vertical, and rotational axis of the display. If the display is misaligned use the following procedure to correct.

a. Set the 3580A controls as follows:

| ADAPTIVE SWEEP | OFF |
|-----------------------|------------------|
| DISPLAYAll Push | buttons Released |
| Amplitude MODE | .Log 1 dB/DIV |
| Amplitude REF LEVEL | NORMAL |
| INPUT SENSITIVITY | CAL |
| VERNIER | CAL (Fully CW) |
| FREQUENCY | 00000 Hz |
| START-CTR | START |
| RESOLUTION BANDWIDTH. | 300 Hz |
| DISPLAY SMOOTHING | MIN |
| FREQ. SPAN/DIV | 5 KHz |
| SWEEP TIME/DIV | 0.1 SEC |
| SWEEP MODE | REP |
| | , |

Use an alignment tool inserted through the DISPLAY ADJUST access holes in the rear panel to make the following adjustments:

- b. Adjust the rotation control so the display baseline is parallel to the bottom of the graticule line.
- c. Adjust the vertical position control so the display falls on the bottom graticule line.
- d. Adjust the horizontal position control so the CAL signal peaks fall on the third, fifth, and seventh vertical graticule lines.
- e. Repeat b. through d. until the display is properly adjusted.

SECTION IV THEORY OF OPERATION

4-1. INTRODUCTION.

4-2. This section contains a Simplified Block Diagram Description and a Functional Description of the 3580A Spectrum Analyzer.

4-3. SIMPLIFIED BLOCK DIAGRAM DESCRIPTION. Δ 16

4-4. Refer to the Simplified Block Diagram (Figure 4-1) for the following discussion.

The 3580A can be divided into five major sections:

- 1) Amplitude Section
- 2) Frequency and Sweep Section
- 3) Digital Storage Section
- 4) Display
- 5) Frequency Display

4-5. Amplitude Section.

- 4-6. The Amplitude Section consists of an Input Circuit, an Overload Detector an Input Mixer, an IF Filter, Log and Linear IF Amplifiers, a Video Detector, a Video Filter and a Video Output Circuit.
- **4-7.** Input Circuits. The Input Circuits, controlled by the front panel INPUT SENSITIVITY switch, provide the gain or attenuation needed to maintain the proper signal level at the input of the Mixer. The Input Circuits also contain a 50 kHz low-pass filter which prevents image frequencies (200 kHz and above) from reaching the Mixer.
- 4-8. Overload Detector. The Overload Detector at the input of the Mixer senses when the input level exceeds the design limits and, in turn, lights the front panel OVERLOAD indicator. This is an important function since signals that overdrive the mixer can produce harmonic and spurious mixing products which ultimately appear on the display.
- **4-9.** Input Mixer. The Input Mixer is a double-balanced active mixer in which the 0 Hz to 50 kHz input signal is mixed with a 100 kHz to 150 kHz signal from the Voltage-Tuned Local Oscillator (VTO). The output of the Mixer is a composite signal containing the upper and lower sidebands.
- 4-10. To select a given frequency component present at the input of the Mixer, the VTO frequency is tuned so that the difference between it and the frequency of interest is 100 kHz:

Fvto - Fin = 100 kHz

Where: Fvto = 100 kHz to 150 kHz VTO frequency Fin = 0 Hz to 50 kHz input frequency

The 100 kHz intermediate frequency (IF) is fed through the IF Filter, detected and displayed on the CRT. Signals outside the passband of the IF Filter are rejected.

- 4-11. IF Filter. The IF Filter contains five cascaded crystal filter stages. The center frequency of the filter is 100 kHz and the 3 dB bandwidth varies from 1 Hz to 300 Hz as a function of the front panel BANDWIDTH setting. Since the Input Circuits and Input Mixer are broadband through 50 kHz, the selectivity of the instrument is determined entirely by the bandwidth of the IF Filter.
- 4-12. Log and Linear Amplifiers. The 100 kHz output of the IF Filter is applied to the Video Detector through a Log Amplifier in the Log Amplitude mode or through a Linear Amplifier in the Linear Amplitude mode. The Log Amplifier converts the amplitude of the incoming IF signal to a logarithmic value, providing an 80 dB display range. The Linear Amplifier is a conventional amplifier circuit in which the gain is varied to provide the 20 V, 10 V, 2 V, 1 V ranging sequence used in the Linear mode. Also, the Linear Amplifier contains a variable attenuator which increases the overall gain as the AMPLITUDE REF LEVEL switch is changed from the X1 position.
- **4-13. Video Detector.** The Video Detector is an average-responding, full-wave detector circuit which produces a dc voltage proportional to the amplitude of the 100 kHz log or linear input signal.
- **4-14.** Video Filter. The Video Filter is an RC filter network controlled by the BANDWIDTH and DISPLAY SMOOTHING controls. The purpose of the filter is to smooth-out the ripple and noise riding on the detected video signal.
- 4-15. Video Output Circuit. The Video Output Circuit functions as an output buffer in the Linear mode and as a variable gain amplifier in the Log 10 dB and Log 1 dB modes. In the Log 10 dB mode, a variable dc offset voltage, controlled by the AMPLITUDE REF LEVEL switch, is summed with the video input signal. This allows the entire display to be offset in steps of 10 dB as the AMPLITUDE REF LEVEL setting is change from 0 dB to -70 dB. In the Log 1 dB mode the gain of the

Video Output Circuit is increased to provide an expanded scale of 1 dB per division. Changing the AMPLITUDE REF LEVEL setting then varies the dc offset voltage to select any 10 dB portion of the 80 dB range. The output of the Video Output Circuit, ranging from 0 V to +5 V dc, is applied to the rear panel Y-AXIS output connector and to the Digital Storage Section.

4-16. Frequency and Sweep Section.

- 4-17. The Frequency and Sweep Section consists basically of a Ramp Generator, a Dial Mixing Amplifier, a Voltage-Tuned Local Oscillator (VTO) and a Tracking Oscillator.
- 4-18. Ramp Generator. The Ramp Generator produces a 0 V to +5 V linear ramp which is applied to the Dial Mixing Amplifier and to the Digital Storage Section. The frequency of the ramp is determined by the front panel SWEEP TIME setting. The FREQ SPAN control, located between the Ramp Generator and Dial Mixing Amplifier, determines the amplitude of the ramp applied to the VTO and thus, the overall change in frequency produced by the ramp.
- **4-19. Dial Mixing Amplifier.** In the Dial Mixing Amplifier, the ramp voltage is combined with a variable dc voltage from the front panel FREQUENCY control. This dc voltage establishes the low-frequency limit or "start frequency" of the VTO.
- **4-20.** VTO. The VTO generates a 100 kHz to 150 kHz square wave which is applied to the Input Mixer in the Amplitude Section and to the Tracking Oscillator.
- **4-21. Tracking Oscillator.** In the Tracking Oscillator, the 100 kHz to 150 kHz VTO signal is mixed with a 100 kHz signal from a crystal oscillator. This produces the 0 Hz to 50 kHz tracking signal which is available at the rear panel TRACKING OSC OUT connector.

4-22. Digital Storage Section.

- 4-23. Because of the extremely slow sweep rates used in the 3580A, some form of display storage is required. The most common method for obtaining display storage is to use a storage CRT in which the display is retained by the phosphor or by a "storage mesh" located behind the CRT face. Relatively recent advances in large-scale integrated circuits, however, have made it possible to use a digital storage technique in the 3580A. Digital storage permits the use of a standard oscilloscope CRT and further provides several operating conveniences not available with conventional displays.
- 4-24. In the Digital Storage Section, the 0 V to +5 V "frequency ramp" from the Frequency and Sweep Section is applied to an A to D converter where it is converted to binary and used to address a memory bank. At the same time the detected video information from the

Amplitude Section is converted to binary by an A to D converter and stored in the memory locations addressed by the ramp. The binary video data is then non-destructively read out of memory, converted to dc, processed and applied to the vertical deflection plates of the CRT.

- 4-25. During the read cycle, a "display ramp," generated in the Digital Storage Section, is used to address the memory and drive the horizontal deflection plates of the CRT. The display ramp scans the memory and sweeps the display approximately 50 times each second. This is a much faster rate than that of the frequency ramp used for storing data. The memory contents are, therefore, refreshed at the slow frequency-sweep rate, while data is read-out of memory at the rapid display-sweep rate. The result is a flicker-free, stored presentation.
- 4-26. When the front panel STORE button is pressed, the display currently in memory is processed and stored in one-half of the memory locations. This leaves the other half of the memory available for the refresh trace. During the read cycle, the display ramp first scans the memory locations containing the refresh trace. It then recycles and scans the locations containing the previously stored trace. Due to the rapid scan rate of the display ramp, the stored trace and the refresh trace appear simultaneously on the CRT.

4-27. Frequency Display Section.

- 4-28. The Frequency Display Section consists of: a Limiter, a Time Base Generator, a Counter, a Display Driver, and a Display.
- 4-29. Limiter Circuit. The Limiter converts the Buffered 100 kHz Reference signal into a square wave that clocks the Time Base Generator. The Buffered 100 kHz Reference is first filtered through a 100 kHz bandpass network to eliminate any spurious signals that could cause time base errors. The output of the filter is then changed to a square wave by a comparator circuit. This square wave is called the Time Base Clock.
- 4-30. Time Base Generator. The Time Base Generator is used to generate the timing signals for the Counter. These timing signals start and stop the counter, and load a register with the last state of the counter. The Time Base Clock is the input for the Time Base Generator.
- 4-31. Counter, Display Driver, and Display. The 1 MHz -1.5 MHz L.O. signal from the VTO is first divided by two; the signal is then counted by the counter chip for an amount of the time determined by the Time Base Generator. The last state of the counter is then loaded into an on-chip register. Finally, the BCD contents of the register is scanned and outputted at the same time the corresponding digit strobes are outputted. The BCD output is interfaced with the display via s BCD-to-

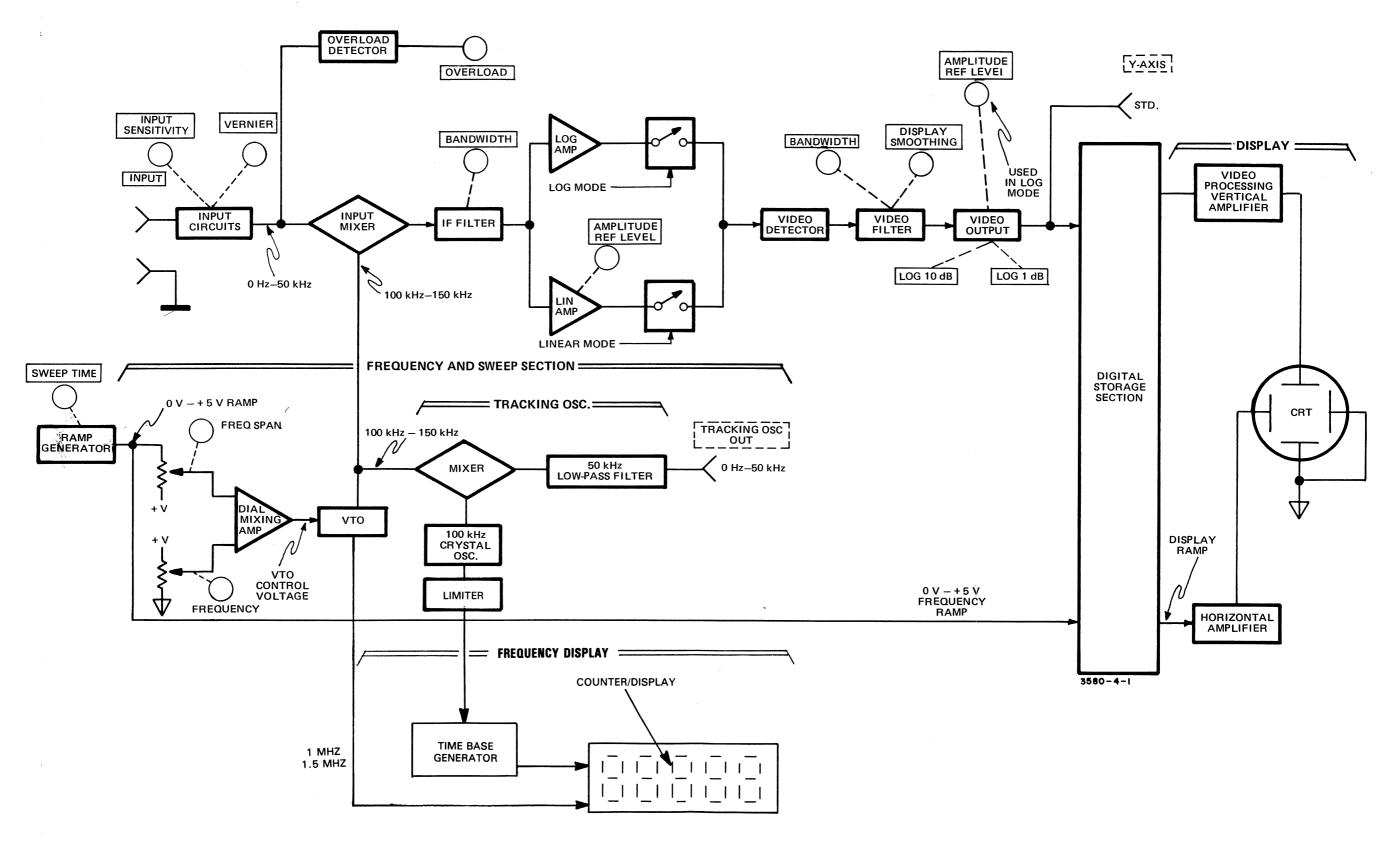


Figure 4-1. Simplified Block Diagram. 4-3/4-4

Theory of Operation

seven-segment-decoder/driver; the digit strobes are buffered and sent to the display. Thus, the contents of the register are displayed.

4-32. FUNCTIONAL DESCRIPTION.

4-33. Amplitude Section.

- 4-34. Refer to the Amplitude Section Detailed Block Diagram (Figure 7-3) for the following discussion.
- 4-35. Input Attenuator. The Input Attenuator, controlled by the front panel INPUT SENSITIVITY switch, serves as an input voltage divider and coupling network between the INPUT connector and the Input Amplifier. The attenuator is comprised of 5 RC divider networks. These networks provide the required signal attenuation for the + 30 dB (20 V) through -10 dB (0.2 V) ranges. On the -20 dB (0.1 V) through -70 dB (0.2 mV) ranges, the Input Attenuator is bypassed by the Input Sensitivity switch and the input signal is applied directly to the Input Amplifier. Table 1 of the Detailed Block Diagram lists the maximum (full scale) input levels, input attenuation and resulting signal levels applied to the Input Amplifier for each INPUT SENSITIVITY setting.
- 4-36. Input Amplifier. The Input Amplifier is a low noise, high input-impedance amplifier circuit which provides variable gain and impedance conversion between the Input Attenuator and the Post Attenuator. The Input Amplifier gain, controlled by the INPUT SENSITIVITY switch, is approximately X1.25 (+1.8 dB) on the +30 dB through -50 dB ranges and is increased to X12.5 (+21.8 dB) on the -60 dB and -70 dB ranges. Table 1 of the Detailed Block Diagram lists the full-scale input levels, Input Amplifier gain and full-scale output levels for each INPUT SENSITIVITY setting.
- 4-37. Post Attenuator. The Post Attenuator is a resistive divider network controlled by the INPUT SEN-SITIVITY switch and by the front panel slide switch that selects dBV/LIN or dBm/600 Ω . With the slide switch in the dBV/LIN position, the post attenuation is -5 dB or -15 dB. With the switch in the dBm/600 Ω position, the attenuation is -2 dB or -12 dB. Table 1 of the Detailed Block Diagram lists the full-scale input levels, post attenuation and output levels for each INPUT SENSITIVITY setting.
- 4-38. Post Amplifier. The output of the Post Attenuator is applied to the Post Amplifier through the wiper of the front panel VERNIER potentiometer, R2. The Post Amplifier provides the final stage of gain and buffering before the signal is applied to the Input Mixer. The Post Amplifier gain, controlled by the INPUT SENSITIVITY switch, is approximately X4.5 (+ 13.2 dB) the + 30 dB through -30 dB ranges and is increased to X45 (+ 33.2 dB) on the -40 dB through -70 dB ranges.

- 4-39. Table 1 of the Detailed Block Diagram lists the full-scale input levels, Post Amplifier gain and full-scale output levels for each INPUT SENSITIVITY setting. In the Log 10 dB or Log 1 dB amplitude mode, the normalized full-scale output of the post Amplifier is 100 mV rms on all ranges. In the Linear mode, the full-scale output is 100 mV rms on the 10 V, 1 V, 0.1 V, etc. ranges and 62 mV rms on the 20 V, 2 V, 0.2 V, etc. ranges. To compensate for this difference in full-scale levels in the Linear mode, the gain of the Linear Amplifier (following IF Section) is increased on the ranges having a lower output voltage.
- **4-40.** Overload Circuit. The Overload Circuit consists of an Overload Detector, an Overload Driver and an LED Overload Indicator. The Overload Detector is a full-wave peak detector designed to sense an overvoltage condition at the output of the Post Amplifier. During normal operation, the full-scale output of the Post Amplifier is 0.1 V rms or 0.14 V peak. If the signal level exceeds 0.14 V peak, the Overload Driver is gated on and the OVERLOAD indicator illuminates.
- 4-41. Note that the Overload Driver has one input labeled "Overload Inhibit." With the INPUT SEN-SITIVITY switch in the CAL position, -10 V dc is applied to the Overload Inhibit line causing the Overload Driver to remain cut off. This prevents the 10 kHz calibration signal from producing an OVERLOAD indication. The calibration signal is a pulse train in which the amplitude of the 10 kHz fundamental-frequency component is set to produce full-scale deflection on the CRT. Because of the rich harmonic content of the pulse train, its overall amplitude is slightly greater than 1.4 V peak.
- 4-42. Low-Pass Filter. To prevent image frequencies (200 kHz and above) from reaching the Input Mixer, the signal from the output of the Post Amplifier is fed through a 50 kHz Low-Pass Filter network. This "Cauer" filter is a 7-pole, passive, LCR filter network. The response of the filter is essentially flat over the 5 Hz to 50 kHz input frequency range. The filter provides 50 dB of rejection at 100 kHz and more than 90 dB of rejection at 200 kHz. Due to the 604-ohm series input resistance (R65) and the 604-ohm terminating resistance (R91), the filter introduces -6 dB of insertion loss. This makes the normalized full-scale input to the Mixer equal to 50 mV rms or 31 mV rms in the Linear mode.
- 4-43. Input Mixer. The Input Mixer Section consists of an active mixer (U2), a gain control circuit and an output buffer (Q14-Q16).
- 4-44. The mixer is a monolithic, double-balanced modulor circuit driven by a 0.8 V p-p, 100 kHz to 150 kHz square wave from the VTO. In the mixer, the square wave from the VTO alternatively gates out positive and negative portions of the 5 Hz to 50 kHz input signal, resulting in full-wave balanced multiplication between the two signals. When the mixer is proper-

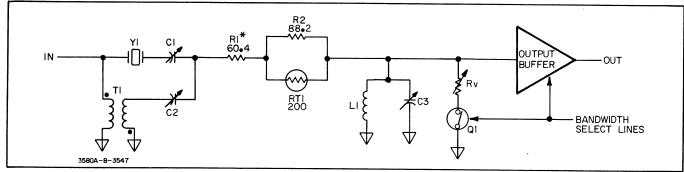


Figure 4-2. Typical Crystal Filter Stage.

ly balanced, the VTO and input frequencies are supressed and the composite output signal is predominately the upper and lower sidebands.

- 4-45. The gain control circuit at the output of the mixer is a resistive attenuator controlled by transistor switches Q12 and Q13. Transistor switch Q12 is energized on the 1 Hz and 3 Hz BANDWIDTH settings and Q13 is energized on the 10 Hz and 30 Hz BANDWIDTH settings. The result is that the signal level is decreased as the bandwidth in narrowed. The reasons for this are:
- a. On the wider bandwidths, the noise floor in the IF Filter rises. A larger signal is, therefore, required to maintain the required signal-to-noise ratio.
- b. On the narrow bandwidths, the IF Filter becomes non-linear when high-level signals are applied. Since the noise floor is lower, the non-linearity can be minimized by lowering the signal level.
- 4-46. The output buffer is a 3-stage amplifier circuit which provides gain and isolation between the Mixer and the IF Filter. The gain of the output buffer can be varied by adjusting the front panel CAL 10 kHz potentiometer.
- **4-47. IF Filter.** The IF Filter consists of 5 synchronously-tuned crystal filter stages. Each stage (Figure 4-2) can be divided into 6 major sections:
 - 1. Crystal (Y1) and Pulling Capacitor (C1)
 - 2. Capacitive Compensating Network (T1, C2)
 - 3. Resistive Compensating Network (R1, R2, RT1)
 - 4. Parallel Resonant Circuit (L1, C3)
 - 5. Variable Q Switch (Rv, Q1)
 - 6. Output Buffer
- 4-48. Crystal. The crystals used in the IF Filter are preaged at the factory and are selected for a center frequency between 99,991 Hz and 99,993 Hz. The reason for selecting a frequency slightly lower than the required 100.00 kHz, is to allow the frequency to be adjusted by placing a "pulling" capacitor (C1) in series with the crystal (see Figure 4-3A).

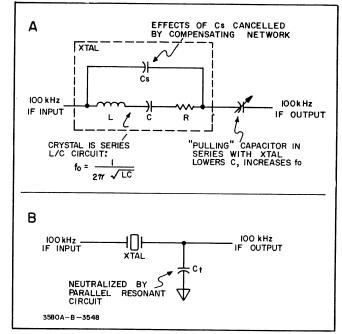


Figure 4-3. Crystal Filter.

- 4-49. Capacitive Compensating Network. The purpose of the capacitive compensating network is to neutralize the shunt capacitance (Cs) of the crystal and any stray capacitance introduced by the component leads and circuit board. Transformer T1 functions as an inverter, producing a voltage that is equal in amplitude and 180 degrees out of phase with the signal applied to the crystal. With the value of C2 set equal to the shunt capacitance, the circulating current flowing through the shunt capacitance is cancelled.
- 4-50. Resistive Compensating Network. The resistive compensating network compensates for variations in the series resistance (Rs) of the crystal. The value of R1 is factory selected so that:

$$R1 + Rs = 200 \text{ ohms}$$

The nominal value of Rs varies from crystal to crystal and is derived from the crystal manufacturer's specifications. The parallel network consisting of R2 and thermistor RT1 compensates for variations in Rs due to temperature.

4-51. Parallel Resonant Circuit. Stray capacitance to ground at the output of the crystal (Figure 4-3B) is neutralized by including it in the parallel resonant circuit formed by L1 and C3. The parallel resonant circuit peaks up the high end and balances out the symmetry of the filter skirts.

4-52. Variable "Q" Switching. The bandwidth and "Q" of each crystal filter stage can be defined by two basic equations:

$$BW = \frac{Fo}{Q}$$

Where: BW = 3 dB Bandwidth

Fo = 100 kHz Resonant Frequency

Q = Figure of Merit

$$Q = \frac{X}{R}$$

Where: Q = Figure of Merit

 $X = Reactance (X_1 \text{ or } X_c) \text{ of crystal at Fo}$

R = Sum of compensated series resistance of crystal (Rs) and variable resistance to ground (Rv)

From these equations it can be noted that bandwidth is inversely proportional to Q and Q is inversely proportional to resistance. Thus, decreasing the resistance increases the Q and narrows the bandwidth; increasing the resistance decreases the Q and widens the bandwidth.

- 4-53. The bandwidth of the filter is varied by switching in the appropriate values of resistance (Rv) to ground. This is accomplished by transistor and diode switches controlled by lines from the front panel BANDWIDTH switch. The resistor values range from 127 K on the 300 Hz bandwidth to 120 ohms on the 1 Hz bandwidth.
- 4-54. With five cascaded stages, the bandwidth of each state must be 2.57 times the required bandwidth. Thus, for a 30 Hz bandwidth each state must have a bandwidth of 77.1 Hz and for a 1 Hz bandwidth each state must have a bandwidth of 2.57 Hz.
- 4-55. Output Buffer. The output buffer is a two-stage amplifier circuit which provides interstage gain and isolation. The buffer has a high impedance FET input state which prevents it from loading the crystal on the wider bandwidths. The gain of the buffer is Unity on the 300 Hz through 10 Hz bandwidths, approximately X1.2 on the 3 Hz bandwidths and X3 on the 1 Hz bandwidth. The gain increase is required on the 1 Hz and 3 Hz bandwidths to compensate for the insertion loss introduced by the low resistance of Ry.
- **4-56. IF Amplifier.** The IF Amplifier section (Figure 4-4) consists of a Gain Control circuit and an LCR-tuned IF Amplifier.
- 4-57. Gain Control Circuit. The gain of the IF Amplifier is determined by the input resistance provided by the Gain Control circuit and by the impedance of the parallel LCR network in the feedback loop. The Gain Control circuit has six resistive input branches. The in-

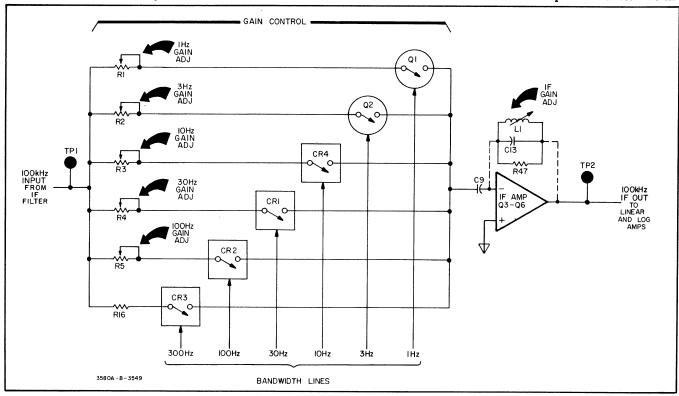


Figure 4-4. IF Amplifier.

put branches are individually switched into the circuit by transistor and diode switches controlled by lines from the BANDWIDTH switch. With the exception of the 300 Hz branch, each section of the Gain Control circuit contains a variable resistor. This provides a separate gain adjustment for each BANDWIDTH setting. The separate gain adjustments compensate for gain variations that occur in the Input Mixer and IF Filter.

4-58. IF Amplifier. The IF Amplifier is a 3-stage amplifier circuit which is tuned to 100 kHz by the parallel resonant tank circuit in the feedback loop. The 3 dB bandwidth of the amplifier is approximately 1.2 kHz. The IF Amplifier has a low-impedance complementary-symmetry output stage which drives the following log and linear amplifier stages. The full-scale signal level at the output (TP2) of the IF Amplifier is approximately 2.8 V rms on all six BANDWIDTH settings.

4-59. Linear Amplifier. The Linear Amplifier (Figure 4-5) consists of an Input Attenuator, an Input Amplifier, an Output Attenuator and an Output Amplifier. The Input Attenuator is controlled by the front panel AMPLITUDE REF LEVEL switch and provides either -40 dB or 0 dB of attenuation. The Input Amplifier provides a fixed gain of approximately 40 dB. The Output Attenuator, also controlled by the AMPLI-TUDE REF LEVEL switch provides -40 dB, -30 dB -20 dB or -10 dB of signal attenuation. Table 4-1 lists the input attenuation, Input Amplifier gain, output attenuation and the resulting gain or attenuation for each AMPLITUDE REF LEVEL setting. Note that as the AMPLITUDE REF LEVEL switch is rotated from X1 (NORMAL) position, the attenuation is decreased and the signal level is increased in steps of 10 dB.

4-60. The Output Amplifier stage provides the variable gain needed to maintain a 0 V rms to 1.2 V rms full-scale output on all input ranges and reference settings. The gain of the Output Amplifier is controlled by both the INPUT SENSITIVITY switch and the AMPLITUDE

REF LEVEL switch. By observing these two front panel controls, it can be noted that the full-scale reference on the INPUT SENSITIVITY switch dial is indicated by a white window that is mechanically linked to the AMPLITUDE REF LEVEL switch. Changing the position of either switch changes the full-scale reference in a 20 V, 10 V, 2 V, 1 V sequence. This sequence differes from the 10 dB/step sequence provided by the A9 Input Circuit and the attenuators in the Linear Amplifier. For this reason, the gain of the Output Amplifier is changed on alternate ranges. With the full-scale reference set to 10 V, 1 V, 0.1 V, etc., the gain of the Output Amplifier is X56. With the reference set to 20 V, 2 V, 0.2 V, etc. the gain is increased to X88.

Table 4-1. Linear Amplifier Gain.

| Ampl Ref Level | Input Atten. | Input Amp Gain | Output Atten. | Net Gain or Atten |
|-------------------|-----------------|-------------------|------------------|----------------------|
| X1 | -40 dB | + 40 dB | -40 dB | -40 dB |
| | -40 dB | + 40 dB | -30 dB | -30 dB |
| X0.1 | -40 dB | + 40 dB | -20 dB | -20 dB |
| | -40 dB | + 40 dB | -10 dB | -10 dB |
| X0.01 | OdB | + 40 dB | -40 dB | OdB |
| | OdB | + 40 dB | -30 dB | + 10 dB |
| X0.001 | O dB | + 40 dB | -20 dB | + 20 dB |
| | OdB | + 40 dB | -10 dB | + 30 dB |

4-61. Log Amplifier. The Log Amplifier (Figure 4-6) is a hybrid circuit consisting of a log amplifier package (U5) and four external control amplifiers (U1 - U4). The log amplifier package contains 12 differential amplifier stages. Each stage has a logarithmic output characteristic over a 10 dB range (Figure 4-7). Internal resistive dividers and the external control amplifiers bias each stage to respond to a different 10 dB portion of the input signal. The outputs of the 12 stages are summed in a common load (R_L), forming the composite output characteristic shown in Figure 4-8.

4-62. From Figure 4-8, the following can be noted:

a. When the input signal is below the range of a given stage, that stage will make essentially no contribution to the output of the log amplifier.

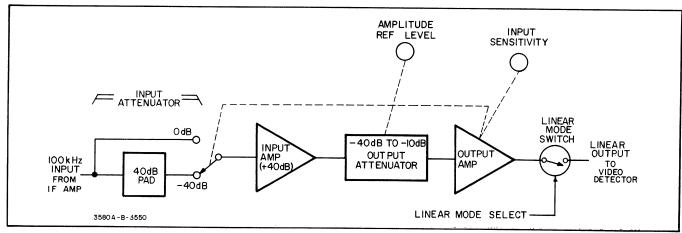


Figure 4-5. Linear Amplifier.

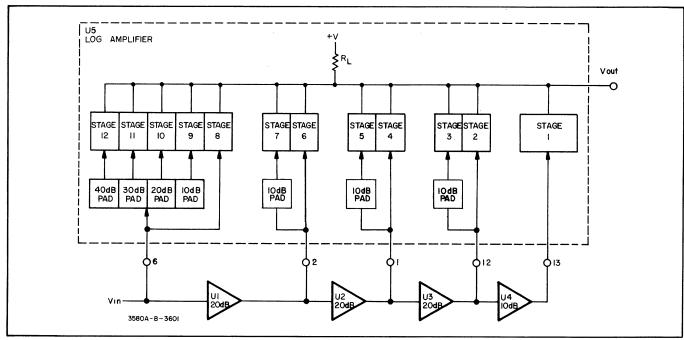


Figure 4-6. Log Amplifier.

- b. When the input signal is above the range of a given stage, that stage will make a constant contribution to the output of the log amplifier.
- c. When the input signal is within the range of a given stage, that stage provides the logrithmic output over a 10 dB range. The logarithmic output is added to the constant output of the more sensitive stages.
- 4-63. Since there are twelve 10 dB stages in the log amplifier package, it will appear that the overall dynamic range is 120 dB. In practice, however, the first and last stages do not produce usable outputs over their entire range. The dynamic range of the device is therefore limited to approximately 100 dB. The 3580A input levels are such that only 80 dB to 90 dB of the 100 dB range is used.
- **4-64. Video Detector.** The Video Detector is an average-responding, active, full-wave detector circuit which produces a dc voltage proportional to the amplitude of the log or linear IF signal. The output of the Video Detector, ranging from 0 V to +2.5 V dc full scale, is applied to the Video Filter.
- **4-65.** Video Filter. The purpose of the Video filter is to smooth out the ripple and random noise riding on the detected video signal. The filter consists of a single-pole RC network followed by an output buffer. The response of the filter is varied by changing the values of the RC elements in the circuit. The amount of filtering is increased as the BANDWIDTH setting is narrowed or as the DISPLAY SMOOTHING control is varied from MIN to MAX.

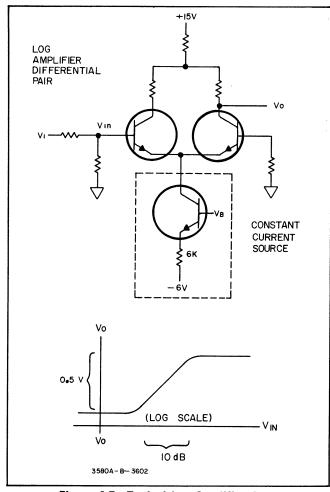


Figure 4-7. Typical Log Amplifier Stage.

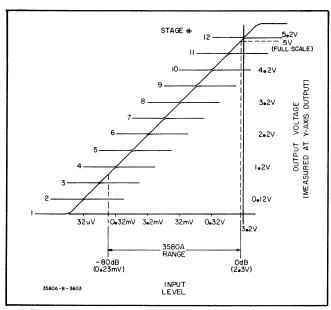


Figure 4-8. Log Amplifier Input and Output Levels.

4-66. Video Output. The Video Output circuits (Figure 4-9) consist of a Reference Divider, a Summing Amplifier and an Output Amplifier.

4-67. The 0 V to + 2.5 V dc video signal from the Video Detector is applied to the inverting (-) port of the Summing Amplifier where it is summed with a negative dc offset voltage from the Reference Divider. In the Log 10 dB and Log 1 dB amplitude modes, the dc offset voltage varies from -2.5V dc to -0.75 V dc as the AMPLITUDE REF LEVEL control is rotated from the 0 dB (NOR-MAL) position to the -70 dB position. This offfsets the display in steps of 10 dB. In the Linear mode, the offset

voltage is fixed at -2.5 V dc and the CRT trace remains at the bottom of the screen.

4-68. In the Log 10 dB and Linear Amplitude modes, the gain of the Summing Amplifier is X2 and an offset of -2.5 V dc produces an output of +5 V dc. This positions the CRT trace at the bottom of the screen. With a video response of +2.5 V dc, the offset voltage is cancelled and the output of the Summing Amplifier drops to 0 V dc for full-scale deflection. When the Log 1 dB mode is selected, the gain of the Summing Amplifier is increased to X20. This expands the CRT scale from 10 dB per division to 1 dB per division.

4-69. With the Summing amplifier gain set to X20 and a video input of 0 V, the dc offset voltage from the Reference Divider drives the output of the Summing Amplifier positive. In this state, the Summing Amplifier output is limited to +6.8 V dc by Zener diode CR29. An output between +5 V and +6.8 V positions the CRT trace at the bottom of the screen. If the positive video level equals the negative offset voltage, the output of the Summing Amplifier drops to 0 V for full-scale deflection. If the video level exceeds the offset voltage, the Summing Amplifier output is driven negative and is limited to -0.6 V by CR29. An output level between 0 V and -0.6 V peaks the display. Table 4-2 lists the offset voltage, displayed video levels, Summing Amplifier output levels and display range for each AMPLITUDE REF LEVEL setting.

4-70. The output of the Summing Amplifier is applied to the A8 Assembly through R125 and to the Output Amplifier. In the A8 Assembly, the video output from the Summing Amplifier is used to detect the presence of a video response for Adaptive Sweep purposes.

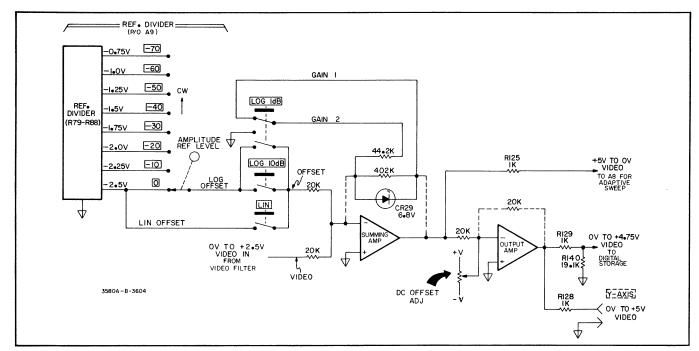


Figure 4-9. Video Output.

| Reference | Offset | Displayed Video | Summing Amp | Display |
|-----------|---------|----------------------|--------------|------------------|
| Level | Voltage | Level | Output | Range |
| 0 dB | -2.50 V | + 2.25 V to + 2.50 V | + 5 V to 0 V | -10 dB to 0 dB |
| -10 dB | -2.25 V | + 2.00 V to + 2.25 V | + 5 V to 0 V | -20 dB to -10 dB |
| -20 dB | -2.00 V | + 1.75 V to + 2.00 V | + 5 V to 0 V | -30 dB to -20 dB |
| -30 dB | -1.75 V | + 1.50 V to + 1.75 V | + 5 V to 0 V | -40 dB to -30 dB |
| -40 dB | -1.50 V | + 1.25 V to + 1.50 V | + 5 V to 0 V | -50 dB to -40 dB |
| -50 dB | -1.25 V | + 1.00 V to + 1.25 V | + 5 V to 0 V | -60 dB to -50 dB |
| -60 dB | -1.00 V | + 0.75 V to + 1.00 V | + 5 V to 0 V | -70 dB to -60 dB |
| -70 dB | -0.75 V | + 0.50 V to + 0.75 V | + 5 V to 0 V | -80 dB to -70 dB |

Table 4-2. Video Output Circuits (Log 1 dB Mode).

4-71. At the inverting port of the Output Amplifier, the +5 V dc to 0 V output from the Summing Amplifier is summed with a -5 V dc offset from the wiper of the DC Offset Adj. potentiometer, R11. The gain of the Output Amplifier is X1 and the resulting output ranges from 0 V dc to +5 V dc, full scale. This output is attenuated to +4.75 V dc full scale by R129 and R140 and applied to the Digital Storage section. The 0 V to +5 V output is also applied to the rear panel Y-AXIS output connector.

4-72. Frequency and Sweep Section.

- 4-73. Figure 4-10 is a functional block diagram of the Frequency and Sweep section. Elements shown on the diagram are described in the following paragraphs.
- **4-74.** Linear Sweep Generator. Because of its Adaptive Sweep capability, the 3580A Linear Sweep generator is considerably more sophisticated than conventional sweep generators. The primary purpose of the Linear

Sweep Generator is to produce a 0 V to +5 V linear ramp which simultaneously sweeps the VTO frequency and the refresh trace on the CRT. In the Adaptive Sweep process, however, it is required to perform a sequence of operations in response to video signals that rise above the baseline threshold set on the CRT display. This sequence or "algorithm" is illustrated and described in Figure 4-11.

- 4-75. Figure 4-12 is a functional block diagram of the Linear Sweep Generator. The major circuit elements include a Digital Controller, a Programmable Ramp Generator, an End of Sweep Comparator a Ramp Comparator and a Delay Circuit.
- 4-76. Digital Controller. The Digital Controller is a simple algorithmic state-machine (ASM) which provides se-

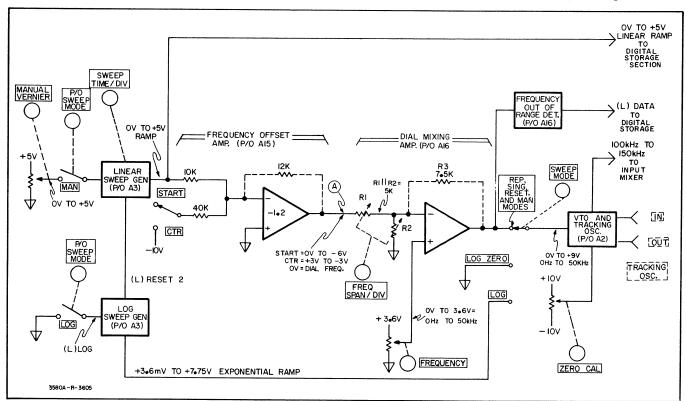


Figure 4-10. Frequency and Sweep Section.

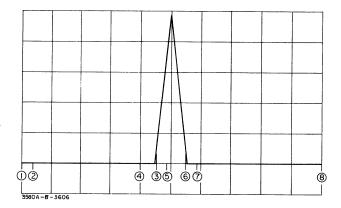
quential instructions that control the Adaptive Sweep process. The six input lines shown on the left-hand side of the controller block are qualifiers which determine the "next state" of the controller. The qualifier lines are listed and defined in Table 4-3. The outputs on the right-hand side of the controller block are instructions which are applied to the Ramp Generator and associated circuitry. The functions of the various instruction lines are described in the following paragraphs.

4-77. The Digital Controller is synchronized by a 55 kHz to 70 kHz pulse train applied to the Clock input. The clock signal is generated by an oscillator in the High Voltage Power Supply section. Even though the clock frequency is 55 kHz to 70 kHz, the Digital Controller does not cycle at a 55 kHz to 70 kHz rate. State times are determined strictly by the qualifier inputs and the clock only ensures that the counting elements within the controller are incremented simultaneously. In order for the digital controller to function properly, the clock frequency must be between 55 kHz and 70 kHz.

4-78. Programmable Ramp Generator. The Programmable Ramp Generator produces a 0 V to +5 V linear

Table 4-3. Qualifier Inputs.

| Qualifier | Description | |
|------------|--|--|
| (L)RESET 1 | Line goes low when SWEEP MODE switch is set to RESET, MAN or LOG ZERO and when CLEAR WRITE button is pressed. When this line initially goes low, the Digital controller is asynchronously reset to State Ø. The controller then increments to State 1 and remains in that state until the (L)RESET 1 line goes high. | |
| (L)SING | Line goes low when SWEEP MODE switch is set to SINGLE position. | |
| (L)RESP | Line goes low when a video response rises above the baseline threshold set on the CRT display. | |
| (H)GEW | Line from Digital Storage section goes high to indicate that the display sweep has reset. | |
| (H)DLYO | Line from Delay Circuit goes high to indicate that the delay period is over. | |
| | NOTE | |
| | The "L" or "H" preceding each qualifier mneumonic indicates the "Low" or "High" assertion state (true or "1" state) of the qualifier line. In some cases, both states of a qualifier are used in the control sequence. For example, a qualifier might be "Response" (H)NRESP. | |



- STOP SWEEPING AND ALLOW TIME FOR THE IF FILTER TO 4. SETTLE. SLOW FORWARD TO POINT 2.
 - During slow sweeps the sweep time is as indicated by the front panel SWEEP TIME setting.
 - The instrument initially sweeps slowly to ensure that the zero response (if present) is properly detected.
 - c . The distance between Points 1 and 2 is equal to 0.75 times the Step Back distance (see Step 3).
- 2. IF A RESPONSE IS NOT PRESENT, SWEEP FAST FORWARD.
 - a. The fast sweep time is 20 to 25 times faster than indicated by the SWEEP TIME setting.
- 3. RESPONSE DETECTED. SWEEP FAST BACKWARD TO POINT 4.
 - a. The distance between Points 3 and 4 is called the "Step Back" distance and is determined by the FREQ SPAN and BANDWIDTH settings.

- 4. STOP SWEEPING AND ALLOW TIME FOR THE IF FILTER TO SETTLE. SWEEP SLOW FORWARD TO POINT 5.
 - a. The distance between Points 4 and 5 is equal to 1.75 times the Step Back distance.
 - b. This step ensures that the instrument sweeps slowly past the point it stepped back from initially.
- CONTINUE TO SWEEP SLOWLY TO POINT 6 WHERE RESPONSE IS NO LONGER DETECTED.
- 6. SWEEP SLOW FORWARD TO POINT 7.
 - a. The distance between Points 6 and 7 is equal to 0.75 times the Step Back distance.
 - b. This step ensures that the response is completely passed before the sweep rate is increased.
- IF RESPONSE IS NOT PRESENT, SWEEP FAST FORWARD TO POINT 8.
- 8. END OF SWEEP, RESET TO POINT 1.

Figure 4-11. Adaptive Sweep Routine.

ramp voltage in response to sequential instructions from the Digital Controller. The instructions applied to the Ramp Generator are listed and defined in Table 4-4.

4-79. End of Sweep Comparator. The EOS Comparator detects when the ramp voltage reaches +5 V and, in turn, produces an End of Sweep (LEOS) command which asynchronously resets the Digital Controller to State \emptyset .

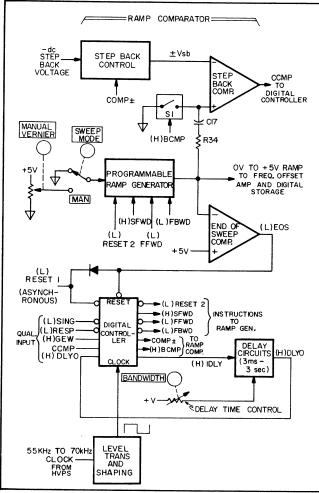


Figure 4-12. Linear Sweep Generator.

4-80. Ramp Comparator. In the Adaptive Sweep routine, the Ramp Comparator measures the forward and reverse excursions of the ramp voltage and informs the Digital Controller when the Ramp Generator has swept the required distance from a given point. The need for this is illustrated in Figure 4-11. At Point 3, for example, a response is initially detected and the Ramp Generator sweeps backward to Point 4. The controller must be informed when the sweep reaches Point 4 so that it can instruct the Ramp Generator to begin sweeping slow forward. Similarly, when a response is no longer detected at Point 6, the Ramp Generator continues to sweep slow forward to Point 7. The controller must be informed when the sweep reaches point 7 so that it can instruct the Ramp Generator to begin sweeping fast.

4-81. The Ramp Comparator consists of a Step Back Control circuit and a Step Back Comparator. Both of these elements operate in response to instructions from the Digital Controller.

Table 4-4. Ramp Generator Instructions.

| INSTR | Description |
|------------|--|
| (L)RESET 2 | In the Single and Repetitive sweep modes, the (L)RESET 2 instruction resets the Ramp Generator. When the Ramp Generator is reset, its output is 0 V. In the Manual sweep mode, the (L)RESET 2 instruction is given continuously. The Ramp Generator then functions as a X1 amplifier and receives its input from the MANUAL VERNIER potentiometer. |
| (H)SFWD | When the (H)SFWD (Slow Forward) instruction is given, the Ramp Generator sweeps in a positive direction from 0 V to + 5 V. The sweep time is as indicated by the SWEEP TIME setting. |
| (L)FFWD | When the (L)FFWD (Fast Forward) instruction is given, the Ramp Generator sweeps in a positive direction at 20 to 25 times the panel-selected rate. |
| (H)FBWD | When the (H)FBWD (Fast Backward) instruction is given, the Ramp Generator sweeps in a negative direction (+5 V to 0 V) at 20 to 25 times the panel-selected rate. |

4-82. The Step Back Control circuit is a "programmable inverter" which receives a negative dc input voltage and provides an inverted or non-inverted output, depending on the state of the COMP instruction line. The negative dc "step-back voltage" applied to the Step Back Control circuit is controlled by the FREO SPAN and BANDWIDTH settings. The magnitude of this voltage determines the "step-back distance" described in Figure 4-11. As the frequency span is narrowed or bandwidth is widened, the magnitude of the step back voltage increases causing the step back distance to increase. When the COMP instruction line is high, the instruction is COMP (-). This means that the output of the Step Back Control circuit is a negative dc voltage that is equal in magnitude to the applied stepback voltage. When the COMP instruction line is low, the instruction is COMP(+). When the COMP(+) instruction is given, the output polarity is changed from negative to positive and the magnitude of the voltage is decreased to 0.75 times the applied step-back voltage. For example, if the applied step-back voltage is -1 V dc and the instruction is COMP (-), the output of the Step Back Control circuit is -1 V dc. If the instruction is changed to COMP (+), the output changes to +0.75 V dc.

4-83. The Step Back Comparator is a high impedance differential amplifier circuit controlled by the (H)BCMP (Begin Comparison) instruction line from the Digital Controller. When the Begin Comparison instruction is *not* given (BCMP line low), switch S1 is closed and the non-inverting (+) port of the comparator is grounded. Capacitor C17 then charges to the ramp voltage through R34. When the Begin Comparison in-

Model 3580A Theory of Operation

struction is given, switch S1 opens and the instantaneous ramp voltage is retained by C17. With S1 open, the polarity of the charge on C17 is such that C17 serves as a bucking supply. Thus, as the Ramp Generator sweeps forward or backward from the point at which S1 opens, only the change in voltage is felt at the noninverting port of the comparator. If, for example, the BCMP instruction is given when the ramp voltage is +4 V and the ramp voltage then decreases to +3 V, the voltage at the non-inverting port is -1 V. When the voltage at the non-inverting port slightly exceeds the positive or negative step-back voltage at the inverting port, the output of the comparator changes states and the CCMP (Comparison Complete) qualifier is met. This indicates to the Digital Controller that the Ramp Generator has swept the required distance from the point at which the comparison began.

4-84. In the Adaptive Sweep routine, the COMP (-) and BCMP instructions are given when the Ramp Generator begins sweeping backward. At the time the BCMP instruction is given, the output of the Step Back Comparator is *high*. As the ramp voltage decreases, the voltage at the non-inverting intput becomes increasingly negative until it slightly exceeds the negative step-back voltage at the inverting port. The output of the comparator then goes *low* and the CCMP qualifier is met.

The COMP (+) and BCMP instructions are given when the Ramp Generator is sweeping forward. In this case, the output of the comparator is *low* when the BCMP instruction is given. As the ramp voltage increases, the voltage at the non-inverting port becomes increasingly positive until it slightly exceeds the positive step-back voltage at the inverting port. At the time, the output of the comparator goes *high* and the CCMP qualifier is met.

4-85. Delay Circuit. The Delay Circuit is a monostable multivibrator which provides a 3 ms to 3 sec. delay period in response to the Initiate Delay (IDLY) instruction from the Digital Controller. At the end of the delay period, the Delay Circuit produces a "delay over" flag (DLYO) which serves as a qualifier input to the Digital Controller.

4-86. The purpose of the 3 ms to 3 sec. delay period is to allow time for the IF Filter to settle between fast and slow sweeps in the Adaptive Sweep routine. The delay period is determined by the BANDWIDTH setting. As the bandwidth is narrowed, the response time of the IF Filter increases and a longer delay period is required.

4-87. Control Sequence. Figure 4-13 is an ASM Chart showing the control sequence for the 8-state Adaptive

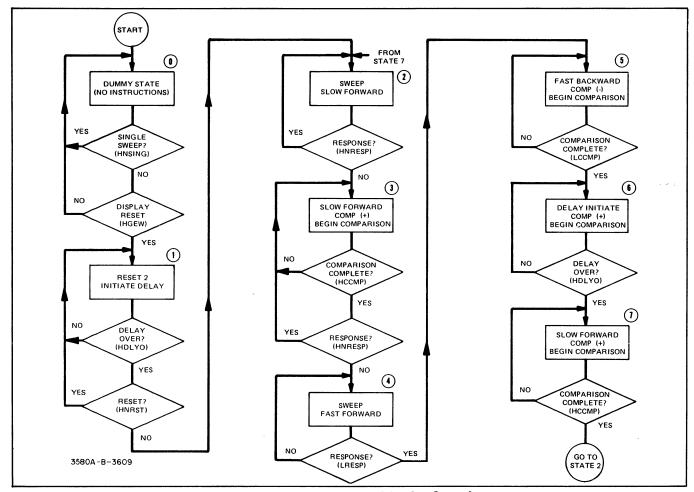


Figure 4-13. ASM Chart (Adaptive Sweep).

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Sweep routine. Each state of the Digital Controller is represented by a rectangular Instruction Block followed by one or two trapezoidal-shpaed Qualifier Blocks. Items listed in the Instruction Block of a given state indicate the instruction(s) given by the controller in that state. Items in the Qualifier Blocks of a given state indicate the qualifiers that must be met before the controller can increment of the next state.

- 4-88. The routine begins with the Digital Controller asynchronously reset the State Ø by an End of Sweep (LEOS) command. The EOS command is momentary and does not prevent the controller from incrementing to the next state. State Ø is a "dummy" state where no instructions are given. The two qualifiers in this state are HNSING (Not Single) and HGEW. The HNSING qualifier is met when the SWEEP MODE switch is not in the SING position. When the Single sweep mode is selected, the controller remains in State Ø following the End of Sweep command. The HGEW qualifier is met when the display sweep in the Digital Storage section resets.
- 4-89. In State 1, the (L) Reset 2 instruction is given to reset the Ramp Generator. At the same time, a delay is initiated to allow the IF Filter to settle. When the delay period is over (DLYO qualifier met) the controller increments to State 2. If the SWEEP MODE switch is in the RESET, MAN or LOG ZERO position, the (L) Reset 1 line is low causing the controller to remain in State 1.
- 4-90. In State 2, the Ramp Generator starts sweeping SLOW FORWARD. The sweep starts out slowly to ensure that any signals on or near the start frequency will be properly detected. If a response is not present or when the initial response is no longer detected, the HNRESP qualifier is met and the controller increments to State 3.
- 4-91. In State 3, the Ramp Generator continues to sweep SLOW FORWARD. At this time, the COMP (+) and BCMP instructions are given and the Ramp Generator must sweep slow forward until the CCMP qualifier is met. If, for some reason, a response is setected in State 3, the controller will not increment to State 4 until the reponse is passed (HNRESP qualifier met).
- 4-92. In State 4, the Ramp Generator sweeps FAST FORWARD until a response is detected. When a response is detected (LRESP qualifier met), the controller increments to State 5.
- 4-93. In State 5, the COMP (-) and BCMP instructions are given and the Ramp Generator sweeps FAST BACKWARD until the CCMP qualifier is met. The CCMP qualifier is met when the ramp voltage decreases by an amount equal to the step-back voltage.
- 4-94. In State 6, a delay is initiated and the Ramp

Generator stops sweeping until the DLYO qualifier is met. The controller then increments to State 7.

- 4-95. In State 7, the Ramp Generator sweeps SLOW FORWARD until the CCMP qualifier is met. The controller then recycles to State 2. Note that the Begin Comparison instruction initiated in State 5 is sustained in States 6 and 7. This means that the ramp voltage stored in State 5 (response initially detected) is still the reference in State 7. Since, in State 7, the instruction applied to the Step Back Control circuit is COMP (+), the Ramp Generator must sweep slow foward past the point it stepped back from initially. (See Steps 3, 4 and 5 of Figure 4-11.)
- 4-96. Non-Adaptive Sweep. When the ADAPTIVE SWEEP control is in the OFF position, the (L)RESP qualifier line is pulled low to simulate a response. As in the Adaptive Sweep routine, the Digital Controller is initially reset to State Ø and is incremented to States 1 and 2. In State 2, however, the (H)NRESP (No response) qualifier is never met and the controller is forced to remain in that state until it is again reset. When the controller is in State 2, the (H)SFWD (Slow Forward) instruction is given and the Ramp Generator sweeps at the rate indicated by the SWEEP TIME setting.
- 4-97. Manual Sweep. When the Manual sweep mode is selected, the (L)RESET 1 line is pulled low causing the Digital Controller to remain in State 1. The (L)RESET 2 instruction given in State 1 converts the Ramp Generator into a X1 amplifier. The 0 V to +5 V dc level from the wiper of the MANUAL VERNIER potentiometer is then present at the output of the Ramp Generator. This dc level determines the VTO frequency and the position of the refresh trace on the CRT.
- 4-98. Frequency Offset Amplifier. The 0 V to +5 V ramp from the Linear Sweep Generator is applied to the inverting port of the frequency Offset Amplifier. The gain of the amplifier is -1.2 and, with the START/CTR switch in the START position, the ramp voltage at the output ranges from 0 V to -6 V. With the START/CTR switch set to the CTR (Center) position, a negative dc offset is summed with the ramp voltage at the inverting port. The ramp voltage at the output then ranges from +3 V to -3 V.
- 4-99. Dial Mixing Amplifier. The output of the Frequency Offset Amplifier is applied to the inverting port of the Dial Mixing Amplifier through a resistive attenuator network (R1, R2) controlled by the FREQ SPAN switch. As the frequency span is narrowed, the attenuation increases and the effective gain of the amplifier (with respect to Point A) decreases. Table 4-5 lists the Dial Mixing Amplifier gain and resulting ramp output levels for each FREQ SPAN setting. Output levels listed in the table are measured with an input ramp of 0 V to -6 V and with the non-inverting port of the amplifier at 0 V.

Table 4-5. Dial Mixing Amplifier Gain.

| Freq Span/Div | Overall Span | Mixing Amp Gain (From Point A) | Output Ramp |
|------------------|-----------------|--------------------------------------|-----------------|
| 5 kHz | 50 kHz | -1.5 | 0 V to +9 V |
| 2 kHz | 20 kHz | -0.6 | 0 V to + 3.6 V |
| 1 kHz | 10 kHz | -0.3 | 0 V to + 1.8 V |
| 0.5 kHz | 5 kHz | -0.15 | 0 V to + 0.9 V |
| 0.2 kHz | 2 kHz | -0.06 | 0 V to + 0.36 V |
| 0.1 kHz | 1 kHz | -0.03 | 0 V to + 0.18 V |
| 50 Hz | 500 Hz | -0.015 | 0 V to + 0.09 V |
| 20 Hz | 200 Hz | -0.006 | 0 V to + 36 mV |
| 10 Hz | 100 Hz | -0.003 | 0 V to + 18 mV |

4-100. A 0 V to + 3 V dc control voltage from the front panel FREQUENCY potentiometer is applied to the non-inverting port of the Dial Mixing Amplifier. The gain at the non-inverting port is determined by the parallel resistance of R1 and R2 and by the feedback resistance, R3. The values of R1 and R2 are such that their parallel resistance is always 5 K. The fixed gain at the non-inverting port is therefore:

$$1 + \frac{7.5 \text{ K}}{5 \text{ K}} = + 2.5$$

With the ramp input at 0 V the output of the Dial Mixing Amplifier varies from 0 V to +9 V as the FRE-QUENCY control is rotated from 0 Hz to 50 kHz. This tunes the analyzer over its entire frequency range. Anytime the ramp at the inverting port is at 0 V, the analyzer frequency is as indicated on the FREQUENCY display.

4-101. The following examples illustrate how the ramp and frequency-dial inputs are combined at the output of the Dial Mixing Amplifier to produce the required frequency sweep.

Example: 1

| FREQUENCY SPAN5K/DIV |
|---|
| START/CENTERSTART |
| GAIN (Point A)1.5 |
| RAMP VOLTAGE (Point A)0 V to -6 V |
| RAMP CONTRIBUTION |
| TO OUTPUT 0 V to +9 V |
| FREQUENCY DISPLAY 0 Hz |
| DISPLAY CONTRIBUTION TO OUTPUTO V |
| OUTPUT RAMP $0 \text{ V to } + 9 \text{ V}$ |
| FREQUENCY SWEEP 0 Hz to 50 kHz |

Example 2:

| FREQUENCY SPAN | 5K/DIV |
|------------------------|----------------|
| START/CENTER | CENTER |
| GAIN (Point A) | |
| RAMP VOLTAGE (Point A) | + 3 V to -3 V |
| RAMP CONTRIBUTION | |
| TO OUTPUT4 | .5 V to +4.5 V |

| FREQUENCY DISPLAY | .25 kHz |
|-------------------------|---------|
| DISPLAY CONTRIBUTION | |
| TO OUTPUT | +4.5 V |
| OUTPUT RAMP 0 V | |
| FREQUENCY SWEEP 0 Hz to | |

Example 3:

| FREQUENCY SPAN | 5K/DIV |
|------------------------|---------------------|
| START/CENTER | START |
| GAIN (Point A) | 1.5 |
| RAMP VOLTAGE (Point A) | 0 V to -6 V |
| RAMP CONTRIBUTION | |
| TO OUTPUT | \dots 0 V to +9 V |
| FREQUENCY DISPLAY | 5 kHz |
| DISPLAY CONTRIBUTION | |
| TO OUTPUT | + 0.45 V |
| OUTPUT RAMP $+ 0.45$ | V to + 9.45 V* |
| FREQUENCY SWEEP5 | kHz > 50 kHz |
| | |

*Out of Range

4-102. Out of Range Detector. As illustrated in Example 3, certain combinations of FREQUENCY and FREQ SPAN settings cause the voltage at the output of the Dial Mixing Amplifier to go below the 0 V (0 Hz) lower limit or above the +9 V (50 kHz) upper limit. When either limit is exceeded, the VTO frequency is driven out of range. This could cause erroneous responses to appear on the display. The Frequency Outof-Range Detector senses when the Dial Mixing Amplifier output is more negative than 0 V or more positive than +9 V and, in turn, generates an (L)Data flag which is applied to the Digital Storage section. In the Digital Storage section, the (L) Data flag clears the memory locations where the frequency is out of range. As a result, a clean baseline appears on the display (Figure 4-14).

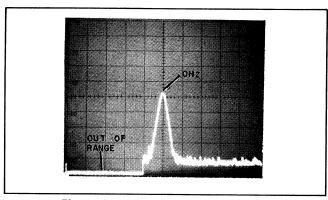


Figure 4-14. Frequency Out of Range.

4-103. VTO and Tracking Oscillator. Refer to Figure 4-15 for the following discussion.

4-104. The 0 V to +9 V ramp from the Dial Mixing Amplifier is applied to the VTO and to the inverting port of the VTO Error Amplifier. At the inverting port of the Error Amplifier, the ramp voltage is summed

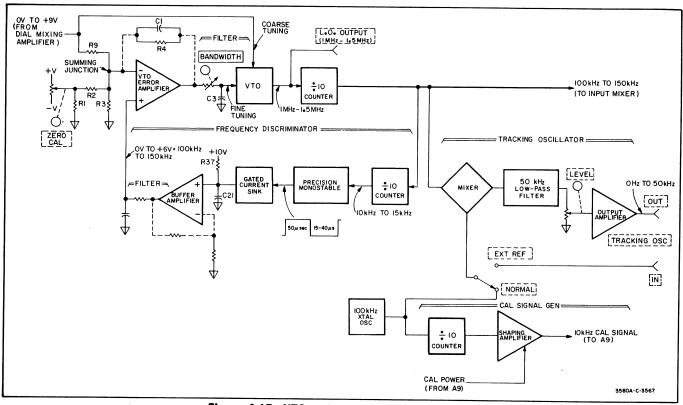


Figure 4-15. VTO and Tracking Oscillator (AS).

with a dc voltage from the front panel ZERO CAL potentiometer. The sum of the two voltages serves as a reference for the frequency control loop.

4-105. The VTO. The VTO is a conventional oscillator circuit that is tuned by changing the dc bias on two varactor diodes which are the capacitive elements in the LC tank circuit. The 0 V to +9 V ramp coarse tunes the VTO frequency from 1 MHz to 1.5 MHz. Fine tuning is provided by the error voltage from the VTO Error Amplifier. The output of the VTO is applied to a Divide-By-Ten Counter and to the rear panel L.O. OUTPUT connector. The output of the Divide-By-Ten Counter is a 100 kHz to 150 kHz square wave which is applied to the Input Mixer (A9) and to the Frequency Discriminator and Tracking Oscillator.

4-106. Frequency Discriminator. Due to the inherent non-linearity of the VTO, an external frequency control loop is required. The frequency control loop is comprised of a Frequency Discriminator and VTO Error Amplifier. The Frequency Discriminator produces a dc voltage that is linearily proportional to the VTO output frequency. This dc voltage is applied to the non-inverting port of the VTO Error Amplifier where it is compared to the reference voltage at the inverting port. Any difference between these two voltages causes the output of the Error Amplifier to increase or decrease to correct the VTO frequency.

4-107. The 100 kHz to 150 kHz VTO output signal is applied to a Divide-By-Ten Counter in the Frequency

Discriminator. The output of the Divide-By-Ten Counter is a 10 kHz to 15 kHz square wave which positive-edge triggers the Precision Monostable Multivibrator. When triggered, the output of the Monostable Multivibrator goes high for exactly 50 µsec This gates off the Current Sink allowing C21 to charge toward + 10 V through R37. At the end of the 50 μ sec charge period, the Current Sink is gated on causing C21 to discharge at a fixed rate. As the VTO frequency increases, the charge period of C21 remains at 50 µsec but the discharge period becomes shorter. As a result, the average charge on C21 increases. The voltage across C21 is amplified, filtered and applied to the noninverting port of the VTO Error Amplifier. This voltage varies from 0 V to +6 V as the VTO frequency is tuned from 100 kHz to 150 kHz.

4-108. Precision Monostable Multivibrator. The magnitude of the dc voltage at the output of the Frequency Discriminator is determined by the duty cycle of the pulse generated by the Precision Monostable Multivibrator. In order for the output voltage to increase linearily with frequency, the width of the positive half cycle of the pulse must be constant reguardless of frequency and the width of the negative half cycle must vary linearily with frequency. This requires precise timing and a high degree of stability not obtainable with conventional R/C-coupled "one-shot" multivibrators.

4-109. Figure 4-16 is a simplified block diagram of the Precision Monostable Multivibrator. In the reset state, the following conditions exist:

- a. The "Q" output of the J-K Flip-Flop is low causing Q13 to cut off. Capacitor C27 then charges to +10 V through R54.
- b. The $\overline{^{''}Q^{''}}$ output of the J-K Flip-Flop is high. This resets the 14-Pulse Counter to State \emptyset .

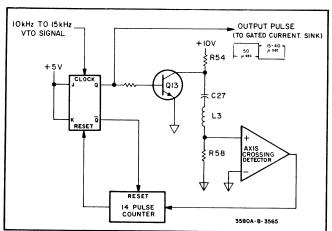


Figure 4-16. Precision Monostable.

- 4-110. The J-K Flip-Flop is clocked by the zero crossing during a low-to-high transition of the VTO input signal. When the Flip-Flop is clocked, the "Q" output goes high, Q13 is gated on and the junction of C27 and R54 is grounded. A series-resonant tank circuit is then formed by C27, L3 and R58. As C27 discharges, the lightly damped tank circuit rings at its resonant frequency (approximately 230 kHz). The 230 kHz signal developed across R58 is squared-up by the Axis Crossing Detector and applied to the 14-Pulse Counter. The 14-Pulse Counter counts 14 pulses and then resets the J-K Flip-Flop to terminate the output pulse.
- 4-111. Tracking Oscillator. In the Tracking Oscillator section, the 100 kHz to 150 kHz output from the VTO is mixed with a 100 kHz signal from a Crystal Oscillator or with an external signal applied to the TRACKING OSC IN connector. The difference frequency at the output of the Mixer is fed through a 50 kHz Low-Pass Filter, amplified and applied to the rear panel TRACKING OSC OUT connector. With the rear panel switch in the NORMAL position, the signal at the Tracking Oscillator Output is a 0 Hz to 50 kHz sinewave which tracks the tuned frequency of the instrument. The amplitude of the signal can be varied from 0 V to 1 V rms by adjusting the rear panel LEVEL control.
- 4-112. Cal. Signal Generation. The 100 kHz output of the Crystal Oscillator is applied to a Divide-By-Ten Counter. The output of the counter is processed and applied to the A9 Input Circuits where it becomes the input signal with the INPUT SENSITIVITY switch set to the CAL position. The calibration signal is a 15/85 duty cycle pulse train which provides a 10 kHz fundamental-frequency component and odd and even harmonic components spaced at 10 kHz intervals. The amplitude of the fundamental-frequency component is such that it

produces full-scale deflection when the instrument is properly calibrated. The amplitudes of the harmonic components are not meaningful.

- 4-113. Log Sweep Generator. In the Log Sweep mode, the 0 V to +5 V linear ramp from the Linear Sweep Generator sweeps the display while a +3.6 mV to +7.75 V exponential ramp from the Log Sweep Generator sweeps the VTO frequency. The frequency range of the log sweep is from 20 Hz to 43 kHz. During log sweeps, the SWEEP TIME control is disabled and the Linear Sweep Generator is automatically set for a 5 second sweep time. The Log Sweep Generator is synchronized by the (L)Reset 2 instruction from the Linear Sweep Generator.
- 4-114. Figure 4-17A shows the basic circuit configuration for the Log Sweep Generator. The major circuit element is a high input-impedance operational amplifier. The gain of the amplifier with respect to Point A is -1 and the gain at the non-inverting port is +2. At the beginning of the log sweep the following conditions exists:
 - a. The (L)Reset 2 line is low.
 - b. FET switch Q32 is closed.
- c. The non-inverting port of the amplifier is grounded through Q32.
 - d. Capacitor C14 is fully discharged.
- e. The output voltage is +3.6 mV dc due to the -3.6 mV dc reference applied to Point A. This sets the analyzer frequency to 20 Hz which is the starting point for the log sweep.

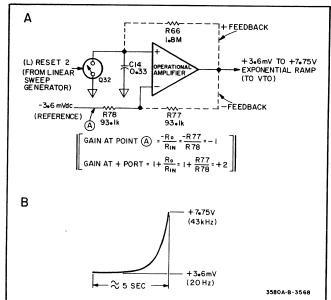


Figure 4-17. Basic Log Sweep Generator.

When the (L)Reset 2 instruction is cleared, switch Q32

opens and C14 charges toward the output voltage through feedback resistor R66. As C14 charges, the output voltage becomes increasingly positive. Due to the bootstrapping effect of the positive feedback through R66, the charge rate of C14 increases exponentially. The exponential ramp at the output is as shown in Figure 4-17B.

4-115. Auto Zero Circuit. An Auto Zero Circuit is included in the Log Sweep Generator to null out any dc offset introduced by the operational amplifier. The overall circuit configuration is shown in Figure 4-18.

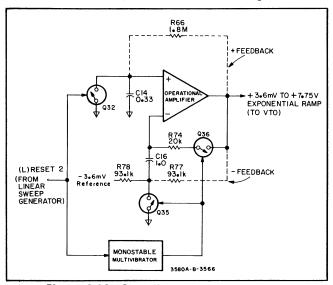


Figure 4-18. Overall Log Sweep Generator.

4-116. When the (L)Reset 2 instruction is initially given, the output of the Monostable Multivibrator goes high for approximately 0.4 seconds. This closes FET switches Q35 and Q36. With switch Q32 also closed, the offset voltage is present at the output of the amplifier and capacitor C16 charges to the offset voltage though R74. At the end of the auto zero period, Q35 and Q36 open and the charge on C16, in series with the input reference voltage, cancels the amplifier offset voltage.

4-117. Digital Storage and Dispaly Sections.

4-118. Introduction to Digital Storage. Low frequency spectrum analyzers require narrow bandwidths and consequently, slow sweep rates. Because of these slow sweep rates, the video cannot be displayed directly on a standard CRT. If, for example, the X and Y axis outputs of the 3580A were applied to a standard CRT, the display would be merely a dot fluctuating up and down while moving slowly across the CRT face. Even with the SWEEP TIME control set to 0.01 SEC/DIV (fastest sweep time), a satisfactory display could not be obtained.

4-119. To retain the slowly scanned video information of the 3580A, some form of display storage is required. As indicated in the Simplified Block Diagram Description (Paragraph 4-22), a storage CRT having long per-

sistance could be used. Recent advances in large-scale integrated circuits and the innovative design efforts of -hp- engineers, however, have made it possible to use a digital storage technique in the 3580A. The major advantages of digital storage are:

- a. Digital storage permits the use of a standard oscilloscope CRT. Standard CRT's are rugged (a must for portable operation) and relatively inexpensive to replace.
- b. A digitally stored trace can be retained indefinitely . . . as long as the instrument is turned on. If a single sweep is made, the trace that is generated will continue to be displayed until it is cleared or updated by a new sweep.
- c. If a trace is needed for future reference, it can be permanently stored in memory by pressing the STORE button. The permanently stored trace and a current or "refresh" trace can be then displayed simultaneously.
- d. Display adjustments are not required when the sweep parameters are changed. The digitally stored trace is automatically cleared and updated at the correct rate. The INTENSITY and FOCUS controls have the same effect as those of a regular oscilloscope. Once they are set, they do not need to be readjusted.
- e. Digital storage provides a bright, crisp flicker-free presentation. There is no blooming or display ambiguity.

4-120. How a Trace is Stored. Refer to Figure 4-19 for the following discussion.

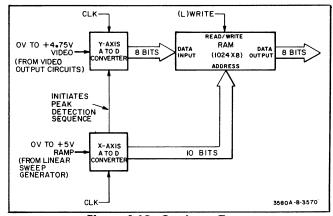


Figure 4-19. Storing a Trace.

4-121. The Digital Memory. The heart of the Digital Storage Section is a Ramdom Access Memory (RAM) comprised of eight 1024 X 1-bit static, MOS memory elements. The RAM has 1024 storage locations or "addresses" (Ø thru 1023). The addresses are selected by a 10-bit binary code applied to the Address lines. Each address is capable of storing an 8-bit binary word applied to the Data Input lines. The input/output function of the RAM is determined by the state of the Read/Write

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control line. When the Read/Write line is low, the 8-bit word present on the Data Input lines is stored or "written" in the memory location selected by the Address lines. When the Read/Write line is high, the 8-bit word stored in the selected address is present on the Data Output lines. In this state, data is non-destructively "read" out of memory.

4-122. X and Y Inputs. The two major inputs to the Digital Storage section are the 0 V to +5 V frequency ramp from the Linear Sweep Generator and the 0 V to +4.75 V video signal from the Video Output Circuits. The magnitude of the ramp voltage at any given time represents a specific frequency and the magnitude of the video signal represents the signal amplitude at the frequency. To store a trace in the Digital Memory, it is first necessary to convert these analog inputs to their corresponding binary codes. This is accomplished by the X-Axis and Y-Axis A to D (Analog to Digital) Converters.

4-123. X-Axis A to D Converter. The 0 V to +5 V frequency ramp is converted to a 10-bit binary code by the X-Axis A to D Converter. This 10-bit binary code is used to address the RAM during the write phase. At the beginning of a frequency sweep, the frequency ramp is at 0 V and the output of the X-Axis A to D Converter is 0000000000, corresponding to RAM address \emptyset . At the end of the frequency sweep, the ramp is at +5 V and the output of the X-Axis A to D Converter is 11111111111, corresponding to RAM address 1023. Thus, during each frequency sweep, the X axis is divided into 1024 discreet segments with each segment corresponding to a given RAM address.

4-124. Y-Axis A to D Converter. The 0 V to +4.75 V video input is converted to an 8-bit binary code by the Y-Axis A to D Converter. During each X-Axis segment, this 8-bit word is written into the memory location ad-

dressed by the X-Axis A to D Converter. As a result the entire memory is filled and its contents are updated by each frequency sweep. Since each address represents a specific frequency and the 8-bit word stored in a given address represents the video amplitude at the frequency, the memory, in effect, contains a point-by-point plot of the amplitude vs. frequency display.

4-125. With 1024 X-Axis segments, the duration of each segment varies from approximately 100 µsec to 1.9 seconds, depending on the SWEEP TIME setting. Since the frequency is continually changing as the ramp voltage increases, the amplitude of the video signal can vary greatly during a given segment. The amount of variation depends on the magnitude of the random noise riding on the video signal and on the slope of the response being traced. Since only one value can be used to represent the video amplitude during each segment, the peak value, being the most important parameter, is the value that is used. The Y-Axis A to D Converter is designed so that it detects and retains the peak value of the video signal during each X-axis segment. The peak detection sequence is initiated by a signal from the X-Axis A to D Converter.

4-126. Displaying a Stored Trace. Refer to Figure 4-20 for the following discussion.

4-127. To obtain a flicker-free stored presentation of the CRT, the memory must be read and the display must be swept at a much faster rate than that of the frequency ramp used for storing data. This rapid scan rate is provided by the Address Counter and Display Ramp Generator.

4-128. Address Counter. During the "read" phase, the X-Axis A to D Converter is disconnected and the Address lines of the RAM are switched to the Address Counter. (The switching operation is performed by a

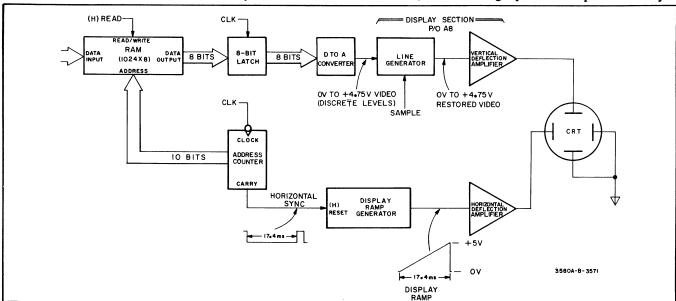


Figure 4-20. Displaying a Stored Trace.

10-bit multiplexer described in following paragraphs.) The Address Counter is a 10-bit binary counter that is incremented at approximately 17 μ sec intervals. The counter continually cycles from state \emptyset (0000000000) to state 1023 (1111111111) and then resets to state \emptyset . As a result, the entire memory is read in periods of approximately 17.4 msec. When the Counter reaches state 1022, its "Carry" output goes high to reset the Display Ramp Generator.

4-129. Display Ramp Generator. The Display Ramp Generator, synchronized by the Address Counter, generates a 0 V to +5 V linear ramp which provides the horizontal sweep for the CRT display. The duration of each sweep is approximately 17.4 msec, corresponding to 1022 increments of the Address Counter. The display sweep is initiated when the Address Counter resets to state Ø and is terminated when the counter reaches state 1022. Addresses 1022 and 1023, are therefore, not displayed.

4-130. 8-Bit Latch. During each increment of the Address counter. The 8-bit word is retained by the Latch until the Address Counter is again incremented.

4-131. D to A Converter. The D to A (Digital to Analog) Converter contains a buffered resistive ladder network which converts the 8-bit word at the output of the Latch to its corresponding dc level. The output of the D to A Coverter, ranging from 0 V to +4.75V full scale,

is applied to the vertical deflection plates of the CRT through the Line Generator and Vertical Deflection Amplifier.

4-132. Line Generator. The output of the D to A Converter is a series of discrete levels which, if applied to the CRT, would produce a display of dots. The Line Generator produces a variable slope ramp which draws lines between the dots to provide a fully reconstructed display.

4-133. The Overall System. During each frequency sweep, the memory contents must be updated by the frequency ramp while the trace currently in memory is being displayed. Since the read and write operations cannot be performed simultaneously, the Address lines of the RAM are rapidly switched between the X-Axis A to D Converter and the Address Counter. Figure 4-21 is a block diagram showing the overall system. Two elements not previously described are the Clock Generator and the Address Multiplexer.

4-134. Clock Generator. The Clock Generator, Driven by a signal from the High Voltage Power Supply, produces ten clock outputs which synchronize the various operations of the system. The frequency of the signal applied to the Clock Generator varies from instrument to instrument and can be anywhere within the range of 55 kHz to 65 kHz. In the following discussion, the input frequency is considered to be 60 kHz which provides a base time period of about $17 \mu \text{sec}$.

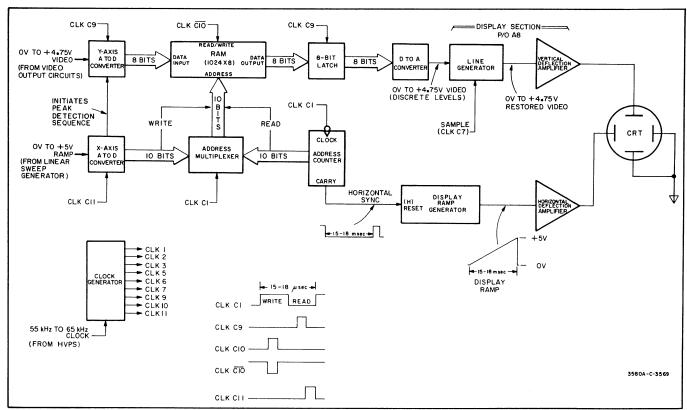


Figure 4-21. Digital Storage and Display Sections.

- 4-135. There are four clocks that are of particular significance in the following discussion. These are: C1, C9, C10 and $\overline{C10}$. The relationship between these clocks is shown on the block diagram. Clock C11 which synchronizes the X-Axis A to D Converter is also shown.
- 4-136. Address Multiplexer. The 10-bit Address Multiplexer switches the RAM Address lines between the X-Axis A to D Converter and the Address Counter. The switching input to the Multiplexer is Clock C1 which is a 60 kHz square wave. The positive half cycle of C1 is to "write" phase and the negative half cycle is the "read" phase. During the write phase of C1, the RAM is addressed by the X-Axis A to D Converter. During the read phase of C1, the RAM is addressed by the Address Counter.
- 4-137. Timing Functions. Before proceeding with the operational sequence, note the following timing functions:
- a. The Address Counter is incremented by the high to low transition that occurs when Clock C1 changes from the write phase to the read phase.
- b. Clock C9 goes high during the read phase of C1. When C9 goes high, the 8-bit word present on the RAM Data Output lines is strobed into the Latch.
- c. The Read/Write line of the RAM is controlled by Clock $\overline{C10}$ which goes low during the write phase of C1. When $\overline{C10}$ goes low, the 8-bit word from the Y-Axis A to D Converter is written into the RAM address selected by the X-Axis A to D Converter.
- d. The A to D Converts are clocked by C9 and C11 during the read phase of C1. This means that their outputs can change *only* during the read phase.
- **4-138. Operational Sequence.** For the operational description, the following initial conditions exist:
- a. The SWEEP MODE switch is set to SING to provide a single frequency sweep.
- b. The SWEEP TIME setting is 0.01 SEC/DIV (fastest sweep time).
- c. The ADAPTIVE SWEEP control is set to the OFF position.
- d. The CLEAR WRITE button has been pressed to clear the display and initiate a new sweep.
 - e. Clock C1 is high (write phase)
- f. The Address Counter is in state 1023 and will reset to \emptyset when C1 goes low.
- 4-139. At the beginning of the frequency sweep, the ramp input to the X-Axis A to D Converter is 0 V and

- the binary code at its output is Ø (000000000). During the write phase of C1, the RAM Address lines are switched to the X-Axis A to D Converter and addressØ is selected. When Clock C10 goes low, the 8-bit word from the Y-Axis A to D Converter is written into address \emptyset . The 8-bit word represents the video amplitude at the start frequency of the sweep. When Clock C1 goes low, the RAM Address lines are switch to the Address Counter and the Address Counter resets to \emptyset . At this point, the Display Ramp Generator is reset, the CRT sweep is at the left-hand side of the screen and RAM address Ø is selected by the Address Counter. When Clock C9 goes high, the 8-bit word stored in Address Ø is strobed into the Latch, converted to dc by the D to A Converter and applied to the vertical deflection plates of the CRT. When Clock C1 goes high, the RAM Address Lines are again switched to the X-Axis A to D Converter. With the SWEEP TIME Control set to 0.01 SEC/DIV, it takes approximately 100 usec for the frequency ramp to increase enough to increment the X-Axis A to D Converter to state 1. In this case, only 17 μ sec have elapsed since the beginning of the sweep so the output of the X-Axis A to D Converter is still Ø. When C10 goes low, the 8-bit word from the Y-Axis A to D Converter is again written into address Ø. This 8-bit word may be the same or may differ from the one previously written into address0. Since the Y-Axis A to D Converter detects and retains the peak value of the video signal during each X-Axis segment, the final word written into address Ø will represent the peak amplitude during the first segment. When C1 again goes low, the Address Counter is incremented to state 1 (000000001) and RAM address 1 is selected. When C9 goes high, the contents of address 1 are strobed into the 8-bit Latch. Since the RAM was cleared at the beginning of the sweep and the X-Axis A to D Converter has not yet incremented to state 1, addresses 1 through 1023 contain all zeros. The output of the D to A Converter is, therefore, 0 V and the CRT trace at this point is at the bottom of the screen.
- 4-140. As the sequence continues, the Address Counter is incremented at 17 μ sec intervals by Clock C1. During each read phase of C1, a new RAM address is selected and a new 8-bit word is strobed into the Latch, converted to dc and applied to the vertical deflection plates of the CRT. As a result, all 1022 addresses are read and the display is swept in approximately 17.4 msec.
- 4-141. At the end of the first display sweep, the frequency ramp will be about +0.81 V and only the first 174 RAM addresses will be filled. Thus, almost six display sweeps will have been made by the time the RAM is completely filled.
- 4-142. At the end of the 0.1 second single sweep, the entire memory will be filled and the frequency ramp at the input of the X-AXIS A to D Converter will remain at +5 V. At that time, the output of the X-AXIS A to D Converter will be 11111111111, corresponding to RAM address 1023. During each write phase of C1, an 8-bit

word will be written into address 1023. This is of no consequence because the Address Counter resets the Display Ramp Generator in state 1022 and addresses 1022 and 1023 are not displayed. The Address Counter will continue to cycle, the memory will be read and the display will be swept at a 17.4 msec rate. The trace stored in memory will, therefore, continue to be displayed until it is cleared or updated by a new frequency sweep.

- 4-143. Clearing a Trace. When the CLEAR/WRITE button is pressed, the following things take place:
- a. The Y-Axis A to D Converter is held in the reset state and its output is 00000000.
- b. The RAM Address lines are switched to the Address Counter during both the read and write phases of C1.
- c. As the Address Counter scans the memory, all zeros are written in each address and the entire memory is cleared in 17.4 msec.
- 4-144. Store Function. A major feature of the Digital Storage section is the "Store Function" which allows a

trace to be permanently stored in memory for future reference. The permanently stored trace can be blanked from the display and then recalled at any time for comparison with the current or "refresh" trace.

- 4-145. To permanenty store a trace, the operator presses the front panel STORE button. This initiates a sequence of operations in which the trace currently in memory is processed and reloaded into 512 of the 1024 memory locations. The remaining half of the memory is used for the refresh trace. To display both traces, the display sweep rate is doubled to provide two 8.7 msec sweeps. During the first display sweep, the Address Counter scans the memory locations containing the refresh trace. It then recycles and scans the memory locations containing the permanently stored trace. As a result, the two traces are displayed alternately in a 17.4 msec period.
- 4-146. Figure 4-22 is an expanded block diagram showing the additional circuitry needed to implement the store function. A 4-state digital controller called the "Store Function Controller" is used to direct the store operation. The ASM chart for the Store Function Controller is shown in Figure 4-23. Other elements used only for the store function are the Store Multiplexer, the 8-Bit Adder and the Write Control circuit.

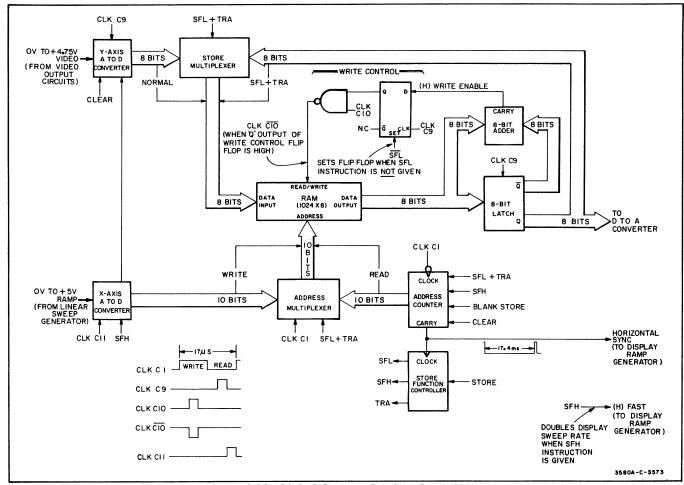


Figure 4-22. Digital Storage Section (Store Mode).

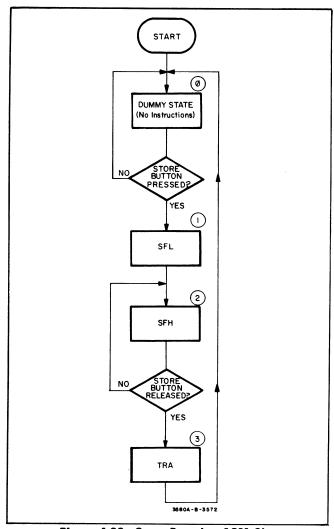


Figure 4-23. Store Function ASM Chart.

4-147. Store Multiplexer. The Store Multiplexer switches the RAM Data Input lines between the Y-Axis A to D Converter and the "Q" outputs of the 8-Bit Latch. The switching inputs to the Store Multiplexer are the SFL and TRA instructions from the Store Function Controller. During normal operation the SFL and TRA instructions are not given and the RAM Data Input lines are always connected to the Y-Axis A to D Converter. When the SFL or TRA instruction is given during the store sequence, the RAM Data Input lines are switched to the "Q" outputs of the 8-Bit Latch and the Y-Axis A to D Converter is disconnected.

4-148. 8-Bit Adder. In State 1 of the store sequence, the Adder is used to compare the 8-bit word on the RAM Data Output lines to the 8-bit word at the output of the Latch. The comparison is made using one's compliment addition i.e., the "Q" outputs of the Latch are the compliments of the "Q" outputs. If the numerical value of the word at the output of the RAM is greater than that of the word at the output of the Latch, the "Carry" output of the Adder goes high, supplying a "Write Enable" command to the Write Control circuit.

4-149. Write Control Circuit. During normal operation the "Set" input to the write control flip-flop is high, forcing the "Q" output to be high. Clock $\overline{C10}$ is then present at the output of the NAND gate and data is written into memory during each write phase of Clock C1. When the SFL instruciton is given during State 1 of the store sequence, the "Set" input of the flip-flop is low and the "Q" output goes high only if the Write Enable line from the 8-Bit Adder is high when the flip-flop is clocked by the positive going edge of C9. If a Write Enable command is given, the "Q" output will be high and data will be written into memory by C10 during the next write phase. If a Write Enable command is not given, the "Q" output of the flip-flop will be low and Clock C10 will be inhibited during the next write phase. 4-150. Store Sequence (State 1). Refer to Figure 4-23. When the STORE button is initially pressed, the Store Function Controller is in State Ø where no instructions are given. It remains in State Ø until the Address Counter completes its current cycle and resets to 0. The Controller then increments to State 1.

4-151. The purpose of State 1 is to condense the trace currently in memory and store it in the 512 memory locations where the Least Significant Bit (LSB) of the address is a logical "0" (addresses \emptyset , 2, 4, 6, etc.) To accomplish this, the Address Counter is incremented from state \emptyset to state 1023. At each increment, the contents of the present address and the preceding address are compared and the larger value is stored in the appropriate memory location. Storing only the larger of the two values ensures that the peak value of each response will be retained in the permanently stored trace.

4-152. In State 1, the Store Function Controller gives the SFL (Sweep Flag) instruction which performs the following functions:

- a. Overrides the Clock C1 input to the Address Multiplexer, causing the Multiplexer to remain switched to the Address Counter. The Address lines of the RAM are, therefore, controlled by the Address Counter during both the read and write phases of Clock C1.
- b. Forces the LSB of the Address Counter to a logical "0" during the write phase of Clock C1. This means that information can only be written into the memory locations where the LSB of the address is a logical "0". The contents of addresses where the LSB is a logical "1" are left unaltered.
- c. Switches the Store Multiplexer so that the Data Input lines of the RAM are connected to the "Q" outputs of the 8-Bit Latch. In this state, the Y-Axis A to D Converter is disconnected.
- d. Enables the Write Control circuit. During normal operation, the Write Control circuit is disabled and Clock C10 is present at the output of the NAND gate. With the Write Control circuit enabled, Clock C10 is inhibited unless a Write Enable command has been generated by the 8-Bit Adder.

- 4-153. Figure 4-24 shows the equivalent circuit during State 1. Elements not shown can be disregarded.
- 4-154. The State 1 sequence begins with the following conditions:
- a. Clock C1 has just completed the transition from the write phase to the read phase.
 - b. The Address Counter is in State 1 (000000001).
- c. The 8-bit word from the preceding address (\emptyset) is at the output of the Latch.
- d. The 8-bit word stored in the present address (1) is on the Data Output lines of the RAM.
- 4-155. Just before Clock C9 goes high, the 8-bit word on the RAM Data Output lines (present address) is compared to the 8-bit word at the output of the Latch (preceding address) by the 8-Bit Adder. If the numerical value of the 8-bit word in the present address is greater than that of the preceding address, the Adder generates a Write Enable command which is applied to the Write Control circuit. If a Write Enable command is generated, data will be written into memory during the next write phase. When Clock C9 goes high, the 8-bit word on the RAM Data Output lines (address 1) is strobed into the Latch.
- 4-156. When Clock C1 goes into the write phase, the Address Counter remains in State 1 (0000000001) but because its LSB is forced to a logical "0", RAM address Ø (0000000000) is selected. If a Write Enable command was generated during the read phase, the 8-bit word

- from address 1 (now at the output of the Latch) is written into address \emptyset . If a Write Enable command was not generated, Clock $\overline{C10}$ is inhibited and the contents of address \emptyset are left unchanged.
- 4-157. When Clock C1 again goes into the read phase, the Address Counter is incremented to State 2 (000000010). At this time, the 8-bit word from address 1 is still at the output of the Latch and the 8-bit word stored in address 2 is on the RAM Data Output lines. If the 8-bit word in address 2 is greater than that of address 1, a Write Enable command will be generated and, during the next write phase of C1, the contents of address 2 will be written back into address 2, leaving address 2 unchanged. Moreover, if the 8-bit word in address 2 is less than that of address 1, a Write Enable command will not be generated and the contents of address 2 will still be left unchanged. This is an important point. Even though the 8-bit word in each address is compared to that of the address that is one count higher, only alternate comparisons have any effect. For example, addresses Ø and 1 are compared and the largest value is written into address Ø addresses 1 and 2 are compared and address 2 is left unchanged, addresses 2 and 3 are compared and the largest value is written into address 2, etc.
- 4-158. The comparison sequence continues until the Address Counter reaches State 1023 and resets. At that time, the Store Function Controller increments to State 2 where it remains until the STORE button is released.
- 4-159. State 2. In State 2, the SFH (Sweep Flag Hold) instruction is given and the system returns to its normal mode of operation with the following exceptions:

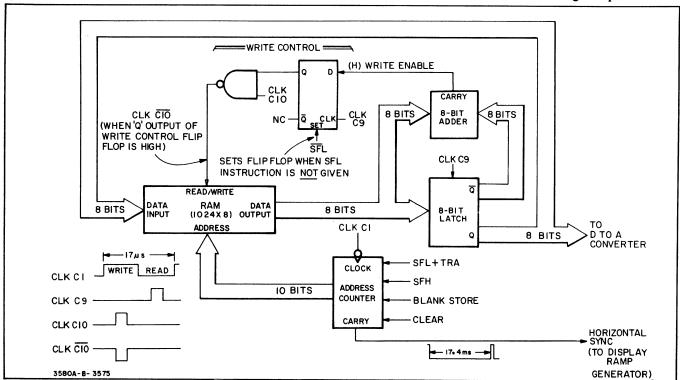


Figure 4-24. Equivalent Circuit (State 1).

- a. The LSB of the X-Axis A to D Converter is forced to a logical "1". Since the X-Axis A to D Converter addresses the RAM during the write phase of C1, new information is written only in addresses where the LSB is a logical "1" (addresses 1, 3, 5, 7, etc.). Addresses containing the permanently stored trace are, therefore, left undisturbed.
- b. The Address Counter is switched so that it first scans the addresses where the LSB is a "1" (refresh trace). It then recycles and scans the addresses where the LSB is a "0" (permanently stored trace).
- c. The SFH instruction (labeled (H)FAST) is applied to the Display Ramp Generator. This doubles the display sweep rate, providing one 8.7 msec sweep for each set of addresses. As a result, both the refresh trace and the permanently stored trace appear on the CRT.
- 4-160. Clear/Write Button. When the CLEAR/-WRITE button is pressed during State 2, the following things take place:
- a. The Y-Axis A to D Converter is held in the reset state and its output is 00000000.
- b. The RAM address lines are switched to the Address Counter during the 8.7 msec periods when it is scanning the addresses containing the refresh trace.
- c. As the Address Counter scans the addresses containing the refresh trace, all zeros are written into memory during the write phases of C1. As a result, the refresh trace is cleared from memory and the permanently stored trace is not disturbed.

- 4-161. Blank Store Button. When the BLANK STORE button is presses, the LSB of the Address Counter is forced to a logical "1". The Address Counter, therefore, continually scans the addresses containing the refresh trace and the permanently stored trace is not displayed. When the BLANK STORE button is released, the permanently stored trace returns to the display.
- 4-162. State 3. When the STORE button is released and the Address Counter resets to \emptyset , the Store Function Controller is incremented to State 3. The purpose of State 3 is to clear the permanently stored trace by filling the memory with the refresh trace. This is accomplished by loading the contents of addresses where the LSB is 1 into addresses where the LSB is \emptyset .
- 4-163. In State 3, the Store Function Controller gives the TRA (Transfer) instruction which performs the following functions:
- a. Overrides the Clock C1 input to the Address Multiplexer causing the RAM Address lines to remain switched to the Address Counter.
- b. Forces the LSB of the Address Counter to a logical "0" during the write phase of Clock C1. This means that information can only be written in addresses where the LSB is a logical "0".
- c. Switches the Store Multiplexer so that the Data Input lines of the RAM are connected to the "Q" outputs of the 8-Bit Latch.
- 4-164. Figure 4-25 shows the equivalent circuit during State 3. Note that the 8-Bit Adder and the Write Control circuit are not used.

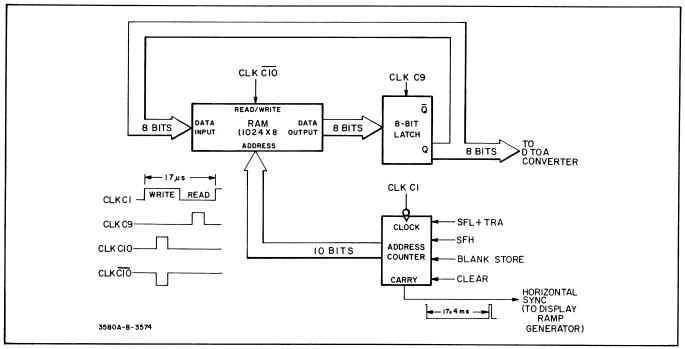


Figure 4-25. Equivalent Circuit (State 3).

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- 4-165. The State 3 sequence begins with the following conditions:
- a. Clock C1 has just made the transition from the write phase to the read phase.
- b. The Address Counter is in State 0 and RAM address 0 is selected.

4-166. When Clock C9 goes high during the read phase, the 8-bit word from address Ø is strobed into the Latch. When Clock C1 goes into the write phase, the Address Counter remains in State \emptyset and, when $\overline{C10}$ goes low, the 8-bit word from address \emptyset is written back into address \emptyset , leaving address Ø unchanged. When Clock C1 again goes into the read phase, the Address Counter is incremented to State 1 and, when C9 goes high, the 8-bit word from address 1 is strobed into the Latch. During the next write phase of C1, the Addresses Counter is still in State 1 (0000000001) but because its LBS is forced to a logical "0", RAM address Ø (000000000) is selected and the 8-bit word from address 1 (now at the output of the Latch) is written into address Ø. As the sequence continues, the contents of address 3 are written into address 2 the contents of address 5 are written into address 4, etc. When the Address Counter reaches State 1023 and resets, the Store Function Controller resets to State Ø and the system returns to its normal mode of operation. At this point, each pair of addresses (0 and 1, 2 and 3, 4 and 5, etc.) contains the same information. Since the addresses are now read sequentially as the Address Counter increments from State Ø to State 1023, the

video amplitude on the display is the same for each pair of X-Axis segments. Because of this redundancy, a series of dots appear on the display (Figure 4-26). The dots are automatically cleared when the memory contents are updated by a new frequency sweep.

4-167. Frequency Counter and Display.

4-168. This section covers the LIMITER, TIME BASE, and COUNTER/DISPLAY blocks of the simplified block diagram. Refer to Figures 4-27 and 4-28 for the following discussions.

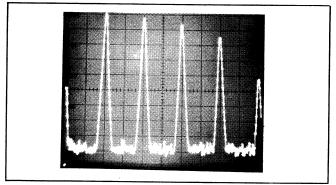


Figure 4-26. Store Button Released.

4-169. Limiter. The Limiter first processes the Buffered 100 kHz Reference signal through a 100 kHz bandpass filter. The filter is simply a parallel LCR network having a resonant frequency of 100 kHz. Next, the Limiter converts the filtered 100 kHz to a square wave which has CMOS logic levels of zero volts (D-ground, logic high)

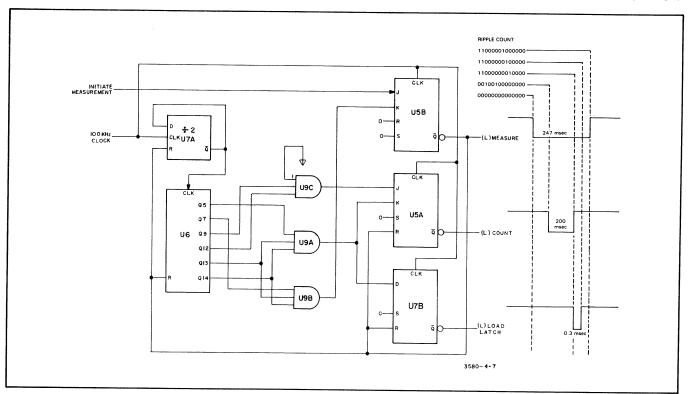


Figure 4-27. Time Base Generator.

and -10 volts (logic low). This convertion is done by comparator U8. The Limiter circuit converts the Buffered 100 kHz Reference to a signal capable of clocking the Time Base Generator accurately.

4-170. Time Base Generator. The Time Base Generator accepts the Initiate Measurement signal from the A36 board and the 100 kHz Clock signal from the Limiter. To run the counter, the Time Base Generator produces three control signals: (L)MEASURE, (L)COUNT, and (L)LOAD Latch. The heart of this signal generation is a 14-stage ripple counter.

4-171. Operation of the Time Base Generator. The Initiate Measurement line going high starts the sequence of events necessary for one frequency measurement. First U5B toggles when its J input goes high and its clock input receives a positive edge of the 100 kHz Clock signal; so, its Q output (H)MEASURE goes high and its Q output (L)MEASURE goes low. Both of these signals leave the board and control the circuits that prevent the VTO from sweeping. (L)MEASURE also drives the reset inputs of U7A, U7B, and U5A low; this enables them to have their J, K, and D inputs utilized. At the same time, (L)MEASURE resets the ripple counter (U6) to zero. Then, since the 100 kHz Clock is divided by two by U7A, the ripple counter begins counting at a rate of 50 kHz. When the binary count reaches 00100100000000 (i.e. the ripple counters Q9 and Q12 outputs go high), the output of AND gate U9C goes high. This forces the J input of U5A high which makes its Q output (L)COUNT go low. Since U5A is a J-K masters/slave flip-flop, its output state won't change

until its K input is driven high; this is done when the ripple counter reaches a binary count of 1100000010000 (i.e. its Q5, Q13, and Q14 outputs have gone high) and the output of AND gate U9A goes high. Also at this time, the D input of U7A is forced high causing its \overline{Q} output (L)LOAD LATCH to go low. When the count reaches 11000000100000 the output of AND gate U9A goes low and the output of U7B goes high. Finally, when the ripple counter reaches 11000001000000 (its Q7, Q13, and Q14 outputs are high), the output of AND gate U9B goes high; this resets U5B via its K input causing its \overline{Q} output (L)MEASURE to go high. In this way, the signals that control the counter are generated.

4-172. Notice that the (L)COUNT line is driven low from 00100100000000 to 1100000010000. The difference of these two binary numbers is 10011100010000; the decimal equivilent of this is 10,000. So, the (L)COUNT line goes low for 10,000 counts. Since the binary number is counting at a rate of 50 kHz (Q1 is a 25 kHz square wave) each count is 20 μ sec in duration. The (L)COUNT line goes low for 10,000 counts times 20 μ sec/count, or 200 msec. The accuracy of this time is dependent on the accuracy of 100 kHz Reference signal. The remaining time durations of the control pulses can be calculated in this manner; they are illustrated in Figure 4-26 for your convenience.

4-173. Counter/Display. Using the 1.0 - 1.5 MHz L.O. signal and the three control signals from the Time Base Generator, the Counter/Display circuitry measures and displays the start, center, or manual vernier frequency. Part of the display circuitry also blanks the display when the BLANK DISPLAY line goes high.

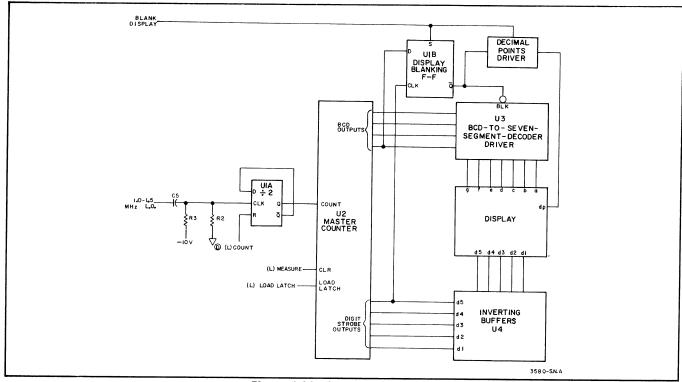


Figure 4-28. Counter and Display.

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4-174. Operation of the Counter. Initially, the 1.0 - 1.5 MHz L.O. signal has any dc component eliminated by C5; then a level shifter, consisting of R2 and R3, assures that the signal has a dc offset of about -5 volts. This signal then goes to the clock input of U1A. The first event occurring in the sequence of counting is the (L)MEASURE line going low. Since this line goes to the clear input of the counter chip, it must change to a low state in order for the counter to be enabled. Next in the sequence, the (L)COUNT line goes low; because this line is connected to the Reset input of U1A, the flip-flop is enabled. Since U1A is connected in a divide-by-two configuration, the count input to the counter is 500-750 kHz (of course, the input to the counter is only one frequency because the VTO has been disabled by the (L)MEASURE and (H)MEASURE lines. (L)COUNT is held low for 200 msec as dictated by the Time Base Generator. At the end of this time, the (L)COUNT line goes high and the (L)LOAD LATCH line goes low. When the (L)COUNT line goes high, U1A is disabled and the counter no longer has an input signal. The (L)LOAD LATCH going low causes the last state of the counter to be stored in an on-chip latch; this line stays low for only 0.3 msec - an adaquate amount of time for the latch to be loaded. The final event of the counting sequence is simply the (L)MEASURE line goine high, disabling the counter and allowing the VTO to sweep again.

4-175. If, for example, the center frequency of the display is 50 kHz, then the 1.0 - 1.5 MHz signal would have a frequency of 1.5 MHz at the time of the frequency measurement. The output of the divide-by-two flipflop would be 750 kHz. The number of counts recorded by the counter during 200 msec would be 750000 countsper-second times 0.2 seconds, or 150000 counts. The latch inside the counter chip contains six digits; since

only the last five digits are displayed, the proper center frequency of 50000 Hz is displayed.

4-176. Operation of the Display. The five digits representing the frequency count are displayed in the following manner. The counted frequency has been stored in the latch in U2. The output of U2 consists of four lines for BCD and five lines to strobe each digit to be displayed. U2 contains all the needed circuitry to insure that for any given digit, the correct BCD will be outputted when its corresponding digit strobe is high. U3 is a BCD-to-seven-segment-decoder-driver; it converts the BCD output of U2 to seven segment information and drives the appropriate anodes of the display high. The digit strobe outputs of U2 are buffered and inverted by U4 before driving the correct common cathodes of the display low. Each digit of the display has seven LEDS with a common cathode; the anodes of identical segments (of all the digits) are also common. As each digit is selected, the correct segments for that digit are also selected. Thus, the five digits from the latch are displayed.

4-177. Whenever the instrument is switched into the Log Zero or Log sweep mode, the Blank Display line goes high. This causes the display-blanking flip-flop (U1B) to drive the blanking input of U3 low; the outputs of U3 are disabled by this. The Blank Display line also causes the decimal points to be blanked.

4-178. If the display attempts to go one count below 00000 Hz, it tries to become 99999 Hz. This is because the number in the latch goes from 100000 to 099999. When the MSD of the attempted display frequency is eight or nine (BCD output of 1000 or 1001, and the D5 output is high), the display is blanked by U1B. U1B also causes the decimal points driver (Q1) to light the decimal points.

SECTION V MAINTENANCE

5-1. INTRODUCTION.

5-2. This section contains Performance Tests (Paragraph 5-5) and Adjustment Procedures (Paragraph 5-48) for the Model 3580A Specturm Analyzer. Troubleshooting information is presented in Section VII, along with the Schematic Diagrams.

5-3. RECOMMENDED TEST EQUIPMENT.

5-4. The test equipment that is recommended for maintaining the Model 3580A is listed in Table 5-1. The equipment is designated as to its use for Performance Tests, Adjustments or Troubleshooting.

5-5. PERFORMANCE TESTS.

- 5-6. The following Performance Tests are procedures that can be used to verify that the 3580A is operating properly and meets the spectifications listed in Table 1-1. These procedures can be used for incoming quality control inspection, to check specifications after a repair or for routine maintenance. Where possible, the Performance Tests call out the proper adjustment in the Adjustment Procedures. Since adjustments interact, it is important to follow the procedures carefully.
 - a. FREQUENCY TESTS (Paragraph 5-9).
 - b. SWEEP TESTS (Paragraph 5-13).
 - c. AMPLITUDE TESTS (Paragraph 5-18).
 - d. BANDWIDTH TESTS (Paragraph 5-28).
- e. DYNAMIC RANGE TESTS (NOISE TESTS) (Paragraph 5-30).
- f. IF FEEDTHRU and ZERO BEAT RESPONSE TSTS (Paragraph 5-36).
 - g. INPUT IMPEDANCE TESTS (Paragraph 5-38).
 - h. OUTPUT TESTS (Paragraph 5-40).
- i. BALANCED INPUT TESTS (Option 002 only) (Paragraph 5-44).

5-7. Test Card.

5-8. A Performance Test Card is provided at the end of this section for your convenience in recording the performance of the Model 3580A during Performance Tests. This card can be removed from the manual and used as a permanent record of the incoming inspection or of a routine performance test. The Performance Test Card may be reproduced without written permission from Hewlett-Packard.

Revised: September 1987

NOTE

Always allow one hour continuous warm-up before attempting any tests.

5-9. Frequency Tests. Δ 16

5-10. These tests verify part of the Frequency Characteristic Specifications listed in Table 1-1. If, for any reason, the instrument will not pass these tests, perform the Sweep Alignment (Paragraph 5-63) of the Adjustment Procedures.

5-11. Range and Frequency Display Accuracy Test.

a. Position the following front panel controls:

AD ADDITION OFFICE

| Αľ | DAPTIVE | E SWEE | P | | OFF |
|----|----------------|-----------|-----------------------------|------------|-------|
| DI | SPLAY. | | All pushbu | ttons rele | eased |
| A۱ | IPLITUI | DE MOD | ELOG | 10 dBv/ | DIV |
| AN | IPLITUI | DE REF | LEVEL | NOR | MAL |
| dB | v/LIN - c | dBm 600 | $\Omega \dots \dots \Omega$ | dBv/ | /LIN |
| IN | PUT SEN | ISITIVI | Γ Υ | | CAL |
| V. | ERNIER | (Ampli | ude) | | CAL |
| | | | | (Fully | CW) |
| FR | EQUEN | CY | | 30000 |) Hz |
| ST | ART-CT | R | | ' | CTR |
| RE | SOLUTI | ON BAN | IDWIDTH | [| 1 Hz |
| D | ISPLAY | SMOOT | HING | | MIN |
| FR | EQ. SPA | N/DIV. | | | 5 Hz |
| SW | EEP TII | ME/DIV | | 2 | SEC |
| SW | EEP MO | DDE | | N | IAN |
| | | | | | |

Option 002: Set dBm 900 Ω /LIN-dBm 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL.

- b. By alternately pressing and releasing DISPLAY-CLEAR WRITE while adjusting the MANUAL VER-NIER, center the display indication.
- c. With the display indication now on the center graticule, vary the FINE FREQUENCY control to peak the display indication.
- d. The FREQUENCY display should read 30000 Hz \pm 3 Hz.
- e. Change the RESOLUTION BANDWIDTH to 3 Hz. Repeat Steps c and d.

Maintenance

Table 5-1. Recommended Test Equipment.

| Digital Multimeter | | Table 5-1. Hecom | | 1 | T | T |
|---|-------------------|---|---|---|-----------------|--|
| Digital Multimeter | INSTRUMENT | REQUIRED CHARACTERISTICS | | | Troubleshooting | |
| Full Scale Range: 1.1, 1.0V, 100V Resolution: 4 digits Input Impedance: ≥ 10 MΩ, ≤ 100 pF Accuracy: ± 1% of reading | | Full scale ranges: 1V, 10V, 100V Resolution: 4 digits Input Impedance: 10-12 MΩ Accuracy: ± .1% of reading AC Function: Response: average | | x | x | -hp- 34740/34702 or |
| Oscilloscope | | Full Scale Range: 1V, 10V, 100V Resolution: 4 digits Input Impedance: ≥ 10 MΩ, ≤ 100 pF | | | | |
| Impedance: 10 MΩ, 10 pF | Oscilloscope | Sensitivity: .005 V/DIV Sweep: .005 μsec/DIV to .1 sec/div Frequency: 0 to 10 MHz Input Impedance: 1MΩ, 25 pF | | × | × | or -hp- 180C/D with -hp- 1801A Vertical Amplifier and |
| Electronic Counter Function: Frequency and Time Interval Frequency Range: 10 Hz to 10 MHz Resolution: 6 digits Sensitivity: 0.1 V rms | for | | | x | x | |
| Frequency Resolution: 6 digits Sensitivity: 0.1 V rms X | | | | | | -hp- 10004D |
| Synthesizer (50 ohms) Amplitude Range: (-67.99 dBm 50 Ω to + 26.99 dBm 50 Ω) Amplitude Accuracy: ± .1 dB Amplitude Resolution: .01 dB Frequency Resolution: .1 Hz Termination for Syntheseizer Distortion Analyzer Distortion Analyzer Distortion Measurement Accuracy: ± 10% for greater than .3% distortion Center of Bandpass at 5kHz, (50 Ω input), Output Distortion: (with Frequency Synthesizer): > 90dB down 1 kΩ Resistor 1% film resistor 1% film resistor 1% film resistor 10% carbon or film resistor X Amplitude Range: (-67.99 dBm 50 Ω to + 26.99 dBm 50 Ω) X X Ax X Ax Ax Ax App. 11048C App. 333A Ang. App. 334A Allen Avionics (special order) Or White Model 2640 Allen Avionics (special order) Or White Model 2640 App. 0757-0280 App. 0757-0280 App. 0757-0242 App. 0757-0442 App. 0757-0344 App. 10528A | | Frequency Range: 10 Hz to 10 MHz Resolution: 6 digits | | x | × | or |
| Termination for Syntheseizer Distortion Analyzer Distortion Analyzer Fundamental Frequency Range: 10 Hz to 100 kHz Distortion Measurement Accuracy: ± 10% for greater than .3% distortion Center of Bandpass at 5kHz, (50 Ω input), Output Distortion: (with Frequency Synthesizer): > 90dB down I kΩ Resistor 1% film resistor X Allen Avionics (special order) or White Model 2640 The O757-0280 -hp- 0757-0280 -hp- 0698-3510 Termination 10 kΩ Resistor 10% carbon or film resistor X 10% carbon or film resistor X -hp- 0757-0442 -hp- 0698-4456 -hp- 0757-0344 Logic Clip Able to detect TTL HIGH and LOW | Synthesizer | Amplitude Range: (-67.99 dBm 50 Ω to + 26.99 dBm 50 Ω) Amplitude Accuracy: ± .1 dB Amplitude Resolution: .01 dB | × | × | x | or |
| Analyzer 10 Hz to 100 kHz Distortion Measurement Accuracy: ± 10% for greater than .3% distortion X | Termination | 1 11 11 11 11 11 11 11 11 11 11 11 11 1 | x | × | x | -hp- 11048C |
| Bandpass Filter Center of Bandpass at 5kHz, (50 Ω input), Output Distortion: (with Frequency Synthesizer): > 90dB down 1 kΩ Resistor 1% film resistor X -hp- 0757-0280 -hp- 0698-3510 resistors 600 Ω Termination 10 kΩ Resistor 10% carbon or film resistor X -hp- 0757-0442 -hp- 0757-0442 -hp- 0698-4456 -hp- 0757-0344 Logic Clip Able to detect TTL HIGH and LOW X (special order) or White Model 2640 X -hp- 0757-0280 -hp- 0757-0280 -hp- 0698-3510 X -hp- 0757-0442 -hp- 0757-0444 | | 10 Hz to 100 kHz Distortion Measurement Accuracy: ± 10% for greater than .3% | × | | | or |
| 1 kΩ Resistor 1% film resistor X -hp- 0757-0280 -hp- 0698-3510 resistors 600 Ω Termination 10 kΩ Resistor 10% carbon or film resistor X -hp- 0757-0442 -hp- 0757-0442 -hp- 0757-0442 -hp- 0757-0344 Logic Clip Able to detect TTL HIGH and LOW X -hp- 10528A | Bandpass Filter* | Output Distortion: | Y | | | (special order) or |
| (2) 453 Ω resistors 600 Ω Termination 10 kΩ Resistor 10% carbon or film resistor X -hp- 0698-3510 X -hp- 0757-0442 The complete of the com | 1 kΩ Resistor | | | | | |
| 600 Ω Termination 10 kΩ Resistor X 11095A 10 kΩ Resistor 10% carbon or film resistor X -hp- 0757-0442 550 Ω Resistor 10% carbon or film resistor X -hp- 0698-4456 1 ΜΩ Resistor 1% film resistor X -hp- 0757-0344 Logic Clip Able to detect TTL HIGH and LOW X -hp- 10528A | | 1% film resistor | | | | • |
| 1 MΩ Resistor | 600 Ω Termination | 10% carbon or film resistor | x | | х | |
| Logic Clip Able to detect TTL HIGH and LOW X -hp- 10528A | 550 Ω Resistor | 10% carbon or film resistor | x | | | -hp- 0698-4456 |
| | 1 MΩ Resistor | 1% film resistor | x | | | -hp- 0757-0344 |
| 16 pins | Logic Clip | levels for DUAL IN-LINE configuration, | | | × | -hp- 10528A |

[★] Optional see ∆ 19

5-12. Display Accuracy Tests.

| a. | Repostion the following front pa | nel controls. |
|----|----------------------------------|---------------|
| | START-CTR | START |
| | SWEEP MODE | |
| | FREQUENCY | .00000 Hz |
| | RESOLUTION BANDWIDTH. | 300 Hz |
| | FREQ. SPAN/DIV | 5 kHz |
| | SWEEP MODE | REP |

b. The 10 kHz CAL signal and its harmonics should be repetitively swept and appear on the display as shown by Figure 5-1. The separation between the Zero Response and 50 kHz harmonic should be 10 major divisions \pm 1 minor division. The separation between any two adjacent responses should be 2 major divisions \pm .2 minor divisions. Momentarily push and release DISPLAY - STORE, watching the dislay to verify that the STORE and NON-STORE traces appear in the same position.

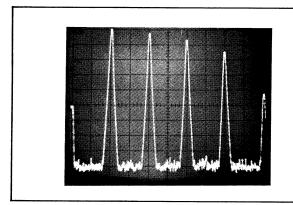


Figure 5-1. 10 kHz CAL Signal.

5-13. Sweep Tests. Δ 16

5-14. These tests verify the Sweep Characteristics Specifications given in Table 1-1. If the instrument fails the Frequency Span Tests (Paragraph 5-15), perform the Sweep Alignment (Paragraph 5-63) of the Adjustment Procedures. If it fails only the Log Sweep Test (Paragraph 5-16), perform only the Log Sweep Adjustments (Paragraph 5-67) of the Sweep Alignment. All sweep time calibration is done with a factory selected resistor. If the instrument will not pass the Sweep Time Tests (Paragraph 5-17), refer to Section VII for additional information.

5-15. Frequency Span Test.

a. Position the following front panel controls: (Only those controls printed in **BOLD** require a change from the previous test.)

| ADAPTIVE SWEEP | OFF |
|----------------------|----------------|
| DISPLAY All pushbu | ttons released |
| AMPLITUDE MODELOG | 10 dBV/DIV |
| AMPLITUDE REF LEVEL. | |
| dRV/I IN _ dRm 600 0 | |

| INPUT SENSITIVITY 0 dBV |
|-------------------------------------|
| VERNIER (Amplitude)CAL |
| (Fully CW) |
| FREQUENCY00000 Hz |
| RESOLUTION BANDWIDTH300 Hz |
| START-CTR START |
| DISPLAY SMOOTHINGMIN |
| FREQ. SPAN/DIV5Hz |
| SWEEP TIME/DIV0.2 SEC |
| SWEEP MODEManual |
| Option 002: Set dBm 900 Ω/LIN - dBm |
| 600 Ω switch to dBm 900Ω; set INPUT |
| MODE switch to UNBAL. |

- b. Adjust MANUAL VERNIER full CCW.
- c. Adjust the FINE FREQUENCY control for a 00000 Hz indication on the FREQUENCY display.
- d. Adjust MANUAL VERNIER full CW. The FRE-QUENCY display should be 00050 Hz \pm 1 Hz.
- e. Readjust MANUAL VERNIER full CCW. Reposition FREQ.SPAN/DIV 10 Hz.
- f. Readjust the FINE FREQUENCY control for a 00000 Hz indication on the FREQUENCY display.
- g. Adjust MANUAL VERNIER full CW. The FRE-QUENCY display should be 00100 Hz \pm 2 Hz.
- h. Continue this procedure for the remaining FREQ. SPAN/DIV settings. Refer to Table 5-2 for the proper tolerances.

Table 5-2. Frequency Span Test.

| | COUNTER READING | | |
|----------------|-----------------|-------------------|--|
| FREQ. SPAN/DIV | MANUAL VERNIER | MANUAL VERNIER | |
| | FULL CCW | FULL CW | |
| 5 Hz | 00000 Hz | 00050 Hz ± 1 Hz | |
| 10 Hz | 00000 Hz | 00100 Hz ± 2 Hz | |
| 20 Hz | 00000 Hz | 00200 Hz ± 4 Hz | |
| 50 Hz | 00000 Hz | 00500 Hz ± 10 Hz | |
| .1kHz | 00000 Hz | 01000 Hz ± 20 Hz | |
| .2kHz | 00000 Hz | 02000 Hz ± 40 Hz | |
| .5kHz | 00000 Hz | 05000 Hz ± 100 Hz | |
| 1 kHz | 00000 Hz | 10000 Hz ± 200 Hz | |
| 2 kHz | 00000 Hz | 20000 Hz ± 400 Hz | |
| 5 kHz | | | |
| (checked in | | | |
| Para 5-16) | | | |

5-16. Log Sweep Test.

a. Reposition the controls as follows:

| INPUT SENSITIVITY | CAL |
|----------------------|--------|
| RESOLUTION BANDWIDTH | .30 Hz |
| SWEEP MODELOG | ZERO |

- b. Momentarily press DISPLAY CLEAR WRITE.
- c. Adjust the FINE FREQUENCY control for a frequency indication of 20 Hz.
 - d. Reposition the controls as follows:

RESOLUTION BANDWIDTH.....300 Hz SWEEP MODE......LOG

Allow time for three complete sweeps.

e. Verify that the 20 kHz harmonic of the internal CAL signal falls on the proper graticule (± 1 minor division). If the instrument will not pass this test, but passes all previous tests, perform only the Log Sweep Adjustments (Paragraph 5-67) of the Adjustment Procedures.

5-17. Sweep Time Tests.

a. Reposition the controls as follows:

SWEEP TIME/DIV......0.01 SEC SWEEP MODE.....SING

Momentarily press:

DISPLAY.....CLEAR WRITE

b. The display should be erased, and then swept once. Remembering the sweep time, reposition the controls as follows:

SWEEP TIME/DIV......0.02 SEC

c. Again, press:

DISPLAY......CLEAR WRITE

The sweep time should appear slower.

d. Repeat this procedure for all sweep times, always looking for progressively slower sweep rates. On the slowest sweep rates, it will not be necessary to complete a full sweep before switching to the next SWEEP TIME/DIV. Let the instrument sweep only as long as is necessary to monitor the rate. A more accurate method for measuring sweep time is given in Paragraph 7-39.

5-18. Amplitude Tests. Δ 16

5-19. These tests verify the Amplitude Specifications given in Table 1-1. Amplitude accuracy must be determined before the Bandwidth Specifications can be tested. Since the IF Filter Alignment (Paragraph 5-70) interacts with the Amplitude Accuracy, it is important that the IF Filter Alignment be performed first if the instrument will not pass any of the Amplitude Accuracy Tests. The Amplitude Tests should then be repeated, and if the instrument still fails these tests, then perform the Amplitude Adjustments (Paragraph 5-74) of the Adjustment Procedures.

NOTE

There are no adjustments for Amplitude Reference Level Tests (Linear and Log Mode). If the instrument passes all Amplitude Tests except one or both of these, refer to Section VII for troubleshooting information.

Equipment Required:

Frequency Synthesizer (-hp- Model 3320B, 50 ohms) 50 Ohm Termination (-hp- 11048C) Digital Multimeter (-hp- Model 34740/34702)

5-20. Bandwidth Switching Accuracy Tests.

a. Position the following front panel controls: (Only those controls printed in **BOLD** require a change from the previous tests.)

| ADAPTIVE SWEEP | OFF |
|------------------------------------|----------|
| DISPLAYAll pushbuttons | released |
| AMPLITUDE MODELOG 1 dl | BV/DIV |
| AMPLITUDE REF LEVELNO | DRMAL |
| $dBV/LIN-dBm$ $600\Omegadl$ | BV/LIN |
| INPUT SENSITIVITY | 0 dBV |
| VERNIER (Amplitude) | CAL |
| | lly CW) |
| START-CTR | CTŔ |
| RESOLUTION BANDWIDTH | .300 Hz |
| DISPLAY SMOOTHING | MIN |
| FREQUENCY SPAN/DIV | 5 Hz |
| SWEEP TIME/DIV | 1.1 SEC |
| SWEEP MODE | RESET |
| FREQUENCY10 | 000 Hz |
| SWEEP MODEMA | NUAL |
| Ontion 002: Sat dBm 0000 /I IN JB. | |

Option 002: Set dBm $900\Omega/LIN-dBm$ 600 Ω switch to dBm 900Ω ; set INPUT MODE switch to UNBAL.

b. Connect a properly terminated frequency synthesizer to the 3580A INPUT and adjust the source for a 10 kHz, 0 dBV output level (0 dBm 900 Ω for instruments with Option 002).

NOTE

See Table 5-3 for the proper level to use with your source. See Figure 5-2 for the proper hookup with an -hp- 3320B Frequency Synthesizer.

- c. By alternately pressing and releasing DISPLAY-CLEAR WRITE while adjusting MANUAL VER-NIER, center the display indication (a narrow spike).
- d. Adjust the FINE FREQUENCY control for a peak of this spike.

| Table | 5.3. | Conversion | Table |
|-------|------|------------|-------|
| | | | |

| 3580A INPUT Signal Level | 3320B or OTHER 50 OHM SOURCE | ABSOLUTE Voltage | | |
|---|---|---|--|--|
| + 10 dBV + 10 dBm 900Ω 0 dBV 0 dBm 900Ω - 10 dBV - 10 dBm 900Ω - 20 dBV - 20 dBm 900Ω - 30 dBW - 30 dBm 900Ω - 40 dBV - 40 dBW - 50 dBW - 50 dBW - 50 dBW - 60 dBW - 70 dBM 900Ω - 70 dBW - 70 dBm 900Ω - 80 dBV - 80 dBM | + 23.01 dBm + 22.55 dBm + 13.01 dBm + 12.55 dBm + 3.01 dBm + 2.55 dBm - 6.99 dBm - 7.45 dBm - 16.99 dBm - 17.45 dBm - 26.99 dBm - 36.99 dBm - 37.45 dBm - 46.99 dBm - 47.45 dBm - 56.99 dBm - 57.45 dBm - 57.45 dBm - 57.45 dBm | 3.162 volts 3 volts 1 volts .949 volts .3162 volts .3000 volts .1 volts .0949 volts .03162 volts .03 volts .01 volts .095 volts 3162 mV 3 mV 1 mV .95 mV .3162 mV .3 mV .1 mV | | |

- e. Adjust the front panel VERNIER (Amplitude) for a-1 dB display indication. Note: The display indication is calibrated 1.0 dB per major division.
 - f. Reposition the following front panel control: RESOLUTION BANDWIDTH.....100 Hz
- g. Adjust the front panel FINE FREQUENCY control for a peak display indication. The display indication should be -1 dB \pm .5 dB.
 - h. Repostion the following front panel control: RESOLUTION BANDWIDTH.....30 Hz
- i. Slowly adjust MANUAL VERNIER for a peak display indiction. The peak indication should be -1 dB \pm .5 dB. Momentarily press DISPLAY-CLEAR WRITE.

- j. Reposition the following front panel control: RESOLUTION BANDWIDTH.....10 Hz
- k. Readjust MANUAL VERNIER for a peak display indication. The peak indication should be -1 dB \pm .5 dB. Momentarily press DISPLAY-CLEAR WRITE.
 - Reposition the following front panel control:
 RESOLUTION BANDWIDTH......3 Hz
- m. Slowly readjust MANUAL VERNIER for a peak display indication. The peak indication should be -1 dB \pm .5 dB. Momentarily press DISPLAY-CLEAR WRITE.
 - n. Reposition the following front panel control: RESOLUTION BANDWIDTH......1 Hz
- o. Very slowly readjust MANUAL VERNIER for a peak display indication. The peak indication should be $-1 \text{ dB} \pm 1 \text{ dB}$.

5-21. Log Amplitude Display Accuracy Tests.

- a. Reposition the following front panel controls:

 VERNIER (Amplitude)..........CAL

 AMPLITUDE MODE....LOG 10 dB/DIV

 RESOLUTION BANDWIDTH.....10 Hz
- b. By alternately pressing and releasing DISPLAY-CLEAR WRITE while adjusting MANUAL VER-NIER, center the display indication (a narrow spike).
- c. Adjust the CAL 10 kHz for a full scale 0 dB display indiction of the spike.
- d. Adjust the signal source to the levels indicated by Table 5-4. Check the display for proper level. (See Table 5-3 for the proper Input Level setting to use on your signal source.)

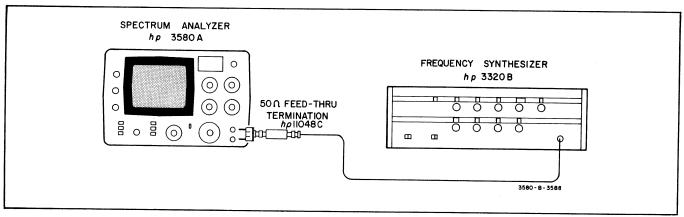


Figure 5-2. Proper Hookup.

NOTE

For the -60 dB, -70 dB and -80 dB readings, readjust MANUAL VERNIER for a peak display indication.

Table 5-4. Log Amplitude Tests.

| INPUT LEVEL (10 k | | | |
|---|----------|--|--|
| STANDARD OPTION 002 INSTRUMENT 900 Ω INSTRUMENT | | DISPLAY INDICATION (0 dB = full scale) | |
| – 10 dBV | - 10 dBm | 10 40 . 0 40 | |
| | | $-10 \text{ dB} \pm 2 \text{ dB}$ | |
| – 20 dBV | – 20 dBm | -20 dB ± 2 dB | |
| – 30 dBV | – 30 dBm | -30 dB ± 2 dB | |
| -40 dBV | -40 dBm | -40 dB ± 2 dB | |
| - 50 dBV | – 50 dBm | -50 dB ± 2 dB | |
| - 60 dBV | - 60 dBm | -60 dB ± 2 dB | |
| - 70 dBV | – 70 dBm | -70 dB ± 2 dB | |
| - 80 dBV | - 80 dBm | -80 dB ± 2 dB | |

5-22. Linear Amplitude Display Accuracy Tests.

- a. Reposition the following front panel controls:
 - AMPLITUDE MODE.....LINEAR
- b. Adjust the signal source for a 1 volt (0 dBV) output (See Table 5-3 for the proper setting to use on your source). Adjust MANUAL VERNIER for a peak display indication.
- c. Adjust the CAL 10 kHz for a full scale 1 volt display indication. Momentarily press DISPLAY-CLEAR WRITE.
- d. Adjust the signal source to the levels indicated by Table 5-5. Check that the display is accurate for each setting.

Table 5-5. Linear Amplitude Tests.

| INPUT LEVEL (10 kHz) | 3320B or OTHER 50 OHM SOURCE | DISPLAY INDICATION (1 volt = full scale) |
|--|--|--|
| .9 V .8 V .7 V .6 V .5 V .4 V .3 V .2 V | 12.10 dBm 11.07 dBm 9.91 dBm 8.51 dBm 6.99 dBm 5.05 dBm 2.55 dBm 97 dBm | .9 V ± .02 V .8 V ± .02 V .7 V ± .02 V .6 V ± .02 V .5 V ± .02 V .4 V ± .02 V .3 V ± .02 V .2 V ± .02 V .1 V ± .02 V |

5-23. Amplitude Reference Level Test (Linear Mode).

- a. Reposition the following front panel control:
 - DISPLAY SMOOTHING.....MAX RESOLUTION BANDWIDTH.....30 Hz

- b. Adjust the synthesizer for a 10 kHz, 1 volt output (+ 13.01 dBm, 50 ohm).
- c. Readjust MANUAL VERNIER for a peak display indication. Adjust VERNIER (Amplitude) for a display indication at 90% of full scale.
- d. Adjust the frequency synthesizer and AMPLITUDE REF LEVEL to the values given in Table 5-6. Check for proper display level.

NOTE

If the instrument fails this test, see Section VII for troubleshooting information. There are no adjustments for this specification.

Table 5-6. Amplitude Ref Level Tests (Linear Mode).

| INPUT 3320B or OTHER LEVEL 50 OHM SOURCE (10 kHz) | | AMP REF LEVEL | DISPLAY INDICATION (% of Full Scale) | |
|---|-------------|------------------|--|--|
| 1 V | + 13.01 dBm | Normal | 90% (CAL) | |
| 200 mV | 99 dBm | -10 | 90% ± 1.5 minor div. | |
| 100 mV | - 6.99 dBm | - 20 | 90% ± 1.5 minor div. | |
| 20 mV | - 20.99 dBm | -30 | 90% ± 1.5 minor div. | |
| 10 mV | - 26.99 dBm | -40 | 90% ± 1.5 minor div. | |
| 2 mV | -40.99 dBm | - 50 | 90% ± 1.5 minor div. | |
| 1 mV | -46.99 dBm | -60 | 90% ± 1.5 minor div. | |
| .2 mV | ~60.99 dBm | - 70 | 90% ± 1 major div. | |

5-24. Amplitude Reference Level Tests (Log Mode).

NOTE

If the instrument fails this test, see Section VII for troubleshooting information. There are no adjustments for this specification.

- a. Reposition the following front panel controls:
 - AMPLITUDE MODE....LOG 10 dB/DIV AMPLITUDE REF LEVEL....NORMAL Input Sensitivity.....-10 dBV
- b. Conect the digital multimeter (DC mode, 100 volt range) to the Y AXIS output of the 3580A.
- c. Adjust the signal source for a -70 dBV output (-70 dBm 900 Ohm for Option 002). (See Table 5-3 for proper levels.) Adjust the MANUAL VERNIER and the FINE FREQUENCY control for a peak display. Adjust VERNIER (Amplitude) for a 1.50 volt \pm .01 volt reading on the multimeter.
- d. Adjust the AMPLITUDE REF LEVEL switch to the settings given in Table 5-7. Check for the proper multimeter reading.

NOTE

MANUAL VERNIER may have to be readjusted to insure a peak display indication.

Table 5-7. Amplitude Ref. Level Tests (Log Mode).

| INPUT LEVEL (10 kHz) | | | |
|---|--|---|---|
| STANDARD INSTRUMENT | OPTION 002 900 Ω | AMPLITUDE REF. LEVEL | MULTIMETER READING |
| - 70 dBV | - 70 dBm - 70 dBm - 70 dBm - 70 dBm - 70 dBm - 70 dBm | - 10 dB - 20 dB - 30 dB - 40 dB - 50 dB - 60 dB - 70 dB | 2.00 V ± 0.02 V 2.50 V ± 0.03 V 3.00 V ± 0.04 V 3.50 V ± 0.06 V 4.00 V ± 0.08 V 4.50 V ± 0.10 V 5.00 V ± 0.12 V |

e. Disconnect the multimeter from the 3580A.

5-25. Input Attenuator Tests.

a. Reposition the following front panel controls:

VERNIER (Amplitude)......CAL
AMPLITUDE MODE.....LINEAR
AMPLITUDE REF LEVEL....-30 dB
INPUT SENSITIVITY
(according to white marker).....1 V
DISPLAY SMOOTHING.....MIN

b. Adjust the signal source for a 1 volt 10 kHz output (See Table 5-8). Adjust MANUAL VERNIER for a peak display indication. Adjust CAL 10 kHz for a full scale display. Momentarily press DISPLAY-CLEAR WRITE.

- c. Adjust the signal source and INPUT SENSITIVI-TY switch to the levels given in Table 5-8. Check for the proper display indication.
 - d. Reposition the following front panel control:

 AMPLITUDE REF LEVEL....NORMAL

Table 5-8. First Input Attenuator Test.

| INPUT LEVEL (10 kHz) | 3320B or OTHER 50 OHM SOURCE | INPUT SENSITIVITY (according to white marker) | DISPLAY INDICATION | |
|----------------------------|--|--|---|--|
| 1 V | + 13.01 dBm | 1 V | Full Scale (CAL) | |
| .2 V .1 V 20mV | .99 dBm6.99 dBm20.99 dBm | .2 V .1 V 20mV | Full Scale (± 3%) Full Scale (± 3%) Full Scale (± 3%) | |

- e. Adjust the signal source for a 1 volt 10 kHz output (See Table 5-9). Adjust the CAL 10 kHz for a full scale display.
- f. Adjust the signal source and INPUT SENSITIVI-TY switch to the values given in Table 5-9. Check for the proper display indication.

Table 5-9. Second Input Attenuator Test.

| INPUT Level (10 kHz) | E L 50 OHM SOURCE SENSITIVITY | | DISPLAY Indication | |
|--|---|--|---|--|
| 1 V | + 13.01 dBm | 1 V | Full Scale (CAL) | |
| .2 V .1 V 20mV 10mV 2mV 1mV | 99 dBm - 6.99 dBm - 20.99 dBm - 26.99 dBm - 40.99 dBm - 46.99 dBm - 60.99 dBm | .2 V .1 V 20mV 10mV 2mV 1mV | Full Scale (± 3%) Full Scale (± 6%) | |

5-26. Frequency Response Tests.

a. Reposition the following front panel controls:

AMPLITUDE MODE...LOG 10 dB/DIV AMPLITUDE REF LEVEL.....-30 dB INPUT SENSITIVITY (according to white marker).....0 dB RESOLUTION BANDWIDTH.....3 Hz

- b. Adjust the signal source for a 10 kHz 0 dBV output (0 dBm 900 Ω for Option 002). (See Table 5-3).
- c. By alternately pressing and releasing DISPLAY-CLEAR WRITE while adjusting MANUAL VERNIER, center the display indiction (a narrow spike). Adjust the FINE FREQUENCY control for a peak of this spike.
- d. Reposition the following front panel control:

AMPLITUDE MODE.....LOG 1 dB/DIV

- e. Readjust MANUAL VERNIER for a peak display.
- f. Adjust VERNIER (Amplitude) for a-1 dB display (1 dB/div).
- g. Adjust the signal source to the frequencies given in Table 5-10 (refer to Table 5-3) for an INPUT SEN-SITIVITY of 0 dB. At each frequency, adjust the FRE-QUENCY controls to that of the source. Then, slowly adjust the FINE FREQUENCY control for a peak display. Momentarily press DISPLAY-CLEAR WRITE. Check for proper level as given in Table 5-10. Note: The display is calibrated 1 dB per major division.
- h. Repeat Steps e through g for an INPUT SEN-SITIVITY and source levels of $-10 \, \mathrm{dB}$, $-20 \, \mathrm{dB}$ and $-40 \, \mathrm{dB}$ (according to white marker and with a $-30 \, \mathrm{dB}$ AMPLITUDE REF LEVEL). Consult Table 5-3 and Table 5-10 for the proper input level and frequencies to use. At the start of each new INPUT SENSITIVITY, always recalibrate the instrument at $10 \, \mathrm{kHz}$ with CAL $10 \, \mathrm{kHz}$.

5-27. Internal Calibrator Test.

a. Reposition the following front panel controls:

| Amplitude ModeLOG 1 dB/DIV |
|----------------------------|
| VERNIER (Amplitude)CAL |
| AMPLITUDE REF LEVELNORMAL |
| INPUT SENSITIVITY – 20 dB |
| START-CTR START |
| RESOLUTION BANDWIDTH300 Hz |
| FREQ. SPAN/DIV 5 kHz |
| SWEEP TIME/DIV0.2 SEC |
| SWEEP MODEREP |
| FREQUENCY00000 Hz |
| |

- b. Adjust the signal source for a 10 kHz -20 dBV (-20 dBm 900 Ω if Option 002) output. (See Table 5-3 for proper level.)
- c. Adjust the FINE FREQUENCY control for a display response on the 10 kHz graticule (2 major divisions from left graticule). (After each trial adjustment, allow 2 seconds for the next sweep before verifying the accuracy of the adjustment.)
- d. Adjust the CAL 10 kHz for a full scale 0 dB display. (After each trial adjustment, allow 2 seconds for the next sweep before verifying the accuracy of the adjustment.)
- f. Verify that the 10 kHz harmonic of the CAL signal appears 2 major divisions from left graticule with a full scale 0 dB level (\pm .15 dB). (1 dB = 1 major division.)

5-28. Bandwidth Tests. Δ 16

5-29. This test verifies the bandwidth specifications of Table 1-1. If the instrument will not pass this test, per-

form the IF Filter Alignment (Paragraph 5-70) of the Adjustment Procedures.

Equipment Required:

Frequency Synthesizer (-hp- Model 3320B, 50 ohms) 50 Ohm Termination (-hp- 11048C)

a. Position the following front panel controls. (Only those controls printed in **BOLD** require a change from the previous test.)

| ADAPTIVE SWEEP | OFF |
|---------------------------------------|------|
| DISPLAYAll pushbuttons relea | ased |
| AMPLITUDE MODELOG 10 dBv/l | DIV |
| AMPLITUDE REF LEVELNORM | IAI. |
| $dBV/LIN - dBm 600\Omega \dots dBV/I$ | LIN |
| INPUT SENSITIVITY20 | dB |
| VERNIER (Amplitude) | AL |
| START-CTR | TR |
| RESOLUTION BANDWIDTH300 | Hz |
| DISPLAY SMOOTHING | 4IN |
| FREQUENCY SPAN/DIV50 | Hz |
| SWEEP TIME/DIV0.2 S | EC |
| SWEEP MODERES | ET |
| FREQUENCY10000 | Hz |
| SWEEP MODEMANU | AL |
| | |

Option 002: Set dBm 900 $\Omega/\text{LIN}-\text{dBm}$ 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL.

- b. By alternately pressing and releasing DISPLAY-CLEAR WRITE while adjusting MANUAL VER-NIER, center the display indication (a narrow spike).
- c. Connect a properly terminated frequency synthesizer to the input of the 3580A. Adjust the synthesizer for a 10 kHz -20 dBV signal (-20 dBm 900Ω for Option 002) output. (See Table 5-3.) Momentarily press DISPLAY-CLEAR WRITE.
- d. Adjust the FINE FREQUENCY control for a peak display indication.

Table 5-10. Frequency Response Tests.

| INPUT INPUT LEVEL SENSITIVITY | | | DISPLAY INDICATION (0 dB = full scale; 1 dB/DIV) | | | | | |
|-------------------------------|----------|-------------------|--|-------------------|-------------------|-------------------|-------------------|-------------------|
| (according to white marker) | STD. | OPT. 002 900 Ω | 10 kHz | 10 Hz | 20 Hz | 1 kHz | 20 kHz | 50 kHz |
| O dB | O dBV | O dBm | CAL | -1 dB ± .3 dB | - 1 dB ± .3 dB | - 1 dB ± .3 dB | – 1dB ± .3 dB | - 1 dB ± .3 dB |
| – 10 dB | - 10 dBV | – 10 dBm | CAL | -1 dB ± .5 dB | - 1 dB ± .3 dB | - 1 dB ± .3 dB | - 1 dB ± .3 dB | – 1 dB ± .5 dB |
| – 20 dB | - 20 dBV | – 20 dBm | CAL | - 1 dB ± .5 dB | -1 dB ±.3 dB | -1 dB ± .3 dB | - 1 dB ± .3 dB | – 1 dB ± .5 dB |
| – 30 dB | - 30 dBV | – 30 dBm | CAL | - 1 dB ± .5 dB | -1 dB ± .3 dB | -1 dB ±.3 dB | - 1 dB ± .3 dB | – 1 dB ± .5 dB |
| – 40 dB | - 40 dBV | – 40 dBm | CAL | −1 dB ± .5 dB | - 1 dB ± .3 dB | -1 dB ± .3 dB | - 1 dB ± .5 dB | – 1 dB ± .5 dB |

- e. Reposition the following front panel controls:
 AMPLITUDE MODE....LOG 1 dB/DIV
- f. Readjust the FINE FREQUENCY control for a peak display indication of the 10 kHz input. Adjust CAL 10 kHz for a full scale 0 dB display, if not already so adjusted.
- g. Slowly rotate MANUAL VERNIER CW until the display dot has dropped 3 dB in amplitude. (Remember, the display is calibrated 1 dB/DIV). This is the upper 3 dB point of the filter.
- h. Read and record the frequency in the FREQUEN-CY DISPLAY. Rotate the MANUAL VERNIER CCW until the display dot has reached the lower 3 dB point of the filter.
- i. Subtract the frequency now in the FREQUENCY DISPLAY from the previously recorded frequency. The result should be 300 Hz \pm 45 Hz.
- j. Repeat Steps f through i for the 100 Hz, 30 Hz and 10 Hz filters. See Table 5-11 for the start frequency of the source, FREQUENCY setting, RESOLUTION BANDWIDTH, FREQ. SPAN/DIV, and the test limits. At the start of each new bandwidth setting, always center the display with MANUAL VERNIER, and adjust the FREQUENCY controls, and CAL 10 kHz for a full scale, peak display at the appropriate start frequency. Then make the appropriate adjustments for the upper and lower 3 dB points.

Table 5-11. 300 Hz thru 10 Hz Bandwidth Tests.

| SOURCE START RESOLUTION FREQ. and 3580A BANDWIDTH FREQUENCY | | FREQ. and 3580A BANDWIDTH | |
|---|--------|---------------------------|----------------|
| 10 kHz | 300 Hz | 50 Hz | 300 Hz ± 45 Hz |
| 1 kHz | 100 Hz | 50 Hz | 100 Hz ± 15 Hz |
| 1 kHz | 30 Hz | 10 Hz | 30 Hz ± 4.5 Hz |
| 1 kHz | 10 Hz | 5 Hz | 10 Hz ± 1.5 Hz |

k. Using Table 5-12 and the same technique used for the 300 Hz, 100 Hz, 30 Hz, and 10 Hz Bandwidths, test the 60 dB Bandpass of the 3 Hz and 1 Hz filters. However, use AMPLITUDE MODE.....LOG 10 dB/DIV and measure the frequency difference between the 60 dB points. As before, always adjust the FINE FREQUENCY control and CAL 10 kHz for a peaked full scale display before attempting to measure the 60 dB bandwidths. If the display is noisy at the 60 dB points, use DISPLAY SMOOTHING.....MAX. Note: The display is now calibrated 10 dB/DIV.

Table 5-12. 3 Hz and 1 Hz Bandwidth Tests.

| SOURCE START FREQ. and 3580A FREQUENCY | RESOLUTION BANDWIDTH | FREQ. SPAN/DIV | 60 db Bandpass Test Limits |
|--|-------------------------|----------------|-------------------------------|
| 1 kHz | 3 Hz | 5 Hz | 30 Hz ± 4.5 Hz |
| 1 kHz | 1 Hz | 5 Hz | 10 Hz ± 1.5 Hz |

Revised: September 1987

5-30. Dynamic Range Tests (Noise Tests). Δ 16

5-31. Dynamic range is the ability of the instrument to detect large and small signals and display them simultaneously. The range and accuracy of the amplifiers is a determing factor. This specification was tested in the Amplitude Tests (Paragraph 5-18). The instrument noise and spurious responses are the other determining factors of dynamic range. These tests verify these parameters. If the instrument will not pass any of these tests, see Section VII for troubleshooting information. There are no adjustments for these specifications.

Equipment Required:

Frequency Synthesizer (-hp- Model 3320B, 50 ohms) 50 Ohm Termination (-hp- 11048C) Bandpass Filter (White Model 2640) Proper input resistor for filter (550 Ω \pm 10%, Part No. 0698-4456) 1% 1 k Ω film resistor (-hp- Part No. 0757-0280)

5-32. Noise Level Tests.

- a. Connect the 1 $k\Omega$ resistor across the INPUT terminals of the 3580A. Disconnect all signal sources.
- b. Position the following front panel controls: (Only those controls printed in **BOLD** require a change from the previous tests).

| ADAPTIVE SWEEPOFF |
|---|
| DICTION DWELL |
| DISPLAYAll pushbuttons released |
| AMPLITUDE MODELOG 10 dBV/DIV |
| AMPLITUDE REF LEVELNORMAL |
| dDV/LIN dD coop |
| $dBV/LIN-dBm 600\OmegadBV/LIN$ |
| INPUT SENSITIVITY -70 dB |
| VERNIER (Amplitude)CAL |
| |
| (Fully CW) |
| START - CTR START |
| RESOLUTION BANDWIDTH300 Hz |
| DISPLAY SMOOTHINGMAX |
| FREQ. SPAN/DIV |
| CHIEFED WITH COLVE |
| SWEEP TIME/DIV 5 SEC |
| SWEEP MODERESET |
| FREQUENCY00000 Hz |
| next |
| CHIEFD MODE |
| SWEEP MODE MANUAL |
| SWEEP MODEMANUAL Option 002: Set dRm 900 0/LIN_dRm |
| Option 002: Set dBm 900 $\Omega/LIN-dBm$ |
| Option 002: Set dBm 900 Ω /LIN-dBm 600 Ω switch to dBm 900 Ω ; set INPUT |
| Option 002: Set dBm 900 Ω /LIN – dBm 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL; turn the TRACK- |
| Option 002: Set dBm 900 $\Omega/LIN-dBm$ |

- c. Adjust the MANUAL VERNIER full CCW. Adjust the FINE FREQUENCY control for a peak display.
- d. Adjust the MANUAL VERNIER for a display indiction at 10 kHz (2 major divisions from left graticule). Momentarily press the following control:

DISPLAY......CLEAR WRITE

e. The display indication should always be less than -130 dB (6 major divisions down from top graticule, since Full Scale = -70 dB).

| f. Reposition the following front panel control: RESOLUTION BANDWIDTH30 Hz | t. Adjust FINE FREQUENCY control for a peak response at the leftmost graticule. Reposition the following front panel control: |
|--|--|
| g. Momentarily press the following control: | DISPLAY SMOOTHINGMAX |
| DISPLAYCLEAR WRITE | u. Adjust the MANUAL VERNIER for a display in- |
| The display indication should be less than -140 dB (7 major divisions down from top graticule). | dication at 10 Hz (2 major divisions from leftmost graticule). Momentarily press the following control: |
| h. Reposition the following control: | DISPLAYCLEAR WRITE |
| FREQ. SPAN/DIV0.1 kHz | v. The display indication should be less than -135 dB (6.5 major divisions down from top graticule). |
| i. Adjust MANUAL VERNIER full CCW. Adjust the FINE FREQUENCY control for a peak display indication. | Remove the 1 k Ω resistor from the input terminals. |
| dication. | 5-33. Noise Sideband Test. |
| j. Adjust MANUAL VERNIER for a display indication at 100 Hz (1 major division from leftmost graticule). Momentarily press the following control: | a. Reposition the following controls: |
| DISPLAYCLEAR WRITE | SWEEP MODERESET INPUT SENSITIVITYCAL |
| | FREQUENCY10000 Hz |
| k. The display indication should be less than -132 dB (6.2 major divisions down from top graticule). | START-CTRCTR DISPLAY SMOOTHINGMIN FREQ. SPAN/DIV5 Hz |
| l. Adjust MANUAL VERNIER for a display indication of 1 kHz (far right graticule). Momentarily press the following control: | SWEEP TIME/DIV10 SEC SWEEP MODEMAN |
| DISPLAYCLEAR WRITE | b. By alternately pressing and releasing DISPLAY-CLEAR WRITE while adjusting MANUAL VER-NIER, center the display indication (a narrow spike). |
| m. The display indication should be less than -140 dB (7 major divisions down from top graticule). | c. Adjust the FINE FREQUENCY control for a peak display of this spike. |
| n. Reposition the following control: | • |
| RESOLUTION BANDWIDTH1 Hz | d. Reposition the following controls: |
| Momentarily press the following control: | SWEEP MODESING |
| DISPLAYCLEAR WRITE | e. After waiting for the sweep to be completed (100 sec.), verify that the noise on the display \pm 10 Hz (\pm 2 major divisions) away from the 10 kHz CAL signal (in |
| o. The display indication should be less than -150 dB (8 major divisions down from top graticule). | center of display) is at least 70 dB below the CAL signal. 5-34. Spurious Response Test. \triangle 19 |
| p. Readjust MANUAL VERNIER for a display indication at 100 Hz (1 major division from leftmost graticule). Momentarily press the following control: | NOTE: Not required for instruments having serial number $> 2030A05376$ or older units which have been updated. |
| DISPLAYCLEAR WRITE | a. Reposition the following controls: |
| q. The indication should be less than -143 dB (7.3 major divisions down from top graticule). | INPUT SENSITIVITY |
| r. Reposition the following controls: | SWEEP TIME/DIV5 SEC SWEEP MODERESET |
| DISPLAY SMOOTHINGMIN FREQ. SPAN/DIV5 Hz | FREQUENCY00000 Hz |
| s. Adjust MANUAL VERNIER FULL CCW. Momentarily press the following front panel control: | b. Momentarily press: DISPLAYCLEAR WRITE |
| DISPLAYCLEAR WRITE | |
| DISTERS | c. Reposition the following controls: |

SWEEP MODE......MANUAL and momentarily press:

DISPLAY......CLEAR WRITE

- d. Connect the frequency synthesizer (use proper output impedance needed for the bandpass filter) to the input of the bandpass filter. Adjust the filter for a 5 kHz center frequency and adjust the synthesizer for a 5 kHz output. (For a 50 ohm source and the White 2640 filter, connect a 550 Ω resistor (\pm 10%) in series between the filter and synthesizer. This gives the 600 Ω source impedance required by the White filter (See Figure 5-3).
- e. Connect the output of the filter to the input of the 3580A. Always terminate properly if required. (The White Model 2640 filter requires no output termination. See Figure 5-3).
- f. Adjust MANUAL VERNIER for a display indication at 5 kHz ($2\frac{1}{2}$ major divisions from left graticule). Adjust the source level for a -20 dBV (full scale) input to the 3580A (For the White 2640 filter and a 50Ω source, this corresponds to -16.99 dBm 50 Ω level on the source). Readjust MANUAL VERNIER for a peak display. Adjust CAL 10 kHz for a full scale display.
 - g. Reposition the following controls:

SWEEP MODE......SING

h. After waiting for one complete sweep (50 sec.) verify that all responses other than the zero response are at least 80 dB below the 5 kHz response.

5-35. Line Related Spurious Test.

Specification:

>80dB below input reference level or -135 dBV (0.18 μ V).

a. Disconnect the Synthesizer and Bandpass Filter from the 3580A Input. Turn off all unnecessary equipment located near the 3580A. This especially includes large current users such as soldering irons, blowers, motors, etc.

- b. Using a short piece of wire, connect a short across the 3580A INPUT terminals.
 - c. Reposition the following controls:

| INPUT SENSITIVITY | |
|----------------------|------|
| RESOLUTION BANDWIDTH | 3 Hz |
| FREQ. SPAN/DIV | 5 Hz |
| SWEEP MODE | MAN |
| MANUAL VERNIER | |
| DISPLAY SMOOTHING | MAX |
| START-CTR | |

NOTE

If the power-line frequency is 50 Hz, substitute the following 3580A frequencies for Steps d and f.

Step d: 50 Hz Step f: 100 Hz Step f: 150 Hz

- d. With the FINE FREQUENCY control, tune the 3580A frequency to 60 Hz.
- e. Press CLEAR WRITE, then slowly turn the MANUAL VERNIER to obtain a peak reading. The peak should be more than 65 dB below full scale (-135 dB).
- f. Repeat Steps d and e substituting 120 Hz, and 180 Hz for the 3580A frequencies.

NOTE

If the instrument fails this test double check that the input short is as small as possible, that all power line current is kept at a minimum, and that all covers are tightly secured on the 3580A

5-36. IF Feedthru and Zero Beat Response Tests. Δ 16

5-37. These tests verify the ability of the instrument to reject a 100 kHz signal at the input and also how well the Zero Beat Response is suppressed. Proceed to the

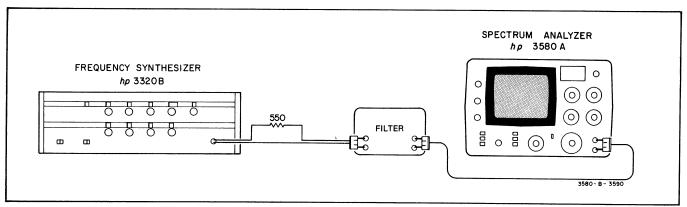


Figure 5-3. Spurious Response Test.

Mixer Balance Adjustments (Paragraph 5-81) of the Adjustment Procedures if the Zero Beat Response is too large. Proceed to Section VII for troubleshooting information if there is too much IF Feedthru.

Equipment Required:

Frequency Synthesizer (-hp- Model 3320B, 50 ohm)

- a. Reconnect the synthesizer to the 3580A. Do not terminate. Adjust the source for a 10 volt 100 kHz output (+ 26.99 dBm 50 ohms setting on 3320B and unterminated).
- b. Position the following front panel controls: (Only those controls printed in **BOLD** require a change from the previous test).

| SWEEP MODE | RESET |
|---------------------|--------------|
| ADAPTIVE SWEEP | OFF |
| DISPLAYAll pushbutt | ons released |
| AMPLITUDE MODELOG 1 | 0 dBV/DIV |
| AMPLITUDE REF LEVEL | .NORMAL |
| dBV/LIN-dBm 600 Ω | dBV/LIN |
| INPUT SENSITIVITY | + 20 dB |
| VERNIER (Amplitude) | CAL |
| | |

| START-CTR | START |
|-----------------------------|-------|
| RESOLUTION BANDWIDTH | 3 Hz |
| DISPLAY SMOOTHING | MIN |
| FREQ. SPAN/DIV | 20 Hz |
| SWEEP TIME/DIV | 5 SEC |
| FREQUENCY | |
| nevt | |

SWEEP MODEMANUAL Option 002: Set dBm 900 Ω /Lin-dBm 600 Ω switch to dBm 900 Ω; set INPUT MODE switch to UNBAL.

- c. Adjust MANUAL VERNIER for a response in the center of the screen. The display indication should be at least 70 dB below full scale to verify the IF Feedthru specification of Table 1-1. If the instrument fails this test, see Section VII for troubleshooting information.
- d. Disconnect the synthesizer. Reposition the following front panel controls:

| RESOLUTION BANDWIDTH | 300 Hz |
|----------------------|--------|
| FREQ. SPAN/DIV | 5 kHz |
| SWEEP MODE | .RESET |

e. Momentarily press the following front panel control:

DISPLAY......CLEAR WRITE

f. Adjust FINE FREQUENCY control for a maximum display indication on the left graticule. This display should be at least 30 dB (3 major divisions) below full scale to verify the Zero Beat Response specification of Table 1-1. If the instrument fails this test, go to the Mixer Balance Adjustments (Paragraph 5-81) of the Adjustment Procedures.

5-38. Input Impedance Tests. Δ 16

5-39. These tests verify the Input Impedance characteristics of Table 1-2. Since there is no adjustment for this parameter, see Section VII for troubleshooting information if the instrument fails this test.

Equipment required:

 $1 M\Omega \pm 1\%$ film resistor (-hp- Part No. 0757-0344)

a. Position the following front panel controls. (Only those controls printed in BOLD require a change from the previous tests.)

| ADAPTIVE SWEEPOFF DISPLAYAll pushbuttons released AMPLITUDE MODELOG 1 dBV/DIV AMPLITUDE REF LEVELNORMAL dBV/LIN-dBm 600 \(\OmegadBV/LIN \) INPUT SENSITIVITY0 dB |
|---|
| TERMED (A. II. I.) |
| VERNIER (Amplitude)CAL |
| , . , |
| START-CTR START |
| DISPLAY SMOOTHINGMIN |
| RESOLUTION BANDWIDTH10 Hz |
| FREQ. SPAN/DIV1 kHz |
| SWEEP TIME/DIV 5 SEC |
| SWEEP MODE |
| FREQUENCY00000 Hz |
| next |
| SWEEP MODEMANUAL |
| Option 002: Set dBm 900Ω/LIN-dBm 600Ω |
| switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL. |

b. Connect the rear panel TRACKING OSC OUT to the front INPUT terminals of the 3580A. Adjust the rear panel TRACKING OSC LEVEL control fully CW. Adjust MANUAL VERNIER for a 1 kHz display indication (1 major division from left graticule). Readjust the TRACKING OSC LEVEL control for a full scale 0 dB display. Momentarily press the following control:

DISPLAY......CLEAR WRITE

- c. Connect the 1 M Ω resistor in series between the TRACKING OSC OUT and front panel INPUT terminals. The display indication should drop 6 dB \pm .3 dB (6 major divisions \pm .3 major divisions) to verify an input impedance of 1 M Ω .
 - d. Reposition the following front panel control:

INPUT SENSITIVITY.....-10 dB

e. Readjust the rear panel TRACKING OSC LEVEL control for a full scale display. Adjust MANUAL VER-NIER for a display indication at 10 kHz (far right display graticule). DO NOT REMOVE 1 $M\Omega$ RESISTOR. Momentarily press the following front panel control:

DISPLAY......CLEAR WRITE

- f. 1) Std. 3580A: The amplitude should drop 3 dB ± 1 dB, verifying that the input shunt capacitance is 30 pF, nominal.
 - 2) Option 002: The amplitude should drop 4 dB ± 1 dB, verifying that the input shunt capacitance is 40 pF, nominal.
- g. Disconnect the cable connected between the TRACKING OSC OUT and the front panel INPUT terminals.

5-40. Output Tests.

5-41. These tests verify the Output specifications of the 3580A listed in Table 1-1.

Equipment Required:

Digital Multimeter (-hp- Model 34740/34702) Distortion Analyzer (-hp- Model 333A)

5-42. TRACKING OSC OUTPUT Tests.

a. Position the following front panel controls. (Only those controls printed in **BOLD** require a change from the previous tests).

| ADAPTIVE SWEEPO | FF |
|--|-----|
| DISPLAYAll pushbuttons release | |
| AMPLITUDE MODELOG 10 dBV/D | |
| AMPLITUDE REF LEVELNORMA | ΑL |
| dBV/LIN-dBm 600 ΩdBV/L | IN |
| INPUT SENSITIVITY+20 | dB |
| VERNIER (Amplitude)CA | ٩L |
| (Fully C | |
| START-CTR STAI | |
| RESOLUTION BANDWIDTH10 | Ηz |
| DISPLAY SMOOTHINGM | |
| FREQ. SPAN/DIV5 kl | H7 |
| SWEEP TIME/DIV5 SI | 30 |
| SWEEP MODERESI | DT. |
| EDECTIENCY 00000 | LT. |
| FREQUENCY00000 1 | ΠZ |
| Option 002: Set dBm 900 Ω/LIN-dBm 600 | Ω |
| switch to dBm 900 Ω ; set INPUT MOI | ÞΕ |
| | |

- b. Connect the multimeter (AC mode 100 volt range) and a 600 Ω termination (11095A) to the rear panel TRACKING OSC OUT. Adjust the FREQUENCY dial for 50Hz (300 Hz for Option 002). Adjust the rear panel TRACKING OSC LEVEL control for a 1.00 volt reading on the multimeter.*
- c. Adjust the FREQUENCY control to 50.0 kHz (20.0 kHz for Option 002 instruments). Verify that the multimeter reads $1.00 \text{ volt } \pm .06 \text{ volts}$.
 - d. Reposition the following front panel controls:

| AMPLITUDE MODE | LIN |
|----------------------|-----------|
| INPUT SENSITIVITY | 2 V |
| FREQUENCY | .00000 Hz |
| RESOLUTION BANDWIDTH | 30 Hz |
| SWEEP MODE | MANUAL |

e. Connect the rear panel TRACKING OSC OUT to the front panel INPUT terminals. Momentarily press the following control:

DISPLAY.....CLEAR WRITE

- f. By alternately pressing and releasing DISPLAY-CLEAR WRITE while adjusting MANUAL VER-NIER, center the display indication (a narrow spike).
- g. Adjust the rear panel TRACKING OSC LEVEL control for a full scale 2 V display. Reposition the following front panel control:

RESOLUTION BANDWIDTH......3 Hz

- h. The display indication should drop no lower than 1 V (5 major divisions) to verify the frequency accuracy of the tracking oscillator. If the tracking oscillator frequency is out of tolerance, remove the top cover and adjust A2C4 for a peak display indication.
 - i. Reposition the following front panel control:

FREQ. SPAN/DIV...........0.1 kHz

j. Adjust MANUAL VERNIER for a 1 kHz display indication (indication on far right display graticule). Momentarily press the following front panel control:

DISPLAY......CLEAR WRITE

- k. Connect the TRACKING OSC OUT to the IN-PUT of the distortion analyzer. Adjust the TRACKING OSC LEVEL control fully CW.
- l. Reference the TRACKING OSC OUT to 0 dB on the distortion analyzer. (For the -hp- 333A Distortion Analyzer, position the following controls:

| FUNCTION | ET LEVEL |
|------------------|----------|
| METER RANGE | 0 dB |
| FREQUENCY RANGE | |
| FREQUENCY | |
| HIGH PASS FILTER | |

Adjust the SENSITIVITY and VERNIER controls of the distortion analyzer for a 0 dB meter indication. Set the distortion analyzer FUNCTION switch to DISTORTION.)

- m. Measure the distortion in dB by nulling the distortion analyzer.
- n. Adjust the FREQUENCY and BALANCE controls for a meter null. Use automatic nulling if available.

switch to UNBAL.

^{*}For measurements below 50 Hz, use a low frequency Digital Voltmeter such as the -hp- Model 3480/3484 with true rms.

o. The total distortion indication should be at least 40 dB below the reference level. If it is not, perform the Mixer Balance Adjustments (Paragraph 5-81). Disconnect the distortion analyzer from the 3580A.

5-43. RECORDER Output Tests.

- a. Connect the multimeter (DC mode, 100 volt range) to the rear panel X-AXIS RECORDER output. Adjust MANUAL VERNIER fully CCW.
 - b. The multimeter should read 0 Vdc \pm .15 V.
- c. Adjust the MANUAL VERNIER fully CW. The multimeter reading should be 5 Vdc ± .15 V.
 - d. Reposition the following front panel control:

RESOLUTION BANDWIDTH.....30 Hz

- e. Reconnect the TRACKING OSC OUTPUT to the INPUT terminals of the 3580A and readjust the rear panel LEVEL control for a full scale display (on the far right graticule). Use DISPLAY-CLEAR WRITE, if necessary, to clear all unwanted data from the display.
- f. Connect the multimeter (DC mode, 100 volt range) to the rear panel Y-AXIS RECORDER output. The multimeter reading should be 5.00 Vdc ± .15 V.*
- g. Disconnect the TRACKING OSC OUT from the INPUT terminals. The voltmeter should now read 0 volts dc ± .15 V. Disconnect the multimeter from the 3580A.*

5-44. Balanced Input Tests (Option 002 only). Δ 16

5-45. These tests verify the Balanced Input specifications for the Option 002 instrument. If the instrument fails these tests, see Section VII for troubleshooting information since there are no adjustments for the parameters tested.

Equipment Required:

Frequency Synthesizer (-hp- Model 3320B, 50 ohm) 50 Ohm Termination (-hp- 11048C)

Two 453 ohm 1% resistors (-hp- Part No. 0698-3510)

5-46. Common Mode Rejection Test.

a. Position the following front panel controls:

ADAPTIVE SWEEP.....OFF DISPLAY.....All pushbuttons released AMPLITUDE MODE..LOG 10 dBV/DIV AMPLITUDE REF LEVEL....NORMAL dBm 900Ω /

LIN-dBm 600 ΩdBm 900 Ω /LIN INPUT SENSITIVITY..... 0 dB

| VERNIER (Amplitude) | CAL |
|-----------------------------|----------|
| (F | ully CW) |
| INPUT MODE | BRDG |
| START-CTR | |
| RESOLUTION BANDWIDTH | 3 Hz |
| DISPLAY SMOOTHING | MIN |
| FREQ. SPAN/DIV | 10 Hz |
| SWEEP TIME/DIV | 5 SEC |
| SWEEP MODE | |
| FREQUENCY | 00000 Hz |
| next | |
| SWEEP MODE | MAN |

- b. Adjust the frequency synthesizer for a 60 Hz, +5 dBm 900 Ω output (+ 17.55 dBm/50 ohms). Connect the synthesizer (properly terminated) to the INPUT of the 3580A.
- c. Slowly adjust MANUAL VERNIER to the 60 Hz signal which will appear as a peak on the sixth major division from the left. Momentarily press the following front panel control:

DISPLAY......CLEAR WRITE

- d. Adjust the VERNIER (Amplitude) for a full scale 0 dB display.
- e. Disconnect the synthesizer from the 3580A and connect two 453 ohm resistors in series between the IN-PUT terminals. (See Figure 5-4.)
- f. Connect the synthesizer to the junction of the two resistors and to the chassis on the rear panel as shown in Figure 5-4. (Do not change the synthesizer amplitude setting.)
- g. The display indication on the 3580A should be at least 70 dB below full scale (10 dB/DIV).

5-47. Frequency Response Test.

- a. Disconnect the resistors from the 3580A INPUT terminals and reconnect the synthesizer (properly terminated in 50 ohms). Adjust the source for a 0 dBm 900 Ω (+12.55 dBm 50 Ω) 10 kHz signal.
 - b. Reposition the following front panel controls:

| SWEEP MODE | RESET |
|---------------------|----------|
| FREQUENCY | 10000 Hz |
| START-CTR | CTR |
| VERNIER (Amplitude) | |
| SWEEP MODE | MAN |

c. By alternately pressing and releasing DISPLAY-CLEAR WRITE while adjusting MANUAL VER-NIER, center the display indication (a narrow spike). Adjust the FINE FREQUENCY control for a peak display of the 10 kHz input signal.

- d. Reposition the following front panel control:

 AMPLITUDE MODE....LOG 1 dB/DIV
- e. Readjust the FINE FREQUENCY control for a peak display indication. Adjust VERNIER (Amplitude) for a full scale -1 dB display indication (1 major division down from full scale).
- f. Adjust the frequency synthesizer and 3580A FRE-QUENCY to the frequencies given by Table 5-13. Always peak the display indication with the FINE FRE-QUENCY control and check for proper amplitude accuracy.

Table 5-13. Balanced Input Frequency Response Tests.

| FREQUENCY | INPUT 900 Ω | DISPLAY ACCURACY |
|----------------------|----------------|--|
| 10 kHz | 0 dBm | CAL to -1 dB |
| 40 Hz Δ ₁ | 0 dBm | − 1 dB ± .5 dB (± .5 major divisions) |
| 300 Hz | 0 dBm | -1 dB ± .5 dB |
| 1 kHz | 0 dBm | -1 dB ± .5 dB |
| 20 kHz | 0 dBm | -1 dB ± .5 dB |

Δ₁ See Backdating.

5-48. ADJUSTMENT PROCEDURE.

5-49. This portion of Section V contains complete Adjustment Procedures for the Model 3580A Specturm Analyzer:

POWER SUPPLY CHECKS AND AD-JUSTMENTS (Paragraph 5-53).

DISPLAY ADJUSTMENTS (Paragraph 5-68).

SWEEP ALIGNMENT (Paragraph 5-63). \triangle 16 LINE GENERATOR ADJUSTMENTS (Paragraph 5-68).

I.F. FILTER ALIGNMENT (Paragraph 5-70).

AMPLITUDE CALIBRATION (Paragraph 5-74).

MIXER BALANCE ADJUSTMENTS (Paragraph 5-81).

ADAPTIVE SWEEP MARKER ADJUST-MENT (Paragraph 5-84).

5-50. TEST POINT AND ADJUSTMENT LOCATIONS.

- 5-51. Test point and adjustment locations are shown in Figure 5-9 at the end of Section V. Most of the test points and adjustments are easily accessible with the outer covers removed. In some cases it will be necessary to remove the inner cover and place the appropriate pc boards on extenders. Set the 3580A POWER switch to OFF when removing or replacing a pc assembly.
- 5-52. The Adjustment Procedure is written in a logical sequence. If the instrument is known to be completely out of calibration, the sequence should be strictly followed. Many times, however, only certain adjustments need to be made. The Performance Tests have been written in such a manner that they will lead you to the proper adjustment. In addition, a brief description of each adjustment is given. Read through the procedures carefully, doing only those that are necessary. Take careful note of any previous adjustments which may affect a future adjustment.

NOTE

Always test the low voltage power supply before performing any calibration. All test measurements should be made with respect to circuit ground, which is available at any point on the instrument chassis. Adjustments should not be made until the instrument has had one hour of continuous warm-up.

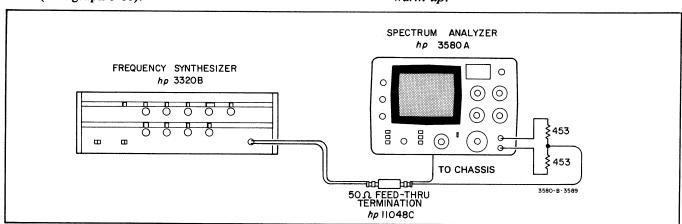


Figure 5-4. Common Mode Rejection Test.

5-53. POWER SUPPLY TESTS AND ADJUSTMENTS.

5-54. These tests and adjustments check the operation of the low voltage +10 Vdc and -10 Vdc regulated power supplies and set the level of the high voltage -2915 Vdc regulated power supply. The low voltage power supply tests should be performed prior to all other adjustments. In addition, the High Voltage -2915 Vdc power supply voltage should be tested if any of its components were changed or if the instrument will not pass the Frequency Tests (Paragraph 5-9) or Amplitude Tests (Paragraph 5-18) of the Performance Tests.

5-55. Recommended Test Equipment:

AC/DC Digital Multimeter (-hp- Model 34740A and 34702A plug-on)

High Voltage DC Probe for above multimeter, calibrated to 1000 V DC Standard (-hp- Model 11045A Probe and -hp- Model 740B DC Standard)

or

Precision .1% High Voltage Probe and appropriate DVM (-hp- 3440A-K05 High Voltage Probe and -hp- Model 3440A DVM)

5-56. ± 10 Volt Power Supply Tests.

- a. Connect the digital multimeter (DC mode 10 volt range) to the red lead (pin 12) at the A13 board connector. The multimeter reading should be \pm 10.000 V \pm .050 V. If it is not, refer to the Factory-Selected Components information in Section VII.
- b. Connect the digital multimeter (DC mode 10 volt range) to the violet lead (pin 10) at the A13 board connector. The dc voltage present should be $-10.000~\rm V~\pm~.050~\rm V.$
- c. Test the ac ripple voltage present on the above two leads with the digital multimeter. There should be less than .1 mV ac difference between the reading obtained on each lead and that obtained with a short circuit to the multimeter.

5-57. High Voltage Power Supply Tests.

WARNING

The voltages involved in the following measurments may cause serious injury or even death. USE EXTREME CAUTION.

- a. Turn the 3580A POWER switch to the OFF position.
- b. Connect up the multimeter's high lead to the green wire at J3 pin 4 on the rear of the A13 board. The low lead is connected to the black wire at J3 pin 2.

- c. Turn the 3580A POWER switch to the ON (AC) position. Caution should now be used as potentially dangerous voltages are exposed.
- d. Check that the multimeter is reading between 150V and 158V. If the reading is outside this range then adjust A8R1 (HV ADJ) until the multimeter reads within this range.

NOTE

This adjustment affects the Sweep Alignment (Pargraph 5-63), as well as the Amplitude Calibration (Pargraph 5-74). Repeat the Frequency Tests (Paragraph 5-9) and Amplitude Tests of the Performance Tests to determine if these additional adjustments need to be made.

e. Turn the 3580A POWER switch to the OFF position and remove the multimeter leads from the unit.

5-58. Display Adjustments.

5-59. These adjustments set the proper intensity limits, astigmatism, and trace alignment on the CRT. In many cases, these display parameters will require no adjustments.

5-60. Intensity Limit Adjustment.

a. Turn the 3580A power switch to OFF. Unplug the A13J3 connector. Remove the nylon access screw from the top of the high voltage power supply box. Turn the front panel INTENSITY control to the "9 o'clock" position. Turn the 3580A POWER switch back to ON (ac).

WARNING

The voltages present inside the high voltage power supply box can cause serious injury. Never place an uninsulated conductive tool or object inside this box while the instrument is turned on.

- b. Using an insulated non-metallic tuning wand, such as -hp- Part No. 8710-0033, adjust A11R1 (INTENSI-TY LIMIT, inside high voltage power supply box) so that the dot on the CRT just disappears.
- c. Replace the nylon screw in the high voltage power supply box.

5-61. Astigmatism Adjustment.

- a. Adjust the front panel focus fully CCW. Turn the front panel INTENSITY adjust to about 10 or 11 o'clock so that the dot on the CRT is bright enough to see, but does not form a "halo".
- b. Adjust A8R2 (ASTIG. ADJ.) for the largest circular dot.
- c. Turn the 3580A POWER switch to OFF. Reconnect the connector to A13J3. Turn the 3580A POWER switch back to ON (ac).

5-62. Trace Alignment Adjustment.

a. Position the 3580A front panel controls as follows:

ADAPTIVE SWEEP......Centered DISPLAY.....All pushbuttons released AMPLITUDE MODE....LOG 10 dB/DIV APLITUDE REF LEVEL.....NORMAL $dBV/LIN-dBm\ 600\ \Omega.....dBV/LIN$ INPUT SENSITIVITY..... - 30 dB VERNIER (Amplitude).....CAL (Fully CW) FREQUENCY......00000 Hz START-CTR START RESOLUTION BANDWIDTH.....300 Hz DISPLAY SMOOTHING.....MIN FREQ. SPAN/DIV......0.2 kHz SWEEP TIME/DIV......0.1 SEC SWEEP MODE......REP Option 002: Set dBm 900 Ω /LIN-dBm 600 Ω

switch to dBm 900 Ω; set INPUT MODE switch to UNBAL.

- b. Adjust the front panel ADAPTIVE SWEEP for a line in the middle of the display. Adjust the front panel FOCUS control for the narrowest and sharpest line.
- c. Set the rear panel ROTATION DISPLAY ADJUST (A40R39) initially to its center position. Δ 21 Adjust A13R5 (TRACE ALIGN) for a level trace. If unable to achieve this, readjust the rear panel DISPLAY ADJUST rotation control.

5-63. Sweep Alignment.

5-64. These adjustments calibrate the frequency sweep limts. They should be done if the Frequency Tests (Paragraph 5-9) or Sweep Tests (Paragraph 5-13) of the Performance Tests cannot be passed by the instrument.

In addition, the adjustment should be made if the high voltage supply was previously adjusted.

5-65. Recommended Test Equipment.

Digital Multimeter (-hp- Model 34740A and 34702A plug-on)

Oscilloscope (-hp- Model 180A with 1801A and 1820A plug-ins)

5-66. Linear Sweep Adjustments.

a. Position the following front panel controls: (Only those controls printed in BOLD require a change from the previous adjustments).

| ADAPTIVE SWEEPOFF |
|--|
| DISPLAYAll pushbuttons released |
| AMPLITUDE MODELOG 10 dB/DIV |
| $dBV/Lin-dBm 600 \OmegadBV/LIN$ |
| INPUT SENSITIVITYCAL |
| VERNIER (Amplitude)CAL |
| (Fully CW) |
| START-CTR START |
| RESOLUTION BANDWIDTH300 Hz |
| DISPLAY SMOOTHINGMIN |
| FREQ. SPAN/DIV0 kHz |
| SWEEP TIME/DIV0.1 SEC |
| SWEEP MODERESET |
| FREQUENCY000000 Hz |
| |
| Option 002: Set dBm 900 Ω /LIN-dBm 600 Ω |
| switch to dBm 900 Ω ; set INPUT MODE |
| switch to UNBAL. |

- b. Remove the inner circuit board shield (covering A2-A5). Connect the multimeter (DC mode 10.0 volt range) to A2TP4.
- c. Change the Sweep mode to LOG ZERO. Turn the fine FREQUENCY control fully CCW; turn the fine FREQUENCY control five revolutions CW. (This assures the control of being close to its center positions.)
- d. Adjust A2L3 (100 kHz FREQ. ADJ.) for a FRE-QUENCY display of 00000 Hz ± 10 Hz; this is not a critical adjustment.
- e. Adjust A2L1 (100 kHz VCO ADJ.) for a voltage reading on the multimeter between -1.4 V and -1.5 V. Record the reading.
- f. Set the SWEEP MODE control to MANUAL and turn the MANUAL VERNIER control fully CCW.
 - g. Reposition the following front panel control:

FREQ. SPAN/DIV 5 kHz

- h. Adjust A3R54 (INTEGRATOR BALANCE) for a frequency display of 00000 Hz \pm 1 Hz.
- i. Position the front panel MANUAL VERNIER control fully CW.
- j. Adjust A2r75 (BUFFER AMP GAIN ADJ.) for a display of 50000 Hz \pm 1 Hz.
- k. Adjust A2R100 (VCO RANGE SET) for a reading on the multimeter equal to -1.45 V to -1.75 V.
- 1. Repeat Steps j and k as necessary to meet the frequency and voltage specifications.
- m. Position the front panel MANUAL VERNIER fully CCW.
 - n. Reposition the following front panel controls:

| RESOLUTION BANDWIDTH | 300 Hz |
|----------------------|----------|
| SWEEP TIME/DIV | 2 SEC |
| SWEEP MODE | REP |
| FREQUENCY | 00000 Hz |

- o. Adjust A13R1 (HORIZONTAL GAIN ADJ.) and A13R2 (HORIZONTAL POSITION ADJ.) for a full 10 cm display. The 10 kHz signal and its harmonics should fall on the proper graticule marking $\pm \frac{1}{2}$ minor divisions (2nd, 4th, 6th, 8th and 10th graticule from the left).
- p. Connect the input of the oscilloscope to A3TP11. Set the oscilloscope input to dc coupling. Connect a jumper between A3TP3 and A3TP4.
- q. Adjust the A3R14 (RAMP COMPARATOR BALANCE) so that the output of the ramp comparator (on scope) just changes states.
 - r. Remove the jumpers from the A3 board.
 - s. Reposition the following front panel control:

SWEEP TIME/DIV0.1 sec

- t. Alternately press and release the STORE pushbutton, adjusting A8R4 (RAMP SIZE ADJ.) so that the 40 kHz harmonic of the CAL signal falls on the same point for both the STORE and non-STORE display modes.
 - u. Reposition the following front panel controls:

| SWEEP MODE | RESET |
|------------|----------|
| FREQUENCY | 50000 Hz |
| START-CTR | |

v. Record the reading in the FREQUENCY display.

- w. Reposition the following front panel control:
 - START-CTRCTR
- x. Adjust A16R52 so that the FREQUENCY reading is the same as that in Step v.
- y. Repeat Steps u through x until there is no change in the FREQUENCY display when switching from START to CTR.

5-67. Log Sweep Adjustments.

ADADTIVE CWEED

a. Position the following front panel controls: (Only those controls printed in **BOLD** require a change from the previous adjustements).

| ADAPTIVE SWEEPOFF |
|---|
| DISPLAYAll pushbuttons released |
| AMPLITUDE MODELOG 10 dB/DIV |
| AMPLITUDE REF LEVELNORMAL |
| $dBV/LIN-dBm 600 \OmegadBV/LIN$ |
| INPUT SENSITIVIYCAL |
| VERNIER (Amplitude)CAL |
| (Fully CW) |
| START-CTR START |
| RESOLUTION BANDWIDTH300 Hz |
| DISPLAY SMOOTHINGMIN |
| FREQ. SPAN/DIV 5 kHz |
| SWEEP TIME/DIV0.5 SEC |
| SWEEP MODELOG ZERO |
| FREQUENCY00020 Hz |
| Option 002: Set dBm 900 Ω/LIN-dBm 600 Ω |

option 002: Set dBm 900 Ω /LIN-dBm 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL.

b. Momentarily push:

DISPLAY......CLEAR WRITE

c. Reposition the following front panel control:

SWEEP MODE.....LOG

- d. Adjust the rear panel HORIZONTAL DISPLAY DISPLAY position control to the center of its range.
- e. Allow the 3580A to make three complete sweeps. Then adjust A3R76 (20 kHz LOG SWEEP ADJ.) so that the 20 kHz harmonic of the CAL signal falls on the 20 kHz LOG SWEEP graticule.

NOTE

After each adjustment of A3R76, wait for the 3580A to sweep through 20 kHz before attempting to readjust the setting.

5-68. Line Generator Adjustments. Δ 16

5-69. This adjustment properly aligns the line generator circuitry. The adjustment is usually not necessary, but should be done if components in the high voltage power supply are changed, or if the display exhibits overshoot to abrupt level changes.

a. Position the following front panel controls: (Only those controls printed in **BOLD** require a change from the previous adjustments).

| SWEEP MODERE ADAPTIVE SWEEP | OFF eased DIV MAL |
|--|----------------------------|
| $dBV/LIN-dBm 600 \Omega \dots dBV/dBV$ | 'LIN |
| INPUT SENSITIVITY | CAL |
| VERNIER (Amplitude) | CAL |
| (Fully | |
| START-CTR | |
| RESOLUTION BANDWIDTH30 | 0 Hz |
| DISPLAY SMOOTHING | MIN |
| FREQ. SPAN/DIV0.2 | kHz |
| SWEEP TIME/DIV0.1 | SEC |
| FREQUENCY |) Hz |
| SWEEP MODEMAN | JAL |

Option 002: Set dBm 900 $\Omega/\text{LIN-dBm }600~\Omega$ switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL.

- b. Adjust MANUAL VERNIER for a peak display signal. Note: The Amplitude VERNIER may have to be adjusted to keep the signal within the display limits.
 - c. Momentarily press:

DISPLAY......CLEAR WRITE

d. Adjust A8C1 (LINE GENERATOR ADJ.) for a single round dot in the top center of the screen.

5-70. I.F. Filter Alignment. Δ 16

5-71. This adjustment aligns the I.F. crystal filters for proper center frequency and symmetry. The TRACK-ING OSC is also precisely adjusted to 100 kHz. This adjustment should be done if the Bandwidth Tests (Paragraph 5-28) of the Performance Tests cannot be passed by the instrument. This adjustment will interact with the Amplitude Calibration (Paragraph 5-74). If it is performed, the Amplitude Tests (Paragraph 5-18) of the Performance Tests should be redone to verify whether any amplitude calibration is necessary.

Recommended Test Equipment:

Timer/Counter (-hp- Model 5328A)

5-72. Tracking Oscillator and Center Frequency Adjustments.

a. Position the following front panel controls: (Only those controls printed in **BOLD** require a change from the previous adjustments).

| SWEEP M | ODE | RESET |
|-----------------|-------|-------|
| ADAPTIVE | SWEEP | OFF |

| DISPLAYAll pushbuttons released |
|--|
| AMPLITUDE MODELINEAR |
| AMPLITUDE REF LEVELNORMAL |
| $dBV/LIN-dBm 600 \Omega \dots dBV/LIN$ |
| INPUT SENSITIVITY $\dots + 20 \text{ dB}$ |
| VERNIER (Amplitude)CAL |
| (Fully CW) |
| FREQUENCY10000 Hz |
| START-CTR START |
| DISPLAY SMOOTHINGMIN |
| RESOLUTION BANDWIDTH1 Hz |
| FREQ. SPAN/DIV0.5 kHz |
| SWEEP TIME/DIV0.5 SEC |
| SWEEP MODEMANUAL |
| Option 002: Set dBm 900 Ω/LIN-dBm 600 Ω |
| switch to dBm 900 Ω; set INPUT MODE switch to UNBAL. |

b. Set the counter to the Frequency Mode and adjust the time base/multiplier for a measurement of 100 kHz with six digits of resolution (100.000 kHz). Adjust for maximum input sensitivity and either a zero trigger level or Preset. Select ac coupling on input. If using the -hp-Model 5328A Counter, the controls should be set to:

Sample Rate: CCW
Function: FREQ A
Multiplier: 106
Channel A: Slope +
AC

Atten: 1 Level: Preset

BNC Input: Sep.

 $1 M\Omega$

- c. Connect the counter Channel A input to A2TP3. Adjust the 3580A rear panel TRACKING OSC LEVEL fully CW. Connect the rear panel TRACKING OSC OUT to the front panel INPUT.
- d. Adjust A2C4 (TRACKING OSCILLATOR 100 kHz FREQUENCY ADJ.) for a counter reading of 99.999 kHz to 100.001 kHz.
 - e. Center the CRT Trace with MANUAL VERNIER.
- f. Remove the blue lead between A5TP1 and A5TP2 and connect a clip lead between A5TP1 and A5TP6. Momentarily press:

DISPLAY......CLEAR WRITE

- g. Adjust A5C13 (STAGE 5 100 kHz ADJ.) for a maximum display indication. Remove the clip lead on A5TP6 and connect to A5TP5. Adjust A5C10 (STAGE 4 100 kHz ADJ.) for a maximum display indication. Repeat this procedure for A5TP4 (adjust A5C7), A5TP3 (adjust A5C4), and A5TP2 (adjust A5C1).
- h. Remove the cable between the TRACKING OSC OUT and the 3580A INPUT.

5-73. Symmetry Adjustments.

a. Position the following front panel controls: (Only those controls printed in **BOLD** require a change from the previous adjustments).

| SWEEP MODE | RESET |
|--------------------------|--------------|
| ADAPTIVE SWEEP | OFF |
| DISPLAYAll pushbutt | ons released |
| AMPLITUDE MODELOG | 10 dB/DIV |
| AMPLITUDE REF LEVEL | .NORMAL |
| $dBV/LIN-dBm 600 \Omega$ | dBV/LIN |
| INPUT SENSITIVITY | CAL |
| VERNIER (Amplitude) | CAL |
| (Fully | CW) |
| FREQUENCY | 10000 Hz |
| START-CTR | CTR |
| RESOLUTION BANDWIDTH. | 30 Hz |
| DISPLAY SMOOTHING | MIN |
| FREQ. SPAN/DIV | 0.5 kHz |
| SWEEP TIME/DIV | 0.1 SEC |
| SWEEP MODE | .MANUAL |

Option 002: Set dBm 900 Ω /LIN-dBm 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL.

- b. Reposition the internal circuit board shield over circuit boards A2-A4 (leave A5 partially uncovered). Reconnect A5TP1 to A5TP6 and adjust MANUAL VERNIER while pressing and releasing DISPLAY-CLEAR WRITE to obtain a spike display indication in the center of the CRT screen.
- c. Fine tune the fine FREQUENCY control for a maximum display indication.
 - d. Reposition the following front panel controls:

| RESOLU | UTION BANDWIDTH \dots | 300 Hz |
|--------------|-------------------------|--------|
| SWEEP | MODE | REP |

- e. Adjust A5C14 (STAGE 5 CRYSTAL BALANCE ADJ.) for equal and symmetrical skirts on the right and left halves of the CRT display.
- f. Adjust A5C15 (STAGE 5 PEAK RESPONSE ADJ.) to move the peak to the center of the CRT screen. Recheck Step e and adjust A5C14 and A5C15 if necessary. See Figure 5-5 for a properly adjusted display.
- g. Repeat Steps e and f for stages 4, 3, 2 and 1. Connect the appropriate test points and adjust the appropriate capacitors:

| Stage Test Point Connection | | Balance Cap | Peak Cap |
|-----------------------------|------------------------|----------------|-------------|
| Stage 4 | Connect A5TP1 to A5TP5 | C11 | C12 |
| Stage 3 | Connect A5TP1 to A5TP4 | C8 | C9 |
| Stage 2 | Connect A5TP1 to A5TP3 | C5 | C6 |
| Stage 1 | Connect A5TP1 to A5TP2 | C2 | C3 |

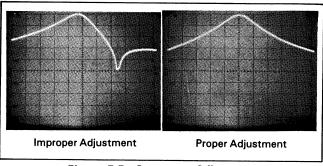


Figure 5-5. Symmetry Adjustment.

NOTE

For the narrower bandwidth displays, positioning the internal circuit board shield completely over A2-A5 eliminates noise and improves symmetry.

h. Reposition the following front panel controls:

| AMPLITUDE MODE | .LINEAR |
|----------------------|---------|
| RESOLUTION BANDWIDTH | 30 Hz |
| FREQ. SPAN/DIV | |
| SWEEP MODE | MANUAL |

- i. Adjust MANUAL VERNIER while pressing and releasing DISPLAY-CLEAR WRITE for a spike display indication in the center of the screen. Adjust the front panel FINE FREQUENCY control for a maximum display indication.
 - j. Reposition the following front panel controls:

 RESOLUTION BANDWIDTH.....300 Hz
 SWEEP MODE.......REP
- k. Disconnect the clip lead between A5TP1 and A5TP2 and reconnect it between A5TP1 and A5TP6.
- 1. Readjust A5C15 for apeak at the center of the display.
- m. Repeat Steps k and l, adjusting A5C12, A5C9, A5C6, and A5C3 with the clip lead connected to the same test points used in Step g for these same capacitors.
- n. Remove the clip lead and reconnect the standard blue lead between A5TP1 and A5TP2.

5-74. Amplitude Calibration. Δ 16

5-75. These adjustments properly calibrate the amplitude section of the 3580A. These adjustments should be made if the instrument fails the Amplitude Tests (Paragraph 5-18) of the Performance Tests. In addition, if the I.F. Filter Alignment (Paragraph 5-70), or the High Voltage Power Supply Adjustments (Paragraph 5-57) have been made, the Amplitude Tests should be performed again to determine if any amplitude calibration is necessary.

5-76. Recommended Test Equipment.

Frequency Synthesizer (-hp- Model 3320B, 50 ohms) Digital Multimeter (-hp- 34740/34702) 50 Ohm Termination (-hp- 11048C)

5-77. Linear and Log Gain Adjustments.

a. Position the following front panel controls: (Only those controls printed in **BOLD** require a change from the previous adjustments).

| ADAPTIVE SWEEP | OFF |
|----------------------|----------------|
| DISPLAYAll pushbu | ttons released |
| AMPLITUDE MODELO | G 10 dB/DIV |
| AMPLITUDE REF LEVEL. | NORMAL |
| dBV/LIN-dBm 600 Ω | dBV/LIN |
| INPUT SENSITIVITY | 20 dB |
| VERNIER (Amplitude) | |
| | (Fully CW) |
| FREQUENCY | 10000 Hz |
| START-CTR | CTR |
| RESOLUTION BANDWIDTH | H300 Hz |
| DISPLAY SMOOTHING | MIN |
| FREQ. SPAN/DIV | 0.5 kHz |
| SWEEP TIME/DIV | 0.1 SEC |
| SWEEP MODE | MANUAL |
| | |

Option 002: Set dBm 900 Ω /LIN-dBm 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL.

- b. Adjust the front panel CAL 10 kHz pot fully CCW.
- c. Adjust the frequency synthesizer to -20~dBV 10.0 kHz output (Option 002, adjust to -20~dBm 900 Ω) and connect the synthesizer to the INPUT terminals of the 3580A.

NOTE

Always terminate your source properly and consult Table 5-14 for the setting needed for a signal source calibrated in dBm 50 Ω . Figure 5-6 shows the proper hookup for use with a 50 ohm frequency synthesizer such as the 3320B.

Table 5-14. Conversion Table.

| 3580A INPUT Signal Level | 3320B or OTHER 50 OHM SOURCE | ABSOLUTE Voltage |
|-------------------------------|---------------------------------|---------------------|
| + 10 dBV | + 23.01 dBm | 3.162 volts |
| + 10 dBm 900 Ω | + 22.55 dBm | 3 volts |
| 0 dBV | + 13.01 dBm | 1 volts |
| 0 dBm 900 Ω | + 12.55 dBm | .949 volts |
| – 10 dBV | + 3.01 dBm | .3162 volts |
| – 10 dBm 900 Ω | + 2.55 dBm | .3000 volts |
| - 20 dBV | -6.99 dBm | .1 volts |
| −20 dBm 900 Ω | -7.45 dBm | .0949 volts |
| - 30 dBV | - 16.9 dBm | .03162 volts |
| – 30 dBm 900 Ω | - 17.45 dBm | .03 volts |
| -40 dBV | - 26.99 dBm | .01 volts |
| −40 dBm 900 Ω | - 27.45 dBm | .095 volts |
| -50 dBV | - 36.99 dBm | 3162 mV |
| -50 dBm $900~\Omega$ | - 37.45 dBm | 3 mV |
| -60 dBV | - 46.99 dBm | 1 mV |
| $-60 \text{ dBm } 900 \Omega$ | -47.45 dBm | .95 mV |
| -70 dBV | - 56.99 dBm | .3162 mV |
| -70 dBm $900~\Omega$ | - 57.45 dBm | .3 mV |
| -80 dBV | - 66.99 dBm | .1 mV |
| $-80 \text{ dBm } 900\Omega$ | - 67.45 dBm | .095 mV |

- d. Adjust MANUAL VERNIER for a maximum display indication.
- e. Connect the multimeter (AC mode, 10 volt range) to A4TP1. Note the reading with the front panel 10 kHz CAL pot fully CCW.
- f. Adjust front panel CAL 10 kHz pot for 1.26 times the reading obtained in Step e.

Examples:

| $100 \text{mV} \times 1.26 = 126 \text{m}$ | V | 117mV | x 1.26 | = | 147mV |
|--|----------------|---------|--------|---|--------|
| $101 \text{mV} \times 1.26 = 127 \text{m}$ | \mathbf{V} | 118mV | x 1.26 | = | 149mV |
| $102mV \times 1.26 = 129m$ | \mathbf{V} | 119mV | x 1.26 | = | 150mV |
| $103 \text{mV} \times 1.26 = 130 \text{m}$ | V | 120mV | x 1.26 | = | 151mV |
| $104 \text{mV} \times 1.26 = 131 \text{m}$ | V | 121mV | x 1.26 | = | 152mV |
| $105 \text{mV} \times 1.26 = 132 \text{m}$ | • | 122mV | x 1.26 | = | 154mV |
| $106 \text{mV} \times 1.26 = 134 \text{m}$ | | 123mV | x 1.26 | = | 155 mV |
| $107 \text{mV} \times 1.26 = 135 \text{m}$ | | 124mV : | x 1.26 | = | 156mV |
| $108 \text{mV} \times 1.26 = 136 \text{m}$ | | 125mV : | x 1.26 | = | 158mV |
| $109 \text{mV} \times 1.26 = 137 \text{m}$ | \mathbf{V} 1 | 126mV | x 1.26 | = | 159mV |
| | | | | | |

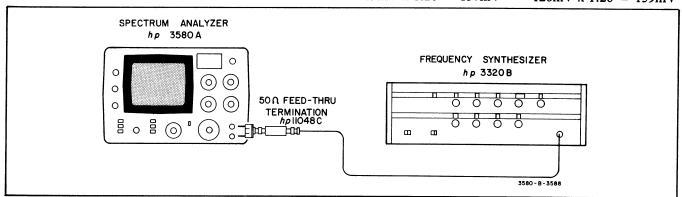


Figure 5-6. Proper Hookup.

Example Cont'd

| $110 \text{mV} \times 1.26 = 139 \text{mV}$ | $127 \text{mV} \times 1.26 = 160 \text{mV}$ |
|---|---|
| $111 \text{mV} \times 1.26 = 140 \text{mV}$ | $128 \text{mV} \times 1.26 = 161 \text{mV}$ |
| $112mV \times 1.26 = 141mV$ | $129 \text{mV} \times 1.26 = 163 \text{mV}$ |
| $113 \text{mV} \times 1.26 = 142 \text{mV}$ | $130 \text{mV} \times 1.26 = 164 \text{mV}$ |
| $114mV \times 1.26 = 144mV$ | $131 \text{mV} \times 1.26 = 165 \text{mV}$ |
| $115 \text{mV} \times 1.26 = 145 \text{mV}$ | $132 \text{mV} \times 1.26 = 166 \text{mV}$ |
| $116 \text{mV} \times 1.26 = 146 \text{mV}$ | $133 \text{mV} \times 1.26 = 168 \text{mV}$ |
| | |

g. Turn 3580A POWER SWITCH to OFF. Place A4 on extender boards. Turn the power switch back to ON and reposition the following front panel control:

AMPLITUDE MODE.....LINEAR

- h. Push DISPLAY CLEAR WRITE momentarily. Adjust A4L1 (I.F. AMP GAIN ADJ.) for a maximum screen display. Remove the source from the 3580A INPUT.
- i. Set the controls of the multimeter for DC mode, 1 volt range. Connect the multimeter to the rear panel Y-AXIS output and adjust A4R11 (DC OFFSET ADJ.) for 0 volt Y-AXIS output level (± 10 mV).
- j. Turn POWER switch to OFF, replace A4 into card nest of 3580A, and turn power switch back to ON (AC). Reconnect the frequency synthesizer (with proper termination) to the 3580A INPUT. Push CLEAR WRITE momentarily. Adjust the source to the same level as in Step c (-20 dBV for standard instrument or -20 dBm 900 Ω if Option 002).

k. Adjust the rear panel VERTICAL DISPLAY ADJUST POSITION control to the center of its range: Δ21

A4R7, A4R8, A4R9, A4R10, A4R3 and A4R4,

1. Reposition the following front panel controls:

| INPUT SENSITIVITY | CAL |
|-------------------|-----|
| SWEEP MODE | REP |

NOTE

If the peak of the waveform is beyond display limits, slightly readjust A13R3 or A13R4 to bring to into view.

- m. Alternately switch between the LOG 1 dB/DIV and LOG 10 dB/DIV AMPLITUDE MODEs and adjust A4R8 (DETECTOR GAIN ADJ.) until the peak amplitude of both waveforms is equal.
- n. Alternately switch between the LOG 10 dB/DIV and LINEAR AMPLITUDE MODEs and adjust A4R6 (LINEAR GAIN ADJ.) until the peak amplitude of both waveforms is equal.
- o. Repeat Steps m and n until the peak amplitude of all three waveforms is equal.

p. Reposition the following front panel control:

AMPLITUDE MODE....LOG 10 dB/DIV

Press and hold the DISPLAY-CLEAR WRITE button to obtain a base line trace. Press the DISPLAY-STORE button to store the base line trace. Release the CLEAR WRITE button.

q. Adjust A13R3 (VERTICAL GAIN ADJ.) and A13R4 (VERTICAL ZERO ADJ.) for a full scale and base line screen display (waveform peak at 0 dB and base line at -100 dB). Press and release the DISPLAY-STORE button.

NOTE

There may be some non-symmetry in the bottom corners of the CRT display. Use the center portion of the base line trace for the above calibration.

- r. Reposition the following front panel controls:
 - INPUT SENSITIVITY.....-20 dB
- s. Adjust the frequency synthesizer output level to -80~dBV for standard instruments or -80~dBm 900 Ω for instruments equipped with Option 002. See Table 5-14 for proper level setting of source.
- t. Adjust A4R7 (LOG GAIN ADJ.) so display peak is at the proper level. (-60 dB graticule $\pm 1 \text{ dB}$ on CRT display, since full scale equals -20 dB).

NOTE

This is a very low level signal. Always slide the cover shield over the A5 assembly after making an adjustment; then verify the results.

- u. Increase the signal level back to full scale $(-20 \text{ dBV for standard instruments or } -20 \text{ dBm } 900 \Omega$ for Option 002) and adjust A4R8 (DETECTOR GAIN ADJ.) for a full scale (0 dB) indication on the display.
- v. Repeat Steps r, s, t, and u until the 0 dB and -60 dB points on the display are calibrated properly.
- w. Alternate the input signal level between -80~dBV and -60~dBV (-80~dBm to -60~dBm 900 Ω for Option 002). See Table 5-14 for proper level. The indication should fall on the -60~dB and -40~dB ($\pm~1~\text{dB}$) graticule lines of the display. If not, adjust A4R10 (BOTTOM END LINEARITY ADJ.) to bring these two points as close into tolerance as possible.
- x. Alternate the input signal level between -20~dBV and -40~dBV (-20~dBm to -40~dBm, $900~\Omega$ for Option 002). See Table 5-14 for proper level setting. These levels should give 0 dB and -20~dB ($\pm~1~\text{dB}$) indications on the display. If not, adjust A4R9 (TOP END

LINEARITY ADJ.) to bring these two points as close into calibration as possible.

- y. Adjust the input signal level to -20 dBV (-20 dBm, 900Ω for Option 002). Switch the AMPLITUDE MODE pushbuttons between LOG 10 dB/DIV and LOG 1 dB/DIV. Adjust A4R8 (DETECTOR GAIN ADJ.) or A4R7 (LOG GAIN ADJU.) to make the levels for the two AMPLITUDE MODE settings equal.
 - z. Reposition the following front panel controls:

AMPLITUDE MODE....LOG 10 dB/DIV

aa. Step the input signal level in 10 dB steps from a full scale 0 dB indication (-20 dBV or -20 dBm 900 Ω input signal) to a -60 dB indication (-80 dBV or -80 dBm 900 Ω input signal). The display should fall within \pm 2 dB of the proper graticule marking to meet specifications.

NOTE

Remember to position the inner circuit board shield over A2-A5 when making low level measurements,

- bb. Repeat Steps r thru aa to bring the log amplifier into the desired test limits.
- cc. Adjust the input signal level to -20 dBV (Instruments with Option 002 should also have this same input level.)
 - dd. Reposition the following front panel control:

AMPLITUDE MODE.....LINEAR

ee. Adjust A4R6 (LINEAR GAIN ADJ.) for a full scale screen display.

5-78. Bandwidth Gain Switching Adjustments.

a. Position the following front panel controls: (Only those controls printed in **BOLD** require a change from the previous adjustments.

| AD ADDITION OWNERD OFF |
|----------------------------------|
| ADAPTIVE SWEEPOFF |
| DISPLAYAll pushbuttons released |
| AMPLITUDE MODELINEAR |
| |
| AMPLITUDE REF LEVELNORMAL |
| $dBV/LIN-dBm 600 \Omega dBV/LIN$ |
| INPUT SENSITIVITY20 dB |
| VERNIER (Amplitude)CAL |
| (Fully CW) |
| START-CTRCTR |
| DISPLAY SMOOTHINGMIN |
| RESOLUTION BANDWIDTH300 Hz |
| FREQ. SPAN/DIV50 Hz |
| SWEEP TIME/DIV0.1 SEC |
| FREQUENCY10000 Hz |
| SWEEP MODEMANUAL |

Option 002: Set dBm 900 Ω /Lin - dBm 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL.

- b. Replace the inner cover shield over A2-A5 and screw down tightly.
- c. Adjust the frequency synthesizer to a 10 kHz, -20 dBV output (same level for Option 002). See Table 5-14 for proper level setting on the frequency synthesizer.
- d. Adjust MANUAL VERNIER and the fine FRE-QUENCY control for a peak reading in the center of the display. Make the following full scale adjustments on the appropriate bandwidth setting.

NOTE

The fine FREQUENCY control may have to be readjusted after each Bandwidth/Freq. Span setting for a peak reading in the center of the screen.

| RESOLUTION | FREQ. | GAIN | SETTING |
|--|--|--------------------------------------|---|
| Bandwidth | Span/div | POT ADJ. | |
| 100 Hz 30 Hz 10 Hz 3 Hz 1 Hz | 50 Hz 10 Hz 5 Hz 5 Hz 5 Hz | A4R5 A4R4 A4R3 A4R2 A4R1 | Full scale O dB display indi- cation. |

5-79. Frequency Response Adjustments.

a. Position the following front panel controls: (Only those controls printed in **BOLD** require a change from the previous adjustments.

| SWEEPMODE RESET |
|---|
| ADAPTIVE SWEEPOFF |
| DISPLAY All pushbuttons released |
| AMPLITUDE MODE1 dB/DIV |
| AMPLITUDE REF LEVELNORMAL |
| dBV/LIN-dBm 600 ΩdBV/LIN |
| INPUT SENSITIVITY 20 dB |
| VERNIER (Amplitude)CAL |
| (Fully CW) |
| START-CTR CTR |
| RESOLUTION BANDWIDTH300 Hz |
| DISPLAY SMOOTHINGMIN |
| FREQ. SPAN/DIV0.2 kHz |
| SWEEP TIME/DIV0.1 SEC |
| FREQUENCY01000 Hz |
| SWEEP MODEREP |
| Option 002: Set dBm 900 Ω/LIN-dBm 600 Ω |
| switch to dBm 900 Ω ; set INPUT MODE |
| Switch to abili 700 M, Set 1141 OT MODE |

switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL.

b. Adjust the frequency synthesizer for 1 kHz output at -20 dBV (-20 dBm 900Ω for Option 002), and connect it to the 3580A INPUT (properly terminated). Ad-

just the front panel CAL 10 kHz for a full scale (0 dB) display.

c. Reposition the following front panel controls:

FREQUENCY......40000 Hz INPUT SENSITIVITY..... - 10 dB

- d. Adjust the signal source for a 40 kHz 10 dBV signal ($-10 \text{ dBm } 900 \Omega \text{ for Option } 002$).
- e. Adjust A9C2 (40 kHz 10 dB AMP ADJ.) for a full scale (0 dB) display (± 1 minor division). In a similar manner, perform the following adjustments:

| SIGNAL SOURCE OUTPUT LEVEL | INPUT Sensitivity | ADJUST | DISPLAY READING |
|-------------------------------|----------------------|--------|--------------------|
| O dBV (or - O dBm 900 Ω) | 0 | A9C3 | OdB ± .2dB |
| + 10 dBV (or 0 dBm 900 Ω) | +10 | A9C4 | OdB ± .2dB |

f. Adjust the ac signal source to 1 kHz at +10 dBV(+ 10 dBm 900 Ω for Option 002). Reposition the following front panel controls:

AMPLITUDE REF LEVEL - 10 dB

INPUT SENSITIVITY

+20 dB (According to MAX INPUT indicator, not white underlay on INPUT SENSITIVITY dial).

FREQUENCY

01000 Hz

g. Store the screen display level by pushing:

DISPLAY STORE

h. Adjust the signal source for a 40 kHz output (same level as in Step f). Reposition the following front panel controls:

FREQUENCY......40000 Hz

- i. Adjust A9C5 (40 kHz + 20 dB AMP ADJ.) for the same level stored in Step g.
 - j. Reposition the following front panel controls:

AMPLITUDE REF LEVEL..... - 20 dB INPUT SENSITIVITY.....+30 dB (According to MAX INPUT indicator) FREQUENCY......01000 Hz

k. Adjust the signal source to a 1 kHz + 10 dBV (+10 dBm for Option 002) output. (Note the screen display level by releasing and then depressing:

DISPLAY STORE).

1. Adjust the signal source for a 40 kHz output (same level as in Step k). Reposition the following front panel control:

FREQUENCY......40000 Hz

m. Adjust A9C6 (40 kHz + 30 dB AMP ADJ.) for the same level stored in Step k).

5-80. Internal Calibrator Adjustment.

a. Position the following front panel controls: (Only those controls printed in BOLD require a change from the previous adjustments).

| ADAPTIVE SWEEPOFF |
|---|
| DISPLAYAll pushbuttons released |
| AMPLITUDE MODE1 dB/DIV |
| AMPLITUDE REF LEVELNORMAL |
| $dBV/LIN-dBm 600 \OmegadBV/LIN$ |
| INPUT SENSITIVITY20 dB |
| VERNIER (Amplitude)CAL |
| (Fully CW) |
| FREQUENCY10000 Hz |
| START-CTRCTR |
| RESOLUTION BANDWIDTH300 Hz |
| DISPLAY SMOOTHINGMIN |
| FREQ. SPAN/DIV0.2 kHz |
| SWEEP TIME/DIV0.1 SEC |
| SWEEP MODEREP |
| Option 002: Set dBm 900 Ω/LIN-dBm 600 Ω |
| switch to dBm 900 Ω; set INPUT MODE |
| switch to UNBAL. |

- b. Connect a properly terminated frequency synthesizer to the 3580A INPUT and adjust the synthesizer for a 10 kHz -20 dBV output level (-20 dBm 900Ω for instruments with Option 002). See Table 5-14 for proper settings.
- c. Adjust front panel CAL 10 kHz for a full scale (0 dB) peak on the display.
 - d. Reposition the following front panel control: INPUT SENSITIVITY......CAL
- e. Remove the input signal source. Adjust A2R5 (CAL LEVEL ADJ.) for a full scale (0 dB) screen display.

5-81. Mixer Balance Adjustments.

5-82. These adjustments balance the input mixer and tracking oscillator mixer. These adjustments should be done if the zero beat response of the instrument under calibration is too large (> -30 dB) or if the TRACK-ING OSC OUTput is distorted (> -40 dB distortion).

5-83. Recommended Test Equipment:

Oscilloscope (-hp- Model 180A with 1801A and 1820A plug-ins)

- a. Disconnect all signal sources from the 3580A.
- b. Position the following front panel controls: (Only those controls printed in **BOLD** require a change from the previous adjustments).

| ADAPTIVE SWEEPOFF |
|---|
| DISPLAYAll pushbuttons released |
| AMPLITUDE MODELOG 10 dB/DIV |
| AMPLITUDE REF LEVELNORMAL |
| $dBV/LIN-dBm 600 \OmegadBV/LIN$ |
| INPUT SENSITIVITY20 dB |
| VERNIER (Amplitude)CAL |
| (Fully CW) |
| FREQUENCY000000 Hz |
| START-CTR START |
| RESOLUTION BANDWIDTH300 Hz |
| DISPLAY SMOOTHINGMIN |
| FREQ. SPAN/DIV0.2 kHz |
| SWEEP TIME/DIV0.1 SEC |
| SWEEP MODEREP |
| Option 002: Set dBm 900 Ω/LIN-dBm 600 |
| Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL. |

- c. Adjust A9R1 (MIXER BALANCE) for a minimum screen display (less than -30 dB to meet specifications).
 - d. Reposition the following front panel controls:

FREQUENCY......0.5000 Hz SWEEP MODE.....MANUAL

- e. Adjust the rear panel TRACKING OSC LEVEL fully CW, and set the EXT REF-NORMAL switch to NORMAL.
- f. Connect the oscilloscope to the rear panel TRACKING OSC OUT connector and monitor the output.
- g. Adjust A2R113 (T.O. MIXER BALANCE) for the cleanest signal. See Figure 5-7.

5-84. Adaptive Sweep Marker Adjustment. Δ 16

5-85. This adjustment properly positions the ADAP-TIVE SWEEP marker. If the marker (blank spot on

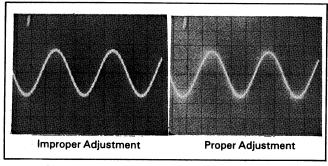


Figure 5-7. Oscillator Output Adjustment.

screen) does not appear at the same point on the display as new information being written onto the display, do this adjustment:

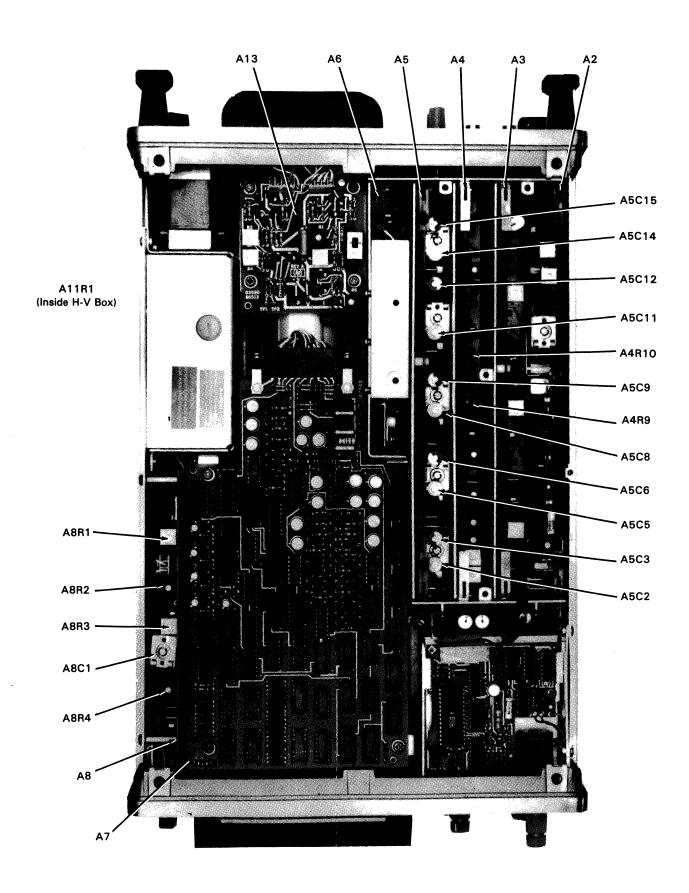
a. Position the following front panel controls: (Only those controls printed in **BOLD** require a change from the previous adjustments).

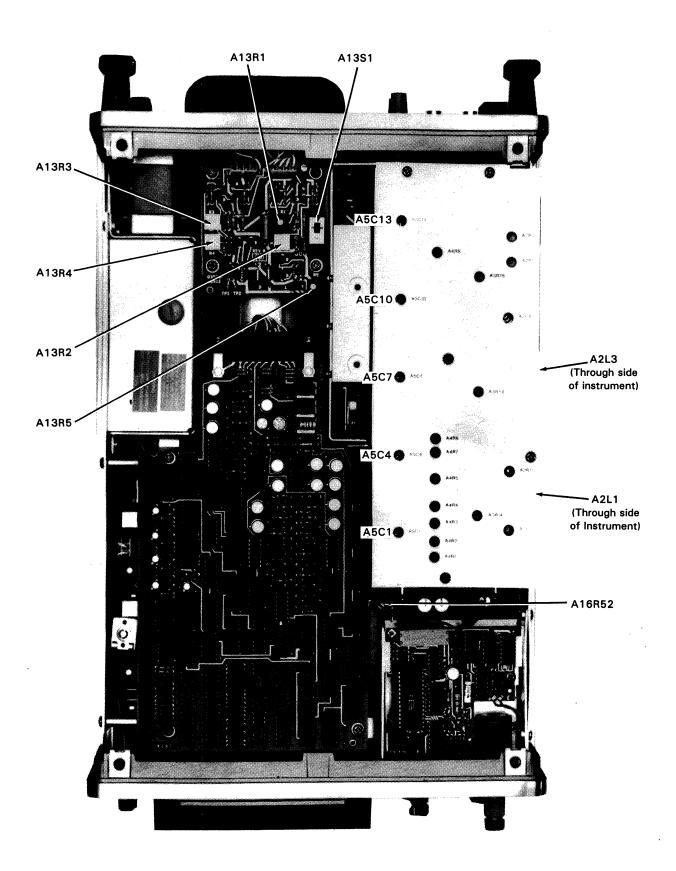
| SWEEP MODERESET |
|--|
| ADAPTIVE SWEEPOFF |
| (Fully CCW) |
| DISPLAYAll pushbuttons released |
| AMPLITUDE MODELOG 10 dBV/DIV |
| AMPLITUDE REF LEVELNORMAL |
| dBV/LIN-dBm 600 ΩdBV/LIN |
| INPUT SENSITIVITYCAL |
| VERNIER (Amplitude)CAL |
| (Fully CW) |
| START-CTR START |
| RESOLUTION BANDWIDTH300 Hz |
| DISPLAY SMOOTHINGMIN |
| FREQ. SPAN/DIV2 kHz |
| SWEEP TIME/DIV1 SEC |
| FREQUENCY00000 Hz |
| SWEEP MODEMAN |
| Option 002: Set dBm 900 ΩLIN-dBm 600 Ω |
| switch to dBm 900 Ω; set INPUT MODE |

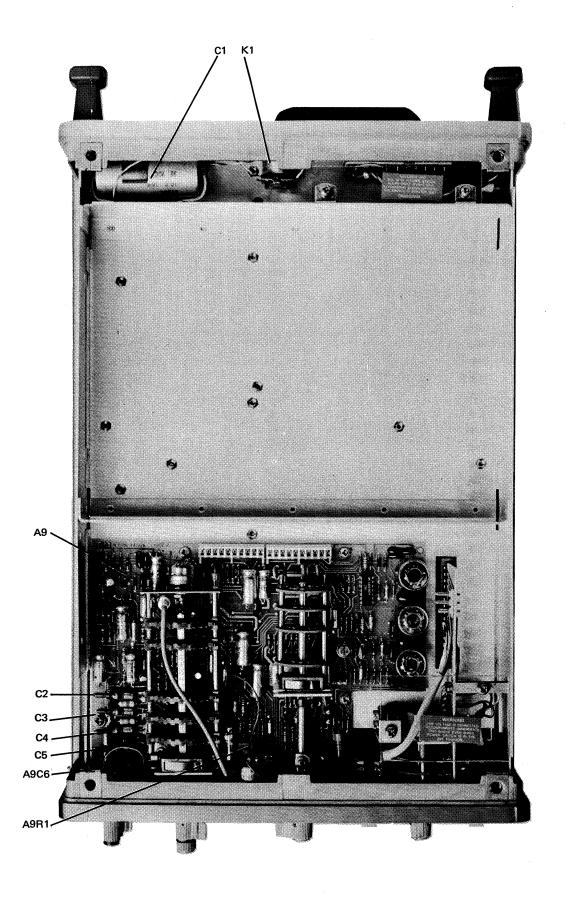
c. Momentarily press the DISPLAY-CLEAR WRITE button. A dot should remain at the top of the scope.

switch to UNBAL.

- b. Adjust the MANUAL VERNIER control until the trace is at the peak of the 10 kHz signal. The FRE-QUENCY display should be about 10 kHz.
- d. Turn the ADAPTIVE SWEEP on and adjust A8R3 (SWEEP MARKER ADJ.) until the sweep marker (blank spot in trace) blanks out the dot at the top of the scope.







 Δ_{16} Figure 5-8. Test Point and Adjustment Locations. 5-27/5-28

PERFORMANCE TEST CARD

| Hewlett-Packard Model 3580A Spectrum Analyzer Serial No. | Tests Performed By Date | | |
|---|---|--|------------------|
| RANGE AND FREQUENCY ACCUR | ACY TESTS | | |
| Ideal Frequency Display Reading | Actual | | Test Limits |
| 30000 Hz 30000 Hz | Hz Hz | | ± 3 Hz ± 3 Hz |
| DISPLAY ACCURACY TESTS | | | |
| The separation between the Zero Responsarior should be 10 div. ± .2 div. between any two adjacent responses sl ± .04 div. | The separation | Pass | Fail |
| FREQUENCY SPAN TESTS | . | | |
| Frequency Span/Div. | Frequency Display Reading (Manual Vernier Fully CW) | Test Limits | |
| 5 Hz 10 Hz 20 Hz 50 Hz . 1 kHz . 2 kHz . 5 kHz 1 kHz 2 kHz | Huny Cw) Hz Hz Hz Hz Hz Hz Hz Hz Hz H | 00050 Hz ± 1 Hz 00100 Hz ± 2 Hz 00200 Hz ± 4 Hz 00500 Hz ± 10 Hz 01000 Hz ± 20 Hz 02000 Hz ± 40 Hz 05000 Hz ± 100 Hz 10000 Hz ± 200 Hz 20000 Hz ± 400 Hz | |
| LOG SWEEP TEST | | | |
| The 20 kHz harmonic of the internal CAl on the 20 kHz LOG SWEEP graticule (± 2 | | Pass | Fail |
| SWEEP TIME TEST | | | |
| All sweep rates must work properly. | | Pass | Fail |
| BANDWIDTH SWITCHING ACCURACY | TEST | | |
| Bandwidth | Display Indication (0 dB full scale) (1 dB/div) | Test Limits | |
| 100 Hz 30 Hz 10 Hz 3 Hz | dB dB dB dB | -1.0 dB ± .5 dB -1.0 dB ± .5 dB -1.0 dB ± .5 dB -1.0 dB ± .5 dB -1.0 dB ± .1 dB | |

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LOG AMPLITUDE DISPLAY ACCURACY TESTS

| LOG AMI LITOI | DE DISI LAT ACCORACT | ILSIS | |
|-------------------------|-----------------------|--|--|
| Input | Level | | |
| Standard | Option 002 900 Ω | Display Indication (0 dB full scale) (10 dB/div) | Test Limits |
| - 10 dBV | - 10 dBm 900 Ω | dB | $-10 \text{ dB} \pm 2 \text{ dB}$ |
| - 20 dBV | - 20 dBm | dB | $-20 \text{ dB} \pm 2 \text{ dB}$ |
| - 30 dBV | - 30 dBm | dB | $-30 dB \pm 2 dB$ |
| - 40 dBV | - 40 dBm | dB | $-40 \text{ dB} \pm 2 \text{ dB}$ |
| - 50 dBV | - 50 dBm | dB | $-50 \text{ dB} \pm 2 \text{ dB}$ |
| - 60 dBV | - 60 dBm | dB | $-60 \text{ dB} \pm 2 \text{ dB}$ |
| - 70 dBV | - 70 dBm | dB | $-70 \text{ dB} \pm 2 \text{ dB}$ |
| - 80 dBV | - 80 dBm | dB | $-80 \text{ dB} \pm 2 \text{ dB}$ |
| LINEAR AMPLI | TUDE DISPLAY ACCUR. | ACY TESTS | |
| Input Level | | Display Indication (1 V full scale) (10 dB/div) | Test Limits |
| .9 V | | (10 dD/div) V | .90 V ± .02 V |
| .9 V .8 V | | V | .90 V ± .02 V |
| .0 V .7 V | | v | .70 V ± .02 V |
| .7 V .6 V | | V | .60 V ± .02 V |
| .5 V | | v | .50 V ± .02 V |
| .5 V .4 V | | V | .50 V ± .02 V .40 V ± .02 V |
| .4 V .3 V | | v | $.40 \text{ V} \pm .02 \text{ V}$ $.30 \text{ V} \pm .02 \text{ V}$ |
| .3 V .2 V | | V | .30 V ± .02 V .20 V ± .02 V |
| .2 V .1 V | | v | .10 V ± .02 V |
| .1 V | | | .10 V ± .02 V |
| AMPLITUDE RE | EFERENCE TESTS (Linea | r Mode) | |
| 3580A Input (10 kHz) | Amp Rev Level | Display Indication (% of full scale) | Test Limits |
| 200 mV | - 10 | % | 90% ± 3% |
| 100 mV | - 20 | % | (\pm .3 major divisions) 90% \pm 3% |
| 20 mV | - 30 | % | (\pm .3 major divisions) 90% \pm 3% |
| 10 | 40 | at. | (± .3 major divisions) |
| 10 mV | - 40 | % | $90\% \pm 3\%$ (± .3 major divisions) |
| 2 mV | - 50 | % | $90\% \pm 3\%$ (± .3 major divisions) |
| 1 mV | - 60 | % | $90\% \pm 3\%$ (± .3 major divisions) |
| .2 mV | - 70 | % | 90% ± 10% |
| AMDITTIDE DI | EEEDENCE LEVEL TECT | (Low Mode) | (± 1 major division) |
| | EFERENCE LEVEL TEST | | |
| Amp Rev Level | | Multimeter Reading | Test Limits |
| - 10 dB | | <u>V</u> | $2.00 \text{ V} \pm .02 \text{ V}$ |
| - 20 dB | | V | $2.50 \text{ V} \pm .02 \text{ V}$ |
| - 30 dB | | V | $3.00 \text{ V} \pm .03 \text{ V}$ |
| - 40 dB | | V | $3.50 \text{ V} \pm .03 \text{ V}$ |
| - 50 dB - 60 dB | | V | 4.00 V ± .04 V 4.50 V ± .04 V |
| - 70 dB | | v | 5.00 V ± .04 V |
| | | | 5.55 / <u>-</u> .05 / |

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INPUT ATTENUATOR TESTS

| Input | Amp Ref Level | Input Sensitivity (according to white marker) | Display Indication (% of full scale) | Test Limits (full scale ± .3 major div) |
|-------|------------------|---|--------------------------------------|---|
| .2 V | - 30 dB | .2 V | % | 100% ± 3% |
| .1 V | - 30 dB | .1 V | % | 100% ± 3% |
| 20 mV | - 30 dB | 20 mV | % | 100% ± 3% |
| .2 V | normal | .2 V | % | $100\% \pm 3\%$ |
| .1 V | normal | .1 V | % | 100% ± 3% |
| 20 mV | normal | 20 mV | % | $100\% \pm 3\%$ |
| 10 mV | normal | 10 mV | <u> </u> | 100% ± 3% |
| 2 mV | normal | 2 mV | % | 100% ± 3% |
| 1 mV | normal | 1 mV | % | $100\% \pm 3\%$ |
| .2 mV | normal | .2 mV | % | 100% ± 3% |
| | | | | |

FREQUENCY RESPONSE TESTS

| | Input L | evel | | | | |
|---|--|--|---|---|---|--|
| | Standard | Option 002 (900 Ω) | Input Sensitivity (according to white marker) | Frequency | Display Indication (0 dB = full scale 1 dB/div) | Test Limits |
| | 0 dBV 0 dBV 0 dBV | 0 dBM 0 dBM 0 dBM | 0 dB 0 dB 0 dB | 10 Hz 20 Hz 1 kHz | dB dB dB | 0 dB ± .5 dB 0 dB ± .3 dB 0 dB ± .3 dB |
|) | 0 dBV 0 dBV | 0 dBM 0 dBM | 0 dB 0 dB | 20 kHz 50 kHz | dB | 0 dB ± .3 dB 0 dB ± .5 dB |
| | - 10 dBV - 10 dBV - 10 dBV - 10 dBV - 10 dBV | - 10 dBM - 10 dBM - 10 dBM - 10 dBM - 10 dBM | - 10 dB - 10 dB - 10 dB - 10 dB - 10 dB | 10 Hz 20 Hz 1 kHz 20 kHz 50 kHz | dB | 0 dB ± .5 dB 0 dB ± .3 dB 0 dB ± .3 dB 0 dB ± .3 dB 0 dB ± .5 dB |
| | - 20 dBV - 20 dBV - 20 dBV - 20 dBV - 20 dBV | - 20 dBM - 20 dBM - 20 dBM - 20 dBM - 20 dBM | - 20 dB - 20 dB - 20 dB - 20 dB - 20 dB | 10 Hz 20 Hz 1 kHz 20 kHz 50 kHz | dB dB dB dB | 0 dB ± .5 dB 0 dB ± .3 dB 0 dB ± .3 dB 0 dB ± .3 dB 0 dB ± .5 dB |
| | - 30 dBV - 30 dBV - 30 dBV - 30 dBV | - 30 dBM - 30 dBM - 30 dBM - 30 dBM | - 30 dB - 30 dB - 30 dB - 30 dB - 30 dB | 10 Hz 20 Hz 1 kHz 20 kHz 50 kHz | dB dB dB dB dB | 0 dB ± .5 dB 0 dB ± .3 dB 0 dB ± .3 dB 0 dB ± .3 dB 0 dB ± .5 dB |
| | - 40 dBV - 40 dBV - 40 dBV - 40 dBV | - 40 dBM - 40 dBM - 40 dBM - 40 dBM - 40 dBM | - 40 dB - 40 dB - 40 dB - 40 dB - 40 dB | 10 Hz 20 Hz 1 kHz 20 kHz 50 kHz | dB dB dB dB | 0 dB ± .5 dB 0 dB ± .3 dB 0 dB ± .3 dB 0 dB ± .3 dB 0 dB ± .5 dB |

INTERNAL CALIBRATOR TEST

Display Indication (0 dB = full scale) Test Limit 10 kHz Cal. Signal Level dB 0 dB \pm .15 dB

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BANDWIDTH TESTS

| Resolution Bandwidth | Lower 3 dB Frequency | Upper 3 dB Frequency | Test Limits |
|--|---|----------------------------------|---|
| 300 Hz 100 Hz 30 Hz 10 Hz | 10 kHz 1 kHz 1 kHz 1 kHz | kHz kHz kHz | 10.3 kHz ± 45 Hz 1.1 kHz ± 15 Hz 1.030 kHz ± 4.5 Hz 1.010 kHz ± 1.5 Hz |
| Resolution Bandwidth | Lower 60 dB Frequency | Upper 60 dB Frequency | Test Limits |
| 3 Hz 1 Hz | 1 kHz 1 kHz | kHz kHz | 1.030 kHz ± 4.5 Hz 1.010 kHz ± 1.5 Hz |
| NOISE LEVEL TESTS | | | |
| Bandwidth | Frequency | Noise (- 70 dB = full scale) | Test Limits |
| 300 Hz 30 Hz 30 Hz 30 Hz 1 Hz 1 Hz 1 Hz | 10 kHz 10 kHz 100 Hz 1 kHz 1 kHz 100 Hz | dB dB dB dB dB dB | <- 130 dB <- 140 dB <- 132 dB <- 140 dB <- 150 dB <- 143 dB <- 135 dB |
| NOISE SIDEBAND TEST | | | |
| Noise Sidebands must be a wave signal, ± 10 Hz away. | t least 70 dB below continuous | Pass | Fail |
| SPURIOUS RESPONSE TES | ST Δ 19 | | |
| All non-line-related spuriou 80 dB below a full scale refe | us responses must be at least rence. | Pass | Fail |
| LINE-RELATED SPURIOU | S RESPONSE TEST | | |
| All line-related spurious -135 dBV $(0.18 \mu V)$. | responses must be less than | Pass | Fail |
| IF FEEDTHRU TEST | | | |
| IF Feedthru must be at leareference. | ast - 70 dB below the full scale | Pass | Fail |
| ZERO BEAT RESPONSE T | EST | | |
| The zero beat response mu full scale reference. | st be at least 30 dB below the | Pass | Fail |
| INPUT IMPEDANCE TESTS | S | | |
| Frequency | Display Indication (0 dB = full scale) Without 1 M Ω With 1 M Ω | Test | Limit |
| 1 kHz | 0 dBdB | - 3 dl | 3 ± 1 dB |

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Revised: September 1987

INPUT IMPEDANCE TESTS (cont'd) Frequency Display Indication Test Limit (with 1 M Ω Resistor) 1 kHz 0 dB10 kHz ____ dB $-3 dB \pm 1 dB$ TRACKING OSCILLATOR OUTPUT TESTS FREQUENCY RESPONSE: Instrument Frequency Multimeter Reading Test Limits Standard 50 Hz 1.00 volt rms $1.00 \text{ volt } \pm .06 \text{ volts}$ 50 kHz ___ volts rms Option 002 300 Hz 1.00 volt rms 20 kHz ____ volts rms $1.00 \text{ volt } \pm .06 \text{ volts}$ FREQUENCY ACCURACY: Resolution Bandwidth Display Indication Test Limit 30 Hz 2 V (full scale) 3 Hz ____ V 1 V - 2 V (half to full scale) **DISTORTION:** Distortion: _____dB Test Limit: less than - 40 dB **RECORDER OUTPUT TESTS** Recorder Output Display Indication Multimeter Reading **Test Limits** X-Axis Manual Vernier fully CCW $0 \text{ V dc} \pm .15 \text{ V}$ Manual Vernier fully CW 5 V dc ± .15 V Y-Axis Full Scale 5 V dc ± .15 V **Bottom Graticule** $0 \text{ V dc} \pm .15 \text{ V}$ **COMMON MODE REJECTION TEST (Option 002 only)** Display Indication Common Mode Input (full scale = $0 \text{ dBM } 900 \Omega$) Test Limit 60~Hz - $0~dBM~900~\Omega$ __ dBM 900 Ω Less than - 60 dBm 900 Ω FREQUENCY RESPONSE TEST (Option 002 only) Display Indication Test Limit

 $(-1 \text{ dB} = 0 \text{ dBM } 900 \Omega, 1 \text{ dB/div})$

_ dBM 900 Ω

 \perp dBM 900 Ω

_____ dBM 900 Ω

Revised: September 1987

Frequency

1 kHz

20 kHz

300 Hz

(± .5 major div)

 $-1 dB \pm .5 dB$

 $-1 dB \pm .5 dB$

 $-1 dB \pm .5 dB$

SECTION VI REPLACEABLE PARTS

6-1. INTRODUCTION.

- 6-2. This section contains information for ordering replacement parts. Table 6-3 lists parts in alphameric order of their reference designators and indicates the description, -hp- part number of each part, together with any applicable notes, and provides the following:
- a. Total quantity used in the instrument (Qty column). The total quantity of a part is given the first time the part number appears.
- b. Description of the part (see list of abbreviations in Table 6-1).
- c. Typical manufacturer of the part in a five-digit code (see Table 6-2 for list of manufacturers).
 - d. Manufacturers part number.
- 6-3. Miscellaneous parts are listed at the end of Table 6-3.

6-4. ORDERING INFORMATION.

6-5. To obtain replacement parts, address order or inquiry to your local Hewlett-Packard Field Office. (See Appendix B for list of office locations.) Identify parts by their Hewlett-Packard part numbers. Include instrument model and serial numbers.

6-6. NON-LISTED PARTS.

- 6-7. To obtain a part that is not listed, include:
 - a. Instrument model number.
 - b. Instrument serial number.
 - c. Description of the part.
 - d. Function and location of the part.

6-8. PROPRIETARY PARTS.

6-9. Items marked by a dagger (†) in the reference designator column are available only for repair and service of Hewlett-Packard instruments.

Table 6-1. List of Abbreviations.

| | ABBRE | VIATIONS | |
|-------------------------------|--|--|--|
| Ag silver | Hz hertz (cycle(s) per second) | NPO negative positive zero | sl slid |
| N aluminum | | (zero temperature coefficient) | SOUT |
| \ ampere(s) | ID inside diameter | nsnanosecond(s) = 10 - 9 seconds | SPDT single-pole double-throw |
| Au | impg impregnated | nsr not separately replaceable | SPST single-pole single-throw |
| • | incd incandescent | not separately replaceable | T |
| Capacitor | ins insulation(ed) | Ω ohm(s) | Tatantalur |
| er ceramic | | obd order by description | TCtemperature coefficier |
| oef coefficient | $k\Omega \dots klohm(s) = 10 + 3 \text{ ohms}$ | OD | TiO2titanium dioxid |
| omcommon | kHzkilohertz = 10+3 hertz | ODoutside diameter | tog toggi |
| comp composition | Kriz | _ | tol |
| conn | L inductor | P. · · · · · · peak | trim trimme |
| | | pA picoampere(s) | TSTR transisto |
| lep deposited | linlinear taper | pcprinted circuit | |
| nep deposited | log logarithmic taper | pFpicofarad(s) 10 - 12 farads | V |
| OPDT double-pole double-throw | | pivpeak inverse voltage | vacw alternating current working voltag |
| OPST double-pole single-throw | mA milliampere(s) = 10 - 3 amperes | p/opart of | var variabl |
| | MHZ megahertz = 10+6 hertz | pos position(s) | vdcwdirect current working voltag |
| llect electrolytic | $M\Omega \dots megahm(s) = 10+6$ ohms | poly polystyrene | vocav |
| ncap encapsulated | met fim metal film | pot potentiometer | W |
| | mfr manufacturer | p-p peak-to-peak | W |
| farad(s) | ms millisecond | ppm parts per million | w/wit |
| ET field effect transistor | mtg mounting | prec precision (temperature coefficient, | wiv working inverse voltage |
| xd fixed | mVmillivolt(s) = $10-3$ volts | precision (temperature coefficient, | w/o |
| Na | μFmicrofarad(s) | long term stability and/or tolerance) | ww wirewound |
| GaAs | | _ | |
| GHzgigahertz = 10 + 9 hertz | μs microsecond(s) | R resistor | |
| | $\mu V \dots \dots \dots \dots \dots \dots $ microvolt(s) = 10 - 6 volts | Rh rhodium | |
| d guard(ed) | my | rms root-mean-square | * optimum value selected at factory |
| Ge germanium | | rot rotary | average value shown (part may be omitted |
| nd ground(ed) | nA nanoampere(s) = 10 _ g amperes | | **no standard type number assigned |
| | NC normally closed | Se selenium | no standard type number assigned |
| i | Ne neon | sect section(s) | selected or special type |
| lg mercury | NO normally open | Si silicon | • |
| · | | Silicon | Dupont de Nemours |
| | DESIGN | IATORS | - |
| | FLfilter | Q transistor | |
| motor | HR heater | OCP transistor | TSterminal strip |
| Tbattery | ICintegrated circuit | QCR transistor-diode | U microcircui |
| | | R(p) resistor(pack) | V vacuum tube, neon bulb, photocell, etc |
| | Jjack | RT thermistor | W |
| Rdiode or thyristor | K relay | S switch | X socke |
| DLdelay line | L inductor | T transformer | XDS lampholde |
| OS | M meter | TBterminal board | XF fuseholde |
| misc electronic part | MPmechanical part | TC thermocouple | Ycrysti |
| | P plug | TPtest point | |

Table 6-2. Code List of Manufacturers.

| Mfr. No. | Manufacturer Name | Address |
|-------------|---------------------------------------|-------------------------|
| | | |
| 01121 | Allen-Bradley Co. | Milwaukee, WI 53204 |
| 01295 | Texas Instr. Inc. Semicond Cmpnt Div. | Dallas, TX 75222 |
| 0192B | RCA Corp. Solid State Div. | Somerville, NJ 08876 |
| 02111 | Spectrol Electronics Corp. | City of Ind, CA 91745 |
| 03888 | KDI Pyrofilm Corp. | Whippany, NJ 07981 |
| 04713 | Motorola Semiconductor Products | Phoenix, AZ 85062 |
| 07263 | Fairchild Semiconductor Div. | Mountain View, CA 94042 |
| 13606 | Sprague Elect. Co. Semiconductor Div. | Concord, NH 03301 |
| 19701 | Mepco/Electra Corp. | Mineral Wells, TX 76067 |
| 20940 | Micro-Ohm Corp. | El Monte, CA 91731 |
| 24046 | Transitron Electronic Corp. | Wakefield, MA 01880 |
| 24546 | Corning Glass Works (Bradford) | Bradford, PA 16701 |
| 27014 | National Semiconductor Corp. | Santa Clara, CA 95051 |
| 28480 | Hewlett-Packard Co. Corporate HQ | Palo Alto, CA 94304 |
| 32997 | Bourns Inc. Trimpot Prod. Div. | Riverside, CA 92507 |
| 34371 | Harris Semicon Div. Harris-Intertype | Melbourne, FL 32901 |
| 50088 | Mostek Corp. | Carrollton, TX 75006 |
| 52763 | Stettner-Trush Inc. | Cazenovia, NY 13035 |
| 56289 | Sprague Electric Co. | North Adams, MA 01247 |
| 72136 | Electro Motive Corp. Sub. IEC | Willimantic, CT 06226 |
| 72982 | Erie Technological Products Inc. | Erie, PA 16512 |
| 75042 | TRW Inc. Philadelphia Div. | Philadelphia, PA 19108 |
| 75915 | Littelfuse Inc. | Des Plaines, IL 60016 |
| 91637 | Dale Electronics Inc. | Columbus, NE 68601 |
| 99515 | Marshall Ind. Capacitor Div. | Monrovia, CA 91016 |

Table 6-3. Replaceable Parts.

| Reference Designation | HP Part Number | Qty | Description | Mfr Code | Mfr Part Number |
|---|---|-------------------------|---|--|--|
| A1 | 03580+66501 | 1 | BOARD ASSEMBLY-MOTHER | 28480 | 03580-66501 |
| AICRI | 1901-0040 | 95 | DIODE-SWITCHING 30V 50MA 2Ng DO-35 | 28480 | 1901-0040 |
| A 1 R 1 A 1 R 2 A 1 R 3 A 1 R 2 4 | 0757-0280 0757-0280 0698-3228 0698-4489 | 1 a 2 | RESISTOR 1K 1X 125W F TC=0+=100 RESISTOR 1K 1X 125W F TC=0+=100 RESISTOR 49,9K 1X 125W F TC=0+=100 RESISTOR 28K 1X 125W F TC=0+=100 | 54246 58480 54246 54246 | C4-1/8-T0-1001-F C4-1/8-T0-1001-F 0698-3228 C4-1/8-T0-2802-F |
| A2 | 03581-66512 | 1 | BOARD ASSY,OSCILLATOR (DOES NOT INCLUDE A12Y1 OR A12R65,SEE PARAGRAPH 7-19) | 28480 | 03581-66512 |
| A2'C1 A2'C2 | 0160-4812 0160-0162 0180-1714 | 3 1 1 | CAPACITOR-FXD 220pf .05 CAPACITOR-FXD .022UF +/- 10% 200WVDC CAPACITOR-FXD; 330UF+-10% 6VDC TA-SOLID | 28480 56289 56289 | 0160-4812 292P22392 150D337X9006S2 |
| A2C4 A2C5 A2C6 A2C7 A2C8 | 0121-0426 0160-4571 0160-4571 0140-0149 0160-0154 | 6 17 1 3 | CAPACITOR, VAR, TRMR, MICA, 50/380PF CAPACITOR-FXD.1uF +80-20 CAPACITOR-FXD.1uF +80-20 CAPACITOR-FXD 470PF+-5% 300MVDC CAPACITOR-FXD →0022UF←10% 200MVDC | 72136 28480 28480 72136 56289 | T52517-7 0160-4571 0160-4571 DML5F471J0300WV1CR 292P22292 |
| A2C9 A2C11 A2C12 A2C13 A2C14 | 0150-0029 0160-4807 0160-4822 0160-3847 0160-4814 | 17 1 2 40 2 | CAPACITOR-FXD 1.0 PF +- 10% 500WVDC CAPACITOR-FXD 33pF 100V CAPACITOR-FXD 1000pF 100V CAPACITOR-FXD .01uF 50V CAPACITOR-FXD 150pF 100V | 28480 28480 28480 28480 | 0160-4807 0160-4822 0160-3847 0160-4814 |
| A2C15 A2C10 A2C17 | 0160-4812 0140-0176 0160-4833 | 4 20 | CAPACITOR-FXD 220pF .05 CAPACITOR-FXD 100PF+-2% 300WVDC CAPACITOR-FXD .022uF 100V | 28480 72136 28480 | 0160-4812 DM15F101G0300WV1CR 0160-4833 |
| A2 C19 | C180-0106 | 7 | CAPACITOR-FXD; 60UF+-20% 6VDC TA-SOLID | 56289 | 1500606X000682 |
| A2C20 A2C21 A2C22 A2C23 A2C24 | 0160-0162 C16C-0160 0180-0228 O160-4814 0160-4833 | 2 13 | CAPACITOR-FXD .022UF+-10% 200MVDC CAPACITOR-FXD .0082UF+-10% 200MVDC CAPACITOR-FXD; 22UF+-10% 15VDC TA-SOLID CAPACITOR-FXD 150pF 100V CAPACITOR-FXD 150pF 100V | 562 89 562 89 562 89 28480 28480 | 292 P22392 292 P82292 1500226 X901582 0160-4814 0160-4833 |
| A2C25 A2C26 A2C27 A2C28 A2C29 | 0160-4571 0160-4571 0160-2939 0150-0116 0180-1701 | 1 1 2 | CAPACITOR-FXD .1uF +80 -20 CAPACITOR-FXD .1uF +80 -20 CAPACITOR-FXD 420pF 500V CAPACITOR-FXD 47pF .10N750 CAPACITOR-FXD; 6-8UF+-20% 6VDC TA-SOLID | 28480 28480 28480 28480 56289 | 0160-4571 0160-4571 0160-2939 0150-0116 1500685X0006A2 |
| A2C31 A2C32 A2C33 A2C34 A2C35 | 0160-4833 0160-4571 0140-3200 0160-3847 0160-0940 | 6 1 | CAPACITOR-FXD .022uF 100V CAPACITOR-FXD .1uF +80 -20 CAPACITOR-FXD .390PF+-5% 300WVDC CAPACITOR-FXD .01uF 50V CAPACITOR-FXD .2400pF | 28480 28480 72136 28480 28480 | 0160-4833 0160-4571 DM15F391J0300WV1CR 0160-3847 0160-0940 |
| A2C36 A2C37 A2C38 A2C39 A2C41 | 0160-3847 0180-0210 0160-3847 0180-0061 0160-2585 | 10 11 1 | CAPACITOR-FXD .01uF 50V CAPACITOR-FXD; 3.3UF+-20% 15VDC TA CAPACITOR-FXD .01uF 50V CAPACITOR-FXD; 100UF+75-10% 16VDC AL CAPACITOR-FXD .002UF+-1% 100WVDC | 28480 56289 28480 56289 28480 | 0160-3847 150D335x0015A2 0160-3847 30D107G016DC2 0160-2585 |
| A2C44 A2C45 A2C45 A2C45 | 016C-2206 0140-0233 0160-2587 0160-0841 018C-0106 | 2 3 1 1 | CAPACITOR-FXD 160PF+-5% 300WVDC CAPACITOR-FXD 480PF+-1% 300WVDC CAPACITOR-FXD .004UF+-1% 100WVDC CAPACITOR-FXD .00174UF+-1% 300WVDC CAPACITOR-FXD; 60UF+-20% 6VDC TA-SOLID | 28480 72136 23480 28480 56289 | 0160-2206 DM15F481F0300WV1C 0160-2587 0160-0841 1500606X0006B2 |
| A2C47 A2C48 A2C49 A2C5 ₁ A2C52 | 0180-0210 0160-4801 0180-4571 0180-0210 0160-4812 | 18 | CAPACITOR-FXD; 3.3UF+-20% 15VDC TA CAPACITOR-FXD 100 pF 100V CAPACITOR-FXD .1uF +80.20 CAPACITOR-FXD; 3.3UF+-20% 15VDC TA CAPACITOR-FXD 220pF.05 | 56289 28480 28480 56289 28480 | 1500335x0015A2 0160-4801 0160-4571 1500335x0015A2 0180-4812 |
| A2C53 A2C54 A2C55 A2C56 A2C56 | 0180-0228 0150-0022 0160-4801 0180-0063 0180-0228 | 1 | CAPACITOR-FXD; 22UF+-10% 15VDC TA-SOLID CAPACITOR-FXD 3.3PF+-10% 500WVDC CAPACITOR-FXD 100PF 100V CAPACITOR-FXD; 500UF+75-10% 3VDC AL CAPACITOR-FXD; 22UF+-10% 15VOC TA-SOLID | 56289 95121 28480 56289 56269 | 1500226X901582 TYPE QC 0180-4801 300507G003DF2 1500226X901582 |
| A2C58 A2C59 A2C61 A2C62 A2C63 | C16C-G174 O18G-J228 O18G-G228 O18G-G106 C18G-G106 | 5 | CAPACITOR-FXD .47UF+80-20% 25WVDC CAPACITOR-FXD; 22UF+-10% 15VDC TA-SULID CAPACITOR-FXD; 22UF+-10% 15VDC TA-SULID CAPACITOR-FXD; 60UF+-20% 6VDC TA-SULID CAPACITOR-FXD; 60UF+-20% 6VDC TA-SULID | 28440 56289 56289 56289 56289 | 0160-0174 1500226x9015t2 1500226x9015t2 1500606x3036f2 1500606x30306f2 |
| A2C64 A2C65 A2C66 A2C67 A2C68 | 016G-0174 018C-0106 0180-4571 0180-0228 0180-0210 | | CAPACITOR-FXD .47UF+80-20% 25WVDC CAPACITOR-FXD; 60UF+-20% 6VDC TA-SOLID CAPACITOR-FXD: 10F +80-20 CAPACITOR-FXD; 22UF+-10% 15VDC TA-SOLID CAPACITOR-FXD; 3.3UF+-20% 15VDC TA | 28480 56289 28480 56289 56289 | 0160-0174 1500606x000682 0160-4571 1500226x901582 1500335x0015A2 |

Table 6-3. Replaceable Parts (Cont'd).

| Reference | HP Part Number | | e 6-3. Replaceable Parts (Cont'd). Description | Mfr | Mfr Part Number |
|--|---|------------------------|---|---|--|
| Designation | rait itallisel | y | Description | Code | wiii i ait ivuilibei |
| A2C69 A2C70,72-75 A2Ck2 A2Ck3 A2Ck4 | 0180-0228 0160-3847 0122-0162 0122-0162 1901-0040 | 2 | CAPACITOR-FXD; 22UF+-10% 15VDC TA-SOLID CAPACITOR-FXD.01uF 50V DIODE: VOLTAGE VARIABLE CAPACITANCE DIODE: VOLTAGE VARIABLE CAPACITANCE DIODE; SWITCHING; ; 30V MAX VRM 50MA | 56289 28480 28480 28480 28480 28480 | 150D226X9015B2 0160-3847 0122-0162 0122-0162 1901-0040 |
| 42CK5 | 1901-0040 | | DIODE; SWITCHING; ; 30V MAX VRM 50MA | 28480 | 1901-2040 |
| A2CR7 A2CR8 A2CR9 | 1901-0040 1901-0040 1902-0041 | 5 | DIODE; SWITCHING; ; 30V MAX VRM 50MA DIODE; SWITCHING; ; 30V MAX VRM 50MA DIODE; ZENER; 5-11V VZ; -4W MAX PD | 28480 28480 04713 | 1901-0040 1901-0040 SZ 10939-98 |
| A2CR11 A2CR12 A2CR13 A2CR14 | 1901-9040 1901-0040 1902-0041 1902-0041 | | DIODE; SWITCHING; ; 30V MAX VRM 50MA DIODE; SWITCHING; ; 30V MAX VRM 50MA DIODE; ZENER; 5-11V VZ; -4W MAX PD DIODE; ZENER; 5-11V VZ; -4W MAX PD | 28480 28480 04713 04713 | 1901-0040 1901-0040 SZ 10939-98 SZ 10939-98 |
| A2CR15 A2CR16 A2L1 A2L2 A2L3 | 1901-0040 1901-0040 9100-3288 914C-0210 9100-0543 | 1 8 1 | DIODE; SWITCHING; ; 30V MAX VRM 50MA DIODE; SWITCHING; ; 30V MAX VRM 50MA INDUCTOR:POT CORE 330 UH COIL; FXD; MOLDED RF CHOKE; 100UH 5% COIL:VAR 1000 UH 10% | 28480 28480 28480 24226 28480 | 1901-0040 1901-0040 9100-3288 15/103 9100-0543 |
| A2L4 A2L5 A2L6 A2L7 A2L8 | 9140-0137 \$100-3278 9100-3277 \$140-0210 9140-0210 | 8 1 4 | COIL; FXD; MOLDED RF CHOKE; 1MH 5% INDUCTOR:POT CORE INDUCTOR:POT CORE COIL; FXD; MOLDED RF CHOKE; 100UH 5% COIL; FXD; MOLDED RF CHOKE; 100UH 5% | 24226 28480 28480 24226 24226 | 19/104 9100-3278 9100-3277 15/103 15/103 |
| A2L9 A2L11 A2L12 A2L13 A2L14 | 9140-0210 9140-0210 9140-0210 9140-0210 9140-0210 | | COIL; FXD; MOLDED RF CHOKE; 100UH 5% | 24226 24226 24226 24226 24226 | 15/103 15/103 15/103 15/103 15/103 |
| A2 MP1 A2 MP2 A2 MP3 A2 Q1 A2 Q2 | 4040-0750 C3580-00609 C3580-00610 1855-0081 1853-0010 | 2 1 1 5 30 | EXTRACTOR:PC BOARD, RED SHIELD, OSCILLATOR SHIELD, CRYSTAL TRANSISTOR; J-FET N-CHAN, D-MODE SI TRANSISTOR PNP SI CHIP PD=360MW | 28480 28480 28480 01295 28480 | 4040-0750 03580-00609 03580-90610 2N5245 1853-9010 |
| A2Q3 A2Q4 A2Q5 A2Q6 A2Q7 | 1853-0010 1854-0071 1654-0071 1855-0234 1855-0081 | 63 1 1 | TRANSISTOR PNP SI CHIP PD=360MW TRANSISTOR NPN SI PD=300MW FT=200MHZ TRANSISTOR NPN SI PD=3)0MW FT=200MHZ TRANSISTOR; FET DUAL ITS30623 TRANSISTOR; J-FET N-CHAN, D-MODE SI | 28480 28480 28480 28480 01295 | 1853-0010 1854-0071 1854-0071 1855-0234 2N5245 |
| A2 Q9 A2 Q11 A2 Q12 A2 Q13 | 1853-001J 1853-0010 1854-0071 1854-0345 | 1 | TRANSISTOR PNP SI CHIP PD=360MW TRANSISTOR PNP SI CHIP PD=360MW TRANSISTOR NPN SI PD=300MW FT=200MHZ TRANSISTOR NPN 2N5179 SI PD=200MW | 28480 28480 28480 04713 | 1853-0010 1853-0010 1854-0071 2N5179 |
| A2 414 A2 415 A2 416 A2 417 A2 418 | 1654-0351 1653-0010 1853-0010 1854-0071 1854-0071 | 1 | TRANSISTOR NPN SI PD=360MW FT=300MHZ TRANSISTOR PNP SI CHIP PD=360MW TRANSISTOR PNP SI CHIP PD=360MW TRANSISTOR NPN SI PD=300MW FT=200MHZ TRANSISTOR NPN SI PD=300MW FT=200MHZ | 28480 28480 28480 28480 28480 | 1854-0351 1853-3010 1853-0010 1854-0071 1854-0071 |
| A2G19 A2G21 A2G23 A2G23 A2G26 A2G26 A2G2 A2G2 A2G2 A2G3 A2G3 A2G3 A2G4 | 1654-0071 1854-6071 1853-3010 1854-0071 1853-0010 1854-0071 1853-0010 C757-0457 C757-0477 C698-5542 C757-0468 | 1 1 2 1 | TRANSISTOR NPN SI PD=300MW FT=200MHZ TRANSISTOR NPN SI PD=300MW FT=200MHZ TRANSISTOR NPN SI CHIP PD=360MW TRANSISTOR NPN SI CHIP PD=360MW FT=200MHZ TRANSISTOR NPN SI CHIP PD=360MW FT=200MHZ TRANSISTOR NPN SI CHIP PD=360MW FT=200MHZ XSTR-PNP SMA713 RESISTOR-FXD 47.5K 1% .125W F TUBULAR RESISTOR-FXD 302K 1% .125W F TUBULAR RESISTOR-FXD 20K 1% .125W F TUBULAR RESISTOR-FXD 909K 1% .125W F TUBULAR | 28480 28480 28480 28480 28480 28480 24546 30583 19701 | 1854-0071 1854-3071 1853-0010 1854-0071 1853-3010 1854-3071 1853-0010 C4-1/8-T0-4752-F MF4C1/8-T9-3323-F MF4C1/8-T9-2002-F MFF-1/8,T-1 |
| A2R5 A2R0 A2R7 A2R6 A2R6 A2R9 | 2100-3352 0698-4536 0757-0430 0757-0440 C698-3274 | 4 1 7 2 4 | RESISTOR, VAR, TRMR, 1KOHM 10% C RESISTOR-FXD 340K 1% -125M F TUBULAR RESISTOR-FXD 2-21K 1% -125M F TUBULAR RESISTOR-FXD 7-5K 1% -125M F TUBULAR RESISTOR-FXD 10K 1% -125M F TUBULAR | 73138 19701 24546 24546 19701 | 72XF132 MF4C1/3-T0-3403-F C4-1/3-T0-2211-F C4-1/3-T0-7501-F MF4C1/9-T9-1002-F |
| A2 R10 A2 R11 A2 R12 A2 R13 A2 R14 | 0757-0430 0757-0438 0757-0438 0757-0438 0757-0416 | 25 5 | RESISTOR-FXD 2-21K 1% -125M F TUBULAR RESISTOR-FXD 5-11K 1% -125M F TUBULAR RESISTOR-FXD 5-11K 1% -125M F TUBULAR RESISTOR-FXD 5-11K 1% -125M F TUBULAR RESISTOR-FXD 511 OHM 1% -125M F TUBULAR | 24546 24546 24546 24546 24546 | C4-1/3-T0-2211-F C4-1/8-T0-5111-F C4-1/3-T0-5111-F C4-1/8-T0-5111-F C4-1/8-T0-5118-F |
| A 2R 15 A 2R 16 A 2R 17 A 2R 18 A 2R 19 | 0698-4481 0683-1055 0757-0427 0698-3497 0698-4443 | 5 2 6 6 2 | RESISTOR-FXD 16.5K 1% .125M F TUBULAR RESISTOR-FXD 1M.05 1/4W RESISTOR-FXD 1.5K 1% .125M F TUBULAR RESISTOR-FXD 6.04K 1% .125M F TUBULAR RESISTOR-FXD 4.53K 1% .125M F TUBULAR | 24546 01121 24546 16299 16299 | C4-1/3-T0-1652-F CB1055 C4-1/3-T0-1501-F C4-1/8-T0-604F-F C4-1/8-T0-4531-F |
| A 2H 21 A 2R 22 A 2R 23 A 2R 24 A 2R 25 | 0757-0430 0757-0280 0757-0442 0757-0427 0757-0415 | 17 25 2 | RESISTOR-FXD 2-21K 1% .125M F TUBULAR RESISTOR-FXD 1K 1% .125M F TUBULAR RESISTOR-FXD 10K 1% .125M F TUBULAR RESISTOR-FXD 1.5K 1% .125M F TUBULAR RESISTOR-FXD 475 OHM 1% .125M F TUBULAR | 24546 24546 24546 24546 24546 | C4-1/8-T0-2211-F C4-1/8-T0-1001-F C4-1/8-T0-1002-F C4-1/8-T0-1501-F C4-1/8-T0-475R-F |

6-4

Table 6-3. Replaceable Parts (Cont'd).

| Reference Designation | HP Part Number | | Description | Mfr Code | Mfr Part Number |
|---|---|---------------------|--|--|---|
| A 2R 26 A 2R 27 A 2R 28 A 2R 29 A 2R 31 | C757-0407 0883-1045 0767-0442 0757-0442 C757-0449 | 13 28 19 2 | RESISTOR-FXD 200 OHM 1% -125W F TUBULAR RESISTOR-FXD 100K .05 1/4W RESISTOR-FXD 10K .01 1/8 RESISTOR-FXD 10K .01 1/8 RESISTOR-FXD 20K 1% -125W F TUBULAR | 24546 01121 24546 24546 24546 | C4-1/8-T0-201-F CB1045 CR-1/8-T0-1002-F CR-1/8-T0-1002-F |
| A 2R 32 A 2R 33 A 2R 34 A 2R 35 A 2R 36 | 0757-0449 0698-3274 0698-3450 0698-3274 0698-3274 | 1 | RESISTOR-FXD 20K 1% .125M F TUBULAR RESISTOR-FXD 10K 1% .125M F TUBULAR RESISTOR-FXD 42.2K 1% .125M F TUBULAR RESISTOR-FXD 10K 1% .125M F TUBULAR RESISTOR-FXD 10K 1% .125M F TUBULAR | 24546 19701 16299 19701 19701 | C4-1/8-T0-2002-F C4-1/8-T0-2002-F MF4C1/8-T9-1002-F C4-1/8-T0-4222-F MF4C1/8-T9-1002-F MF4C1/8-T9-1002-F |
| A 2R 37 A 2R 38 | 0698-5542 C698-6338 | 1 | RESISTOR-FXD 20K 1% -125W F TUBULAR RESISTOR-FXD 5K 1% -125W F TUBULAR | 19701 19701 | MF4C1/8-T9-2002-F MF4C1/8-T9-5001-F |
| A 2R 43 A 2R 44 A 2R 45 A 2R 46 A 2R 46 | 0757-0438 0757-0438 0698-0064 0757-0442 0757-0442 | 1 | RESISTOR-FXD 5-11K 1% -125W F TUBULAR RESISTOR-FXD 5-11K 1% -125W F TUBULAR RESISTOR-FXD 9-31K 1% -125W F TUBULAR RESISTOR-FXD 10K .01 1/8 RESISTOR-FXD 10K .01 1/8 | 24546 24546 91637 24546 24546 | C4-1/8-T0-5111-F C4-1/8-T0-5111-F CMF-1/8-T1-9311-F C4-1/8-T0-1002-F C4-1/8-T0-1002-F |
| A 2R 48 A 2R 49 A 2R 51 A 2R 52 A 2R 53 | C757-0446 0757-0446 C698-4447 0757-0427 0698-4447 | 8 2 | RESISTOR-FXD 15K 1% -125M F TUBULAR RESISTOR-FXD 15K 1% -125M F TUBULAR RESISTOR-FXD 280 OHM 1% -125M F TUBULAR RESISTOR-FXD 1-5K 1% -125M F TUBULAR RESISTOR-FXD 280 OHM 1% -125M F TUBULAR | 24546 24546 24546 24546 24546 | C4-1/8-T0-1502-F C4-1/8-T0-1502-F C4-1/8-T0-280R-F C4-1/8-T0-1501-F C4-1/8-T0-280R-F |
| A 2R 54 A 2R 55 A 2R 56 A 2R 57 A 2R 58 | 0757-0442 C698-4435 0757-0438 0757-0381 0683-0825 | 3 1 1 | RESISTOR-FXD 10K 1% .125M F TUBULAR RESISTOR-FXD 2.49K 1% .125M F TUBULAR RESISTOR-FXD 5.11K 1% .125M F TUBULAR RESISTOR-FXD 15 0HM 1% .125M F TUBULAR RESISTOR-FXD 8-2 0HM 5% .25M CC TUBULAR | 24546 16299 24546 30983 01121 | C4-1/8-T0-1002-F C4-1/8-T0-2491-F C4-1/8-T0-5111-F MF4C1/8-T0-15R0-F C88265 |
| A2R59 A2R61 A2R62 A2R63 A2R64 | 0757-0438 0757-0438 0757-0416 0757-0280 0698-3449 | 2 | RESISTOR-FXD 5-11K 1% -125W F TUBULAR RESISTOR-FXD 5-11K 1% -125W F TUBULAR RESISTOR-FXD 511 OHM 1% -125W F TUBULAR RESISTOR-FXD 1K 1% -125W F TUBULAR RESISTOR-FXD 28-7K 1% -125W F TUBULAR | 24546 24546 24546 24546 24546 16299 | C4-1/8-T0-5111-F C4-1/8-T0-5111-F C4-1/8-T0-511R-F C4-1/8-T0-1001-F C4-1/8-T0-2872-F |
| A2R65* A2R66 A2R67 A2R68 | 0698-4387 C698-4505 0757-0442 C757-0446 | 6 | RESISTOR-FXD 60.4 OHM 1% -125M F FACTORY SELECTED PART (SEE PARAGRAPH 7-19) RESISTOR-FXD 71.5K 1% -125M F TUBULAR RESISTOR-FXD 10K 1% -125M F TUBULAR RESISTOR-FXD 15K 1% -125M F TUBULAR | 16299 24546 24546 24546 | C4-1/8-T0-60R4-F C4-1/8-T0-7152-F C4-1/8-T0-1002-F C4-1/8-T0-1502-F |
| A 2R 69 A 2R 71 A 2R 72 A 2R 73 A 2R 74 | 0883-2235 0757-0416 0883-4745 0698-3558 0698-3558 | 4 3 5 | RESISTOR-FXD 22K.05 1/4W RESISTOR-FXD 511 0HM 1% -125W F TUBULAR RESISTOR-FXD 470K.05 1/4W RESISTOR-FXD 4-02K 1% -125W F TUBULAR RESISTOR-FXD 4-02K 1% -125W F TUBULAR | 01121 24546 01121 16299 16299 | CB2235 C4-1/8-T0-511R-F CB4745 C4-1/8-T0-4021-F C4-1/8-T0-4021-F |
| A2R75 A2R76 A2R77 A2R78 A2R78 | 2100-3054 0698-4486 0757-0280 0698-4486 0757-0416 | 1 13 | RESISTOR, VAR, TRMR 50K OHM 10% C RESISTOR-FXD 24.9K 1% .125M F TUBULAR RESISTOR-FXD 1K 1% .125M F TUBULAR RESISTOR-FXD 24.9K 1% .125M F TUBULAR RESISTOR-FXD 511 OHM 1% .125M F TUBULAR | 32997 24546 24546 24546 24546 24546 | 3006P-1-503 C4-1/8-T0-2492-F C4-1/8-T0-1001-F C4-1/8-T0-2492-F C4-1/8-T0-511R-F |
| A2R81 A2R82 A2R83 A2R84 A2R85 | 0757-0416 C757-0280 0757-0280 0757-0421 0757-0446 | 1 | RESISTOR-FXD 511 OHM 1% -125M F TUBULAR RESISTOR-FXD 1K 1% -125M F TUBULAR RESISTOR-FXD 1K 1% -125M F TUBULAR RESISTOR-FXD 825 OHM 1% -125M F TUBULAR RESISTOR-FXD 15K 1% -125M F TUBULAR | 24546 24546 24546 24546 24546 | C4-1/8-T0-511R-F C4-1/8-T0-1001-F C4-1/8-T0-1001-F C4-1/8-T0-825R-F C4-1/8-T0-1502-F |
| A2R86 A2R87 A2R88 A2R88 A2R89 A2R81 | Co98-3497 0698-4425 0683-2235 0683-1045 0683-1005 | 1 | RESISTOR-FXD 6.04K 1% .125W F TUBULAR RESISTOR-FXD 1.54K 1% .125W F TUBULAR RESISTOR-FXD 22K .05 1/4W RESISTOR-FXD 10K .05 1/4W RESISTOR-FXD 10 .05 1/4W | 16299 16299 01121 01121 01121 | C4-1/8-T0-604R-F C4-1/8-T0-1541-F CB2235 CB1045 CB1005 |
| A 2R 92 A 2R 93 A 2R 94 A 2R 95 A 2R 96 | 0757-0442 0698-4484 0757-0430 0683-3925 0698-4461 | 2 2 2 | RESISTOR-FXD 10K.01 1/8W RESISTOR-FXD 19.1K 1% .125W F TUBULAR RESISTOR-FXD 2.21K 1% .125W F TUBULAR RESISTOR-FXD 300.05 RESISTOR-FXD 300 0HM 1% .125W F TUBULAR | 24546 24546 24546 01121 24546 | C4-1/8-T0-1002-F C4-1/3-T0-1912-F C4-1/3-T0-2211-F CB3925 C4-1/3-T0-6989-F |
| A2R57 A2R98 A2R59 A2R100 A2R101 | 0698-4461 0757-0458 0683-1015 2100-3207 0757-0427 | 4 4 1 | RESISTOR-FXO 698 OHM 1% .125W F TUBULAR RESISTOR-FXO 51-1K 1% .125W F TUBULAR RESISTOR-FXO 100.05 1/4W RESISTOR- VAR, TRMR, 5KOHM 10% C RESISTOR-FXO 1.5K 1% .125W F TUBULAR | 24546 24546 01121 28480 24546 | C4-1/8-T0-698R-F C4-1/3-T0-5112-F CB1015 2100-3207 C4-1/8-T0-1501-F |
| A2K102 A2R103 A2R104 A2R105 A2R106 | 0757-0446 0757-0280 0698-3488 0757-0448 0757-0401 | 5 3 9 | RESISTOR-FXD 15K 1% -125W F TUBULAR RESISTOR-FXD 1K 1% -125W F TUBULAR RESISTOR-FXD 442 OHM 1% -125W F TUBULAR RESISTOR-FXD 18-2K 1% -125W F TUBULAR RESISTOR-FXD 100 OHM 1% -125W F TUBULAR | 24546 24546 16299 24546 24546 | C4-1/8-T0-1502-F C4-1/8-T0-1001-F C4-1/8-T0-422R-F C4-1/8-T0-1822-F C4-1/8-T0-101-F |

Table 6-3. Replaceable Parts (Cont'd).

| Reference Designation | HP Part Number | C D | Qty | Description | Mfr Code | Mfr Part Number |
|--|--|-----------------------|---------------------------------|--|---|---|
| A2R107 A2R108 A2R109 A2R111 A2R112 | C757-0401 0698-4459 0757-0442 0683-1015 0683-4705 | | 1 | RESISTOR-FXD 100 OHM 1% -125M F TUBULAR RESISTOR-FXD 634 OHM 1% -125M F TUBULAR RESISTOR-FXD 10K .01 1/8W RESISTOR-FXD 10C .05 1/4W RESISTOR-FXD 47 .05 1/4W | 24546 24546 24546 01121 01121 | C4-1/8-T0-101-F C4-1/8-T0-634R-F C4-1/8-T0-1002-F CB1015 CB4705 |
| A2R113 R114 A2R115 A2R116 A2R117 A2R118 A2R118 A2R119 | 2100-3357 0683-5635 0683-2215 0757-0442 0698-3495 8150-3375 0757-0416 | | 1 1 1 | RESISTOR, VAR, TRMR, 500KOHM 10% C RESISTOR-FXD 56K.05 1/4W RESISTOR-FXD 220.05 RESISTOR-FXD 10K.01 1/8 RESISTOR-FXD 866.01 1/8 JUMPER WIRE ELEC. RESISTOR-FXD 511.01 1/8W | 73138 01121 01121 24546 24546 75042 24546 | 72 XR 504 CB5635 CB2215 C4-1/8-T0-1002-F C4-1/8-T0-866 8150-3375R-F C4-1/8-T0-511R-F |
| 4201 | 1826-0043 | | 15 | IC;LIN;GPERATIONAL AMPLIFIER | 27014 | LM307H |
| A2U2 A2U3 A2U4 A2U5 A2U6 A2U7 A2U8 A2U8 A2U11 A2U12 A2U113 A2Y1 | 1820-1442 1826-0043 1826-0043 1820-1442 1820-1112 1820-112 1820-1442 1820-0058 1820-1202 1820-1443 1820-0475 | | 1 3 3 3 1 1 1 | IC CNTR SN74LS290N IC;LIN;OPERATIONAL AMPLIFIER IC;LIN;OPERATIONAL AMPLIFIER IC CNTR SN74LS290N IC;F SN74LS74AN IC;LIN;MISCELLANEOUS (LINEAR) IC CNTR SN74LS290N IC;LIN;OPERATIONAL AMPLIFIER IC GATE-DIG SN74LS10N TTL CTR 74LS293N INTEGRATED CIRCUIT, DGTL, VOLTAGE CRYSTAL: NOT FIELD REPLACEABLE (SEE PARAGRAPH 7-25) | 01295 27014 27014 01295 01295 04713 01295 07263 01295 01295 27014 | SN74LS290N LM307H SN74LS290N SN74LS290N SN74LS74AN MC1496G SN74LS290N 709HC SN74LS10N SN74LS293N LM306H |
| A3 | 03580-66503 | 9 | 1 | BOARD ASSEMBLY-SWEEP | 28480 | 03580-66503 |
| A3C1 A3C2 A3C3 A3C4 A3C5 | 0180-1743 0160-2930 0180-0197 0150-0050 0150-0050 | 2 0 8 9 | 3 18 | CAPACITOR=FxD .1UF+=10x 35VDC TA CAPACITOR=FXD .01UF +80=20x 100VDC CER CAPACITOR=FXD 2.2UF+=10x 20VDC TA CAPACITOR=FXD 1000PF +80=20x 1kVDC CER CAPACITOR=FXD 1000PF +80=20x 1kVDC CER | 56289 28480 56289 28480 26480 | 1507104X9035A2 0150=0093 150225X9020A2 0150=0050 0150=0050 |
| A3C6 A3C7 A3C8 A3C9 A3C10 | 0150-0050 0180-1701 0150-0050 0160-2930 0160-2150 | 92905 | | CAPACITOR-FXD 1000PF +80-20X 1KVDC CER CAPACITOR-FXD 6.8UF+-20X 6VDC TA CAPACITOR-FXD 1000PF +80-20X 1KVDC CER CAPACITOR-FXD .01UF +80-20X 100VDC CER CAPACITOR-FXD 33PF +-5X 300VDC MICA | 28480 56289 28480 28480 28480 | 0150-0050 150De85X0006A2 0150-0050 0150-0093 0160-2150 |
| A3C11 A3C12 A3C13 A3C14 A3C15 | 0150-0050 0180-0197 0160-2150 0170-0042 0180-1743 | 9 8 5 1 2 | 5 | CAPACITOR=FXD 1000PF +80=20% 1KVDC CER CAPACITOR=FXD 2.2UF+=10% 20VDC TA CAPACITOR=FXD 33PF +=5% 300VDC MICA CAPACITOR=FXD .33UF +=5% 100VDC POLYE CAPACITOR=FXD .1UF+=10% 35VDC TA | 28480 56289 28480 99515 56289 | 0150-0050 1500225x9020A2 0160-2150 E1-334D 1500104x9035A2 |
| A3C16 A3C17 A3C18 A3C19 A3C20 | 0160-2611 0160-0168 0150-0050 0180-0197 0150-0050 | 3 1 9 8 9 | 1 | CAPACITOR-FXD 1UF +-10% SOVDC MET-POLYE CAPACITOR-FXD 1000PF +80-20% 1KVDC CER CAPACITOR-FXD 2.2UF+-10% 20VDC TA CAPACITOR-FXD 1000PF +80-20% 1KVDC CER | 28480 28480 28480 56289 28480 | 0160-2611 0160-0168 0150-0050 1500225x9020A2 0150-0050 |
| A3C21 A3C22 A3C23 A3C24 | 0160-0170 0160-0170 0160-0170 0140-0199 | 5 5 6 | 3 | CAPACITOR-FXD .22UF +80-20% 25VDC CER CAPACITOR-FXD .22UF +80-20% 25VDC CER CAPACITOR-FXD .22UF +80-20% 25VDC CER CAPACITOR-FXD 24UPF +-5% 300VDC MICA | 28480 28480 28480 72136 | 0160-0170 0160-0170 0160-0170 DM15F241J0300WY1CR |
| A3CR1 A3CR2 A3CR3 A3CR4 A3CR5 | 1901-0040 1902-3128 1910-0016 1910-0016 1910-0016 | 14000 | 3 4 | DIDDE-SHITCHING 30V 50MA 2NS DO-35 DIDDE-ZNR 7,32V 5% DO-35 PD=,4W DIDDE-GE 60V 60MA 1US DO-7 DIDDE-GE 60V 60MA 1US DO-7 DIDDE-GE 60V 60MA 1US DO-7 | 58480 58480 58480 58480 | 1901-0040 1902-3128 1910-0016 1910-0016 1910-0016 |
| A3CR6 A3CR7 A3CR6 A3CR0 A3CR11 | 1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 | 1 1 1 1 1 | | DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 | 28480 28480 28480 28480 | 1901-0040 1901-0040 1901-0040 1901-0060 1901-0040 |
| A3CR12 A3CR13 A3CR14 A3CR15 A3CR16 | 1910-0016 1901-0040 1901-0040 1901-0040 1901-0586 | 0 1 1 1 0 | 1 | DIODE-GE 60V 60MA 1US DO-7 DIODE-SMITCHING 30V 50MA 2NS DO-35 DIODE-SMITCHING 30V 50MA 2NS DO-35 DIODE-SMITCHING 30V 50MA 2NS DO-35 DIODE-GEN PRP 30V 25MA TO-72 | 28480 28480 28480 28480 28480 | 1910-0016 1901-0040 1901-0040 1901-0040 1901-0586 |
| A3CR17 A3CR18 A3CR19 A3CR21 A3CR22 | 1901-0040 | 0 1 1 4 1 | 2 | DIODE-ZNR 12.1V 5% DO-35 PDm.4W DIODE-SWITCHING 30V SOMA 2NS DO-35 DIODE-SWITCHING 30V SOMA 2NS DO-35 DIODE-ZNR 7.32V 5% DD-35 PDm.4W DIODE-SWITCHING 30V SOMA 2NS DO-35 | 28480 28480 28480 28480 28480 | 1902-3182 1901-0040 1901-0040 1902-3128 1901-0040 |
| A3CR23 A3CR24 | 1901-0040 1902-3085 | 1 2 | , | DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-ZNR 4.75V 5% DO-35 PDB.4W | 26460 26460 | 1901-0040 1902-3065 |
| A301 A302 A303 A304 A305 | 1855-0368 1853-0010 | 9 7 7 2 2 | 2 4 1 | TRANSISTOR-JFET DUAL N-CHAN D-MODE TO-78 TRANSISTOR NPN SI PD=300Mm FT=200Mmz TRANSISTOR J-FET N-CHAN D-MODE TO-72 SI TRANSISTOR PNP SI TO-18 PD=360MW TRANSISTOR J-FET P-CHAN D-MODE SI | 28480 28480 28480 28480 28480 | 1855-0237 1854-0071 1855-0368 1853-0010 1855-0082 |

Replaceable Parts

Table 6-3. Replaceable Parts (Cont'd)

| Table 6-3. Replaceable Parts (Cont'd). | | | | | | | | |
|---|--|-----------------------|----------|---|---|--|--|--|
| Reference Designation | HP Part Number | CD | Qty | Description | Mfr Code | Mfr Part Number | | |
| A3G6 A3G7 A3G8 A3G9 A3G11 | 1854-0071 1854-0071 1854-0071 1854-0071 1854-0087 | 7 7 7 7 5 | 4 | TRANSISTOR NPN SI PD=300MH FT=200MHZ TRANSISTOR NPN SI PD=360MH FT=75MHZ | 28480 28480 28480 28480 28480 | 1854-0071 1854-0071 1854-0071 1854-0071 1854-0087 | | |
| A3G12 A3G13 A3G14 A3G15 A3G16 | 1854-0071 1854-0071 1855-0308 1855-0386 1853-0010 | 7 7 5 9 2 | 1 | TRANSISTOR NPN SI PD=300MM FT=200MHZ TRANSISTOR NPN SI PD=300MM FT=200MHZ TRANSISTOR-JFET DUAL N=CHAN D=MDDE SI TRANSISTOR J=FET 2NU392 N=CHAN D=MODE TRANSISTOR PNP SI TO=18 PD=360MM | 28480 28480 28480 04713 28480 | 1854-0071 1854-0071 1855-0308 2M4392 1853-0010 | | |
| A3Q17 A3Q19 A3Q19 A3Q21 A3Q22 | 1854-0071 1854-0071 1854-0071 1853-0010 1854-0087 | 7 7 7 2 5 | | TRANSISTOR NPN SI PDE300MM FTE200MMZ TRANSISTOR NPN SI PDE300MM FTE200MMZ TRANSISTOR NPN SI PDE300MM FTE200MMZ TRANSISTOR NPN SI TP-18 PDE360MM TRANSISTOR NPN SI PDE360MM FTE75MMZ | 28480 28480 28480 | 1854-0071 1854-0071 1854-0071 1853-0010 1854-0087 | | |
| A3Q23 A3Q24 A3Q25 A3Q26 A3Q27 | 1853-0010 1854-0071 1854-0354 1853-0010 1853-0010 | 27922 | | TRANSISTOR PNP SI TU-18 PD=360MW TRANSISTOR NPN SI PD=360MM FT=260MM2 TRANSISTOR NPN SI TO-52 PD=360MM TRANSISTOR NPN SI TO-18 PD=360MM TRANSISTOR PNP SI TO-18 PD=360MM | 28480 28480 28480 28480 | 1853-0010 1854-0071 1854-0354 1853-0010 1853-0010 | | |
| A3Q28 A3Q39 A3Q31 A3Q32 A3Q33 | 1853-0010 1853-0010 1853-0010 1855-036A 1855-0453 | 2 2 7 9 | | TRANSISTOR PNP 31 TO-18 PD=360Mm TRANSISTOR PNP 31 TU-18 PD=360Mm TRANSISTOR PNP 31 TO-18 PD=360Mm TRANSISTOR PFT N-C-FAN D-MODE TO-72 SI TRANSISTOR-JFET DUAL N-CHAN D-MODE TO-78 | 28480 28480 28480 28480 28480 | 1853-0010 1853-0010 1853-0010 1855-0368 1855-0237 | | |
| A3Q34 A3Q35 A3Q36 A3Q37 A3Q38 | 1854-0071 1855-0368 1855-0368 1853-0016 1854-0071 | 7 7 7 8 7 | 5 | TRANSISTOR NPN 31 PD=300MM FT=200MMZ TRANSISTOR J=FET N=CHAN D=MODE TO=72 SI TRANSISTOR J=FET N=CHAN D=MODE TO=72 SI TRANSISTOR PP 31 TU=92 PD=300MM TRANSISTOR NPN SI PD=300MM FT=200MMZ | 28480 28480 28480 28480 28480 | 1854-0071 1855-0368 1855-0368 1853-0016 1854-0071 | | |
| V30n1 V30n3 V30n1 V3036 | 1853-0016 1853-0016 1854-0087 1854-0087 1853-0016 | 8 5 5 | | TRANSISTOR PNP SI TO-92 PD=300MH TRANSISTOR PNP SI TO-92 PD=300MH TRANSISTOR NPN SI PD=300MH FT=75MHZ TRANSISTOR NPN SI PD=300MH FT=75MHZ TRANSISTOR PNP SI TO-92 PD=300MH | 28480 28480 28480 28480 | 1853-0016 1853-0016 1854-0087 1854-0087 1853-0016 | | |
| A3R1 A3R2 A3R3 A3R4 A3R5 | 0698-4479 0757-0426 0698-4479 0757-0272 0684-1031 | 4 9 4 3 9 | 3 | RESISTOR 14K 1% .125W F TC=0+-100 RESISTOR 1.3K 1% .125W F TC=0+-100 RESISTOR 14K 1% .125W F TC=0+-100 RESISTOR 52.3K 1% .125W F TC=0+-100 RESISTOR 52.3K 1% .125W F TC=0+-100 RESISTOR 10K 10% .25W FC TC=-400/+700 | 01151 54249 54249 54249 | C4-1/8-T0-1402-F C4-1/8-T0-1301-F C4-1/8-T0-1402-F C4-1/8-T0-5232-F C81031 | | |
| A3R6 A3R7 A3R8 A3R9 A3R11 | 0594-5641 0694-1041 0684-1041 0684-3331 0757-0457 | 5 1 1 6 | 12 | RESISTOR 560K 10% .25W FC TC=-800/+900 RESISTOR 100K 10% .25W FC TC=-400/+800 RESISTOR 100K 10% .25W FC TC=-400/+800 RESISTOR 33K 10% .25W FC TC=-400/+800 RESISTOR 47.5K 1% .125W F TC=0++100 | 01121 01121 01121 01121 24546 | C85641 C81041 C81041 C83331 C4-1/8-T0-4752-F | | |
| A3R12 A3R13 A3R14 A3R15 A3R16 | 0698-3228 0698-4496 2100-3273 0757-0483 0683-3325 | 9 3 1 8 6 | 5 | RESISTOR 49.9K 1% .125M F TC=0+=100 RESISTOR 24.9K 1% .125M F TC=0+=100 RESISTOR=TRMR 2K 10% C SIDE=ADJ 1=TRN RESISTOR=504K 1% .125M F TC=0+=100 RESISTOR 3.3K 5% .25M FC TC==400/+700 | 28480 28480 28480 01121 | 0698-3228 C4-1/8-T0-2492-F 2100-3273 0757-0483 CB3325 | | |
| 43R17 43R18 43R19 43R21 43R22 | 0684-4731 0684-6831 0684-1041 0757-0442 0698-4486 | 2 7 1 9 3 | 5 | RESISTOR 47K 10% 25K FC TC=-400/+800 RESISTOR 66K 10% 25K FC TC=-400/+800 RESISTOR 100K 10% 25K FC TC=-400/+800 RESISTOR 10K 10% 25K F TC=0+-100 RESISTOR 10K 1% 125K F TC=0+-100 RESISTOR 24.9K 1% 125K F TC=0+-100 | 01121 01121 01121 24546 24546 | C84731 C86831 C81041 C4-1/8-T0-1002-F C4-1/8-T0-2492-F | | |
| A3R23 A3R24 A3R25 A3R26 A3R27 | 0684-1061 0757-0442 0757-0442 0644-1041 0664-1041 | 5 9 1 1 | 1 | RESISTOR 10M 10% 25% FC TC==900/+1100 RESISTOR 10K 1% 125% F TC=0+-100 RESISTOR 10K 1% 125% F TC=0+-100 RESISTOR 10K 10% 25% FC TC==400/+800 RESISTOR 100K 10% 25% FC TC==400/+800 | 01121 24546 24546 01121 01121 | C81001 C4-1/8-T0-1002-F C4-1/8-T0-1002-F C81041 C81041 | | |
| A3R26 A3R29 A3R31 A3R32 A3R33 | 0684-1041 0696-4484 0698-4484 0684-1031 | 1 1 1 9 6 | | RESISTOR 100K 10% .25W FC TC=-400/+800 RESISTOR 19.1K 1% .125W F TC=0+-100 RESISTOR 19.1K 1% .125W F TC=0+-100 RESISTOR 10K 10% .25W FC TC=-400/+700 RESISTOR 28K 1% .125W F TC=0+-100 | 01121 24546 24546 01121 24546 | C81041 C4-1/8-T0-1912-F C4-1/8-T0-1912-F C81031 C4-1/8-T0-2802-F | | |
| A3R34 A3R35 A3R36 A3R37 A3R38 | 0684-1011 0684-1041 0684-1041 0684-4731 0684-2251 | 5 1 1 2 7 | 1 | RESISTOR 100 10% .25% FC TC=-400/+500 RESISTOR 100K 10% .25% FC TC=-400/+800 RESISTOR 100K 10% .25% FC TC=-400/+800 RESISTOR 47K 10% .25% FC TC=-400/+800 RESISTOR 2.2% 10% .25% FC TC=-900/+1100 | 01121 01121 01121 01121 | C81011 C81041 C81041 C84731 C82251 | | |
| A3R39 A3R41 A3R42 A3R43 A3R44 | 0684 -1041 0644-1531 0644-5621 0684-1041 0684-1031 | 6 4 1 1 9 | 16 10 | RESISTOR 100K 1% .125W RESISTOR 15K 10% .25W FC TC==400/+800 RESISTOR 5.6K 10% .25W FC TC==400/+700 RESISTOR 10% 10% .25W FC TC==400/+800 RESISTOR 10K 10% .25W FC TC==400/+700 | 28480 01121 01121 01121 01121 | C83331 C81531 C85621 C81041 C81031 | | |
| A3R45 A3R46 A3R47 A3R48 A3R49 | 0684-1041 0684-4731 0684-1031 0684-1041 0684-4731 | 1 29 1 2 | | RESISTOR 100K 10% .25W FC TC=-400/+800 RESISTOR 47K 10% .25W FC TC=-400/+800 RESISTOR 10K 10% .25W FC TC=-400/+700 RESISTOR 10K 10% .25W FC TC=-400/+800 RESISTOR 47K 10% .25W FC TC=-400/+800 | 01121 01121 01121 01121 | C81041 C84731 C81031 C81041 C84731 | | |

Table 6-3. Replaceable Parts (Cont'd).

| Table 6-3. Replaceable Parts (Cont'd). | | | | | | | | |
|--|--|-----------------------------|------------------|---|--|---|--|--|
| Reference Designation | HP Part Number | C D | Qty | Description | Mfr Code | Mfr Part Number | | |
| A3R51 A3R52 A3R53 A3R54 A3R55 | 0684-4731 0683-1635 0684-1041 2109-3273 0684-1041 | 2 7 1 1 | 1 | RESISTOR 47K 10% .25W FC TC=-400/+800 RESISTOR 16K 5% .25W FC TC=-400/+800 RESISTOR 100K 10% .25W FC TC=-400/+800 RESISTOR-TRMR 2K 10% C SIDE-ADJ 1-TRN RESISTOR 100K 10% .25W FC TC=-400/+800 | 01121 01121 01121 28480 01121 | C84731 C81635 C81041 2100-3273 C81041 | | |
| A3R56 A3R57 A3R58 A3R59 A3R61 | 0684-4731 0684-1041 0684-1041 0684-1041 0684-4731 | 1 1 2 | - | RESISTOR 47K 10% ,25W FC TC=-400/+800 RESISTOR 100K 10% ,25W FC TC=-400/+800 RESISTOR 100K 10% ,25W FC TC=-400/+800 RESISTOR 100K 10% ,25W FC TC=-400/+800 RESISTOR 47K 10% ,25W FC TC=-400/+800 | 01121 01121 01121 01121 | C84731 C81041 C81041 C81041 C84731 | | |
| A3R62 A3R63 A3R64 A3R65 A3R66 | 0684-4741 0684-4731 0684-1041 0684-1041 0698-5922 | 4 2 1 1 4 | 1 | RESISTOR 470K 10% .25W FC TC=-800/+900 RESISTOR 47K 10% .25W FC TC=-400/+800 RESISTOR 100K 10% .25W FC TC=-400/+800 RESISTOR 100K 10% .25W FC TC=-400/+800 RESISTOR 1.8M 1% .5W F TC=0+-100 | 01121 01121 01121 01121 28480 | C84741 C84731 C81041 C81041 0698-5922 | | |
| A3R67 A3R68 A3R69 A3R71 A3R72 | 0698-3572 0698-3499 0757-0449 0757-0449 0757-0426 | 6 6 6 9 | 5 | RESISTOR 60.4K 1% .125W F TC=0+=100 RESISTOR 40.2K 1% .125W F TC=0+=100 RESISTOR 20K 1% .125W F TC=0+=100 RESISTOR 20K 1% .125W F TC=0+=100 RESISTOR 1.3K 1% .125W F TC=0+=100 | 24546 24546 24546 24546 | C4-1/8-T0-6042=F C4-1/8-T0-4022=F C4-1/8-T0-2002=F C4-1/8-T0-2002=F C4-1/8-T0-1301=F | | |
| A3R73 A3R74 A3R75 A3R76 A3R77 | 0757-0272 0757-0449 0684-1041 2100-3357 0698-0077 | 3 6 1 2 0 | 5 | RESISTOR 52.3K 1% .125W F TC=0+=100 RESISTOR 20K 1% .125W F TC=0+=100 RESISTOR 100K 10% .25W FC TC==400/+800 RESISTOR=TRMR 500K 10% C SIDE=ADJ 1=TRN RESISTOR 93.1K 1% .125W F TC=0+=100 | 24546 24546 01121 28480 03888 | C4-1/8-T0-5232-F C4-1/8-T0-2002-F C81041 2100-3357 PME55-1/8-T0-9312-F | | |
| A3R78 A3R79 A3R81 A3R82 A3R83 | 0698-0077 0757-0277 0757-0475 0757-0346 0698-4497 | 9 8 8 0 | 1 1 2 1 | RESISTOR 93.1K 1% .125M F TC=0+-100 RESISTOR 49.9 1% .125M F TC=0+-100 RESISTOR 274K 1% .125M F TC=0+-100 RESISTOR 10 1% .125M F TC=0+-100 RESISTOR 48.7K 1% .125M F TC=0+-100 | 03888 24546 24546 24546 24546 | PME55-1/8-T0-9312-F C4-1/8-T0-4992-F C4-1/8-T0-2743-F C4-1/8-T0-10R0-F C4-1/8-T0-4872-F | | |
| A3R84 A3R85 A3R86 A3R87 A3R89 | 0684-2231 0684-1041 0684-1031 0698-0077 0684-2231 | 3 1 9 0 3 | | RESISTOR 22K 10% 25W FC TC=-400/+800 RESISTOR 100K 10% 25W FC TC=-400/+800 RESISTOR 10K 10% 25W FC TC=-400/+700 RESISTOR 93.1K 1% 125W F TC=0+-100 RESISTOR 22K 10% 25W FC TC=-400/+800 | 01121 01121 01121 03888 01121 | C82231 C81041 C81031 PME55=1/8=T0=9312=F C82231 | | |
| A3R91 A3R92 A3R93 A3R94 A3R95 | 0684-1731 0684-1041 0684-3331 0684-1041 0698-3279 | 2 1 6 1 0 | 6 | RESISTOR 47K 10% .25W FC TC=-400/+800 RESISTOR 100K 10% .25W FC TC=-400/+800 RESISTOR 33K 10% .25W FC TC=-400/+800 RESISTOR 100K 10% .25W FC TC=-400/+800 RESISTOR 4.99K 1% .125W F TC=0+-100 | 01121 01121 01121 01121 24546 | C84731 C81041 C83331 C81041 C4-1/8-T0-4991-F | | |
| A3R96 A3R97 A3R98 A3R99 A3R101 | 0684-1041 0684-1041 0757-0442 0684-2231 0684-2231 | 1 1 9 3 3 | | RESISTOR 100% 10% .25% FC TC==400/+800 RESISTOR 100% 10% .25% FC TC==400/+800 RESISTOR 10% 1% .125% F TC=0+0100 RESISTOR 22% 10% .25% FC TC==400/+800 RESISTOR 22% 10% .25% FC TC==400/+800 | 01121 01121 24546 01121 01121 | CB1041 CB1041 C4-1/8-T0-1002-F CB2231 CB2231 | | |
| A3R102 A3R103 A3R104 A3R105 A3R106 | 0684-1041 0684-1041 0684-2231 0684-5641 0684-1041 | 1 1 3 5 1 | | RESISTOR 100K 10% .25W FC TC==400/+800 RESISTOR 100K 10% .25W FC TC==400/+800 RESISTOR 22K 10% .25W FC TC==400/+800 RESISTOR 560K 10% .25W FC TC==800/+900 RESISTOR 100K 10% .25W FC TC==400/+800 | 01121 01121 01121 01121 | C81041 C81041 C82231 C85641 C81041 | | |
| A3R107 A3R108 A3R109 A3R110 A3R112 | 0684-1041 0684-1041 0684-1041 0696-3279 0684-1041 | 1 1 1 0 1 | | RESISTOR 100K 10% .25W FC TC=-400/+800 RESISTOR 100K 10% .25W FC TC=-400/+800 RESISTOR 100K 10% .25W FC TC=-400/+800 RESISTOR 4,99K 1% .125W F TC=0+-100 RESISTOR 100K 10% .25W FC TC=-400/+800 | 01121 01121 01121 24546 01121 | CB1041 CB1041 CB1041 C4-1/6-T0-4991-F CB1041 | | |
| A301 A3U1 A3U2 A3U3 A3U4 A3U5 | 3101-1312 1826-0043 1820-0223 1826-0043 1820-0223 1820-1418 | 8. 4 0 4 0 7 | 1 4 | SHITCH-SLIDE SPOT NS IC OP AMP GP TO-99 IC OP CAP GP TO-99 IC DCDR TTL LS SCD-TO-DEC 4-TO-10-LINE | 26480 01928 04713 01928 04713 01295 | 3101-1312 CA307T MLM301AG CA307T MLM301AG 8N74L842N | | |
| A3U6 A3U7 A3U8 A3U9 A3U11 | 1820=1574 1820=0594 1820=1197 1820=1204 1820=1144 | 68996 | 1 1 1 3 | IC FF TTL LS J-K PULSE CLEAR DUAL IC FF TTL L J-K M/S PULSE PRESET/CLEAR IC GATE TTL LS NAND QUAD 2-INP IC GATE TTL LS NAND DUAL 4-INP IC GATE TTL LS NOR QUAD 2-INP | 01295 27014 01295 01295 01295 | 3N74L873AN DM74L72N 3N74L30N 3N74L82ON 3N74L82ON 3N74L302N | | |
| A3U12 A3U13 A3U14 A3U15 | 1820-1202 1820-1204 1820-1204 1826-0043 | 7 9 9 | | IC GATE TTL L8 NAND TPL 3-INP IC GATE TTL L8 NAND DUAL 4-INP IC GATE TTL L3 NAND DUAL 4-INP IC OP AMP GP TO-99 | 01295 01295 01295 01295 | 3N74L810N 3N74L820N 3N74L320N CA307T | | |
| A4 | 03581-66504 | 0 | 1 | BOARD ASSEMBLY-DETECTOR | 28480 | 03580-66504 | | |
| A4C1 A4C2 A4C3 A4C4 A4C5 | 0180-0210 0180-0210 0150-0093 0150-0093 0150-0093 | 6 0 0 0 | | CAPACITOR=FXD 3.3UF+=20x 15VDC TA CAPACITOR=FXD 3.3UF+=20x 15VDC TA CAPACITOR=FXD .01UF +80-20X 100VDC CER CAPACITOR=FXD .01UF +80-20X 100VDC CER CAPACITOR=FXD .01UF +80-20X 100VDC CER | 56289 56289 28480 28480 28480 | 150D335x0015A2 150D335x0015A2 0150-0093 0150-0093 | | |

Table 6-3. Replaceable Parts (Cont'd).

| | Number | | Qty | Description | Mfr Code | Mfr Part Number |
|---|---|---|----------|---|---|--|
| Designation A4C6 A4C7 A4C9 A4C11 | 0150-0093 0150-0093 0150-0093 0180-1735 0166-0363 | 8 0 0 0 | 1 3 | CAPACITOR-FXD .01UF +80=20% 100VDC CER CAPACITOR-FXD .01UF +80=20% 100VDC CER CAPACITOR-FXD .01UF +80=20% 100VDC CER CAPACITOR-FXD .22UF+=10% 35VDC TA CAPACITOR-FXD 620PF +=5% 300VDC MICA | 28480 28480 28480 56289 28480 | 0150-0093 0150-0093 0150-0093 150022449035A2 0160-0363 |
| A4C12 A4C13 A4C14 A4C15 A4C16 | 0150-0093 0140-0159 0150-0093 0180-0197 0160-0153 | 08084 | 1 | CAPACITOR=FXD .01UF +RU=20X 100VDC CER CAPACITOR=FXC 3000PF +=2% 300VDC MICA CAPACITOR=FXD .01UF +RU=20X 100VDC CER CAPACITOR=FXD 2.2UF=10% 20VDC TA CAPACITOR=FXD 1000PF +=10% 200VDC POLYE | 28480 72136 28480 56289 28480 | 0150-0093 DM19F302G0300mv1CR 0150-0093 1500225x9020A2 0160-0153 |
| A4C17 A4C18 A4C19 A4C21 A4C22 | 0160-4557 0160-0763 0160-2204 0160-4557 | 9 2 0 9 9 | 1 1 8 | CAPACITOR=FXD .1UF +8C=20x 100VDC CER CAPACITOR=FXO 5PF +-10X 500VDC MICA CAPACITOR=FXD 100PF +-5x 300VDC MICA CAPACITOR=FXD .1UF +8O=20x 100VDC CER CAPACITOR=FXD .1UF +8O=20x 100VDC CER | 28480 28480 28480 28480 28480 | 0150=0084 0160=0763 0160=2204 0150=0084 0150=0084 |
| A4C23 A4C24 A4C25 A4C26 A4C26 | 0160-0763 0160-2204 0160-4557 0160-0763 | 5 6 6 | | CAPACITOR=FXD 5PF +=10% 500VDC MICA CAPACITOR=FXD 100PF +=5% 300VDC MICA CAPACITOR=FXD ,1UF +80=20% 100VDC CER CAPACITOR=FXD ,1UF +80=20% 100VDC CER CAPACITOR=FXD 5PF +=10% 500VDC MICA | 28480 28480 28480 28480 | 0160-0763 0160-2204 0150-0084 0150-0084 0160-0763 |
| A 4 C 2 A A 4 C 2 P A 4 C 3 1 A 4 C 3 2 A 4 C 3 3 | 0160-2204 0160-4557 0160-0763 0160-2204 | 0 9 9 0 | | CAPACITOR-FXD 100PF +=5% 300VDC MICA CAPACITOR-FXD .1UF +80-20% 100VDC CER CAPACITOR-FXD .1UF +80-20% 100VDC CER CAPACITOR-FXD 5PF +=10% 500VDC MICA CAPACITOR-FXD 100PF +-5% 300VDC MICA | 28480 28480 28480 28480 | 0160-2204 0150-0084 0150-0084 0160-0763 0160-2204 |
| A4C34 A4C35 A4C36 A4C37 A4C38 | 0160-3094 0160-0210 0160-2960 0160-0106 0160-0197 | 8 6 5 9 6 | 1 | CAPACITOR-FXD .1UF +=10% 100VDC CER CAPACITOR-FXD 3.3UF+=20% 15VDC TA CAPACITOR-FXD .05UF +=20% 100VDC CER CAPACITOR-FXD 60UF+=20% 6VDC TA CAPACITOR-FXD 2.2UF+=10% 20VDC TA | 28480 56289 28480 56289 56289 | 0160=3094 1500335×0015A2 0160=2960 1500606×0006B2 1500225×902042 |
| A 4C 4 9 A 4C 4 1 A 4C 4 3 A 4C 4 3 | 0160-2605 0160-2605 0150-0093 0160-2204 0150-0022 | 5 0 0 5 | | CAPACITOR-FXD .02UF +A0-20% 25VDC CER CAPACITOR-FXD .02UF +80-20% 25VDC CER CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD 10UPF +5% 300VDC MICA CAPACITOR-FXD 3,3PF +-10% 500VDC TI DIOX | 28480 28480 28480 28480 | 0160-2605 0160-2605 0150-0093 0160-2204 0150-0022 |
| 14C45 14C46 14C47 14C48 14C49 | 0150-0093 0150-0093 0150-0093 0150-0093 0180-0291 | 0 0 0 0 3 | 16 | CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD 1UF++10% 35VDC TA | 28480 28480 28480 28480 56289 | 0150-0093 0150-0093 0150-0093 0150-0093 150D105X9035A2 |
| 14C51 14C52 14C53 14C54 14C55 | 0180-0291 0180-0210 0160-2605 0160-2204 0150-0022 | 3 6 5 0 5 | | CAPACITOR-FXD 1UF+-10% 35VDC TA CAPACITOR-FXD 3.3UF+-20% 15VDC TA CAPACITOR-FXD .02UF +80-20% 25VDC CER CAPACITOR-FXD 100PF +-5% 300VDC MICA CAPACITOR-FXD 3.3PF +-10% 500VDC TI DIOX | 56289 56289 28480 28480 28480 | 1500105xe035A2 1500335x0015A2 0160-2605 0160-2204 0150-0022 |
| 14C56 14C57 14C58 14C59 14C61 | 0180-1743 0150-0093 0150-0093 0160-2605 0160-2605 | 2 0 0 5 5 | | CAPACITOR-FXD .1UF+=10% 35VDC TA CAPACITOR-FXD .01UF +80=20% 100VDC CER CAPACITOR-FXD .01UF +80=20% 100VDC CER CAPACITOR-FXD .02UF +80=20% 25VDC CER CAPACITOR-FXD .02UF +80=20% 25VDC CER | 56289 28480 28480 28480 28480 | 1500104X9035A2 0150-0093 0150-0093 0160-2605 0160-2605 |
| 14C96 14C93 14C93 14C93 | 0160-2960 0160-0763 0160-4557 0160-2960 0160-0154 | 5 9 5 5 | | CAPACITOR=FXD .05UF +=20% 100VDC CER CAPACITOR=FXD 5PF +=10% 500VDC MICA CAPACITOR=FXD .1UF +80=20% 100VDC CER CAPACITOR=FXD .05UF +=20% 100VDC CER CAPACITOR=FXD 2200PF +=10% 200VDC POLYE | 28480 28480 28480 28480 28480 | 0160-2960 0160-0763 0150-0084 0160-2960 0160-0154 |
| 14C67 4C68 4C69 4C70 4C71 | 0160-0157 0140-0198 0160-2960 | 5 5 5 6 | 1 1 | CAPACITOR-FXD 2200PF +-10% 200VDC POLYE CAPACITOR-FXD 4700PF +-10% 200VDC POLYE CAPACITOR-FXD 200PF +-5% 300VDC MICA CAPACITOR-FXD 05UF +-20% 100VDC CER CAPACITOR-FXD 2.2UF++10% 20VDC TA | 28480 28480 72136 28480 56289 | 0160-0154 0160-0157 DM15F201J0300WY1CR 0160-2960 150D225X9020A2 |
| 4C72 4C73 4C74 4C75 4C76 | 0180=1746 0180=0197 0180=0228 | 5 6 6 8 | 15 | CAPACITOR=FXD 15UF+=10X 20VDC TA CAPACITOR=FXD 15UF+=10X 20VDC TA CAPACITOR=FXD 2.2UF+=10X 20VDC TA CAPACITOR=FXD 22UF+=10X 15VDC TA CAPACITOR=FXD 2.2UF+=10X 20VDC TA | 56289 56289 56289 56289 56289 | 150D156X9020B2 150D156X9020B2 150D225X9020A2 150D226X9015B2 150D225X9020A2 |
| 4C77 4C78 4C79 4C81 4C82 | 0180-1746 0180-0197 | 55586 | | CAPACITOR=FXD 15UF+=10% 20VDC TA CAPACITOR=FXD 15UF+=10% 20VDC TA CAPACITOR=FXD 15UF+=10% 20VDC TA CAPACITOR=FXD 22UF+=10% 20VDC TA CAPACITOR=FXD 22UF+=10% 15VDC TA | 56289 56289 56289 56289 56289 | 150D156X902082 150D156X902082 150D156X902082 150D225X9020A2 150D226X901582 |
| 4CR1 4CR2 4CR3 4CR4 4CR5 | 1901-0040 1901-0040 1901-0040 | 1 | | DIODE-SWITCHING 30V 50MA 2NS D0-35 DIODE-SWITCHING 30V 50MA 2NS D0-35 DIODE-SWITCHING 30V 50MA 2NS D0-35 DIODE-SWITCHING 30V 50MA 2NS D0-35 DIODE-SWITCHING 30V 50MA 2NS D0-35 | 28480 28480 28480 28480 28480 | 1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 |

Table 6-3. Replaceable Parts (Cont'd).

| Table 0.3. Replaceable Parts (Cont'd). | | | | | | | | | |
|--|---|----------------------------|--------------|---|---|--|--|--|--|
| Reference Designation | HP Part Number | C D | Qty | Description | Mfr Code | Mfr Part Number | | | |
| AUCR6 AUCR7 AUCR8 AUCR9 AUCR11 | 1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 | 1 1 1 1 1 1 | | DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 | 28480 28480 28480 28480 28480 | 1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 | | | |
| AUCR12 AUCR13 AUCR14 AUCR15 AUCR16 | 1901-0040 1901-0179 1901-0179 1901-0179 1901-0179 | 1 7 7 7 7 | 8 | DIODE-SWITCHING 30V SOMA 2NS DO-35 DIODE-SWITCHING 15V SOMA 750PS DO-7 DIODE-SWITCHING 15V SOMA 750PS DO-7 DIODE-SWITCHING 15V SOMA 750PS DO-7 DIODE-SWITCHING 15V SOMA 750PS DO-7 | 28480 28480 28480 28480 28480 | 1901=0040 1901=0179 1901=0179 1901=0179 1901=0179 | | | |
| A4CR18 A4CR18 A4CR19 A4CR21 A4CR22 | 1901-0179 1901-0179 1901-0179 1901-0179 1901-0040 | 7 7 7 7 1 | | DIODE-SWITCHING 15V SOMA 750PS DO-7 DIODE-SWITCHING 15V SOMA 2NS DO-35 | 28480 28480 28480 28480 28480 | 1901-0179 1901-0179 1901-0179 1901-0179 1901-0040 | | | |
| A4CR24 A4CR24 A4CR25 A4CR26 A4CR27 | 1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 | 1 1 1 1 1 | | DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 | 28480 28480 28480 28480 28480 | 1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 | | | |
| A4CR28 A4CR29 A4CR31 A4CR32 | 1901-0040 1902-3128 1901-0347 1901-0347 | 1 4 1 1 | 2 | DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-ZNR 7.32V 5X DO-35 PD=.4M DIODE-SCHOTTKY BY DIODE-SCHOTTKY BY | 28480 28480 26480 28480 | 1901-0040 1902-3128 1901-0347 1901-0347 | | | |
| A 4 L 1 A 4 L 3 A 4 L 5 | 9100-3261 9100-0541 9140-0129 9140-0129 9140-0129 | 1 1 1 | 1 4 10 | INDUCTORRE-CH-MLD 846UH 5% .8D G=200 INDUCTORRE-CH-MLD 250UH 10% .255%.5LG INDUCTORRE-CH-MLD 220UH 5% .166D%.385LG INDUCTORRE-CH-MLD 220UH 5% .166D%.385LG INDUCTORRE-CH-MLD 220UH 5% .166D%.385LG | 28480 28480 28480 28480 28480 | 9100-3261 9100-0541 9140-0129 9140-0129 | | | |
| A4L6 A4L7 A4L8 A4L9 A4L11 | 9100-0541 9140-0129 9140-0129 9140-0129 | 7 1 1 1 1 | | INDUCTORRF-CH-MLD 250UH 10% .250%.5LG INDUCTORRF-CH-MLD 220UH 5% .1060%.385LG INDUCTORRF-CH-MLD 220UH 5% .1060%.385LG INDUCTORRF-CH-MLD 220UH 5% .1060%.385LG INDUCTORRF-CH-MLD 220UH 5% .1060%.385LG | 28480 28480 28480 28480 28480 | 9100-0541 9140-0129 9140-0129 9140-0129 9140-0129 | | | |
| A401 A402 A403 A404 A405 | 1854-0071 1854-0071 1854-0071 1853-0010 1854-0071 | 7 7 7 2 7 | | TRANSISTOR NPN SI PD=300MW FT=200MHZ TRANSISTOR NPN SI PD=300MW FT=200MHZ TRANSISTOR NPN SI PD=300MW FT=200MMZ TRANSISTOR NPN SI TO=18 PD=360MW TRANSISTOR NPN SI PD=300MW FT=200MMZ | 28480 28480 28480 28480 28480 | 1854-0071 1854-0071 1854-0071 1853-0010 1854-0071 | | | |
| A406 A407 A408 A409 A4011 | 1853-0010 1854-0071 1854-0071 1854-0071 1854-0071 | 2 7 7 7 7 | | TRANSISTOR PNP SI TO-18 PD=360MW TRANSISTOR NPN SI PD=300MW FT=200MMZ | 28480 28480 28480 28480 | 1853-0010 1854-0071 1854-0071 1854-0071 1854-0071 | | | |
| A4012 A4013 A4014 A4015 A4016 | 1854-0071 1854-0071 1854-0071 1854-0071 1853-0010 | 7 7 7 7 7 2 | | TRANSISTOR NPN SI PD=300Mm FT=200MmZ TRANSISTOR PNP SI TO=18 PD=360Mm | 28480 28480 28480 28480 28480 | 1854-0071 1854-0071 1854-0071 1854-0071 1853-0010 | | | |
| AUR1 AUR2 AUR3 AUR4 AUR5 | 2100-3350 2100-3349 2100-3352 2100-3352 2100-3353 | 5 2 7 7 8 | 1 1 | RESISTOR-TRMR 200 10% C SIDE-ADJ 1-TRN RESISTOR-TRMR 100 10% C SIDE-ADJ 1-TRN RESISTOR-TRMR 1K 10% C SIDE-ADJ 1-TRN RESISTOR-TRMR 1K 10% C SIDE-ADJ 1-TRN RESISTOR-TRMR 20K 10% C SIDE-ADJ 1-TRN | 28480 28480 28480 28480 32997 | 2100-3350 2100-3352 2100-3352 3386x-446-203 | | | |
| AUR6 AUR7 AUR8 AUR9 AUR10 | 2100-3351 2100-3273 2100-3273 2100-3354 2100-3354 | 1 1 9 9 | 3 | RESISTOR-TRMR 500 10% C SIDE-ADJ 1-TRN RESISTOR-TRMR 2K 10% C SIDE-ADJ 1-TRN RESISTOR-TRMR 2K 10% C SIDE-ADJ 1-TRN RESISTOR-TRMR 50K 10% C SIDE-ADJ 1-TRN RESISTOR-TRMR 50K 10% C SIDE-ADJ 1-TRN | 28480 28480 28480 28480 28480 | 2100-3351 2100-3273 2100-3273 2100-3354 2100-3354 | | | |
| A4R11 A4R12 A4R13 A4R14 A4R15 | 2100-3273 0757-0449 0757-0449 0757-0274 0757-0438 | 1 6 5 3 | 1 | RESISTOR_TRMR 2K 10% C SIDE_ADJ 1_TRN RESISTOR 20K 1% .125W F TC=0+-100 RESISTOR 20K 1% .125W F TC=0+-100 RESISTOR 1,21K 1% .125W F TC=0+-100 RESISTOR 5.11K 1% .125W F TC=0+-100 | 24546 24246 24246 26480 | 2100-3273 C4-1/8-T0-2002-F C4-1/8-T0-2002-F C4-1/8-T0-1213-F C4-1/8-T0-5111-F | | | |
| A4R16 A4R17 A4R18* A4R19 A4R20 | 0698-3449 0698-4436 0757-0282 0684-1031 | 6 3 5 9 | 3 | RESISTOR 28.7K 1% .125W F TC=0+-100 RESISTOR 2.8K 1% .125W F TC=0+-100 RESISTOR 2.81 1% .125W F TC=0+-100 RESISTOR 10K 10% .25W FC TC=+400/+700 RESISTOR 10K 10% .25W FC TC=+400/+700 | 24546 24546 24546 01121 | C4-1/8-T0-2872-F C4-1/8-T0-2801-F C4-1/8-T0-221R-F C81031 C81031 | | | |
| A4R21 = A4R23 A4R23 A4R24 A4R25 | 0757-0280 0757-0469 0757-0469 | 0 3 0 0 3 | 1 4 | RESISTOR 340 1% .125W F TC=0+-100 RESISTOR 1K 1% .125W F TC=0+-100 RESISTOR 150K 1% .125W F TC=0+-100 RESISTOR 150K 1% .125W F TC=0+-100 RESISTOR 1K 1% .125W F TC=0+-100 | 24546 24546 24546 24546 | C4=1/8=T0=287R=F C4=1/8=T0=1001=F C4=1/8=T0=1503=F C4=1/8=T0=1503=F C4=1/8=T0=1001=F | | | |
| | | | | | | | | | |
| | | L_ | | | | | | | |

Table 6-3. Replaceable Parts (Cont'd).

| Table 6-3. Replaceable Parts (Cont'd). | | | | | | | | |
|--|---|-----------------------|-------------|---|---|---|--|--|
| Reference Designation | HP Part Number | C D | Qty | Description | Mfr Code | Mfr Part Number | | |
| A4R26 A4R27 A4R28 A4R29 A4R31 | 0757-0449 0757-0449 0684-3331 0684-1031 | 00000 | | RESISTOR 20K 1% .125W F TC=0+-100 RESISTOR 20K 1% .125W F TC=0+-100 RESISTOR 33K 10% .25W FC TC=-400/+800 RESISTOR 10K 10% .25W FC TC=-400/+700 RESISTOR 10K 10% .25W FC TC=-400/+700 | 24546 24546 01121 01121 | C4-1/8-T0-2002-F C4-1/8-T0-2002-F C83331 C81031 | | |
| A4R32 A4R33 A4R34 A4R35 A4R36 | 0684-3331 0684-1031 0684-1031 0684-3331 0684-3331 | 69966 | | RESISTOR 33K 10% .25W FC TC==400/+80U RESISTOR 10K 10% .25W FC TC==400/+700 RESISTOR 10K 10% .25W FC TC==400/+800 RESISTOR 33K 10% .25W FC TC==400/+800 RESISTOR 33K 10% .25W FC TC==400/+800 | 01121 01121 01121 01121 | C83331 C81031 C81031 C83331 C83331 | | |
| 44R37 44R38 44R39 44R41 44R42 | 0684-1831 0684-1531 0757-0426 0757-0394 0757-0401 | 7 4 9 0 0 | 5 | RESISTOR 18K 10% .25W FC TC=-400/+800 RESISTOR 15K 10% .25W FC TC=-400/+800 RESISTOR 1.3K 1% .125W F TC=0+-100 RESISTOR 51.1 1% .125W F TC=0+-100 RESISTOR 100 1% .125W F TC=0+-100 | 01121 01121 24546 24546 24546 | CB1831 CB1531 C4-1/8-T0-1301-F C4-1/8-T0-51R1-F C4-1/8-T0-101-F | | |
| A4R43 A4R44 A4R45 A4R46 A4R47 | 0698-3488 0757-0401 0757-0401 0698-4483 0757-0465 | 3 0 0 0 6 | 5 11 | RESISTOR 442 1% .125W F TC=0+-100 RESISTOR 100 1% .125W F TC=0+-100 RESISTOR 100 1% .125W F TC=0+-100 RESISTOR 18.7K 1% .125W F TC=0+-100 RESISTOR 100K 1% .125W F TC=0+-100 | 54249 54249 54249 54249 54249 | C4-1/8-T0-422R-F C4-1/8-T0-101-F C4-1/8-T0-101-F C4-1/8-T0-187Z-F C4-1/8-T0-1003-F | | |
| A4R48 A4R49 A4R51 A4R52 A4R53 | 0698-4483 0664-5641 0684-1531 0683-2225 0684-1031 | 0 5 4 3 9 | 3 | RESISTOR 18.7K 1% .125W F TC=0+-100 RESISTOR 560K 10% .25W FC TC=-800/+900 RESISTOR 15K 10% .25W FC TC=-400/+800 RESISTOR 2.2K 5% .25W FC TC=-400/+700 RESISTOR 10K 10% .25W FC TC=-400/+700 | 24546 01121 01121 01121 01121 | C4-1/8-T0-1872-F C85641 C81531 C82225 C81031 | | |
| A4R54 A4R55 A4R56 A4R57 A4R58 | 0684-4731 0684-1031 0698-4434 0757-0346 0757-0280 | 2 1 2 3 | 1 | RESISTOR 47K 10% _25W FC TC==400/+800 RESISTOR 10K 10% _25W FC TC==400/+700 RESISTOR 2.32K 1% _125W F TC=0+=100 RESISTOR 10 1% _125W F TC=0+=100 RESISTOR 1K 1% _125W F TC=0+=100 | 01121 01121 24546 24546 24546 | C84731 C81031 C4-1/8-T0-2321-F C4-1/8-T0-10R0-F C4-1/8-T0-1001-F | | |
| A4R59 A4R62* A4R63 A4R64* A4R65* | 0757-0273 0098-4488 0757-0273 0698-4488 0698-4488 | 4 5 4 5 5 | 5 | RESISTOR 3.01K 1X .125w F TC=0+=100 RESISTOR 20.7K 1X .125w F TC=0+=100 RESISTOR 3.01K 1X .125w F TC=0+=100 RESISTOR 20.7K 1X .125w F TC=0+=100 RESISTOR 20.7K 1X .125w F TC=0+=100 | 24246 54246 54246 54246 | C4-1/8-T0-3011-F C4-1/8-T0-2072-F C4-1/8-T0-3011-F C4-1/8-T0-2072-F C4-1/8-T0-2672-F | | |
| A4R66 A4R67* A4R68 A4R69 A4R71* | 0757-0273 0698-3245 0698-3279 0757-0273 0698-4482 | 40040 | ú | RESISTOR 3.01K 1X .125W F TC=0+-100 RESISTOR 20.5K 1X .125W F TC=0+-100 RESISTOR 4.99K 1X .125W F TC=0+-100 RESISTOR 3.01K 1X .125W F TC=0+-100 RESISTOR 17.4K 1X .125W F TC=0+-100 | 24546 24546 24546 24546 03888 | C4-1/8-T0-3011-F C4-1/8-T0-2052-F C4-1/8-T0-2052-F C4-1/8-T0-3011-F PME55-1/8-T0-1742-F | | |
| A4R72 A4R73 A4R74 A4R75 A4R76 | 0698-3558 0698-3497 0757-0430 0698-3228 0698-3516 | 8 4 5 9 8 | 6 | RESISTOR 4.02K 1% .125W F TC=0+=100 RESISTOR 6.04K 1% .125W F TC=0+=100 RESISTOR 2.21K 1% .125W F TC=0+=100 RESISTOR 49.9K 1% .125W F TC=0+=100 RESISTOR 6.34K 1% .125W F TC=0+=100 | 24546 24546 24546 28480 24546 | C4-1/8-T0-4021-F C4-1/8-T0-604R-F C4-1/8-T0-6221-F 0698-3228 C4-1/8-T0-6341-F | | |
| A4R77 A4R78 A4R79 A4R81 A4R82 | 0757-0434 0757-0449 0683-1515 0683-1515 0757-0442 | 9 6 2 2 9 | 5 | RESISTOR 3,65K 1% .125W F TC=0+-100 RESISTOR 20K 1% .125W F TC=0+-100 RESISTOR 150 5% .25W FC TC=-400/+600 RESISTOR 150 5% .25W FC TC=-400/+600 RESISTOR 10K 1% .125W F TC=0+-100 | 24546 24546 01121 01121 24546 | C4-1/8-T0-3651-F C4-1/8-T0-2002-F C81515 C81515 C4-1/8-T0-1002-F | | |
| 4483* A4884 A4885 A4886 A4887 | 0698-4403 0664-3331 0684-3331 0757-0465 0757-0427 | 4 6 6 6 0 | 1 | RESISTOR 102 1% .125W F TC=0+-100 RESISTOR 33K 10% .25W FC TC=-400/+800 RESISTOR 33K 10% .25W FC TC=-400/+800 RESISTOR 100K 1% .125W F TC=0+-100 RESISTOR 1.5K 1% .125W F TC=0+-100 | 24546 01121 01121 24546 24546 | C4-1/8-T0-102R-F C63331 C63331 C4-1/8-T0-1003-F C4-1/8-T0-1501-F | | |
| A4R88 A4R89 A4R91 A4R92 A4R93 | 0698-3557 0757-0465 0757-0449 0684-3331 0684-3331 | 7 6 6 6 | 2 | RESISTOR 806 1% .125W F TC=0+-100 RESISTOR 100K 1% .125W F TC=0+-100 RESISTOR 20K 1% .125W F TC=0+-100 RESISTOR 33K 10% .25W FC TC=-400/+800 RESISTOR 33K 10% .25W FC TC=-400/+800 | 24546 24546 24546 01121 01121 | C4-1/8-T0-806R-F C4-1/8-T0-1003-F C4-1/8-T0-2002-F C83331 C83331 | | |
| A4R94 A4R95 A4R96 A4R97 A4R98 | 0684-4741 0684-4741 0684-1041 0684-3331 0757-0442 | 1 6 9 | | RESISTOR 470K 10% .25W FC TC==800/+900 RESISTOR 470K 10% .25W FC TC==800/+900 RESISTOR 100K 10% .25W FC TC==400/+800 RESISTOR 33K 10% .25W FC TC==400/+800 RESISTOR 10K 1% .125W F TC=0+=100 | 01121 01121 01121 01121 24546 | C84741 C84741 C81041 C83331 C4-1/8-T0-1002-F | | |
| A4R99 A4R101 A4R102 A4R103+ A4R104+ | 0757-0442 0757-0442 0698-4475 0698-4442 0698-4466 | 9 9 0 1 9 | 1 1 1 | RESISTOR 10K 1% .125W F TC=0+-100 RESISTOR 10K 1% .125W F TC=0+-100 RESISTOR 9.76K 1% .125W F TC=0+-100 RESISTOR 4.42K 1% .125W F TC=0+-100 RESISTOR 976 1% .125W F TC=0+-100 | 24546 24546 03888 24546 24546 | C4=1/8-T0=1002=F C4=1/8-T0=1002=F PME55=1/8-T0=9761=F C4=1/8-T0=4421=F C4=1/8-T0=976R=F | | |
| A4R105+ A4R106 A4R107 A4R108 A4R109+ | 0698-3441 0757-0401 0757-0465 0698-4435 0698-4429 | 20624 | 1 | RESISTOR 215 1% .125W F TC=0+=100 RESISTOR 100 1% .125W F TC=0+=100 RESISTOR 100K 1% .125W F TC=0+=100 RESISTOR 2.49K 1% .125W F TC=0+=100 RESISTOR 1.87K 1% .125W F TC=0+=100 | 24546 24546 24546 24546 | C4-1/8-T0-210R-F C4-1/8-T0-101-F C4-1/8-T0-103-F C4-1/8-T0-2491-F C4-1/8-T0-1871-F | | |
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Revised: September 1987

Table 6-3. Replaceable Parts (Cont'd).

| Table 6-3. Keplaceable Parts (Cont'd). | | | | | | | | | |
|--|---|-----------------------|------------------------|---|---|---|--|--|--|
| Reference Designation | HP Part Number | C D | Qty | Description | Mfr Code | Mfr Part Number | | | |
| A4R111 A4R112 A4R113 A4R114 A4R115 | 0698-3279 0684-2241 0757-0465 0757-0446 0757-0427 | 0 5 6 3 0 | 1 | RESISTOR 4.99K 1X .125W F TC=0+=100 RESISTOR 220K 10X .25W FC TC==800/+900 RESISTOR 100K 1X .125W F TC=0+=100 RESISTOR 15K 1X .125W F TC=0+=100 RESISTOR 1.5K 1X .125W F TC=0+=100 | 24546 01121 24546 24546 24546 | C4-1/8-T0-4991-F C82241 C4-1/8-T0-1003-F C4-1/8-T0-1502-F C4-1/8-T0-1501-F | | | |
| A4R116 A4R117 A4R118 A4R119 A4R120 | 0757-0407 0694-1531 0684-1031 0684-3341 0684-4721 | 64000 | 1 | RESISTUR 200 1% .125W F TC=0+-100 RESISTOR 15K 10% .25W FC TC=-400/+800 RESISTOR 10K 10% .25W FC TC=-400/+700 RESISTOR 330K 10% .25W FC TC=-800/+700 RESISTOR 4,7K 10% .25W FC TC=-400/+700 | 24546 01121 01121 01121 01121 | C4-1/8-T0-201-F C81531 C81031 C83341 C84721 | | | |
| A4R121 A4R122A A4R122B A4R123 A4R124 | 0698-3499 0698-4509 0757-0465 0698-4539 0757-0442 | 6 1 7 7 | 2 1 1 | RESISTOR 40.2K 1% .125W F TC=0+-100 RESISTOR 80.6K 1% .125W F TC=0+-100 RESISTOR 100K 1% .125W F TC=0+-100 RESISTOR 402K 1% .125W F TC=0+-100 RESISTOR 10K 1% .125W F TC=0+-100 | 24546 24546 03888 28480 24546 | C4-1/8-T0-4022-F C4-1/8-T0-8062-F PME55-1/8-T0-7762-F 0698-4539 C4-1/8-T0-1002-F | | | |
| A4R125 A4R126 A4R127 A4R128 A4R129 | 0757-0280 0757-0449 0757-0449 0757-0280 0757-0280 | 3 6 6 3 3 | | RESISTOR 1K 1% .125W F TC=0+=100 RESISTOR 20K 1% .125W F TC=0+=100 RESISTOR 20K 1% .125W F TC=0+=100 RESISTOR 1K 1% .125W F TC=0+=100 RESISTOR 1K 1% .125W F TC=0+=100 | 24546 24546 24546 24546 24546 | C4-1/8-T0-1001-F C4-1/8-T0-2002-F C4-1/8-T0-2002-F C4-1/8-T0-1001-F C4-1/8-T0-1001-F | | | |
| A4R130 A4R131 A4R132 A4R133 A4R134 | 0698-3499 0698-3499 0698-4473 0757-0458 0698-3279 | 6 6 8 7 0 | 4 | RESISTOR 40.2K 1X .125W F T(=0+-100 RESISTOR 40.2K 1X .125W F T(=0+-100 RESISTOR 6.06K 1X .125W F T(=0+-100 RESISTOR 51.1K 125W F T(=0+-100 RESISTOR 4.99K 1X .125W F T(=0+-100 | 24546 24546 24546 24546 24546 | C4-1/8-T0-4022-F C4-1/8-T0-4022-F C4-1/8-T0-8081-F C4-1/8-T0-5112-F C4-1/8-T0-4991-F | | | |
| Auri35 Auri36 Auri37 Auri38 Auri39 | 0757-0317 0698-3264 0757-0280 0757-0438 0757-0288 | 7 3 3 3 1 | 1 1 | RESISTOR 1,33K 1% .125M F TC=0+-100 RESISTOR 11,8K 1% .125M F TC=0+-100 RESISTOR 1K 1% .125M F TC=0+-100 RESISTOR 5,11K 1% .125M F TC=0+-100 RESISTOR 9,09K 1% .125M F TC=0+-100 | 24546 24546 24546 24546 19701 | C4-1/8-T0-1331-F C4-1/8-T0-1182-F C4-1/8-T0-1001-F C4-1/8-T0-5111-F MF4C1/8-T0-9091-F | | | |
| AUR140 AUR141 AUR142 AUR143 AUR144 | 0698-4484 0757-0453 0757-0458 0757-0439 0698-3268 | 1 2 7 4 7 | 2 1 5 | RESISTOR 19.1K 1% .125W F TC=0+=100 RESISTOR 30.1K 1% .125W F TC=0+=100 RESISTOR 51.1K 1% .125W F TC=0+=100 RESISTOR 6.81K 1% .125W F TC=0+=100 RESISTOR 11.5K 1% .125W F TC=0+=100 | 24546 24546 24546 24546 | C4-1/8-T0-1912-F C4-1/8-T0-3012-F C4-1/8-T0-5112-F C4-1/8-T0-6811-F C4-1/8-T0-1152-F | | | |
| A4R145 A4R146 A4R147 A4R148 A4R148 | 0757-0438 0684-6831 0684-5621 0698-4307 0757-0444 | 3 7 1 7 1 | 1 1 | RESISTOR 5.11K 1% .125W F TC=0+=100 RESISTOR 68K 10% .25W FC TC==400/+800 RESISTOR 5.6K 10% .25W FC TC==400/+700 RESISTOR 14.3K 1% .125W F TC=0+=100 RESISTOR 12.1K 1% .125W F TC=0+=100 | 24546 01121 01121 24546 24546 | C4-1/8-T0-5111-F C8-831 C8-5621 C4-1/8-T0-1432-F C4-1/8-T0-1212-F | | | |
| A4R150 | 0684-1531 | 4 | | RESISTOR 15K 10% .25W FC TC==400/+800 | 01121 | CB1531 | | | |
| A4RT1 | 0837-0050 | 5 | 1 | THERMISTOR DISC 1K-DHM TC=-4.4%/C-DEG | 28480 | 0837-0050 | | | |
| A4U1 A4U3 A4U4 A4U5 | 1826-0109 1826-0109 1826-0109 1826-0109 1813-0017 | 3 3 3 5 | 1 | IC OP AMP WB TO-99 LOGIC AMPLIFIER | 34371 34371 34371 34371 26460 | HA2-2625-80593 HA2-2625-80593 HA2-2625-80593 HA2-2625-80593 1813-0017 | | | |
| A4U6 A4U7 A4U6 A4U9 A4U10 | 1620-0058 1820-0058 1826-0043 1826-0043 1826-0043 | 00444 | | IC OP AMP GP TO=99 | 24046 24046 01928 01928 | TOA 2709V TOA 2709V CA307T CA307T | | | |
| A4U1 1 A5 A5 ²² A5 ²²² | 1826-0043 03580-66505 03580-69515 03580-69505 | 1 | 1 | IC OP AMP GP TO=09 BOARD ASSEMBLY-IF FILTER KIT:BOARD ASSY:IF FILTER REBUILT EXCHANGE ASSEMBLY | 01928 28480 28480 28480 | CA307T 03580-66505 03580-69515 03580-69505 | | | |
| ASC1 ASC2 ASC3 ASC4 ASC5 | 0121-0426 0121-0059 0121-0105 0121-0426 0121-0059 | 2 7 4 2 7 | 5 5 | CAPACITOR=V TRMR=MICA 50-380PF 175V CAPACITOR=V TRMR=CER 2-8PF 350V PC=MTG CAPACITOR=V TRMR=CER 9-35PF 200V PC=MTG CAPACITOR=V TRMR=MICA 50-380PF 175V CAPACITOR=V TRMR=CER 2-8PF 350V PC=MTG | 72136 52763 52763 72136 52763 | T52517-7 304324 2/6PF NPO 304324 9/35PF N650 T52517-7 304324 2/8PF NPO | | | |
| A5C6 A5C7 A5C8 A5C9 A5C10 | 0121-0105 0121-0426 0121-0059 0121-0105 0121-0426 | 4 2 7 4 2 | | CAPACITOR-V TRMR-CER 9-35PF 200V PC-MTG CAPACITOR-V TRMR-MICA 50-380PF 175V CAPACITOR-V TRMR-CER 2-8PF 350V PC-MTG CAPACITOR-V TRMR-CER 9-35PF 200V PC-MTG CAPACITOR-V TRMR-MICA 50-380PF 175V | 52763 72136 52763 52763 72136 | 304324 9/35PF N650 T52517-7 304324 2/8PF NPO 304324 9/35PF N650 T52517-7 | | | |
| A5C11 A5C12 A5C13 A5C14 A5C15 | 0121-0059 0121-0105 0121-0426 0121-0059 0121-0105 | 7 4 2 7 4 | | CAPACITOR-V TRMR-CER 2-8PF 350V PC-MTG CAPACITOR-V TRMR-CER 9-35PF 200V PC-MTG CAPACITOR-V TRMR-MICA 50-380PF 175V CAPACITOR-V TRMR-CER 2-8PF 350V PC-MTG CAPACITOR-V TRMR-CER 9-35PF 200V PC-MTG | 52763 52763 72136 52763 52763 | 304324 2/8PF NPO 304324 9/35PF Ne50 152517-7 304324 2/8PF NPO 304324 9/35PF Ne50 | | | |
| A5C17 A5C18 A5C19 A5C21 A5C22 | 0140-0218 0160-5269 | 0 2 0 5 5 | 5 | CAPACITOR-FXD 300FF +-5% 300VDC MICA CAPACITOR-FXD 5PF +-10% 500VDC MICA CAPACITOR-FXD 160FF +-2% 300VDC MICA CAPACITOR-FXD 047UF ±20% 100VDC CER CAPACITOR-FXD 0.02UF +80-20% 25VDC CER | 72136 28480 72136 28480 28480 | DM15F3q1J0300WY1CR 0160-0763 DM15F161G0300WY1CR 0160-5269 0160-2605 | | | |
| *KIT INCLUDES NEW AS (**EXCHANGE KIT INCLUDES | (IF FILTER) ASS'Y S REBUILT A5 (IF. | AND FILT | MATCHED C ER) BOARD | RYSTAL FOR REPLACING A2Y1 (SEE PARAGRAPH 7-25). AND MATCHED CRYSTAL FOR REPLACING A2Y1 (SEE PARAGRAP) | d 7-25). | | | | |

Table 6-3. Replaceable Parts (Cont'd).

| Reference Designation | HP Part Number | C | Qty | Description | Mfr Code | Mfr Part Number |
|--|---|-----------------------|-----|--|---|---|
| ASC23 ASC25 ASC26 ASC27 ASC28 | 0150-0291 0150-0093 0160-2605 0160-5269 0140-0200 | 30550 | | CAPACITOR-FXD 1UF++10% 35VDC TA CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD .02UF +80-20% 25VDC CER CAPACITOR-FXD .07UF ±20% 100VDC CER CAPACITOR-FXD .390PF +-5% 300VDC MICA | 56289 28480 28480 28480 72136 | 1500105X9035A2 0150-0093 0160-2605 0160-269 DM15F391J0300WV1CR |
| A5C31 A5C32 A5C33 A5C33 A5C34 | 0160-0763 0140-0218 0160-3960 0160-2605 0180-0291 | 2 0 7 5 3 | 1 | CAPACITOR-FXD SPF +-10X 500VDC MICA CAPACITOR-FXD 160PF +-2X 300VDC MICA CAPACITOR-FXD 1000PF +-20X 8KVDC CAPACITOR-FXD 02UF +80-20X 25VDC CER CAPACITOR-FXD 1UF+-10X 35VDC TA | 28480 72136 28480 28480 56289 | 0160-0763 DM15F16160300mv1cR 0160-3960 0160-2605 1500105X9035A2 |
| 45C36 45C37 45C38 45C39 45C41 | 0150-0093 0160-2605 0160-5269 0140-0200 0160-0763 | 0 5 5 0 2 | | CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD .02UF +80-20% 25VDC CER CAPACITOR-FXD .047UF ±20% 100VDC CER CAPACITOR-FXD 300PF +-5% 300VDC MICA CAPACITOR-FXD 5PF +-10% 500VDC MICA | 28480 28480 28480 72136 28480 | 0150-0093 0160-2605 0160-5269 pm15f391J0300my1CR 0160-0763 |
| ASC 42 ASC 43 ASC 44 ASC 45 ASC 47 | 0140-7318 0160-52 0160-2605 0180-0291 0150-0093 | 0 5 5 0 | | CAPACITOR-FXD 160PF +=2% 300VDC MICA CAPACITOR-FXD .047UF ±20% 100VDC CER CAPACITOR-FXD .02UF +80-20% 25VDC CER CAPACITOR-FXD 1UF++10% 35VDC TA CAPACITOR-FXD .01UF +80-20% 100VDC CER | 72136 28480 28480 56289 28480 | DM15F161G0300MV1CR 0160-5269 0160-2605 1500105X9035A2 0150-0093 |
| 45C48 45C49 45C51 45C52 45C53 | 0160-2605 0160-5269 0140-0200 0160-0763 0140-0218 | 5 0 2 0 | | CAPACITOR-FXD .02UF +80-20% 25VDC CER CAPACITOR-FXD .047UF ±20% 100VDC CER CAPACITOR-FXD 300FF +-5% 300VDC MICA CAPACITOR-FXD 55FF +-10% 500VDC MICA CAPACITOR-FXD 160FF +-2% 300VDC MICA | 28480 28480 72136 28480 72136 | 0160-2605 0160-5269 DM15F391J0300WY1CR 0160-0763 DM15F161G0300WV1CR |
| A5C54 A5C55 A5C56 A5C58 A5C58 | 0160-5269 0160-2605 0140-0291 0150-0693 0160-2605 | 5 5 3 0 5 | | CAPACITOR-FXD .047UF ±20% 100VDC CER CAPACITOR-FXD .02UF +80-20% 25VDC CER CAPACITOR-FXD 1UF++10% 35VDC TA CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD .02UF +80-20% 25VDC CER | 28480 28480 56289 28480 28480 | 0160-5269 0160-2605 1500105x9035A2 0150-0093 0160-2605 |
| 45Cb1 45Cb2 45Cb3 45Cb4 45Cb5 | 0160-5269 0140-0200 0160-0763 0140-0218 0160-5269 | 5 0 2 0 5 | | CAPACITOR-FXD.047UF ±20% 100VDC CER CAPACITOR-FXD 390PF +=5% 300VDC MICA CAPACITOR-FXD 5PF +=10% 500VDC MICA CAPACITOR-FXD 100PF +=2% 300VDC MICA CAPACITOR-FXD 100PF +=2% 300VDC CER | 28480 72136 28480 72136 28480 | 0160-5269 DM15F391J0300WV1CR 0160-0763 DM15F161G0300WV1CR 0160-5269 |
| A5C66 A5C67 A5C68 A5C69 A5C71 | 0160-0195 0160-0291 0180-0291 0150-0093 0150-0093 | 3 3 0 | 1 | CAPACITOR-FXD 1000PF +-20% 250VAC(RM3) CAPACITOR-FXD 1UF++10% 35VDC TA CAPACITOR-FXD 1UF++10% 35VDC TA CAPACITOR-FXD 01UF +80-20% 100VDC CER CAPACITOR-FXD 01UF +80-20% 100VDC CER | 28480 56289 56289 28480 28480 | 0160-0195 150D105X9035A2 150D105X9035A2 0150-0093 0150-0093 |
| A5C72 A5C73 A5C74 A5C75 A5C76 | 0160-2605 0160-5269 0160-2605 0160-5269 0160-5269 | 5 5 5 5 5 | | CAPACITOR-FXD .02UF +80-20% 25VDC CER CAPACITOR-FXD .047UF ±20% 100VDC CER CAPACITOR-FXD .02UF +80-20% 25VDC CER CAPACITOR-FXD .047UF ±20% 100VDC CER CAPACITOR-FXD .047UF ±20% 100VDC CER | 28480 28480 58480 58480 58480 | 0160=2605 0160-5269 0160-2605 0160-5269 0160-5269 |
| ASC77 ASC78 ASC79 ASC81 ASC82 | 0160-5269 0180-0061 0180-0061 0180-0061 0180-0061 | 5 5 5 5 5 | | CAPACITOR-FXD .047UF ±20% 100VDC CER CAPACITOR-FXD 100UF+75-10% 16VDC AL CAPACITOR-FXD 100UF+75-10% 16VDC AL CAPACITOR-FXD 100UF+75-10% 16VDC AL CAPACITOR-FXD 100UF+75-10% 16VDC AL | 28480 56289 56289 56289 56289 | 0160-5269 30010760160C2 30010760160C2 30010760160C2 30010760160C2 |
| ASCR1 ASCR2 ASCR3 ASCR4 ASCR4 | 1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 | 1 1 1 1 1 | | DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 | 28480 28480 28480 28480 28480 | 1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 |
| ASCR6 ASCR7 ASCR8 ASCR9 ASCR11 | 1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 | 1 1 1 1 1 1 | | DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 | 28480 28480 28480 28480 28480 | 1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 |
| ASCR12 ASCR13 ASCR14 ASCR15 ASCR16 | 1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 | 1 1 1 1 1 1 1 1 | | DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 | 28480 28480 28480 28480 28480 | 1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 |
| ASCR17 ASCR18 ASCR19 ASCR21 ASCR22 | 1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 | 1 1 1 1 1 | | DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 | 28480 28480 28480 28480 28480 | 1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 |
| ASCR23 ASCR24 ASCR25 ASCR26 ASCR27 | 1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 | 1 1 1 1 1 1 1 | | DIODE-BWITCHING 30V 50MA 2NS DO-35 DIODE-BWITCHING 30V 50MA 2NS DO-35 DIODE-BWITCHING 30V 50MA 2NS DO-35 DIODE-BWITCHING 30V 50MA 2NS DO-35 DIODE-BWITCHING 30V 50MA 2NS DO-35 | 28480 28480 28480 28480 | 1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 |

Table 6-3. Replaceable Parts (Cont'd).

| Reference Designation | HP Part Number | C D | Qty | Description | Mfr Code | Mfr Part Number |
|---|---|---|------------------|---|---|--|
| A5CR28 A5CR29 A5CR31 A5CR32 | 1901-0040 1901-0040 1901-0040 1901-0040 | 1 1 1 1 | | DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 | 28480 28480 28480 28480 | 1901-0040 1901-0040 1901-0040 1901-0040 |
| A5L1 A5L2 A5L3 A5L4 A5L5 | 9100-3276 9100-3276 9100-3276 9100-3276 9100-3276 | 1 1 1 1 1 | 5 | INDUCTORRF-CH-MLD 10MH 2% .62D Q=100 | 28480 28480 28480 28480 28480 | 9100-3276 9100-3276 9100-3276 9100-3276 9100-3276 |
| A5L6 A5L7 | 9140-0137 9140-0137 | 1 | | INDUCTORRF-CH-MLD 1MH 5½ .2Dx.45LG Q=60 Inductorrf-ch-MLD 1MH 5% .2DX.45LG Q=60 | 28480 28480 | 9140=0137 9140=0137 |
| A501 A502 A503 A504 A505 | 1855-0081 1853-0010 1854-0071 1854-0071 1855-0081 | 1 2 7 7 1 | | TRANSISTOR J=FET N=CHAN D=MODE SI TRANSISTOR PNP SI TO=18 PD=360MM TRANSISTOR NPN SI PD=300MM FT=200MMZ TRANSISTOR NPN SI PD=300MM FT=200MMZ TRANSISTOR J=FET N=CHAN D=MODE SI | 01295 28480 28480 28480 01295 | 2N5245 1853-0010 1854-0071 1854-0071 2N5245 |
| A508 A507 A508 A509 A5011 | 1853-0010 1854-0071 1854-0071 1855-0081 1853-0010 | 2 7 7 1 2 | | TRANSISTOR PNP SI TO-18 PD=360MW TRANSISTOR NPN SI PD=300MW FT=200MMZ TRANSISTOR NPN SI PD=300MW FT=200MMZ TRANSISTOR J=FET N=CMAN D=MODE SI TRANSISTOR PNP SI TO-18 PD=360MW | 28480 28480 28480 01295 28480 | 1853-0010 1854-0071 1854-0071 285245 1853-0010 |
| A5012 A5013 A5014 A5015 A5016 | 1854-0071 1854-0071 1855-0081 1853-0010 1854-0071 | 7 7 1 2 7 | | TRANSISTOR NPN SI PD=300MM FT=200MHZ TRANSISTOR NPN SI PD=300MM FT=200MHZ TRANSISTOR J=FET N=CMAN D=MODE SI TRANSISTOR PNP SI TO=18 PD=360MM TRANSISTOR NPN SI PD=300MM FT=200MHZ | 28480 28480 01295 28480 28480 | 1854-0071 1854-0071 285245 1853-0010 1854-0071 |
| A5017 A5018 A5019 A5021 A5022 | 1854-0071 1854-0226 1853-0010 1854-0071 1854-0071 | 7 4 2 7 7 | 4 | TRANSISTOR NPN SI PD=300MW FT=200MMZ TRANSISTOR NPN 2N4384 SI TO=18 PD=500MW TRANSISTOR PNP SI TO=18 PD=360MW TRANSISTOR NPN SI PD=300MW FT=200MMZ TRANSISTOR NPN SI PD=300MW FT=200MMZ | 28480 13606 28480 28480 28480 | 1854-0071 2N4384 1853-0010 1854-0071 1854-0071 |
| 45R 45R1+ 45R2 45R3 45R4 | 0837-0086 0698-999P 0698-4399 0696-4517 0698-4486 | 7 9 7 1 3 | 1 5 5 | THERMISTOR DISC 200-OHM TC=-4.4%/C-DEG RESISTOR-PAD VALUE RESISTOR 88.7 1% 125% F TC=0+-100 RESISTOR 127K 1% 125% F TC=0+-100 RESISTOR 24.9K 1% 125% F TC=0+-100 | 28480 28480 24546 24546 24546 | 0837-0086 0698-999P C4-1/8-T0-88R7-F C4-1/8-T0-1273-F C4-1/8-T0-2492-F |
| 4585 4586 4587 4588 4589 | 0698-3382 0757-0283 0698-4481 0684-1041 0757-0460 | 6 8 1 1 | 5 13 5 | RESISTOR 5,49k 1% .125W F TC#0+-100 RESISTOR 2K 1% .125W F TC#0+-100 RESISTOR 10.5K 1% .125W F TC#0+-100 RESISTOR 100K 10% .25W FC TC#-400/+800 RESISTOR 61.9K 1% .125W F TC#0+-100 | 24546 24546 24546 01121 24546 | C4-1/8-T0-5491-F C4-1/8-T0-2001-F C4-1/8-T0-1652-F CB1041 C4-1/8-T0-6192-F |
| A5R10 A5R11 A5R12 A5R13 A5R14 | 0684-1531 0757-0445 0698-4441 0698-3495 0757-0403 | 5 0 5 0 5 0 5 0 5 0 5 0 5 0 5 0 5 0 5 0 | 5 7 6 5 | RESISTOR 15K 10X .25W FC TC==400/+800 RESISTOR 13K 1X .125W F TC=0+=100 RESISTOR 3,74K 1X .125W F TC=0+=100 RESISTOR 866 1X .125W F TC=0+=100 RESISTOR 121 1X .125W F TC=0+=100 | 01121 24546 24546 24546 24546 | C81531 C4-1/8-T0-1302-F C4-1/8-T0-3741-F C4-1/8-T0-868-F C4-1/8-T0-1218-F |
| A5R15 A5R16 A5R17 A5R18 A5R19 | 0698-3516 0698-4462 0684-2731 0664-2731 0684-1531 | 8 5 8 8 4 | 5 11 | RESISTOR 6.34K 1% .125W F TC=0+=100 RESISTOR 768 1% .125W F TC=0+=100 RESISTOR 27K 10% .25W FC TC==400/+800 RESISTOR 27K 10% .25W FC TC==400/+800 RESISTOR 15K 10% .25W FC TC==400/+800 | 24546 24546 01121 01121 | C4-1/8-T0-6341-F C4-1/8-T0-768R-F C82731 C82731 C81531 |
| A5R21 A5R22 A5R23 A5R24 A5R25 | 0684-1531 0684-1041 0683-1025 0683-1025 0683-1025 | 4 1 9 9 | 8 | RESISTOR 15K 10X .25W FC TC==400/+800 RESISTOR 100K 10X .25W FC TC==400/+800 RESISTOR 1K 5X .25W FC TC==400/+600 RESISTOR 1K 5X .25W FC TC==400/+600 RESISTOR 1K 5X .25W FC TC==400/+600 | 01121 01121 01121 01121 | C81531 C81041 C81025 C81025 C81025 |
| A5R27 A5R28 A5R29 A5R31 A5R32 | 0698-4399 0698-4517 0698-4486 0698-3382 0757-0283 | 7 1 3 6 | | RESISTOR 88.7 1% .125w F TC=0+-100 RESISTOR 127K 1% .125w F TC=0+-100 RESISTOR 24.9K 1% .125w F TC=0+-100 RESISTOR 5.49K 1% .125w F TC=0+-100 RESISTOR 2K 1% .125w F TC=0+-100 | 24546 24546 24546 24546 | C4=1/8-T0=88R7=F C4=1/8-T0=1273=F C4=1/8-T0=2492=F C4=1/8-T0=5491=F C4=1/8-T0=2001=F |
| A5R33 A5R34 A5R35 A5R36 A5R37 | 0698-4481 0684-1041 0757-0460 0757-0445 0698-4441 | 8 1 2 0 | | RESISTOR 16.5K 1% .125W F TC=0+=100 RESISTOR 100K 10% .25W FC TC==400/+800 RESISTOR 61.9K 1% .125W F TC=0+=100 RESISTOR 13K 1% .125W F TC=0+=100 RESISTOR 3.74K 1% .125W F TC=0+=100 | 24546 24546 24546 24546 | C4-1/8-T0-1052-F C81041 C4-1/8-T0-6192-F C4-1/8-T0-1302-F C4-1/8-T0-3741-F |
| A5R38 A5R39 A5R41 A5R42 A5R43 | 0698-3495 0757-0403 0698-3516 0698-4462 0684-2731 | 22858 | | RESISTOR 866 1% ,125W F TC=0+-100 RESISTOR 121 1% ,125W F TC=0+-100 RESISTOR 0,34K 1% ,125W F TC=0+-100 RESISTOR 766 1% ,125W F TC=0+-100 RESISTOR 27K 10% ,25W FC TC==400/+800 | 24546 24546 24546 24546 01121 | C4-1/8-T0-866R-P C4-1/8-T0-121R-F C4-1/8-T0-6341-F C4-1/8-T0-768R-P C82731 |
| A5R44 A5R45 A5R46 A5R47 A5R49 | 0684-2731 0684-1531 0684-1531 0684-1041 0696-4399 | 6 4 1 7 | | RESISTOR 27K 10% .25W FC TC=-400/+800 RESISTOR 15K 10% .25W FC TC=-400/+800 RESISTOR 15K 10% .25W FC TC=-400/+800 RESISTOR 100K 10% .25W FC TC=-400/+800 RESISTOR 88.7 1% .125W F TC=0+-100 | 01121 01121 01121 01121 24546 | C82731 C81531 C81531 C81041 C4-1/8-T0-68R7-F |
| | | | | | | |

Table 6-3. Replaceable Parts (Cont'd).

| | Table 0-3. Replaceable Parts (Cont d). | | | | | | | | |
|--|---|-----------------------|-----|---|---|--|--|--|--|
| Reference Designation | HP Part Number | C D | Qty | Description | Mfr Code | Mfr Part Number | | | |
| A5R51 A5R52 A5R53 A5R54 A5R55 | 0698-4517 0698-4486 0698-3382 0757-0283 0698-4481 | 1 3 6 6 8 | | RESISTOR 127K 1% .125W F TC=0+=100 RESISTOR 24.9K 1% .125W F TC=0+=100 RESISTOR 5.49K 1% .125W F TC=0+=100 RESISTOR 2K 1% .125W F TC=0+=100 RESISTOR 16.5K 1% .125W F TC=0+=100 | 24546 24546 24546 24546 24546 | C4-1/8-T0-1273-F C4-1/8-T0-2492-F C4-1/8-T0-2491-F C4-1/8-T0-2001-F C4-1/8-T0-1652-F | | | |
| ASR56 ASR57 ASR58 ASR59 ASR61 | 0684-1041 0757-0460 0757-0445 0698-4441 0698-3495 | 2 0 2 | | RESISTOR 100K 10% .25W FC TC==400/+800 RESISTOR 15.9K 1% .125W F TC=0+=100 RESISTOR 15K 1% .125W F TC=0+=100 RESISTOR 3.74K 1% .125W F TC=0+=100 RESISTOR 866 1% .125W F TC=0+=100 | 01121 24546 24546 24546 24546 | C81041 C4-1/8-T0-6192-F C4-1/8-T0-1302-F C4-1/8-T0-3741-F C4-1/8-T0-866R-F | | | |
| ASR62 ASR63 ASR64 ASR65 ASR66 | 0757-0403 0698-3516 0698-4462 0684-2731 0684-2731 | 2 8 5 8 8 | | RESISTOR 121 1% .125W F TC=0+-100 RESISTOR 6.34K 1% .125W F TC=0+-100 RESISTOR 768 1% .125W F TC=0+-100 RESISTOR 27K 10% .25W FC TC=-400/+800 RESISTOR 27K 10% .25W FC TC=-400/+800 | 24546 24546 24546 01121 01121 | C4-1/8-T0-121R-F C4-1/8-T0-6341-F C4-1/8-T0-768R-F C82731 | | | |
| A5R67 A5R68 A5R69 A5R72 A5R73 | 0684-1531 0684-1531 0684-1041 0698-4399 0698-4517 | 4 4 1 7 1 | | RESISTOR 15K 10% .25W FC TC=-400/+800 RESISTOR 15K 10% .25W FC TC=-400/+800 RESISTOR 100K 10% .25W FC TC=-400/+800 RESISTOR 88.7 1% .125W F TC=0+-100 RESISTOR 127K 1% .125W F TC=0+-100 | 01121 01121 01121 24546 24546 | C81531 CB1531 CB1041 C4-1/8-T0-88R7-F C4-1/8-T0-1273-F | | | |
| A5R74 A5R75 A5R76 A5R77 A5R78 | 0698-4486 0698-3382 0757-0283 0698-4481 0684-1041 | 3 6 6 8 1 | | RESISTOR 24.9K 1% .125W F TC=0+=100 RESISTOR 5.49K 1% .125W F TC=0+=100 RESISTOR 2K 1% .125W F TC=0+=100 RESISTOR 16.5K 1% .125W F TC=0+=100 RESISTOR 16.5K 1% .125W F TC=0+=100 RESISTOR 100K 10% .25W FC TC=-400/+800 | 24546 24546 24546 24546 01121 | C4-1/8-T0-2492=F C4-1/8-T0-5491=F C4-1/8-T0-2001=F C4-1/8-T0-1652=F CB1041 | | | |
| A5R79 A5R81 A5R82 A5R83 A5R84 | 0757-0460 0757-0445 0698-4441 0698-3495 0757-0403 | 1 2 0 2 2 | | RESISTOR 61.9K 1% .125W F TC=0+-100 RESISTOR 13K 1% .125W F TC=0+-100 RESISTOR 3.74K 1% .125W F TC=0+-100 RESISTOR 866 1% .125W F TC=0+-100 RESISTOR 121 1% .125W F TC=0+-100 | 24546 24546 24546 24546 24546 | C4-1/8-T0-6192-F C4-1/8-T0-1302-F C4-1/8-T0-3741-F C4-1/8-T0-868R-F C4-1/8-T0-1218-F | | | |
| ASR85 ASR86 ASR87 ASR88 ASR89 | 0698-3516 0698-4462 0684-2731 0684-2731 0684-1531 | 8 5 8 8 4 | | RESISTOR 6,34K 1% .125W F TC=0+-100 RESISTOR 768 1% .125W F TC=0+-100 RESISTOR 27K 10% .25W FC TC=-400/+800 RESISTOR 27K 10% .25W FC TC=-400/+800 RESISTOR 15K 10% .25W FC TC=-400/+800 | 24546 24546 01121 01121 01121 | C4-1/8-T0-6341-F C4-1/8-T0-768R-F C82731 C82731 C81531 | | | |
| ASR91 ASR92 ASR94 ASR95 ASR96 | 0684-1531 0684-1041 0698-4399 0698-4517 0757-0401 | 4 1 7 1 0 | | RESISTOR 15k 10% .25m FC TC=-400/+800 RESISTOR 100K 10% .25m FC TC=-400/+800 RESISTOR 88.7 1% .125m F TC=0+=100 RESISTOR 127k 1% .125m F TC=0+=100 RESISTOR 100 1% .125m F TC=0+=100 | 01121 01121 24546 24546 24546 | C81531 C81041 C4-1/8-T0-88R7-F C4-1/8-T0-1273-F C4-1/8-T0-101-F | | | |
| ASR97 ASR98 ASR99 ASR101 ASR102 | 0698-4486 0698-3223 0757-0283 0698-3155 0684-1041 | 3 4 6 1 | 1 | RESISTOR 24,9K 1% .125W F TC=0+-100 RESISTOR 1,24K 1% .125W F TC=0+-100 RESISTOR 2K 1% .125W F TC=0+-100 RESISTOR 4.64M 1% .125W F TC=0+-100 RESISTOR 100K 10% .25W FC TC=-400/+800 | 24546 24546 24546 24546 01121 | C4-1/8-T0-2492-F C4-1/8-T0-1241-F C4-1/8-T0-2001-F C4-1/8-T0-4641-F C81041 | | | |
| A5R103 A5R104 A5R106 A5R107 A5R108 | 0757-0460 0757-0445 0698-3495 0757-0403 0698-3516 | 2 2 2 8 | | RESISTOR 61.9K 1% .125W F TC=0+-100 RESISTOR 13K 1% .125W F TC=0+-100 RESISTOR 866 1% .125W F TC=0+-100 RESISTOR 121 1% .125W F TC=0+-100 RESISTOR 6.34K 1% .125W F TC=0+-100 | 24546 24546 24546 24546 24546 | C4-1/8-T0-6192-F C4-1/8-T0-1302-F C4-1/8-T0-866R-F C4-1/8-T0-121Ř-F C4-1/8-T0-654Î-F | | | |
| A5R109 A5R111 A5R112 A5R113 A5R114 | 0698-4462 0684-2731 0684-2731 0684-2731 0684-1531 | 5 8 8 4 | | RESISTOR 768 1% .125W F TC=0+=100 RESISTOR 27K 10% .25W FC TC==400/+800 RESISTOR 27K 10% .25W FC TC==400/+800 RESISTOR 27K 10% .25W FC TC==400/+800 RESISTOR 15K 10% .25W FC TC==400/+800 | 24546 01121 01121 01121 01121 | C4=1/8=T0=768R=F C82731 C82731 C82731 C81531 | | | |
| ASR115 ASR116 ASR117 ASR118 ASR119 | 0684-1531 0757-0442 0684-1041 0757-0394 0757-0394 | 4 9 1 0 0 | | RESISTOR 15K 10x .25w FC TC==400/+800 RESISTOR 10K 1x .125w F TC=0+=100 RESISTOR 10K 10x .25w FC TC==400/+800 RESISTOR 51.1 1x .125w F TC=0+=100 RESISTOR 51.1 1x .125w F TC=0+=100 | 01121 24546 01121 24546 24546 | C81531 C4-1/8-T0-1002-F C81041 C4-1/8-T0-51R1-F C4-1/8-T0-51R1-F | | | |
| A5R405 | 0698-4441 | 0 | | RESISTOR 3.74K 1% .125W F TC=0+-100 | 24546 | C4-1/8-T0-3741-F | | | |
| A5T1 A5T2 A5T3 A5T4 A5T5 | 9100-3262 9100-3262 9100-3262 9100-3262 | 5 5 5 5 5 | 5 | TRANSFORMER Transformer Transformer Transformer Transformer | 28480 28480 28480 28480 28480 | 9100-3262 9100-3262 9100-3262 9100-3262 9100-3262 | | | |
| A5Y1 A5Y2 A5Y3 A5Y4 A5Y5 | 0410=0480 0410=0480 0410=0480 0410=0480 0410=0480 | 1 1 1 1 1 | 5 | CRYSTAL SET:NOT FIELD REPLACEABLE (SEE P. 7-25) | 28480 28480 28480 28480 28480 | 0410-0480 0410-0480 0410-0480 0410-0480 | | | |
| A6 | 03580-66526 | 6 | 1 | BOARD ASSEMBLY-LOW VOLTAGE POWER SUPPLY | 28480 | 03580=66526 | | | |
| | | | | | | | | | |

Table 6-3. Replaceable Parts (Cont'd).

| lable 6-3. Keplaceable Parts (Cont'd). | | | | | | | | | |
|--|---|---|--------|--|--|---|--|--|--|
| Reference Designation | HP Part Number | C D | Qty | Description | Mfr Code | Mfr Part Number | | | |
| A6C1 A6C2 A6C3 A6C4 A6C5 | 0180-1746 0180-0291 0180-2960 0180-2960 0180-0291 | 5 3 7 7 3 | 2 | CAPACITOR-FXD 15UF+-10% 20VDC TA CAPACITOR-FXD 1UF+-10% 35VDC TA CAPACITOR-FXD 1000UF+75-10% 35VDC AL CAPACITOR-FXD 1000UF+75-10% 35VDC AL CAPACITOR-FXD 1UF+-10% 35VDC TA | 56289 56289 28480 28480 56289 | 1500156x902082 1500105x9035A2 0180-2960 0180-2960 1500105x9035A2 | | | |
| A6C6 A6C7 A6C8 A6C9 A6C11 | 0180+1746 0180-0224 0180-0291 0180-0291 0140-0206 | 5 2 3 6 | 1 2 | CAPACITOR-FXD 15UF+-10% 20VDC TA CAPACITOR-FXD 10UF+75-10% 16VDC AL CAPACITOR-FXD 1UF+-10% 35VDC TA CAPACITOR-FXD 1UF+-10% 35VDC TA CAPACITOR-FXD 270PF +-5% 500VDC MICA | 56289 56289 56289 56289 72136 | 150D156X902082 30D106G0168A2 150D105X9035A2 150D105X9035A2 DM15F271J0500WV1CR | | | |
| A6C12 A6C13 A6C14 A6C15 A6C16 A6C16 | 0150-0022 0180-0291 0140-0217 0150-0022 0160-0161 | 539544 | 2 | CAPACITOR-FXD 3.3FF +=10% 500VDC TI DIOX CAPACITOR-FXD 1UF+=10% 35VDC TA CAPACITOR-FXD 140FF +=2% 500VDC MICA CAPACITOR-FXD 3.3FF +=10% 500VDC TI DIOX CAPACITOR-FXD .01UF +=10% 200VDC POLYE CAPACITOR-FXD .01UF +=10% 200VDC POLYE | 28480 56289 72136 28480 28480 28480 | 0150-0022 1500105X9035A2 DM15F141G0300WV1CR 0150-0022 0160-0161 | | | |
| A6C18 A6C19 A6C21 A6C22 A6C24 | 0180-0061 0180-1746 0180-0061 0180-1746 0180-0291 | 55553 | | CAPACITOR-FXD 100UF+75=10% 16VDC AL CAPACITOR-FXD 15UF+=10% 20VDC TA CAPACITOR-FXD 100UF+75=10% 16VDC AL CAPACITOR-FXD 15UF+=10% 20VDC TA CAPACITOR-FXD 1UF+=10% 35VDC TA | 56289 56289 56289 56289 56289 | 30D107G016DC2 150D156X9020B2 30D107G016DC2 150D156X9020B2 150D105X9035A2 | | | |
| A6C30 | 0180-0291 | 3 | | CAPACITOR=FXD 1UF+=10% 35VDC TA | 56289 | 150D105X9035A2 | | | |
| A6CR1 A6CR2 A6CR3 A6CR4 A6CR7 | 1902-3149 1901-0040 1901-0040 1901-0040 1901-0704 | 9 1 1 1 4 | 1 | DIODE-ZNR 9.09V 5% DO=35 PD=.4W DIODE-SWITCHING 30V 50MA 2NS DO=35 DIODE-SWITCHING 30V 50MA 2NS DO=35 DIODE-SWITCHING 30V 50MA 2NS DO=35 DIODE-PWR RECT 1N4002 100V 1A DO=41 | 28480 28480 28480 28480 01295 | 1902-3149 1901-0040 1901-0040 1901-0040 1N4002 | | | |
| A6CR8 A6CR9 A6CR11 A6CR15 A6CR16 | 1901-0704 1901-0704 1901-0704 1902-0025 1902-0777 | 4 4 4 3 | 2 1 | DIODE-PWR RECT 1N4002 100V 1A D0-41 DIODE-PWR RECT 1N4002 100V 1A D0-41 DIODE-PWR RECT 1N4002 100V 1A D0-41 DIODE-ZNR 10V 5X D0-35 PDB-4W TC++.06X DIODE-ZNR 1N825 6,2V 5X D0-7 PDB-4W | 01295 01295 01295 28480 04713 | 1 N 4 0 0 2 1 N 4 0 0 2 1 N 4 0 0 2 1 N 4 0 2 5 1 N 8 2 5 | | | |
| A6CR17 A6CR18 A6CR19 A6CR20 A6CR21 | 1901-0040 1901-0040 1901-0040 1902-3190 1902-0025 | 1 1 0 4 | 2 | DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-ZNR 13V 5% DO-35 PDB-4W TCB+.06% DIODE-ZNR 10V 5% DO-35 PDB-4W TCB+.06% | 28480 28480 28480 28480 28480 | 1901-0040 1901-0040 1901-0040 1902-3190 1902-0025 | | | |
| A6CR22 A6CR23 A6CR24 A6CR26 A6CR27 | 1901-0040 1901-0040 1902-3190 1901-0704 1901-0704 | 1 1 0 4 4 | | DIODE-SWITCHING 30V SOMA 2NS DO-35 DIODE-SWITCHING 30V SOMA 2NS DO-35 DIODE-ZNR 13V 5% DO-35 PDB.4W TC2+.06% DIODE-PWR RECT 1N4002 100V 1A DO-41 DIODE-PWR RECT 1N4002 100V 1A DO-41 | 28480 28480 28480 01295 | 1901-0040 1901-0040 1902-3190 1N4002 1N4002 | | | |
| A6CR28 A6CR29 | 1901-0704 1901-0704 | 4 4 | | DIODE-PWR RECT 1N4002 100V 1A DO-41 DIODE-PWR RECT 1N4002 100V 1A DO-41 | 01295 | 1N4002 1N4002 | | | |
| 46F1 A6F2 | 2110-0490 2110-0297 | 94 | 1 | FUSE .375A 125V .281X.093 FUSE .5A 125V .281X.093 | 75915 28480 | 275.375 2110-0297 | | | |
| 46K1 | 0490+1208 0490+1208 | 1 1 | 5 | RELAY=REED 1A 500MA 200VDC 10VDC=COIL RELAY=REED 1A 500MA 200VDC 10VDC=COIL | 28480 28480 | 0490=1208 0490=1208 | | | |
| A6Q1 A6Q2 A6Q3 A6Q4 A6Q5 | 1853-0010 1854-0404 1853-0052 1854-0404 | 50505 | 7 2 | TRANSISTOR PNP SI TO=18 PD=360MW TRANSISTOR NPN SI TO=18 PD=360MW TRANSISTOR NPN SI TO=18 PD=360MW TRANSISTOR NPN SI TO=16 PD=360MW TRANSISTOR NPN SI TO=18 PD=360MW TRANSISTOR PNP SI TO=18 PD=360MW | 28480 28480 04713 28480 28480 | 1853-0010 1854-0404 2N3740 1854-0404 1853-0010 | | | |
| A6Q6 A6Q7 A6Q8 A6Q9 A6Q11 | 1854-0404 1853-0010 1853-0010 1854-0072 1853-0010 | 8 2 2 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 0 | 2 | TRANSISTOR NPN SI TO=18 PD=360MW TRANSISTOR PNP SI TO=18 PD=360MW TRANSISTOR PNP SI TO=18 PD=360MW TRANSISTOR NPN 2N3054 SI TO=66 PD=25W TRANSISTOR PNP SI TO=18 PD=360MW | 28480 28480 28480 01928 28480 | 1854-0404 1853-0010 1853-0010 283054 1853-0010 | | | |
| A6912 A6914 A6915 A6916 A6917 | 1854-0072 1853-0010 1854-0404 | 9 2 9 2 5 5 | | TRANSISTOR PNP SI TO=18 PD=360MW TRANSISTOR PNP SI TO=18 PD=360MW TRANSISTOR PNP SI TO=18 TO=66 PD=25W TRANSISTOR PNP SI TO=18 PD=360MW TRANSISTOR NPN SI TO=18 PD=360MW | 28480 28480 01928 28480 28480 | 1853-0010 1853-0010 2N3054 1853-0010 1854-0404 | | | |
| A6Q18 A6Q19 A6Q20 A6Q133 | 1854-0404 1853-0052 1854-0404 1854-0404 | 0 0 0 0 | | TRANSISTOR NPN 81 TO-18 PD=360MW TRANSISTOR PNP 2N3740 \$1 TO-66 PD=25W TRANSISTOR NPN 81 TO-18 PD=360MW TRANSISTOR NPN 81 TO-18 PD=360MW | 28480 04713 28480 28480 | 1854-0404 2N3740 1854-0404 1854-0404 | | | |
| A6R1 A6R2 A6R3 A6R4 A6R5 | 0757-0438 0698-4123 | 8 8 3 5 7 | 2 4 1 | RESISTOR 3,32K 1% ,125M F TC=0+-100 RESISTOR 16,9K 1% ,125M F TC=0+-100 RESISTOR 5,11K 1% ,125M F TC=0+-100 RESISTOR 499 1% ,125M F TC=0+-100 RESISTOR 287 1% ,5M F TC=0+-100 | 24546 24546 24546 24546 28480 | C4-1/8-T0-3321-F C4-1/8-T0-1692-F C4-1/8-T0-5111-F C4-1/8-T0-499R-F 0757-1092 | | | |
| | | | | | | | | | |

Table 6-3. Replaceable Parts (Cont'd).

| Reference Designation | HP Part Number | C | Qty | Description | Mfr | |
|---|---|-----------------------|------------------|---|---|--|
| | | | , | Description | Code | Mfr Part Number |
| A6R6 A6R7 A6R8 A6R9 A6R11 | 0757-0282 0757-0388 0698-3559 0757-0388 0698-0001 | 5 2 9 2 0 | 2 1 | RESISTOR 221 1% .125W F TC=0+-100 RESISTOR 30.1 1% .125W F TC=0+-100 RESISTOR 3.9 10% .5W CC TC=0+412 RESISTOR 30.1 1% .125W F TC=0+-100 RESISTOR 4.7 5% .5W CC TC=0+412 | 24546 24546 01121 24546 01121 | C4-1/8-T0-221R-F C4-1/8-T0-30R1-F EB39G1 C4-1/8-T0-30R1-F EB47G5 |
| A6R12 A6R13 A6R14 A6R15 A6R16 | 0757-0433 0698-4308 0757-0438 0698-4123 0757-0465 | 8 8 3 5 6 | | RESISTOR 3.32K 1% .125W F TC=0+-100 RESISTOR 16.9K 1% .125W F TC=0+-100 RESISTOR 5.11K 1% .125W F TC=0+-100 RESISTOR 499 1% .125W F TC=0+-100 RESISTOR 100K 1% .125W F TC=0+-100 | 24546 24546 24546 | C4-1/8-T0-3321-F C4-1/8-T0-1892-F C4-1/8-T0-5111-F C4-1/8-T0-1998-F C4-1/8-T0-1003-F |
| A6R17 A6R18 A6R19 A6R21 A6R22 | 0757-0442 0757-0442 0686-3315 0757-0469 0757-0458 | 9 9 0 0 7 | 5 | RESISTOR 10k 1% .125W F TC=0+=100 RESISTOR 10k 1% .125W F TC=0+=100 RESISTOR 330 5% .5W CC TC=0+529 RESISTOR 150k 1% .125W F TC=0+=100 RESISTOR 51.1K 1% .125W F TC=0+=100 | 24546 24546 01121 24546 24546 | C4-1/8-T0-1002-F C4-1/8-T0-1002-F E83315 C4-1/8-T0-1503-F C4-1/8-T0-5112-F |
| 46R23 46R24 46R25 46R26 46R27 | 0686-3315 0757-0465 0698-3488 0698-4435 0698-3499 | 0 6 3 2 0 | | RESISTOR 330 5% .5W CC TC=0+529 RESISTOR 100K 1% .125W F TC=0+=100 RESISTOR 442 1% .125W F TC=0+=100 RESISTOR 40.2K 1% .125W F TC=0+=100 RESISTOR 40.2K 1% .125W F TC=0+=100 | 01121 24546 24546 24546 24546 | E83315 C4-1/8-T0-1003=F C4-1/8-T0-422R=F C4-1/8-T0-2491=F C4-1/8-T0-4022=F |
| A6R28 A6R29 A6R31 A6R32 A6R33 | 0757-0283 0698-3558 0757-0161 0811-3069 0698-4123 | 68 9 8 5 | 6 | RESISTOR 2K 1% ,125W F TC=0+-100 RESISTOR 4,02K 1% ,125W F TC=0+-100 RESISTOR 604 1% ,125W F TC=0+-100 RESISTOR 15% ,5W PW TC=0+-150 RESISTOR 499 1% ,125W F TC=0+-100 | 24546 24546 24546 75042 24546 | C4-1/8-T0-2001-F C4-1/8-T0-4021-F C4-1/8-T0-604R-F B#20-1-1R0-J C4-1/8-T0-499R-F |
| A6R34 A6R35 A6R36 A6R37 A6R38 | 0757-0283 0698-3245 0757-0442 0698-3245 0698-5323 | 60909 | 1 | RESISTOR 2K 1% ,125W F TC=0+-100 RESISTOR 20,5K 1% ,125W F TC=0+-100 RESISTOR 10K 1% ,125W F TC=0+-100 RESISTOR 20,5% 1% ,125W F TC=0+-100 RESISTOR 4K ,5% ,125W F TC=0+-50 | 24546 24546 24546 28480 | C4-1/8-T0-2001-F C4-1/8-T0-2052-F C4-1/8-T0-1002-F C4-1/8-T0-2052-F 0498-5323 |
| A6R39 A6R41 * A6R42 A6R43 A6R44 | 0698-6846 0698-999P 0698-3279 0698-4509 0757-0283 | 3 9 0 1 6 | 1 | RESISTOR 5,42K .5% .125W F TC=0+-50 RESISTOR-FX0 PAD VALUE RESISTOR 4,99K 1% .125W F TC=0+-100 RESISTOR 80.6K 1% .125W F TC=0+-100 RESISTOR 2K 1% .125W F TC=0+-100 | 24546 24546 24546 24546 | NC55-1/8-T2-5421-D 0698-999P C4-1/8-T0-4991-F C4-1/8-T0-8062-F C4-1/8-T0-2001-F |
| A6R45 A6R46 A6R47 A6R48 A6R49 | 0698-3558 0757-0161 0698-4123 0811-3069 0698+3245 | 8 9 5 8 0 | | RESISTOR 4,02K 1% ,125W F TC=0+-100 RESISTOR 604 1% ,125W F TC=0+-100 RESISTOR 499 1% ,125W F TC=0+-100 RESISTOR 1 5% 5W PW TC=0+-150 RESISTOR 20,5K 1% ,125W F TC=0+-100 | 24546 24546 24546 75042 24546 | C4-1/8-T0-4021-F C4-1/8-T0-604R-F C4-1/8-T0-499R-F Bn20-1-1R0-J C4-1/8-T0-2052-F |
| A6R51 A6R52 A6R53 | 0757-0283 0698-3193 0698-3193 | 6777 | 5 | RESISTOR 2K 1% .125W F TC=0+-100 RESISTOR 10K .25% .125W F TC=0+-50 RESISTOR 10K .25% .125W F TC=0+-50 | 24546 28480 28480 | C4=1/8=T0=2001=F 0698=3193 0698=3193 |
| A6U4 A6U3 A6U4 | 1820-0223 1820-0223 1826-0393 1826-0527 | 0 0 7 9 | 1 1 | IC OP AMP GP TO-99 IC OP AMP GP TO-99 IC V RGLTR TO-220 IC 337 V RGLTR TO-220 | 04713 04713 27014 27014 | MLM301AG MLM301AG LM317T LM337T |
| A7 A7C1 A7C2 A7C3 A7C4 A7C4 | 03580-69507 03580-69507 9180-0291 0160-2530 0160-2012 0160-0127 0160-0297 | 3 5 6 2 7 | 1 1 1 2 | BOARD ASSEMBLY=LOGIC REBUILT EXCHANGE ASSEMBLY CAPACITOR=FXD 1UF==10X 35VDC TA CAPACITOR=FXD 180PF +=2X 300VDC MICA CAPACITOR=FXD 330PF +=5X 500VDC MICA CAPACITOR=FXD 1:UF +=20X 25VDC CER CAPACITOR=FXD 1:20PF +=10X 200VDC PDLYE | 28480 28480 56289 28480 28480 28480 28480 | 03580-66507 03580-69507 1500105x9035A2 0160-2530 0160-2012 0160-0127 |
| A7C b A7C 7 A7C 8 A7C 9 | 0180-1746 0160-0127 0180-0229 0180-1746 | 5 2 7 5 | 3 | CAPACITOR=FXD 15UF+=10% 20VDC TA CAPACITOR=FXD 1UF +=20% 25VOC CER CAPACITOR=FXD 33UF+=10% 10VDC TA CAPACITOR=FXD 15UF+=10% 20VDC TA | 56289 28480 56289 56289 | 150p156x9020g2 0160=0127 150D336x901082 1500156x9020g2 |
| A7CR1 A7CR2 | 1902-0551 1902-0551 | 1 1 | 2 | DIODE-ZNR 6.19V 5% DO-15 PD=1W TC=+.022% DIODE-ZNR 6.19V 5% DO-15 PD=1W TC=+.022% | 28480 28480 | 1902-0551 1902-0551 |
| 17L1 17L2 17L3 | 9100-0541 9140-0129 9100-0541 | 7 1 7 | | INDUCTORRF-CH-MLD 250UH 10% 250%,SLG INDUCTORRF-CH-MLD 250UH 5% ,1865C,SB5CG INDUCTORRF-CH-MLD 250UH 10% ,250%,SLG | 28480 28480 28480 | 9100-0541 9140-0129 9100-0541 |
| 1701 1702 1703 1704 1705 | 1854-0071 1853-0010 1854-0071 1853-0010 1854-0071 | 7 2 7 2 7 | | TRANSISTOR NPN SI PD=300MM FT=200MMZ TRANSISTOR PNP SI TO=18 PD=360MM TRANSISTOR NPN SI PD=300MM FT=200MMZ TRANSISTOR PNP SI TO=18 PD=360MM TRANSISTOR NPN SI PD=300MM FT=200MMZ | 28480 28480 28480 28480 28480 | 1854-0071 1853-0010 1854-0071 1853-0010 1854-0071 |
| 1706 1707 1708 1709 1701 1 | 1853-0010 1854-0071 1853-0010 1853-0010 1854-0071 | 2 2 2 7 | | TRANSISTOR PNP SI TO-16 PD=360MW TRANSISTOR NPN SI PD=300MW FT=200MMZ TRANSISTOR PNP SI TO-16 PD=360MW TRANSISTOR PNP SI TO-16 PD=360MW TRANSISTOR PNP SI TO-16 PD=300MW TRANSISTOR NPN SI PD=300MW FT=200MMZ | 28480 28480 28480 28480 28480 | 1853-0010 1854-0071 1853-0010 1853-0010 1854-0071 |

Table 6-3. Replaceable Parts (Cont'd).

| Table 0-3. Replaceable Parts (Cont d). | | | | | | | | | |
|--|---|---|-------------|---|---|--|--|--|--|
| Reference Designation | HP Part Number | C D | Qty | Description | Mfr Code | Mfr Part Number | | | |
| A7012 A7013 A7014 | 1854-0071 1853-0012 1854-0039 | 7 4 7 | 1 1 | TRANSISTOR NPN SI PD#300MW FT#200MMZ TRANSISTOR PNP 2N2904A SI TO=39 PD#600MW TRANSISTOR NPN 2N30538 SI TO=39 PD#1W | 28480 01295 01928 | 1854-0071 2N2904A 2N30538 | | | |
| A7R1 A7R2 A7R3 A7R4 A7R5 | 0684-1031 0684-3931 0757-0465 0698-0077 0698-3228 | 9 2 6 0 9 | 7 | RESISTOR 10K 10X .25M FC TC=-400/+700 RESISTOR 39K 10X .25M FC TC=-400/+800 RESISTOR 100K 1X .125M F TC=0+=100 RESISTOR 93.1K 1X .125M F TC=0+=100 RESISTOR 49.9K 1X .125M F TC=0+=100 | 01121 01121 24546 03888 28480 | C81031 C83931 C4-1/8-T0-1003-F PME55-1/8-T0-9312-F 0098-3228 | | | |
| A7R6 A7R7 A7R8 A7R9 A7R10 | 0698-0077 0698-3228 0698-5575 0698-3228 0698-1031 | 0 9 3 9 | 6 | RESISTOR 93.1K 1X .125W F TC=0+=100 RESISTOR 49.9K 1X .125W F TC=0+=100 RESISTOR 100K .5X .125W F TC=0+=100 RESISTOR 49.9K 1X .125W F TC=0+=100 RESISTOR 10K 10X .25W FC TC==400/+700 | 03868 28480 24546 28480 01121 | PME55=1/8=TO=9312=F 0698=3228 C4=1/6=TO=1003=D 0698=3228 C81031 | | | |
| A7R11 A7P12 A7R13 A7R14 A7R15 | 0698-5575 0698-5573 0698-3445 0698-6688 0698-5573 | 3 1 2 2 1 | 6 6 7 | RESISTOR 100K .5% .125W F TC=0+=100 RESISTOR 50K .5% .125W F TC=0+=100 RESISTOR 348 1% .125W F TC=0+=100 RESISTOR 99.8K .1% .05W PWW TC=±10 RESISTOR 99.8K .5% .125W F TC=0+=100 | 24546 24546 24546 28480 24546 | C4-1/8-T0-1003-D C4-1/8-T0-5002-D C4-1/8-T0-348R-F 0698-688 C4-1/8-T0-5002-D | | | |
| A7R16 A7R17 A7R18 A7R19 A7R21 | 0698-3445 0698-6688 0698-7973 0698-4158 0698-7973 | 00 00V | 8 5 | RESISTOR 348 1% 125W F TC=0+-100 RESISTOR 99.8K.1%.05W PWW TC=±10 RESISTOR 50K .05% 1.25W F TC=0+-25 RESISTOR 100K .1% .125W F TC=0+-25 RESISTOR 50K .05% .125W F TC=0+-25 | 24546 28480 19701 28480 19701 | C4-1/8-T0-346R-F 0698-6688 MF4C1/8-T9-5002-D 0698-4156 MF4C1/8-T9-5002-D | | | |
| A7R22 A7R23 A7R24 A7R25 A7R26 | 0698-4158 0684-3931 0684-3931 0684-3931 0684-3931 | 5 2 2 9 | | RESISTOR 100K .1% .125M F TC=0+=50 RESISTOR 39K 10% .25M FC TC==400/+800 | 28480 01121 01121 01121 01121 | 0698-4158 C83931 C83931 C83931 C83931 | | | |
| A7R27 A7R28 A7R29 A7R31 A7R32 | 0698-3268 0757-0442 0684-5621 0757-0280 0698-4469 | 7 9 1 3 2 | 4 | RESISTOR 11.5k 1% .125W F TC=0+-100 RESISTOR 10K 1% .125W F TC=0+-100 RESISTOR 5.6K 10% .25W FC TC=-400/+700 RESISTOR 1K 1% .125W F TC=0+-100 RESISTOR 1.15K 1% .125W F TC=0+-100 | 24546 24546 01121 24546 24546 | C4-1/8-T0-1152-F C4-1/8-T0-1002-F C85621 C4-1/8-T0-1001-F C4-1/8-T0-1151-F | | | |
| A7R33 A7R34 A7R35 A7R36 A7R37 | 0698-3268 0757-0442 0684-5621 0757-0280 0698-4469 | 7 9 1 3 2 | | RESISTOR 11.5K 1% .125W F TC=0+=100 RESISTOR 10K 1% .125W F TC=0+-100 RESISTOR 5.6K 10% .25W FC TC==400/+700 RESISTOR 1K 1% .125W F TC=0+=100 RESISTOR 1.15K 1% .125W F TC=0+=100 | 24546 24546 01121 24546 24546 | C4-1/8-T0-1152-F C4-1/8-T0-1002-F C85621 C4-1/8-T0-1001-F C4-1/8-T0-1151-F | | | |
| A7R38 A7R39 A7R41 A7R42 A7R43 | 0698-3268 0757-0442 0684-5621 0757-0280 0698-4469 | 7 9 1 3 2 | | RESISTOR 11.5K 1% .125W F TC=0+=100 RESISTOR 10K 1% .125W F TC=0++100 RESISTOR 5.6K 10% .25W F TC=-400/+700 RESISTOR 1h 1% .125W F TC=0+=100 RESISTOR 1.15K 1% .125W F TC=0+=100 | 24546 01121 24546 24546 | C4-1/8-T0-1152-F C4-1/8-T0-1002-F C85621 C4-1/8-T0-1001-F C4-1/8-T0-1151-F | | | |
| A7R44 A7R45 A7R46 A7R47 A7R48 | 0698-3268 0757-0442 0684-5621 0757-0280 0698-4469 | 7 9 1 3 2 | | RESISTOR 11.5% 1% .125W F TC=0+=100 RESISTOR 10% 1% .125W F TC=0+=100 RESISTOR 5.6% 10% .25W FC TC=-400/+700 RESISTOR 1% 1% .125W F TC=0+=100 RESISTOR 1.15% 1% .125W F TC=0+=100 | 24546 24546 01121 24546 24546 | C4=1/8=T0=1152=F C4=1/8=T0=1002=F C85621 C4=1/8=T0=1001=F C4=1/8=T0=1151=F | | | |
| A7R49 A7R51 A7R52 A7R53 A7R54 | 0757-0464 0698-3226 0757-0978 0698-3228 0757-0978 | 59696 | 1 4 | RESISTOR 90.9K 1X .125W F TC=0+-100 RESISTOR 49.9K 1X .125W F TC=0+-100 RESISTOR 95.3K 1X .125W F TC=0+-100 RESISTOR 49.9K 1X .125W F TC=0+-100 RESISTOR 95.3K 1X .125W F TC=0+-100 | 24546 26460 24546 26460 24546 | C4-1/8-T0-9092-F 0698-3228 C4-1/8-T0-9532-F 0698-3228 C4-1/8-T0-9532-F | | | |
| A7RSS A7RS6 A7RS7 A7RS8 A7RS8 A7RS9 | 0698-5573 0698-5575 | 9 3 1 3 1 | | RESISTOR 49,9K 1% .125W F TC=0+-100 RESISTOR 100K .5% .125W F TC=0+-100 RESISTOR 50K .5% .125W F TC=0+-100 RESISTOR 100K .5% .125W F TC=0+-100 RESISTOR 50K .5% .125W F TC=0+-100 | 28480 24546 24546 24546 | 0698-3228 C4-1/8-T0-1003-D C4-1/8-T0-5002-D C4-1/8-T0-5003-D C4-1/8-T0-5002-D | | | |
| A7R61 A7R62 A7R63 A7R64 A7R65 | 0698-6688 0698-7973 0698-3445 | 5 6 5 5 5 5 5 5 5 5 5 5 5 5 5 5 5 5 5 5 | - | RESISTOR 348 1% 125W F TC=0+=100 RESISTOR 99.8K.1%.05W PWW TC=±10 RESISTOR 50K .05% .125W F TC=0+=25 RESISTOR 348 1% .125W F TC=0+=100 RESISTOR 99.8K.1%.05W PWW TC=±10 | 24546 28480 19701 24546 28480 | C4=1/8=T0=348R=F 0698-6688 Mf4C1/8=T9=5002=D C4=1/8=T0=348R=F 0698-6688 | | | |
| A7R66 A7R67 A7R68 A7R69 A7R71 | 0698-4158 0698-7973 0698-7975 | 9 6 9 1 9 | 2 | RESISTOR 50K .05% .125W F TC=0+-25 RESISTOR 100K .1% .125W F TC=0+-50 RESISTOR 50K .05% .125W F TC=0+-25 RESISTOR 100K .05% .125W F TC=0+-25 RESISTOR 50K .05% .125W F TC=0+-25 | 19701 28480 19701 19701 | MF4C1/8-T9-5002-D 0898-4158 MF4C1/8-T9-5002-D MF4C1/8-T9-1003-D MF4C1/8-T9-5002-D | | | |
| A7R72 A7R73 A7R74 A7R75 A7R76 | 0811-1794 0698-6688 0757-0449 | 1 2 7 6 | 1 | RESISTOR 100K .05% .125W F TC=0+-25 RESISTOR 99.25K .1% .05W PWW TC=0+-10 RESISTOR 99.8K .1% .05W PWW TC=±10 RESISTOR 20K 1% .125W F TC=0+-100 RESISTOR 20K 1% .125W F TC=0+-100 | 19701 20940 28480 24546 24546 | MF4C1/8-T9-1003-D 140-1/40-D-99251-B 0698-6688 C4-1/8-T0-2002-F C4-1/8-T0-2002-F | | | |
| A7R77 A7R78 A7R79 A7R61 A7R62 | 0757-0449 0757-0449 0757-0449 | 00000 | | RESISTOR 20K 1% .125W F TC=0+-100 | 24546 24546 24546 24546 | C4-1/8-T0-2002-F C4-1/8-T0-2002-F C4-1/8-T0-2002-F C4-1/8-T0-2002-F C4-1/8-T0-2002-F | | | |

Table 6-3. Replaceable Parts (Cont'd).

| | Table 0-3. Replaceable Parts (Cont a). | | | | | | | | | |
|---|---|---|-----------------------|---|--|--|--|--|--|--|
| Reference Designation | HP Part Number | C D | Qty | Description | Mfr Code | Mfr Part Number | | | | |
| A7R83 A7R84 A7R85 A7R86 A7R87 | 0757-0449 0757-0449 0757-0449 0757-0449 0757-0449 | 00000 | | RESISTOR 20K 1% .125W F TC=0+-100 | 24546 24246 24246 24546 24546 | C4-1/8-T0-2002-F C4-1/8-T0-2002-F C4-1/8-T0-2002-F C4-1/8-T0-2002-F C4-1/8-T0-2002-F | | | | |
| 47888 47889 47891 47892 47893 | 0757-0449 0757-0449 0757-0449 0757-0449 0757-0449 | 0000 | | RESISTOR 20K 1% .125W F TC=0+=100 | 24546 24546 24546 24546 | C4-1/8-T0-2002-F C4-1/8-T0-2002-F C4-1/8-T0-2002-F C4-1/8-T0-2002-F C4-1/8-T0-2002-F | | | | |
| A7R94 A7R95 A7R96 A7R97 A7R98 | 0757-0449 0757-0449 0757-0449 0684-5621 0684-5621 | 6 6 1 1 | | RESISTOR 20K 1% .125W F TC=0+=100 RESISTOR 20K 1% .125W F TC=0+=100 RESISTOR 20K 1% .125W F TC=0+=100 RESISTOR 20K 10 % .25W FC TC==400/+700 RESISTOR 5.6K 10% .25W FC TC==400/+700 | 24546 24546 24546 01121 | C4-1/8-T0-2002-F C4-1/8-T0-2002-F C4-1/8-T0-2002-F C85621 | | | | |
| A7R99 A7R101 A7R102 A7R103 A7R104 | 0684-5621 0684-5621 0757-0438 0757-0438 0757-0442 | 1 1 3 3 | | RESISTOR 5.6K 10% .25W FC TC==400/+700 RESISTOR 5.6K 10% .25W FC TC==400/+700 RESISTOR 5.11K 1% .125W F TC=0+=100 RESISTOR 5.11K 1% .125W F TC=0+=100 RESISTOR 10K 1% .125W F TC=0+=100 | 01121 01121 24546 24546 24546 | C85621 C85621 C4-1/8-T0-5111-F C4-1/8-T0-5111-F C4-1/8-T0-1002-F | | | | |
| ATR105 ATR106 ATR107 ATR108 ATR109 | 0684-3931 0684-1031 0684-3931 0684-1031 0698-3160 | 9 29 29 29 | 1 | RESISTOR 39% 10% .25% FC TC==400/+800 RESISTOR 10% 10% .25% FC TC==400/+700 RESISTOR 39% 10% .25% FC TC==400/+800 RESISTOR 10% 10% .25% FC TC==400/+700 RESISTOR 31.6% 1% .125% F TC=0+-100 | 01121 01121 01121 01121 01121 24546 | C83931 C81031 C83931 C81031 C4-1/8-T0-3162-F | | | | |
| A7R111+ A7R112 A7R113 A7R114 A7R115 | 0698-4499 0698-3728 0757-0438 0698-3498 0757-0438 | 9 3 5 3 | 1 | RESISTOR 29.4K 1% .125W F TC=0+-100 RESISTOR 49.9K 1% .125W F TC=0+-100 RESISTOR 5.11K 1% .125W F TC=0+-100 RESISTOR 8.66K 1% .125W F TC=0+-100 RESISTOR 5.11K 1% .125W F TC=0+-100 | 24546 24546 24546 24546 | C4-1/8-T0-2942-F 0698-3228 C4-1/8-T0-5111-F C4-1/8-T0-866R-F C4-1/8-T0-5111-F | | | | |
| A7R116 A7R117 A7R116 A7R119 A7R121 | 0757-0978 0757-0465 0698-3228 0757-0978 0698-3228 | 00000 | | RESISTOR 95.3K 1% .125W F TC=0+=100 RESISTOR 100K 1% .125W F TC=0+=100 RESISTOR 49.9K 1% .125W F TC=0+=100 RESISTOW 95.3K 1% .125W F TC=0+=100 RESISTOP 49.9K 1% .125W F TC=0+=100 | 24546 24546 24546 24546 | C4-1/8-T0-9532-F C4-1/8-T0-1003-F 0698-3228 C4-1/8-T0-9532-F 0698-3228 | | | | |
| 47R122 47R123 47R124 47R125 47R120 | 0698-5575 0698-3228 0698-5575 0698-6688 | 3 3 1 2 | | RESISTOR 100K .5% .125W F TC=0+-100 RESISTOR 49.9K 1% .125W F TC=0+-100 RESISTOR 100K .5% .125W F TC=0+-100 RESISTOR 50K .5% .125W F TC=0+-100 RESISTOR 99.8K .1% .05W PWW TC=±10 | 24546 28480 24546 24546 28480 | C4-1/8-T0-1003-D 0698-3228 C4-1/8-T0-1003-D C4-1/8-T0-5002-D 0698-6688 | | | | |
| A7R127 A7R128 A7R129 A7R131 A7R132 | 0698-3445 0698-6688 0698-3445 0698-5573 0698-7973 | 5 | | RESISTOR 348 1% .125% F TC=0+-100 RESISTOR 99.8K.1% .05W PWW TC=±10 RESISTOR 348 1% .125% F TC=0+-100 RESISTOR 50K .5% .125% F TC=0+-100 RESISTOR 50K .05% .125% F TC=0+-25 | 24546 28480 24546 24546 19701 | C4-1/8-T0-348R-F 0698-6688 C4-1/8-T0-348R-F C4-1/8-T0-5002-D MF4C1/8-T9-5002-D | | | | |
| A7R135 A7R134 A7R135 A7R136 A7R137 | 0698-4158 0698-7973 0698-4158 0687-3301 0687-3301 | 69 6 6 | 2 | RESISTOR 100K .1% .125W F TC#0+-50 RESISTOR 50K .05% .125W F TC#0+-25 RESISTOR 100K .1% .125W F TC#0+-50 RESISTOR 33 10% .5W CC TC#0+412 RESISTOR 33 10% .5W CC TC#0+412 | 28480 19701 28480 01121 01121 | 0698-4158 MF4C1/8-T9-5002-D 0698-4158 E83301 E83301 | | | | |
| A7R138 A7R139 A7R140 | 0698-3193 0698-3193 0684-1031 | 7 7 9 | | RESISTOR 10K .25% .125w F TC#0+-50 RESISTOR 10K .25% .125w F TC#0+-50 RESISTOR 10K 10% .25w FC TC#-400/+700 | 28480 28480 01121 | 0698-3193 0698-3193 C81031 | | | | |
| A7U1 A7U2 A7U3 A7U4 A7U5 | 1826-0026 1820-0939 1820-0949 1820-0943 1820-1114 | 3 5 .7 1 0 | 2 3 4 2 6 | IC COMPARATOR PRCN TO-99 IC FF CMOS D-TYPE POS-EDGE-TRIG DUAL IC GATE CMOS NAND GUAD 2-INP IC GATE CMOS NAND TPL 3-INP IC CNTR CMOS BIN SYNCHRO POS-EDGE-TRIG | 01295 01928 01928 01928 04713 | LM311L CD4013AF CD40011AF CD4023AF MC145168CP | | | | |
| A7U6 A7U7 A7U8 A7U9 A7U11 | 1820=1114 1820=1938 1820=1943 1820=1949 1820=1928 | 0 4 1 7 2 | 3 | IC CNTR CMOS BIN SYNCHRO POS_EDGE_TRIG IC FF CMOS J=K M/S POS=EDGE=TRIG DUAL IC GATE CMOS NAND TPL J=INP IC GATE CMOS NAND GUAD 2=INP IC BFR CMOS GUAD | 04713 01928 01928 01928 01928 | MC14516BCP CD4027AE CD4023AF CD4011AF CD4041AE | | | | |
| A7U12 A7U13 A7U14 A7U15 A7U16 | 1820-1145 1820-0949 1820-0203 1820-1601 1820-0938 | 7 7 6 0 4 | 1 5 1 | IC BFR CMOS INV HEX 1-INP IC GATE CMOS MAND QUAD 2-INP IC OP AMP GP TO-99 IC GATE CMOS EXCL-OR QUAD 2-INP IC FF CMOS J-K M/S POS-EDGE-TRIG DUAL | 01928 01928 01928 01928 01928 | CD4049AF CD4011AF CA741CT CD4070BE CD4027AE | | | | |
| A7U17 A7U18 A7U19 A7U21 A7U22 | 1820-0946 1826-0021 1826-0026 1826-0021 1820-0951 | 4 8 3 8 1 | 12 | IC GATE CMOS NOR GUAD 2-INP IC OP AMP GP TO-99 IC COMPARATOR PRCN TO-99 IC OP AMP GP TO-99 IC MUXR/DATA-SEL CMOS 2-TO-1-LINE GUAD | 01928 27014 01295 27014 01928 | CD4001AF LM310H LM311L LM310H CD4019AF | | | | |
| A7U23 A7U24 A7U25 A7U26 A7U27 | 1820-0951 | 4 4 1 1 6 | | IC GATE CMOS NOR QUAD 2=INP IC FF CMOS J=K M/S POS=EDGE=TRIG DUAL IC MUXX/DATA=SEL CMOS 2=TO=1=LINE QUAD IC MUXX/DATA=SEL CMOS 2=TO=1=LINE QUAD IC OP AMP GP TO=99 | 01928 01928 01928 01928 01928 | CD4001AF CD4027AE CD4019AF CD4019AF CA741CT | | | | |
| | | | | | | | | | | |

Table 6-3. Replaceable Parts (Cont'd).

| | ladie b.3. Kepiaceadie Parts (Cont.d). | | | | | | | | | | |
|---|--|------------------------|-------------|--|--|--|--|--|--|--|--|
| Reference Designation | HP Part Number | C D | Qty | Description | Mfr Code | Mfr Part Number | | | | | |
| A7U28 A7U29 A7U31 A7U32 A7U33 | 1820-1114 1820-1114 1820-0958 1820-0926 1820-0958 | 0 0 8 0 8 | 5 | IC CNTR CMOS BIN SYNCHRO POS-EDGE-TRIG IC CNTR CMOS BIN SYNCHRO POS-EDGE-TRIG IC LCH CMOS D-TYPE QUAD IC ADDR CMOS FULL ADDER 4-BIT IC LCH CMOS D-TYPE QUAD | 04713 04713 01928 01928 01928 | MC145168CP MC145168CP C04042AF C04048AF C04042AF | | | | | |
| A7U34 A7U35 A7U36 A7U37 A7U38 | 1820-0926 1818-1756 1818-1756 1818-1756 1818-1756 | 0 5 5 5 | 8 | IC ADDR CMUS FULL ADDER 4-BIT IC RAM MEMORY IC RAM MEMORY IC RAM MEMORY IC RAM MEMORY | 01928 28480 28480 28480 28480 | CD4008AF 1818-1756 1818-1756 1818-1756 1818-1756 | | | | | |
| A7U39 A7U41 A7U42 A7U43 A7U44 | 1818-1756 1818-1756 1818-1756 1818-1756 1826-0021 | 5 5 5 8 | - | IC RAM MEMORY IC OP AMP GP TO=99 | 28480 28480 28480 28480 27014 | 1818-1756 1818-1756 1818-1756 1818-1756 LM310H | | | | | |
| A 7 U 4 5 A 7 U 4 6 A 7 U 4 7 A 7 U 4 8 A 7 U 4 9 | 1826-0021 1826-0021 1826-0021 1820-0939 1820-0951 | 8 8 6 5 | | IC OP AMP GP TO-99 IC FF CMOS D-TYPE POS-EDGE-TRIG DUAL IC MUXR/DATA-SEL CMOS 2-TO-1-LINE QUAD | 27014 27014 27014 01928 01928 | LM310H LM310H LM310H CD4013AF CD4013AF | | | | | |
| A7U51 A7U52 A7U53 A7U54 A7U55 | 1820-0951 1820-0951 1826-0021 1820-0949 1820-0939 | 1 8 7 5 | | IC MUXR/DATA-SEL CMOS 2-TO-1-LINE QUAD IC MUXR/DATA-SEL CMOS 2-TO-1-LINE QUAD IC OP AMP GP TO-99 IC GATE CMOS NAND QUAD 2-INP IC FF CMOS D-TYPE POS-EDGE-TRIG DUAL | 01928 01928 27014 01928 01928 | CD4019AF CD4019AF LM310M CD4011AF CD4013AF | | | | | |
| A7U56 A7U57 A7U58 A7U59 A7U61 | 1820-1114 1820-1114 1820-0951 1820-0730 1826-0021 | 0 0 1 4 8 | 1 | IC CNTR CMOS BIN SYNCHRO POS-EDGE-TRIG IC CNTR (MOS BIN SYNCHRO POS-EDGE-TRIG IC MUXP/DATA-SEL CMOS 2-TO-1-LINE QUAD IC MV TTL L MONOSTBL RETRIG/RESET DUAL IC OP AMP GP TO-99 | 04713 04713 01928 07263 27014 | MC14516BCP MC14516BCP CD4019AF 96L02DC LM310M | | | | | |
| A7U62 A7U63 A7U64 A7U65 | 1826-0021 1820-0928 1820-0928 1820-0203 | 9 5 2 | | IC OP AMP GP TO-99 IC 8FR CMOS GUAD IC 8FR CMOS GUAD IC 0P AMP GP TO-99 | 27014 01928 01928 01928 | LM310H CD4041AE CD4041AE CA741CT | | | | | |
| A 8 | 03580-66508 | 4 | 1 | BOARD ASSEMBLY-CONTROL | 28480 | 03580-66508 | | | | | |
| ABC1 ABC2 ABC3 ARC4 ABC5 | 0121-0426 0160-2940 0150-0093 0160-0945 0160-0945 | 1 0 2 2 | 1 | CAPACITOR-V TRMR-MICA 50-380PF 175V CAPACITOR-FXD 470PF *-5% 300VDC MICA CAPACITOR-FXD 01UF +00-20% 100VDC CER CAPACITOR-FXD 910PF +-5% 100VDC MICA CAPACITOR-FXD 910PF *-5% 100VDC MICA | 72136 28480 28480 28480 28480 | T52517-7 0160-2940 0150-0093 0160-0945 0160-0945 | | | | | |
| ABC6 ABC7 ABC8 ABC9 ABC11 | 0160-0363 0150-0093 0140-0206 0150-0084 0160-5344 | 8 0 6 9 5 | 1 | CAPACITOR=FXD 620PF +=5% 300VDC MICA CAPACITOR=FXD .01UF +80=20% 100VDC CER CAPACITOR=FXD 270PF +=5% 500VDC MICA CAPACITOR=FXD .1UF +80=20% 100VDC CER CAPACITOR=FXD .082 ±5% 100VDC | 28480 28480 72136 28480 84411 | 0160=0363 0150=0093 DM15F271J0500#V1CR 0150=0084 HEW-249 | | | | | |
| ABC12 ARC13 ABC14 ARC15 ABC16 | 0150-0093 0160-0161 0160-0363 0170-0055 0150-0093 | 0 4 8 6 0 | 1 | CAPACITOR=FXD .01UF +80=20X 100VDC CER CAPACITOR=FXD .01UF +=10X 200VDC PDLYE CAPACITOR=FXD 620PF +=5X 300VDC MICA CAPACITOR=FXD .1UF +=20X 200VDC PDLYE CAPACITOR=FXD .01UF +80=20X 100VDC CER | 28480 28480 28480 28480 | 0150-0093 0160-0161 0160-0363 0170-0055 0150-0093 | | | | | |
| ABC18 ABC19 ABC21 ABC22 ABC23 | 0150-0093 0160-0164 0180-0374 0160-0166 0160-5269 | 0 7 3 9 5 | 1 3 1 | CAPACITOR=FXD .01UF +80=20% 100VDC CER CAPACITOR=FXD .039UF +=10% 200VDC POLYE CAPACITOR=FXD .068UF +=10% 200VDC POLYE CAPACITOR=FXD .068UF +=10% 200VDC POLYE CAPACITOR=FXD .047 ±10% 50VDC | 28480 28480 56289 28480 28480 | 0150-0093 0160-0164 1500108X902082 0160-0166 0160-5269 | | | | | |
| APC 24 ABC 25 ABC 26 ABC 27 ABC 28 | 0150-0122 0150-0122 0180-1746 | .5 8 6 6 5 | 2 | CAPACITOR=FXD 47UF+=10X 35VDC TA CAPACITOR=FXD 2,2UF+=10X 20VDC TA CAPACITOR=FXD 2000PF +=20X 500VDC CER CAPACITOR=FXD 2000PF +=20X 500VDC CER CAPACITOR=FXD 15UF+=10X 20VDC TA | 56289 56289 28480 28480 56289 | 150D474X9035A2 150D225X9020A2 0150=0122 150D156X902082 | | | | | |
| A 8C 29 A 8C 31 A 8C 32 A 8C 33 A 8C 34 A 8C R 1 A 8C R 2 A 8C R 3 A 8C R 3 A 8C R 3 | 0180-0141 0180-1746 0180-1746 0180-1746 0160-4571 1901-0040 1901-0040 1901-0040 | 5 2 5 5 1 1 1 1 1 1 4 | 1 | CAPACITOR-FXD 1SUF+-10% 20VDC TA CAPACITOR-FXD 5QUF+75-10% 50VDC AL CAPACITOR-FXD 5UFF+10% 20VDC TA CAPACITOR-FXD 1SUF+=10% 20VDC TA CAPACITOR-FXD 1UF +80 20% 50VDC DIODE-SWITCHING 30V 50MA 2N8 D0-35 DIODE-ZNR 5.11V 5% D0-35 P0-4M | 56289 56289 56289 28480 28480 28480 28480 28480 28480 28480 | 1500156x902082 300506G050002 1500156x902082 1500156x902082 0160-4571 1901-0040 1901-0040 1901-0040 1902-0040 | | | | | |
| A8CR6 A8CR7 A8CR8 A8CR9 A8CR11 | 1902-3182 1901-0040 1901-0040 | 1 0 1 1 1 1 | | DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-ZNR 12.1V 5X DO-35 PD=.un DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 | 28480 28480 28480 28480 28480 | 1901-0040 1902-3182 1901-0040 1901-0040 1901-0040 | | | | | |
| | | | | | | | | | | | |

Table 6-3. Replaceable Parts (Cont'd).

| | | _ | | inie 0.3. nepiaceanie raits (Cont i | -,. | |
|---|---|-----------------------|------------------|---|---|--|
| Reference Designation | HP Part Number | C D | Qty | Description | Mfr Code | Mfr Part Number |
| ABCR12 ABCR13 ABCR14 ABCR15 ABCR16 | 1901-0040 1902-3311 1902-3311 1901-0033 1901-0033 | 1 7 7 2 2 | 2 5 | DIODE-SWITCHING 30V 50MA 2NS DO-35 DIODE-ZNR 38.3V 5% DO-35 PD=.4W DIODE-ZNR 38.3V 5% DO-35 PD=.4W DIODE-GEN PRP 180V 200MA DO-7 DIODE-GEN PRP 180V 200MA DO-7 | 28480 28480 28480 28480 28480 | 1901-0040 1902-3311 1902-3311 1901-0033 1901-0033 |
| ABCR17 ABCR18 ABCR19 ABCR21 ABCR22 | 1901-0033 1901-0033 1901-0033 1901-0050 1901-0040 | 2 2 3 1 | 1 | DIODE-GEN PRP 180V 200MA DO-7 DIODE-GEN PRP 180V 200MA DO-7 DIODE-GEN PRP 180V 200MA DO-7 DIODE-SWITCHING 80V 200MA 2NS DO-35 DIODE-SWITCHING 30V 50MA 2NS DO-35 | 28480 28480 28480 28480 28480 | 1901-0033 1901-0033 1901-0033 1901-0050 1901-0040 |
| ABCR23 | 1901-0040 | 1 | | DIODE-SWITCHING BOV SOMA ZNS DO-35 | 28480 | 1901-0040 |
| A 5 L 1 A 6 L 2 A 5 L 3 A 6 L 4 | 9140-0129 9140-0129 9100-3282 9100-3282 | 1 1 9 9 | 5 | INDUCTORRF=CH=MLD 220UH 5% .166DX.385LG INDUCTORRF=CH=MLD 220UH 5% .166DX.385LG INDUCTOR 200UH 20% .55DX.5LG | 28480 28480 28480 28480 | 9140-0129 9140-0129 9100-3282 9100-3282 |
| A 8 Q 1 A 8 Q 2 A 8 Q 3 A 8 Q 4 A 8 Q 5 | 1854-0071 1854-0071 1855-0081 1855-0081 1853-0010 | 7 1 1 2 | | TRANSISTOR NPN SI PD=300MW FT=200MHZ TRANSISTOR NPN SI PD=300MW FT=200MHZ TRANSISTOR J=FET N=CHAN D=MODE SI TRANSISTOR J=FET N=CHAN D=MODE SI TRANSISTOR PNP SI TO=18 PD=360MW | 28480 28480 01295 01295 28480 | 1854-0071 1854-0071 285245 285245 1853-0010 |
| A 8 Q 6 A 8 Q 7 A 8 Q 8 A 8 Q 9 A 8 Q 1 1 | 1853-0010 1854-0071 1854-0071 1853-0010 1854-0071 | 2 7 7 2 7 | | TRANSISTOR PNP SI TD=18 PD=360MW TRANSISTOR NPN SI PD=300MW FT=200MHZ TRANSISTOR NPN SI PD=300MW FT=200MHZ TRANSISTOR PNP SI TO=18 PD=360MW TRANSISTOR NPN SI PD=300MW FT=200MHZ | 28480 28480 28480 28480 28480 | 1853-0010 1854-0071 1854-0071 1853-0010 1854-0071 |
| A8Q12 A8Q13 A8Q14 A8Q15 A8Q16 | 1853-0086 1854-0071 1854-0071 1854-0019 1853-0016 | 2 7 7 3 8 | 3 | TRANSISTOR PNP SI PD=310MW FT=40MMZ TRANSISTOR NPN SI PD=300MW FT=200MHZ TRANSISTOR NPN SI PD=300MW FT=200MHZ TRANSISTOR NPN SI TO=18 PD=360MW TRANSISTOR NPN SI TO=92 PD=300MW | 27014 26480 26480 26480 26480 | 2N5087 1854-0071 1854-0071 1854-0019 1853-0016 |
| A8G17 A8G19 A8G21 A8G22 | 1854-0019 1854-0232 1854-0474 1854-0476 1853-0010 | 3 2 4 6 2 | 1 1 1 | TRANSISTOR NPN SI TO-18 PD=360MW TRANSISTOR NPN SI TO-39 PD=1W FT=15MHZ TRANSISTOR NPN SI PD=310MW FT=1000HZ TRANSISTOR NPN SI PD=350MW TRANSISTOR NPN SI TO-18 PD=350MW | 26480 28480 04713 01928 28480 | 1854-0019 1854-0232 285551 283079 1853-0010 |
| A8Q23 | 1855-0081 | 1 | | TRANSISTOR J-FET N-CHAN D-MODE SI | 01295 | 2N5245 |
| 48R1 46R2 48R3 48R4 48R5 | 2100-3354 2100-3358 2100-3357 2100-3353 0684-4721 | 9 3 2 8 0 | í | RESISTOR-TRMR 50K 10% C SIDE-ADJ 1-TRN RESISTOR-TRMR 1M 20% C SIDE-ADJ 1-TRN RESISTOR-TRMR 500K 10% C SIDE-ADJ 1-TRN RESISTOR-TRMR 20% 10% C SIDE-ADJ 1-TRN RESISTOR-TRMR 20% 10% C SIDE-ADJ 1-TRN RESISTOR 4.7K 10% .25W FC TC=-400/+700 | 28480 28480 28480 32997 01121 | 2100-3354 2100-3358 2100-3357 3368x-946-203 CB4721 |
| A6R6 A6R7 A6R6 A6R9 A6R11 | 0683-4715 0683-1825 0684-2231 0683-3325 0684-2231 | 0 7 3 6 3 | 3 1 | RESISTOR 470 5% ,25W FC TC==400/+600 RESISTOR 1,8K 5% ,25W FC TC==400/+700 RESISTOR 22K 10% ,25W FC TC==400/+800 RESISTOR 3,3K 5% ,25W FC TC==400/+800 RESISTOR 22K 10% ,25W FC TC==400/+800 | 01121 01121 01121 01121 | C84715 C81825 C82231 C83325 C82231 |
| ABR12 ABR13 ABR14 ABR15 ABR16 | 0684-2231 0698-4483 0698-4483 0757-0449 0757-0449 | 3 0 0 6 6 | | RESISTOR 22K 10% _25W FC TC==400/+800 RESISTOR 18.7K 1% _125W F TC=0+=100 RESISTOR 26K 1% _125W F TC=0+=100 RESISTOR 26K 1% _125W F TC=0+=100 RESISTOR 26K 1% _125W F TC=0+=100 | 01121 24546 24546 24546 24546 | C82231 C4-1/8-T0-1872-F C4-1/8-T0-1872-F C4-1/8-T0-2002-F C4-1/8-T0-2002-F |
| A8R17 A8R18 A8R19 A8R21 A8R22 | 0757-0442 0757-0449 0757-0280 0698-4473 0757-0458 | 9 6 3 8 7 | | RESISTOR 10K 1% .125W F TC=0+-100 RESISTOR 20K 1% .125W F TC=0+-100 RESISTOR 1K 1% .125W F TC=0+-100 RESISTOR 8.00K 1% .125W F TC=0+-100 RESISTOR 51.1K 1% .125W F TC=0+-100 | 24546 24546 24546 24546 24546 | C4=1/8=T0=1002=F C4=1/8=T0=2002=F C4=1/8-T0=1001=F C4=1/8-T0=8061=F C4=1/8=T0=5112=F |
| ABR23 ABR24 ABR25 ABR26 ABR27 | 0757-0451 0683-1555 0684-1041 0757-0458 0757-0458 | 0 0 1 7 7 | 1 1 | RESISTOR 24.3k 1% .125W F TC=0+=100 RESISTOR 1.5M 5% .25W FC TC==900/+1100 RESISTOR 100K 10% .25W FC TC==400/+800 RESISTOR 51.1K 1% .125W F TC=0+=100 RESISTOR 51.1K 1% .125W F TC=0+=100 | 24546 01121 01121 24546 24546 | C4-1/8-T0-2432-F C8:555 C8:1041 C4-1/8-T0-5:1:2-F C4-1/8-T0-5:1:2-F |
| A8R28 A8R29 A8R31 A8R32 A6R33 | 0757-0472 0757-0458 0698-4503 0698-4526 0698-4507 | 5 7 5 2 9 | 1 1 1 1 | RESISTOR 200K 1% .125W F TC=0+-100 RESISTOR 51.1K 1% .125W F TC=0+-100 RESISTOR 66.5K 1% .125W F TC=0+-100 RESISTOR 191K 1% .125W F TC=0+-100 RESISTOR 76.8K 1% .125W F TC=0+-100 | 24546 24546 24546 24546 24546 | C4=1/8=T0=2003=F C4=1/8=T0=5112=F C4=1/8=T0=652=F C4=1/8=T0=1913=F C4=1/8=T0=7682=F |
| A8R34 A8R35 A8R36 A8R38 A8R39 | 0684-1041 0684-5631 0757-0465 0683-2225 0757-0442 | 1 3 6 3 9 | | RESISTOR 100K 10% ,25W FC TC=-400/+800 RESISTOR 56K 10% ,25W FC TC=-400/+800 RESISTOR 100K 1% ,125W F TC=0+-100 RESISTOR 2,2K 5% ,25W FC TC=-400/+700 RESISTOR 10K 1% ,125W F TC=0+-100 | 01121 01121 24546 01121 24546 | C81041 C85631 C4-1/8-T0-1003-F C82225 C4-1/8-T0-1002-F |
| A8R41 A8R42 A8R43 A8R44 A8R45 | 0757-0477 0698-4541 0757-0483 0684-1831 0684-1831 | 0 1 8 7 7 | 1 | RESISTOR 332K 1% 125W F TC=0+-100 RESISTOR 442K 1% 125W F TC=0+-100 RESISTOR 562K 1% 125W F TC=0+-100 RESISTOR 16K 10% 25W FC TC==400/+600 RESISTOR 16K 10% 25W FC TC==400/+600 | 19701 28480 28480 01121 01121 | MF4C1/8-T0-3323-F 0098-4541 0757-0403 C81631 C81631 |
| | | | | | | |

Table 6-3. Replaceable Parts (Cont'd).

| Table 0-3. neplaceable Parts (Collt Q). | | | | | | | | | |
|---|---|-----------------------|-------------|--|---|--|--|--|--|
| Reference Designation | HP Part Number | C D | Qty | Description | Mfr Code | Mfr Part Number | | | |
| A8R46 A8R47 A8R48 A8R49 A8R50 | 0684-1831 0684-1831 0684-1041 0684-5631 0698-3161 | 7 7 1 3 9 | 2 | RESISTOR 18K 10% _25W FC TC==400/+800 RESISTOR 18K 10% _25W FC TC==400/+800 RESISTOR 100K 10% _25W FC TC==400/+800 RESISTOR 56K 10% _25W FC TC==400/+800 RESISTOR 38,3K 1% _125W F TC=0+=100 | 01121 01121 01121 01121 24546 | C81831 C81831 C81041 C85631 C4-1/8-T0-3832-F | | | |
| A8R51 A8R52 A8R53 A8R54 A8R55 | 0684-4731 0683-2225 0684-4731 0757-0449 0757-0465 | 23266 | | RESISTOR 47K 10% .25W FC TC==400/+800 RESISTOR 2_2K 5% .25W FC TC==400/+700 RESISTOR 47K 10% .25W FC TC==400/+800 RESISTOR 20K 1% .125W F TC=0+=100 RESISTOR 100K 1% .125W F TC=0+=100 | 01121 01121 01121 24546 24546 | C84731 C82225 C84731 C4-1/8-T0-2002-F C4-1/8-T0-1003-F | | | |
| A8R56 A8R57 A8R58 A8R59 A8R61 | 0684-4731 0684-4731 0698-3519 0698-3228 0698-3149 | 2 1 9 3 | 1 | RESISTOR 47K 10% _25W FC TC==400/+800 RESISTOR 47K 10% _25W FC TC==400/+800 RESISTOR 12_4K 1% _125W F TC=0+=100 RESISTOR 40_9K 1% _125W F TC=0+=100 RESISTOR 255K 1% _125W F TC=0+=100 | 01121 01121 24546 26460 24546 | C84731 C84731 C4-1/8-T0-1242-F 0698-3228 C4-1/8-T0-2553-F | | | |
| A8R62 A8R63 A8R64 A6R65 A6R66 | 0698-3266 0698-4532 0698-3460 0698-7332 0698-4505 | 5 0 1 4 7 | 1 1 1 | RESISTOR 237K 1% .125W F TC=0+-100 RESISTOR 280K 1% .125W F TC=0+-100 RESISTOR 422K 1% .125W F TC=0+-100 RESISTOR 1M 1% .125W F TC=0+-100 RESISTOR 71.5K 1% .125W F TC=0+-100 | 24546 24546 28480 28480 24546 | C4-1/8-T0-2373-F C4-1/8-T0-2803-F 0698-3460 0698-7332 C4-1/8-T0-7152-F | | | |
| A8R67 A8R68 A8R69 A8R71 A8R72 | 0757-0486 0757-0469 0684-1041 0684-1041 0757-0394 | 1 0 1 1 0 | 1 | RESISTOR 750K 1% .125W F TC=0+-100 RESISTOR 150K 1% .125W F TC=0+-100 RESISTOR 100K 10% .25W FC TC=+400/+800 RESISTOR 100K 10% .25W FC TC=-400/+800 RESISTOR 51.1 1% .125W F TC=0+-100 | 26480 24546 01121 01121 24546 | 0757-0486 C4-1/8-T0-1503-F C81041 C81041 C4-1/8-T0-51R1-F | | | |
| A8R73 A8R74 A8R75 A8R76 A8R77 | 0757-0273 0757-0284 0757-0282 0757-0283 0698-3161 | 4 7 5 6 9 | 1 | RESISTOR 3.01k 1% .125W F TC=0+-100 RESISTOR 150 1% .125W F TC=0+-100 RESISTOR 221 1% .125W F TC=0+-100 RESISTOR 2K 1% .125W F TC=0+-100 RESISTOR 38.3K 1% .125W F TC=0+-100 | 24546 24546 24546 24546 24546 | C4-1/8-T0-3011-F C4-1/8-T0-151-F C4-1/8-T0-221R-F C4-1/8-T0-201-F C4-1/8-T0-3832-F | | | |
| A8R78 A8R95 A8R96 | 0698-3149 0757-0283 0757-0401 | 3 6 0 | | RESISTOR 255K 1% ,125W F TC=0+=100 RESISTOR 2K 1% ,125W F TC=0+=100 RESISTOR 100 1% ,125W F TC=0+=100 | 24546 24546 24546 | C4-1/8-T0-2553-F C4-1/8-T0-2001-F C4-1/8-T0-101-F | | | |
| 48U1 A8U2 A8U3 A8U4 A8U5 | 1826-0021 1826-0021 1826-0043 1826-0021 1826-0043 | 8 8 4 8 4 | | IC OP AMP GP TO-99 | 27014 27014 01928 27014 01928 | LM310H LM310H CA307T LM310H CA307T | | | |
| A8U6 A8U7 A8U8 | 1826-0043 1820-0203 1820-0203 | 4 6 6 | | IC OP AMP GP TO-99 IC OP AMP GP TD-99 IC OP AMP GP TO-99 | 01928 01928 01928 | CA307T CA741CT CA741CT | | | |
| | 1205-0048 1251-0599 | 3 6 | 1 1 | HEAT SINK TO-8-CS CONNECTOR 3-PIN M POST TYPE | 28480 28480 | 1205-0048 1251-0599 | | | |
| 49 | 03580-66509 | 5 | 1 | BOARD ASSEMBLY-INPUT (STD. 3580A ONLY) | 28480 | 03580-66509 | | | |
| A9C1 A9C2 A9C3 A9C4 A9C5 | 0170-0042 0121-0407 0121-0407 0121-0407 0121-0407 | 1 9 9 9 | 5 | CAPACITOR-FXD .33UF +-5% 100VDC POLYE CAPACITOR-V TRMR-PSTN .7-3PF 600V CAPACITOR-V TRMR-PSTN .7-3PF 600V CAPACITOR-V TRMR-PSTN .7-3PF 600V CAPACITOR-V TRMR-PSTN .7-3PF 600V | 99515 72982 72982 72982 72982 | E1-334D 536-016 536-016 536-016 536-016 | | | |
| A9C6 A9C7 A9C8 A9C9 A9C10 | 0121-0407 0150-0022 0140-0162 0150-0011 0160-2207 | 9 5 3 2 3 | 1 1 1 | CAPACITOR-V TRMR-PSTN .7-3PF 600V CAPACITOR-FXD 3.3PF +-10% 500VDC TI DIOX CAPACITOR-FXD 4700PF +-10% 300VDC MICA CAPACITOR-FXD 1.5PF +-20% 500VDC TI DIOX CAPACITOR-FXD 300PF +-5% 300VDC MICA | 72982 28480 72136 28480 28480 | 536-016 0150-0022 DM20F472K0300WV1CR 0150-0011 0160-2207 | | | |
| A9C11 A9C12 A9C13 A9C14 A9C15 | 0150-0022 0150-0022 0160-0356 0150-0022 0150-0022 | 5 9 5 5 | 1 | CAPACITOR=FXD 3.3PF +=10x S00VDC TI DIOX CAPACITOR=FXD 3.3PF +=10X S00VDC TI DIOX CAPACITOR=FXD 18PF +=5X 300VDC MICA CAPACITOR=FXD 3.3PF +=10X 500VDC TI DIOX CAPACITOR=FXD 3.3PF +=10X 500VDC TI DIOX | 28480 28480 28480 28480 28480 | 0150-0022 0150-0022 0160-0356 0150-0022 0150-0022 | | | |
| A9C16 A9C17 A9C18 A9C19 A9C21 | 0180-0229 0180-0229 0140-0210 0160-2198 0180-0060 | 7 7 2 1 4 | 2 1 2 | CAPACITOR=FXD 33UF+=10X 10VDC TA CAPACITOR=FXD 33UF+=10X 10VDC TA CAPACITOR=FXD 270FF +=5X 300VDC MICA CAPACITOR=FXD 20PF +=5X 300VDC MICA CAPACITOR=FXD 20PUF+75=10X 3VDC AL | 56289 56289 72136 28480 56289 | 150D336X9010B2 150D336X9010B2 DM15F271J0300WV1CR 0160-2198 30D207G003CC2 | | | |
| A9C22 A9C23 A9C24 A9C25 A9C26 | 0180-1758 | 0 8 9 5 9 | 3 | CAPACITOR-FXD 100PF +-5% 300VDC MICA CAPACITOR-FXD 2,2UF+-10% 20VDC TA CAPACITOR-FXD 300UF+75=10% 3VDC AL CAPACITOR-FXD 100UF+75=10% 3VDC AL CAPACITOR-FXD 300UF+75=10% 3VDC AL | 28480 56289 56289 56289 56289 | 0160-2204 150D225x9020A2 30D307c003DC2 30D107c016DC2 30D307G003DC2 | | | |
| A9C27 A9C2B A9C29 A9C30 A9C31 | 0180-0210 0140-0210 0180-0060 0160-2204 0160-0763 | 62402 | | CAPACITOR-FX0 3.3UF+-20% 15VDC TA CAPACITOR-FX0 270PF +-5% 300VDC MICA CAPACITOR-FXD 200UF+75-10% 3VDC AL CAPACITOR-FXD 100PF +-5% 300VDC MICA CAPACITOR-FXD 5PF +-10% 500VDC MICA | 56289 72136 56289 28480 28480 | 150D335x0015A2 DM15F271J0300WV1CR 30D2076003CC2 0160-2204 0160-0763 | | | |
| | | | | | | | | | |

Table 6-3. Replaceable Parts (Cont'd).

| lable 6-3. Replaceable Parts (Cont'd). | | | | | | | | | |
|---|---|-----------------------|------------------|--|---|--|--|--|--|
| Reference Designation | HP Part Number | C D | Qty | Description | Mfr Code | Mfr Part Number | | | |
| A9C32 A9C33 A9C34 A9C35 A9C36 | 0180-1758 0180-0061 0180-0137 0160-2724 0140-0217 | 95699 | 1 1 | CAPACITOR-FXD 300UF+75-10% 3VDC AL CAPACITOR-FXD 100UF+75-10% 16VDC AL CAPACITOR-FXD 100UF+20% 10VDC TA CAPACITOR-FXD 3600PF +-2% 500VDC MICA CAPACITOR-FXD 140PF +-2% 300VDC MICA | 56289 56289 56289 28480 72136 | 30D307G003DC2 30D107G016DC2 150D107X0010R2 0160-27724 DM15F141G0300WV1CR | | | |
| A9C37 A9C38 A9C39 A9C41 A9C42 | 0160-3269 0160-0341 0160-3269 0140-0233 | 5 6 5 | 1 | CAPACITOR-FXD 7610PF +=1% 100VDC MICA CAPACITOR-FXD 640PF +=1% 300VDC MICA CAPACITOR-FXD 7610PF +=1% 100VDC MICA CAPACITOR-FXD 480PF +=1% 300VDC MICA CAPACITOR-FXD 3300PF +=5% 300VDC MICA | 28480 28480 28480 72136 28480 | 0160-3269 0160-0341 0160-3269 DM15F481F0300WV1C 0160-2230 | | | |
| 49C43 49C45 49C45 49C45 | 0180-0303 0160-2930 0180-0374 0180-0374 0160-2930 | 8 0 3 3 0 | 1 | CAPACITOR-FXD 100UF+75=10% 3VDC AL CAPACITOR-FXD 01UF+80-20% 100VDC CER CAPACITOR-FXD 10UF+=10% 20VDC TA CAPACITOR-FXD 10UF+=10% 20VDC TA CAPACITOR-FXD 01UF+80-20% 100VDC CER | 56289 28480 56289 56289 28480 | 300107g003cg2 0150-0093 1500106×902082 1500106×902082 0150-0093 | | | |
| 49C47 49C48 49C49 49C51 49C52 | 0180-0197 0160-2605 0150-0093 0160-2035 0180-0197 | 8 5 0 5 8 | 1 | CAPACITUR-FXD 2.2UF++10% 20VDC TA CAPACITOR-FXD .02UF +80-20% 25VDC CER CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD 750PF +-5% 300VDC MICA CAPACITOR-FXD 2.2UF++10% 20VDC TA | 56289 28480 28480 28480 56289 | 1500225X9020A2 0160-2605 0150-003 0160-2035 1500225X9020A2 | | | |
| A9C53 A9C54 A9C55 A9C56 A9C56 | 0160-2930 01 A 0 = 0197 0160 = 2009 01 A 0 = 0197 0160-2930 | 0 8 3 6 0 | 1 | CAPACITOR-FXD .01UF +80-20X 100VDC CER CAPACITOR-FXD 2.2UF+-10X 20VDC TA CAPACITOR-FXD 820PF +-5X 300VDC MICA CAPACITOR-FXD 2.2UF+-10X 20VDC TA CAPACITOR-FXD .01UF +80-20X 100VDC CER | 28480 56289 28480 56289 28480 | 0150=0093 1500225x9020A2 0160-2009 1500225x9020A2 0150=0093 | | | |
| 49C58 49C59 49C61 49C63 | 0160-2930 0180-0197 0180-0228 0180-0197 0180-0339 | 0 8 6 8 0 | 2 | CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD 2.2UF+-10% 20VDC TA CAPACITOR-FXD 22UF+-10% 15VDC TA CAPACITOR-FXD 2.2UF+-10% 20VDC TA CAPACITOR-FXD 50UF+75-10% 16VDC AL | 28480 56289 56289 56289 56289 | 0150-0093 1500225x9020A2 1500226x901592 1500225x9020A2 300506G016C82 | | | |
| A9C64 A9C65 A9C66 | 0180-0197 0180-0228 0180-0339 | 8 6 0 | | CAPACITOR-FXD 2.2UF+-10% 20VDC TA CAPACITOR-FXD 22UF+-10% 15VDC TA CAPACITOR-FXD 50UF+75-10% 16VDC AL | 56289 56289 56289 | 150D225x9020A2 150D226x901582 30D506G016C82 | | | |
| A9CR1 A9CR2 A9CR3 A9CR4 A9CR5 | 1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 | 1 1 1 1 1 | | DIODE-SWITCHING 30V 50MA 2NS 00-35 DIODE-SHITCHING 30V 50MA 2NS 00-35 DIODE-SWITCHING 30V 50MA 2NS 00-35 DIODE-SWITCHING 30V 50MA 2NS 00-35 DIODE-SWITCHING 30V 50MA 2NS 00-35 | \$8480 \$8480 \$8480 \$8480 | 1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 | | | |
| A9J1 | 1251-2969 | 8 | 1 | CONNECTOR-PHONO SINGLE PHONO JACK! DIP | 28480 | 1251-2969 | | | |
| A9L1 A9L2 A9L3 A9L4 A9L5 | 9100-3264 9100-3259 9100-3260 9100-3277 9170-0894 | 7 0 3 2 0 | 1 1 1 | INDUCTORRE-CH-MLD 2.46MH 2% .8D Q=200 INDUCTORRE-CH-MLD 2.46MH 10% Q=200 INDUCTORRE-CH-MLD 2.15H 2% .8D Q=200 INDUCTORRE-CH-MLD 3.1MH 2% .62D Q=150 CORE-SHIELDING BEAD | 28480 28480 28480 28480 28480 | 9100-3264 9100-3259 9100-3260 9100-3277 9170-0894 | | | |
| A 9 0 1 A 9 0 2 A 9 0 3 A 9 0 4 A 9 0 5 | 1855-0377 1854-0226 1853-0086 1854-0071 | 8 4 2 7 7 | 1 | TRANSISTOR J=FET N=CHAN D=MODE TO=18 SI TRANSISTOR NPN 2N4384 SI TO=18 PD=500MW TRANSISTOR PNP SI PD=300MW FT=40MHZ TRANSISTOR NPN SI PD=300MW FT=200MHZ TRANSISTOR NPN SI PD=300MW FT=200MHZ | 28480 13606 27014 28480 28480 | 1855-0377 2N4384 2N5087 1854-0071 1854-0071 | | | |
| A906 A907 A908 A909 A9011 | 1854-0226 1853-0086 1854-0071 1854-0071 | 4 2 7 7 7 | | TRANSISTOR NPN 2N4384 SI TO-18 PDB500MW TRANSISTOR PNP SI PDB310MM FTB40MMZ TRANSISTOR NPN SI PDB300MM FTB200MMZ TRANSISTOR NPN SI PDB300MM FTB200MMZ TRANSISTOR NPN SI PDB300MM FTB200MMZ | 13606 27014 28480 28480 28480 | 2N4384 2N5087 1854-0071 1854-0071 1854-0071 | | | |
| 49012 49013 49014 49015 49016 | 1854-0071 1854-0071 1854-0226 1853-0010 1854-0071 | 7 7 4 2 7 | · | TRANSISTOR NPN SI PD=300MM FT=200MHZ TRANSISTOR NPN SI PD=300MM FT=200MMZ TRANSISTOR NPN 2N4384 SI TO=16 PD=500MM TRANSISTOR NPN SI TO=16 PD=360MM TRANSISTOR NPN SI TO=16 PD=360MM TRANSISTOR NPN SI PD=300MM FT=200MMZ | 28480 28480 13606 28480 28480 | 1854-0071 1854-0071 200384 1853-0010 1854-0071 | | | |
| 49017 49018 49019 | 1854-0071 1854-0071 1853-0010 | 7 7 2 | | TRANSISTOR NPN SI PD=300MM FT=200MHZ TRANSISTOR NPN SI PD=300MW FT=200MHZ TRANSISTOR PNP SI TO=18 PD=360MW | 28480 28480 28480 | 1854-0071 1854-0071 1853-0010 | | | |
| A9R1 A9R2 A9R3 A9R4 A9R5 | 2100-0580 2100-0640 0698-5159 0698-4055 0698-5132 | 7 0 9 2 8 | 1 2 2 2 | RESISTOR-TRMR 500K 10% C TOP-ADJ 1-TRN RESISTOR-VAR W/3W 5K 10% LIN 3PST-NO RESISTOR 1M .5% .25W F TC=0+-100 RESISTOR 1K .25% .125W F TC=0+-100 RESISTOR 990K .5% .25W F TC=0+-100 | 28480 28480 28480 03888 28480 | 2100-0580 2100-0640 0698-5159 PME55-1/8-T0-1001-C 0698-5132 | | | |
| A9R5 A9R7 A9R5 A9R9 A9R11 | 0698-5132 0698-5131 | 2 0 8 7 6 | 2 1 2 1 | RESISTOR 124K 1% .125W F TC=0+-100 RESISTOR 11.11K .25% .125W F TC=0+-100 RESISTOR 990K .5% .25W F TC=0+-100 RESISTOR 900K .5% .25W F TC=0+-100 RESISTOR 127K .25% .125W F TC=0+-100 | 24546 26480 28480 19701 26480 | C4=1/8-T0=1243-F 0698-6661 0698-5132 MF52C1/4-T0=9003-D 0698-6659 | | | |
| A9R12 A9R13 A9R14 A9R15 A9R16 | 0757-0430 0698-3150 0698-5159 | 7 5 6 9 1 | 1 | RESISTOR 900K .5% .25W F TC=0+=100 RESISTOR 2.21K 1% .125W F TC=0+=100 RESISTOR 2.37K 1% .125W F TC=0+=100 RESISTOR 1M .5% .25W F TC=0+=100 RESISTOR 2K 1% .5W F TC=0+=100 | 19701 24546 24546 28480 28480 | MF52C1/4=T0=0003=D C4=1/8=T0=2211=F C4=1/8=T0=2371=F 0698=5159 0757=0824 | | | |

Table 6-3. Replaceable Parts (Cont'd).

| Poforonco UD David - | | | | | | | | | | |
|--------------------------|------------------------|--------|-----|---|-------------------------|---|--|--|--|--|
| Reference Designation | HP Part Number | C D | Qty | Description | Mfr Code | Mfr Part Number | | | | |
| 49R17 A9R18 | 0684-1041 0698-3581 | 1 7 | 2 | RESISTOR 100K 10% .25W FC TC==400/+800 RESISTOR 13.7K 1% .125W F TC=0+=100 | 01121 | C81041 | | | | |
| AGRIG | 0698-3581 | 7 | • | RESISTOR 13.7K 1% .125W F TC#0+=100 | 24546 | C4-1/8-T0-1372-F C4-1/8-T0-1372-F | | | | |
| 15982 15982 | 0698-4473 | 8 9 | | RESISTOR 8.06K 1% .125W F TC=0+-100 RESISTOR 10K 1% .125W F TC=0+-100 | 24546 | C4-1/8-T0-8061-F C4-1/8-T0-1002-F | | | | |
| AGR23 | 0698-4421 | 6 | 2 | RESISTOR 249 1% .125W F TC=0+=100 | 24546 | C4-1/8-T0-249R-F | | | | |
| 49R2U 49R25 | 0698-3193 | 7 3 | 1 | KESISTUR 10K .25% .125W F TC80+-50 | 28480 | 0698-3193 | | | | |
| A9R26 A9R27 | 0698-4486 | 3 | | RESISTOR 1.153K .25% .125W F TC=0+-50 RESISTOR 24.9K 1% .125W F TC=0+-100 RESISTOR 5.49K 1% .125W F TC=0+-100 | 28480 24546 24546 | 0698-6862 C4-1/8-T0-2492-F C4-1/8-T0-5491-F | | | | |
| ASREA | 0757-0407 | 6 | | RESISTOR 200 1x .125W F TC=0+=100 | 24546 | C4-1/8-T0-201-F | | | | |
| 49829 49831 | 0698-4464 | 7 | 1 | RESISTOR 887 1% ,125W F TC=0+=100 RESISTOR 100K 10% ,25W FC TC==400/+800 | 24546 | C4-1/8-T0-887R-F | | | | |
| A9R32 A9R33 | 0757-0448 0684-4701 | 5 6 | | RESISTOR 18.2K 1% .125W F TC=0+-100 RESISTOR 47 10% .25W FC TC=-400/+500 | 01121 24546 01121 | C81041 C4-1/8-T0-1822-F C84701 | | | | |
| APR34 | 0757-0407 | | | RESISTOR 200 1% .125W F TC=0+=100 | 24546 | C4-1/8-T0-201-F | | | | |
| 49835 49836 | 0698-3488 | 3 1 | | RESISTOR 442 1% .125W F TC=0+=100 RESISTOR 100K 10% .25W FC TC==400/+800 | 24546 | C4-1/8-T0-422R-F | | | | |
| A9R37 | 0757-0442 | 9 | | RESISTOR 10K 1% .125W F TC=0+=100 | 01121 | C81041 C4-1/8-T0-1002-F | | | | |
| A9R38 | 0757-0278 | 9 | 5 | RESISTOR 1.78K 1% .125W F TC#0+=100 | 24546 | C4-1/8-T0-1781-F | | | | |
| A9R40 | 0698-6780 0684-1041 | 1 | 1 | RESISTOR 5.62K .25% .125W F TC=0+=50 RESISTOR 100K 10% .25W FC TC==400/+800 | 28480 01121 | 0698-6780 CB1041 | | | | |
| A9R41 A9R42 | 0698-6823 | 8 | 1 | RESISTOR 2.61K .25% .125W F TC#0+=100 RESISTOR 8.06K 1% .125W F TC#0+=100 | 19701 24546 | MF4C1/8-T0-2611-C | | | | |
| A9R43 | 0698-3495 | 2 | | RESISTOR 966 1% .125W F TC=0+=100 | 24546 | C4-1/8-T0-8061-F C4-1/8-T0-866R-F | | | | |
| A9R44 A9R45 | 0757-0424 0757-0442 | 7 | 1 | RESISTOR 1.1K 1% .125w F TC=0+=100 RESISTOR 10K 1% .125w F TC=0+=100 | 24546 24546 | C4-1/8-T0-1101-F | | | | |
| A9R46 A9R47 | 0757-0442 | 9 | | RESISTOR 10K 1% .125W F TC=0+-100 | 24546 | C4-1/8-T0-1002-F C4-1/8-T0-1002-F | | | | |
| AGRUS | 0698-3154 0757-0407 | 6 | 1 | RESISTOR 4.22K 1% .125W F TC=0+-100 RESISTOR 200 1% .125W F TC=0+-100 | 24546 24546 | C4-1/8-T0-4221-F C4-1/8-T0-201-F | | | | |
| 49R49 49R50 | 0698-4483 0683-1025 | 9 | | RESISTOR 18.7K 1% .125W F TC#0+=100 | 24546 | C4-1/6-T0-1872-F | | | | |
| A9R51 | 0698-4421 | 6 | | RESISTOR 1K 5% .25W FC TC==400/+600 RESISTOR 249 1% .125W F TC=0+=100 | 01121 24546 | C81025 C4-1/8-T0-249R-F | | | | |
| 49852 49853 | 0684-1041 0757-0278 | 9 | | RESISTOR 100K 10% .25W FC TC==400/+800 RESISTOR 1.78K 1% .125W F TC=0+=100 | 01121 | C81041 C4-1/8-T0-1781-F | | | | |
| 49854 49855 | 0757-0407 | 6 | | RESISTOR 200 1% .125W F TC=0+=100 RESISTOR 3.92K .5% .125W F TC=0+=100 | 24546 | C4-1/8-T0-201-F | | | | |
| A 9R56 | 0698-3327 0698-4518 | 2 | 1 | RESISTOR 3.92K .5% .125W F TC=0+=100 RESISTOR 137K 1% .125W F TC=0+=100 | 03888 | PME55-1/8-T0-3921-D C4-1/8-T0-1373-F | | | | |
| 49857 49858 | 0698-4492 0698-4055 | 1 2 | 1 | RESISTOR 32.4K 1x .125W F TC=0+=100 RESISTOR 1K .25x .125W F TC=0+=100 | 24546 03888 | C4-1/8-T0-3242-F PME55-1/8-T0-1001-C | | | | |
| AgRSq | 0698-3497 | 4 | j | | 24546 | C4-1/8-T0-604R-F | | | | |
| 49861 49862 | 0698-4488 0698-7417 | 5 | 1 | RESISTOR 6.04K 1% .125W F TC=0+-100 RESISTOR 26.7K 1% .125W F TC=0+-100 RESISTOR 69.8K .25% .125W F TC=0+-100 | 24546 19701 | C4-1/8-T0-2672-F | | | | |
| A9R63 A9R64 | 0757-0407 0757-0442 | 6 9 | İ | RESISTOR 200 1% .125W F TC=0+-100 RESISTOR 10K 1% .125W F TC=0+-100 | 24546 | MF4C1/8-T0-6982-C C4-1/8-T0-201-F | | | | |
| 49R65 | 0757-0161 | 9 | | RESISTOR 604 1% .125W F TC=0+=100 | 24546 | C4-1/8-T0-1002-F | | | | |
| A9R66 A9R67 | 0698-4422 | 7 | 1 | RESISTUR 1.27K 1% .125W F TC#0+-100 | 24546 | C4-1/8-T0-604R-F C4-1/8-T0-1271-F | | | | |
| A 9R 6 B | 0757-0283 0757-0976 | 4 | 2 | RESISTOR 2K 1% ,125W F TC=0+-100 RESISTOR 150K 2% ,125W F TC=0+-100 | 24546 | C4-1/8-T0-2001-F C4-1/8-T0-1502-G | | | | |
| 49R69 | 0698-4202 | 1 | 2 | RESISTOR 8.87K 1% .125W F TC#0+-100 | 24546 | C4-1/6-T0-8871-F | | | | |
| A 9871 A 9872 | 0757-0438 0757-0283 | 6 | | RESISTOR 5.11K 1% .125W F TC=0+-100 RESISTOR 2K 1% .125W F TC=0+-100 | 24546 | C4-1/6-T0-5111-F C4-1/6-T0-2001-F | | | | |
| A 9R 7 3 A 9R 7 4 | 0698-4202 0757-0976 | 1 | l | RESISTOR 8.87K 1% .125W F TC#0+=100 RESISTOR 150K 2% .125W F TC#0+=100 | 24546 | C4-1/8-T0-8871-F C4-1/8-T0-1502-G | | | | |
| 19R75 | | 2 | | RESISTOR 30.1K 1% .125W F TC=0+=100 | 24546 | C4-1/6-T0-3012-F | | | | |
| A9R76 A9R77 | 0757-0438 0757-0438 | 3 | | RESISTOR 5.11K 1% .125W F TC=0+=100 RESISTOR 5.11K 1% .125W F TC=0+=100 | 24546 | C4-1/8-T0-5111-F | | | | |
| 49R78 49R79 | | 9 | | RESISTOR 1K 5% .25W FC TC=-400/+600 | 01121 | C4=1/8=T0=5111=F C81025 | | | | |
| A9R81 | | ž | 3 | RESISTOR 3.65K 1% .125W F TC=0+-100 RESISTOR 133 1% .125W F TC=0+-100 | 24546 | C4-1/6-T0-3651-F C4-1/6-T0-133R-F | | | | |
| 19R82 | | 5 | | RESISTOR 133 1% .125W F TC=0++100 RESISTOR 133 1% .125W F TC=0+-100 | 24546 | C4-1/8-T0-133R-F | | | | |
| 19884 | 0757-0404 | 3 | 4 | RESISTOR 130 1% .125W F TC=0+-100 | 24546 | C4-1/8-T0-133R-F C4-1/8-T0-131-F | | | | |
| 1985 1986 | | 3 | | RESISTOR 130 1x .125W F TC=0+-100 RESISTOR 130 1x .125W F TC=0+-100 | 24546 | C4-1/8-T0-131-F C4-1/8-T0-131-F | | | | |
| 19857 19888 | | 3 | , | RESISTOR 130 1% .125W F TC=0+-100 RESISTOR 383 1% .125W F TC=0+-100 | 24546 | C4-1/8-T0-131-F | | | | |
| 19889 19891 | 0757-0438 | 3 | . | RESISTOR 5.11K 1% .125W F TC=0+=100 | 24546 | C4-1/8-T0-383R-F C4-1/8-T0-5111-F | | | | |
| 19R92 | | 0 | | RESISTOR 604 1% .125W F TC=0+-100 RESISTOR 3.74K 1% .125W F TC=0+-100 | 24546 | C4-1/8-T0-604R-F C4-1/8-T0-3741-F | | | | |
| 19R93 | | 1 | 1 | RESISTOR 9.53K 1% .125W F TC=0+-100 | 24546 | C4-1/8-T0-9531-F | | | | |
| 19R95 | 0757-0161 | ; | 3 | RESISTOR 3,92K 1% .125W F TC#0+=100 RESISTOR 604 1% .125W F TC#0+=100 | 24546 | C4-1/8-T0-3921-F C4-1/8-T0-604R-F | | | | |
| 19896 | | 0 | | RESISTOR 3.92K 1% .125W F TC=0+-100 RESISTOR 1K 1% .125W F TC=0+-100 | 24546 | C4-1/8-T0-3921-F | | | | |
| | | | | HELLOW IN IA STEEL L. ICEO+#100 | 24546 | C4-1/8-T0-1001-F | | | | |
| | | | | 1 | | | | | | |
| | | | | | | | | | | |
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Table 6-3. Replaceable Parts (Cont'd).

| APRIL4 0684-1041 1 RESISTOR 100K 10X .25W FC TC=-400/+800 01121 CB1041 APRIL5 0684-1031 0699-3557 7 APRIL6 0699-3557 7 APRIL7 0683-2045 APRIL7 0684-1041 1 RESISTOR 806 1x .125W F TC=0+0100 RESISTOR 806 1x .25W FC TC=-800/+900 01121 CB1031 C | | i avie 0.3. kehiaceabie Parts (Cont d). | | | | | | | | |
|--|--|---|-------------|-----|---|-------------------------|--|--|--|--|
| ASR99 ASR90 ASR90 ASR90 ASR101 OSS96-4486 ASR102 O757-0271 2 ASR102 O757-0271 2 ASR102 O757-0271 2 ASR103 O757-0401 O757-0402 | | | | Qty | Description | | Mfr Part Number | | | |
| ANRIDO 0757-0435 0 0768-3158 0 1 RESISTOR 3,92K 1X .125W F TC00+-100 24546 C4-1/8-T0-2572-F C81025 | A9R99 A9R101 A9R102 | 0757-0280 0698-4486 0757-0271 | 3 | : | RESISTOR 1K 1% .125W F TC=0+-100 RESISTOR 24.9K 1% .125W F TC=0+-100 RESISTOR 124K 1% .125W F TC=0+-100 | 24546 24546 24546 | C4-1/8-T0-1001-F C4-1/8-T0-2492-F C4-1/8-T0-1243-F | | | |
| ARRI12 | A9R105 A9R106 A9R107 | 0757-0438 0757-0435 0698-3158 | 04 | 1 | RESISTOR 3.92K 1% .125W F TC=0+=100 RESISTOR 23.7K 1% .125W F TC=0+=100 | 24546 24546 24546 | C4-1/8-T0-5111-F C4-1/8-T0-3921-F C4-1/8-T0-2372-F | | | |
| ASPRILO ASPRIL | A9R111 A9R112 A9R113 | 0698-4441 0757-0413 0683-2045 | 0 4 5 | 1 | RESISTOR 392 1x .125W F TC=0+-100 RESISTOR 200K 5x .25W FC TC=-800/+900 | 24546 24546 01121 | C4-1/8-T0-3741-F C4-1/8-T0-392R-F C82045 | | | |
| A9R122 | A9R116 A9R117 A9R118 | 0698-3557 0683-2045 0684-1041 | 7 5 1 | | RESISTOR 806 1% .125W F TC#0+=100 RESISTOR 200K 5v .25W FC TC#=800/+900 | 24546 01121 01121 | C4-1/8-T0-806R-F C82045 C81041 | | | |
| A9R127 0/57-0402 3 0633-1025 9 RESISTOR 1% % 25W FC TC=-400/+600 01121 CB1025 C4-1/8-T0-7502-F C4-1/8-1025 C4-1/8-T0-7502-F C4-1/8-T0-7502-F C4-1/8-T0-7502-F C4-1/8-T0-7502-F C4-1/8-T0-7502-F C4-1/8-T0-7502-F C4-1/8-T0-7502-F C4-1/8-T0-5111-F C4-1/8-T0-5111-F C4-1/8-T0-5111-F C4-1/8-T0-5111-F C4-1/8-T0-5111-F C4-1/8-T0-5111-F C4-1/8-T0-5111-F C4-1/8-T0-7502-F C | A9R121 A9R122 A9R123 A9R124 A9R125 | 0684-1011 0684-4701 0683-1025 | 5 6 9 | 1 | RESISTOR 47 10% .25W FC TC=-400/+500 RESISTOR 1K 5% .25W FC TC=-400/+600 | 01121 | CB1011 CB4701 CB1025 | | | |
| A982 3100-2738 2 1 SWITCH-ROTARY 28480 3100-2738 A9981 03580-61905 5 1 SWITCH ASSEMBLY 28480 03580-61905 A901 1826-0044 5 1 IC OP AMP GP DUAL 14-DIP-C 07263 UA739DC | A9R126 A9R127 A9R128 A9R129 A9R131 | 0757-0462 0683-1025 0684-4701 | 3 9 6 | 2 | RESISTOR 75K 1% 125W F TC=0+-100 RESISTOR 1K 5% 25W FC TC=-400/+600 RESISTOR 47 10% 25W FC TC=-400/+500 | 24546 01121 01121 | C4-1/6-T0-7502-F CB1025 C84701 | | | |
| A9841 03580-61905 5 1 SWITCH A88EMBLY 28480 03580-61905 A9U1 1826-0044 5 1 IC OP AMP GP DUAL 14-DIP-C 07263 UA739DC | A9R132 A9R133 | | | | RESISTOR 470 5% .25W FC TC=-400/+600 RESISTOR 75K 1% .125W F TC=0+-100 | | | | | |
| A9U1 1826-0044 5 1 IC OP AMP GP DUAL 14-DIP-C 07263 UA739DC | A 9S2 | 3100-2738 | 2 | 1 | SWITCH-ROTARY | 28480 | 3100-2738 | | | |
| | A 98 # 1 | 03580-61905 | 5 | 1 | SWITCH ASSEMBLY | 28480 | 03580-61905 | | | |
| | | | | | | | | | | |
| | | | | | | | | | | |

Table 6-3. Replaceable Parts (Cont'd).

| Deference | Potential LID D. 4 | | | | | | | | | |
|---|---|-----------------------|------------------|--|---|--|--|--|--|--|
| Reference Designation | HP Part Number | D | Qty | Description | Mfr Code | Mfr Part Number | | | | |
| 49 | 03580-66519 | 7 | 1 | BOARD ASSEMBLY-INPUT (OPT. OOZ ONLY) | 28480 | 03580-66519 | | | | |
| A9C1 A9C2 A9C3 A9C4 A9C5 | 0170-0042 0121-0407 0121-0407 0121-0407 0121-0407 | 19999 | 1 5 | CAPACITOR=FXD .33UF +=5% 100VDC POLYE CAPACITOR=V TRMR=P8IN .7=3PF 600V CAPACITOR=V TRMR=P8IN .7=3PF 600V CAPACITOR=V TRMR=P8IN .7=3PF 600V CAPACITOR=V TRMR=P8IN .7=3PF 600V | 99515 72982 72982 72982 72982 | E1-334D 536-016 536-016 536-016 536-016 | | | | |
| A9C6 A9C7 A9C8 A9C9 A9C10 | 0121-0407 0150-0022 0140-0162 0150-0011 0160-2207 | 9 5 3 2 3 | 4 1 1 1 | CAPACITOR=V TRMR=PSTN _7=3PF 600V CAPACITOR=FXD 3_3PF +=10X 500VDC TI DIOX CAPACITOR=FXD 4700PF +=10X 300VDC MICA CAPACITOR=FXD 1.5PF +=20X 500VDC TI DIOX CAPACITOR=FXD 300PF +=5X 300VDC MICA | 72982 28480 72136 28480 28480 | 536-016 0150-0022 DM20F472K0300WV1CR 0150-0011 0160-2207 | | | | |
| A9C11 A9C13 A9C14 A9C15 A9C16 | 0150-0022 0160-0356 0150-0022 0150-022 | 5 9 5 5 7 | 1 | CAPACITOR-FXD 3.3PF +=10% 500VDC TI DIOX CAPACITOR-FXD 18PF +=5% 300VDC MICA CAPACITOR-FXD 3,3PF +=10% 500VDC TI DIOX CAPACITOR-FXD 3.3PF +=10% 500VDC TI DIOX CAPACITOR-FXD 33UF+=10% 10VDC TA | 28480 28480 28480 28480 56289 | 0150=0022 0160=0356 0150=0022 0150=0022 150D336×9010B2 | | | | |
| A9C17 A9C18 A9C19 A9C21 A9C22 | 0180-0229 0140-0210 0160-2198 0180-0060 0160-2204 | 7 2 1 4 0 | 2 1 2 | CAPACITOR=FXD 33UF+=10X 10VDC TA CAPACITOR=FXD 270PF +=5X 300VDC MICA CAPACITOR=FXD 20PF +=5X 300VDC MICA CAPACITOR=FXD 200UF+75=10X 3VDC AL CAPACITOR=FXD 100PF +=5X 300VDC MICA | 56289 72136 28480 56289 28480 | 150D336X901082 DM15F271J0300WV1CR 0160-2198 30D207G003CC2 0160-2204 | | | | |
| A9C23 A9C24 A9C25 A9C26 A9C27 | 0180-0197 0180-1758 0180-0061 0180-1758 0180-0210 | 89596 | 8 3 2 1 | CAPACITOR-FXD 2.2UF+-10x 20VDC TA CAPACITOR-FXD 300UF+75-10X 3VDC AL CAPACITOR-FXD 100UF+75-10X 16VDC AL CAPACITOR-FXD 300UF+75-10X 3VDC AL CAPACITOR-FXD 3.3UF+-20X 15VDC TA | 56289 56289 56289 56289 56289 | 150P225X9020A2 30D307G003DC2 30D107G015DC2 30D307G003DC2 150D335X0015A2 | | | | |
| A9C28 A9C29 A9C30 A9C31 A9C32 | 0140-0210 0180-0060 0160-2204 0160-0763 0180-1758 | 24029 | 1 | CAPACITOR-FXD 270PF +-5% 300VDC MICA CAPACITOR-FXD 200UF+75-10% 3VDC AL CAPACITOR-FXD 100PF +-5% 300VDC MICA CAPACITOR-FXD 5PF +-10% 500VDC MICA CAPACITOR-FXD 300UF+75-10% 3VDC AL | 72136 56289 28480 28480 56289 | DM15F271J0300WV1CR 30D207G003CC2 0160-2204 0160-0763 30D307G003DC2 | | | | |
| A9C33 A9C34 A9C35 A9C36 A9C37 | 0180-0061 0180-0137 0160-2724 0140-0217 0160-3269 | 5 6 9 9 | 1 1 2 | CAPACITOR-FXD 100UF+75-10% 16VDC AL CAPACITOR-FXD 100UF+-20% 10VDC TA CAPACITOR-FXD 3600PF +-2% 500VDC MICA CAPACITOR-FXD 140PF +-2% 300VDC MICA CAPACITOR-FXD 7610PF +-1% 100VDC MICA | 56289 56289 28480 72136 28480 | 30D107G016DC2 150D107X0010R2 0160-2724 DM15F141G0300WY1CR 0160-3269 | | | | |
| A9C38 A9C39 A9C41 A9C42 A9C43 | 0160-3269 0140-0233 | 29928 | 1 1 1 1 | CAPACITOR-FXD 640PF +-1X 300VDC MICA CAPACITOR-FXD 7610PF +-1X 100VDC MICA CAPACITOR-FXD 480PF +-1X 300VDC MICA CAPACITOR-FXD 3300PF +-5X 300VDC MICA CAPACITOR-FXD 100UF+75-10X 3VDC AL | 28480 28480 72136 28480 56289 | 0160-0341 0160-3269 DM15F481F0300WY1C 0160-2230 30D107G003C82 | | | | |
| A9C44 A9C45 A9C46 A9C47 A9C48 | 0150-0093 0180-0197 | 0 3 0 8 5 | 1 | CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD 10UF+-10% 20VDC TA CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD 2.2UF+-10% 20VDC TA CAPACITOR-FXD .02UF +80-20% 25VDC CER | 28480 56289 28480 56289 28480 | 0150-0093 150D106X9020B2 0150-0093 150D225X9020A2 0160-2605 | | | | |
| A9C49 A9C51 A9C52 A9C53 A9C54 | 0160-2035 0180-0197 0150-0093 | 0 5 8 0 8 | 1 | CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD 750PF +-5% 300VDC MICA CAPACITOR-FXD 2.2UF+-10% 20VDC TA CAPACITOR-FXD 2.2UF+-10% 20VDC CER CAPACITOR-FXD 2.2UF+-10% 20VDC TA | 28480 28480 56289 28480 56289 | 0150-0093 0160-2035 150D225x9020A2 0150-0093 150D225x9020A2 | | | | |
| A9055 A9056 A9057 A9058 A9059 | 0180-0197 0150-0093 0150-0093 | 3 8 0 0 8 | 1 | CAPACITOR-FXD 820PF +-5% 300VDC MICA CAPACITOR-FXD 2.2UF+=10% 20VDC TA CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD .01UF +80-20% 100VDC CER CAPACITOR-FXD 2.2UF+=10% 20VDC TA | 28480 56289 28480 28480 56289 | 0160-2009 1500225X9020A2 0150-0093 0150-0093 1500225X9020A2 | | | | |
| A9C61 A9C62 A9C63 A9C64 A9C65 | 0180-0197 0180-0339 0180-0197 | 6 8 0 8 6 | 5 | CAPACITOR-FXD 22UF+=10% 15VDC TA CAPACITOR-FXD 2.2UF+=10% 20VDC TA CAPACITOR-FXD 50UF475=10% 16VDC AL CAPACITOR-FXD 2.2UF+=10% 20VDC TA CAPACITOR-FXD 22UF+=10% 15VDC TA | 56289 56289 56289 56289 56289 | 150D226x901582 150D225x902DA2 30D506016682 150D225x902OA2 150D226x901582 | | | | |
| A9C66 | 0180-0339 | 0 | | CAPACITOR=FXD SOUF+75=10X 16VDC AL | 56289 | 30D506G016C82 | | | | |
| A9CR1 A9CR2 A9CR3 A9CR4 A9CR5 | 1901-0040 1901-0040 1901-0040 | 1 1 1 1 1 1 1 1 | 6 | DIODE-SWITCHING 30V 50MA 2NS DO-35 | 28480 28480 28480 28480 28480 | 1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 | | | | |
| A9CR7 | | | | DIODE-SWITCHING SOV SOMA 2NS DO-35 | 28480 | 1901-0040 | | | | |
| A9J1 | 1251-2969 | 8 | 1 | CONNECTOR-PHONO SINGLE PHONO JACK; DIP | 28480 | 1251-2969 | | | | |
| | | | | | | | | | | |
| | | | | | | | | | | |

Table 6-3. Replaceable Parts (Cont'd).

| | Table 0-3. Replaceable Parts (Cont a). | | | | | | | | |
|--|---|------------------------|---|--|---|---|--|--|--|
| Reference Designation | HP Part Number | C D | Qty | Description | Mfr Code | Mfr Part Number | | | |
| A9L1 A9L2 A9L3 A9L4 A9L5 | 9100-3264 9100-3259 9100-3260 9100-3277 9170-0894 | 7 0 3 2 0 | 1 1 1 1 | INDUCTORRF=CH-MLD 2,34MH 2% .8D Q=200 INDUCTORRF=CH-MLD 2,46MH 10% Q=200 INDUCTORRF=CH-MLD 2,15H 2% .8D Q=200 INDUCTORRF=CH-MLD 3,1MH 2% .62D Q=150 CORE-SHIELDING BEAD | 28480 28480 28480 28480 28480 | 9100-3264 9100-3259 9100-3260 9100-3277 9170-0894 | | | |
| A901 A902 A903 A904 A905 | 1855-0377 1854-0226 1853-0086 1854-0071 1854-0071 | 8 4 2 7 7 | 1 3 2 10 | TRANSISTOR J=FET N=CHAN D=MODE TO=18 SI TRANSISTOR NPN 2N4384 SI TO=18 PD=500MW TRANSISTOR PNP SI PD=310MW FT=40MHZ TRANSISTOR NPN SI PD=300MW FT=200MHZ TRANSISTOR NPN SI PD=300MW FT=200MHZ | 28480 13606 27014 28480 28480 | 1855-0377 2N4384 2N5087 1854-0071 1854-0071 | | | |
| A906 A907 A908 A909 A9011 | 1854-0226 1853-0086 1854-0071 1854-0071 1854-0071 | 4 2 7 7 7 | | TRANSISTOR NPN 2N4384 SI TO-18 PD=500MW TRANSISTOR PNP SI PD=310MW FT=40MMZ TRANSISTOR NPN SI PD=300MW FT=200MHZ TRANSISTOR NPN SI PD=300MW FT=200MHZ TRANSISTOR NPN SI PD=300MW FT=200MHZ | 13606 27014 28480 28480 28480 | 2N4384 2N5087 1854-0071 1854-0071 1854-0071 | | | |
| A9012 A9013 A9014 A9015 A9016 | 1854-0071 1854-0071 1854-0226 1853-0010 1854-0071 | 7 7 4 2 7 | 2 | TRANSISTOR NPN SI PD=300MW FT=200MHZ TRANSISTOR NPN SI PD=300MW FT=200MMZ TRANSISTOR NPN 2N4364 SI TO-18 PD=500MW TRANSISTOR PNP SI TO-18 PD=360MW TRANSISTOR NPN SI PD=300MW FT=200MHZ | 28480 28480 13606 28480 28480 | 1854-0071 1854-0071 2N4384 1853-0010 1854-0071 | | | |
| A9017 A9018 A9019 | 1854-0071 1854-0071 1853-0010 | 7 7 2 | | TRÂNSISTOR NPN SI PD=300MW FT=200MHZ TRÂNSISTOR NPN SI PD=300MW FT=200MHZ TRÂNSISTOR PNP SI TO=18 PD=360MW | 28480 28480 28480 | 1854-0071 1854-0071 1853-0010 | | | |
| A9R1 A9R2 A9R4 A9R5 A9R6 | 2100-0580 2100-0640 0698-4055 0698-5132 0757-0271 | 7 0 2 8 2 | 1 2 2 2 | RESISTOR-TRMR 500K 10% C TOP-ADJ 1-TRN RESISTOR-VAR W/8W 5K 10% LIN SPST-NO RESISTOR 1K 25% 125W F TC=0+-100 RESISTOR 990K 5% 25W F TC=0+-100 RESISTOR 124K 1% 125W F TC=0+-100 | 28480 28480 03888 28480 24546 | 2100-0580 2100-0640 PME55-1/8-T0-1001-C 0698-5132 C4-1/8-T0-1243-F | | | |
| A9R7 A9R8 A9R9 A9R10 A9R11 | 0698-6661 0698-5132 0698-5131 0698-3359 0698-6659 | 0 8 7 7 6 | 1 2 1 1 | RESISTOR 11.11K .25% .125W F TC=0+-100 RESISTOR 990K .5% .25W F TC=0+-100 RESISTOR 900K .5% .25W F TC=0+-100 RESISTOR 12.7K 1% .125W F TC=0+-100 RESISTOR 12.7K 1% .125W F TC=0+-100 | 28480 26480 19701 24546 28480 | 0698-6661 0698-5132 MF52C1/4-T0-9003-D C4-1/8-T0-1272-F 0698-6659 | | | |
| A9R12 A9R13 A9R14 A9R16 A9R17 | 0698-5131 0757-0430 0698-4437 0757-0824 0684-1041 | 7 5 4 1 | 1 1 1 6 | RESISTOR 900K .5% .25W F TC=0+-100 RESISTOR 2.21K 1% .125W F TC=0+-100 RESISTOR 2.94K 1% .125W F TC=0+-100 RESISTOR 2K 1% .5W F TC=0+-100 RESISTOR 100K 10% .25W FC TC=-400/+800 | 19701 24546 24546 28480 01121 | MF52C1/4-T0-9003-D C4-1/8-T0-2211-F C4-1/8-T0-2941-F 0757-0824 C81041 | | | |
| A9R18 A9R19 A9R21 A9R22 A9R23 | 0698-3581 0698-3581 0698-4473 0757-0442 0698-4421 | 7 7 8 9 6 | 2 5 2 | RESISTOR 13.7K 1% .125W F TC=0+-100 RESISTOR 13.7K 1% .125W F TC=0+-100 RESISTOR 8.06K 1% .125W F TC=0+-100 RESISTOR 10K 1% .125W F TC=0+-100 RESISTOR 249 1% .125W F TC=0+-100 | 24546 24546 24546 24546 24546 | C4-1/8-T0-1372-F C4-1/8-T0-1372-F C4-1/8-T0-8061-F C4-1/8-T0-1002-F C4-1/8-T0-249R-F | | | |
| A9R24 A9R25 A9R26 A9R27 A9R2R | 0698-3193 0698-6862 0698-4486 0698-3382 0757-0407 | 7 3 3 6 6 | 1 3 2 5 | RESISTOR 10K .25% .125W F TC=0+-50 RESISTOR 1,153K .25% .125W F TC=0+-50 RESISTOR 24.9K 1% .125W F TC=0+-100 RESISTOR 5,49K 1% .125W F TC=0+-100 RESISTOR 200 1% .125W F TC=0+-100 | 28480 28480 24546 24546 24546 | 0698-3193 0698-6862 C4-1/8-T0-2492-F C4-1/8-T0-5491-F C4-1/8-T0-201-F | | | |
| A9R29 A9R31 A9R32 A9R33 A9R34 | 0698-4464 0684-1041 0757-0448 0684-4701 0757-0407 | 7 1 5 6 | 1 1 3 | RESISTOR 887 1% .125W F TC=0+-100 RESISTOR 100K 10% .25W FC TC=-400/+800 RESISTOR 18.2K 1% .125W FC TC=-400/-500 RESISTOR 47 10% .25W FC TC=-400/-500 RESISTOR 200 1% .125W F TC=0+-100 | 24546 01121 24546 01121 24546 | C4-1/8-T0-887R-F C81041 C4-1/8-T0-1822-F C84701 C4-1/8-T0-201-F | | | |
| A9R35 A9R36 A9R37 A9R38 A9R39 | 0698-3488 0684-1041 0757-0442 0757-0278 0698-6780 | 3 1 9 9 | 1 2 1 | RESISTOR 442 1% .125W F TC=0+-100 RESISTOR 100K 10% .25W FC TC=-400/+800 RESISTOR 10K 1% .125W F TC=0+-100 RESISTOR 1.78K 1% .125W F TC=0+-100 RESISTOR 5.62K .25% .125W F TC=0+-50 | 24546 01121 24546 24546 28480 | C4-1/8-T0-422R-F C81041 C4-1/8-T0-1002-F C4-1/8-T0-1781-F 0698-6780 | | | |
| A9R40 A9R41 A9R42 A9R43 A9R44 | 0684-1041 0698-6823 0698-4473 0698-3495 0757-0424 | 1 6 8 2 7 | 1 1 1 | RESISTOR 100K 10% .25W FC TC=-400/+800 RESISTOR 2.61K .25% .125W F TC=0+-100 RESISTOR 8.06K 1% .125W F TC=0+-100 RESISTOR 8.06K 1% .125W F TC=0+-100 RESISTOR 1.1K 1% .125W F TC=0+-100 | 01121 19701 24546 24546 24546 | C81041 MF4C1/8=T0=2611=C C4=1/8=T0=8061=F C4=1/8=T0=808N=F C4=1/8=T0=1101=F | | | |
| A 9R 45 A 9R 46 A 9R 47 A 9R 48 A 9R 49 | 0757-0442 0757-0442 0698-3382 0757-0407 0698-4483 | 9 6 6 0 | 1 | RESISTOR 10K 1% .125W F TC=0+-100 RESISTOR 10K 1% ,125W F TC=00+-100 RESISTOR 5.49K 1% .125W F TC=0+-100 RESISTOR 200 1% ,125W F TC=0+-100 RESISTOR 18,7K 1% ,125W F TC=0+-100 | 24546 24546 24546 24546 24546 | C4=1/8=T0=1002=F C4=1/8=T0=1002=F C4=1/8=T0=3491=F C4=1/8=T0=201=F C4=1/8=T0=201=F | | | |
| A9R50 A9R51 A9R52 A9R53 A9R54 | 0683-1025 0698-4421 0684-1041 0757-0278 0757-0407 | 9 6 1 9 6 | 5 | RESISTOR 1K 5% .25M FC TC==400/+600 RESISTOR 249 1% .125M F TC=04-100 RESISTOR 100K 10% .25M FC TC==400/+800 RESISTOR 1.76K 1% .125M F TC=04-100 RESISTOR 200 1% .125M F TC=04-100 | 01121 24546 01121 24546 24546 | C81025 C4-1/8-T0-249R-F C81041 C4-1/8-T0-1781-F C4-1/8-T0-201-F | | | |
| A9R55 A9R56 A9R57 A9R58 A9R58 | 0698-3327 0698-4518 0698-4492 0698-4055 0698-3497 | 9 2 1 2 4 | 1 1 1 | RESISTOR 3.92K .5% .125W F TC=0+-100 RESISTOR 137K 1% .125W F TC=0+-100 RESISTOR 32.4K 1% .125W F TC=0+-100 RESISTOR 1K .25% .125W F TC=0+-100 RESISTOR 6.04K 1% .125W F TC=0+-100 | 03888 24546 24546 03888 24546 | PME55-1/8-T0-3921-D C4-1/8-T0-1373-F C4-1/8-T0-3242-F PME55-1/8-T0-1001-C C4-1/8-T0-04R-F | | | |
| 49R444 49R444 49R444 49R444 49R444 49R446 49R446 49R446 49R847 49R851 49R853 49R853 49R853 49R853 49R858 | 0684-1041 0698-6823 0698-4473 0698-3495 0757-0424 0757-0442 0757-0407 0698-3382 0757-0407 0698-4483 0683-1025 0698-4421 0684-1041 0684-1041 0684-1047 0698-3327 0698-3327 | 16827 99660 96196 9212 | 1 1 1 5 5 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 | RESISTOR 100K 10% .25% FC TC=-400/+800 RESISTOR 2,61K .25% .125% F TC=0+-100 RESISTOR 8,06K 1% .125% F TC=0+-100 RESISTOR 866 1% .125% F TC=0+-100 RESISTOR 1.1K 1% .125% F TC=0+-100 RESISTOR 1.1K 1% .125% F TC=0+-100 RESISTOR 10K 1% .125% F TC=0+-100 RESISTOR 10K 1% .125% F TC=0+-100 RESISTOR 200 1% .125% F TC=0+-100 RESISTOR 200 1% .125% F TC=0+-100 RESISTOR 18.7K 1% .125% F TC=0+-100 RESISTOR 10K 10% .25% F TC=0+-100 RESISTOR 10K 10% .25% F TC=0+-100 RESISTOR 10K 10% .25% F TC=0+-100 RESISTOR 10K 10% .125% F TC=0+-100 RESISTOR 3.92K .5% .125% F TC=0+-100 RESISTOR 3.94K 1% .125% F TC=0+-100 RESISTOR 3.94K 1% .125% F TC=0+-100 RESISTOR 3.4K 1% .125% F TC=0+-100 RESISTOR 15.4K 1% .125% F TC=0+-100 RESISTOR 15.4K 1% .125% F TC=0+-100 RESISTOR 15.4K 1% .125% F TC=0+-100 | 0177446 17744 17 | C81041 MF4C1/8-T0-2411-C C4-1/8-T0-8041-F C4-1/8-T0-8068-F C4-1/8-T0-101-F C4-1/8-T0-101-F C4-1/8-T0-102-F C4-1/8-T0-102-F C4-1/8-T0-201-F C4-1/8-T0-201-F C4-1/8-T0-201-F C4-1/8-T0-201-F C4-1/8-T0-201-F PME55-1/8-T0-3921-D C4-1/8-T0-3242-F PME55-1/8-T0-3242-F PME55-1/8-T0-3242-F | | | |

Table 6-3. Replaceable Parts (Cont'd).

| | | | | adie 0-3. Kepiaceadie Parts (Cont | u/. | |
|--|---|-----------------------|-----------------------|---|---|--|
| Reference Designation | HP Part Number | C D | Qty | Description | Mfr Code | Mfr Part Number |
| A9R61 A9R62 A9R63 A9R64 A9R65 | 0698-4488 0698-7417 0757-0407 0757-0442 0757-0161 | 56699 | 1 1 | RESISTOR 26.7K 1% .125W F TC=0+-100 RESISTOR 69.8K .25% .125W F TC=0+-100 RESISTOR 200 1% .125W F TC=0+-100 RESISTOR 10K 1% .125W F TC=0+-100 RESISTOR 604 1% .125W F TC=0+-100 | 24546 19701 24546 24546 24546 | C4-1/8-T0-2672-F MF4C1/8-T0-6982-C C4-1/8-T0-201-F C4-1/8-T0-1002-F C4-1/8-T0-604R-F |
| A9R66 A9R67 A9R68 A9R71 | 0698-4422 0698-4202 0757-0283 0757-0976 0757-0438 | 7 1 6 4 3 | 1 2 2 2 7 | RESISTOR 1.27K 1% .125W F TC=0+-100 RESISTOR 8.87K 1% .125W F TC=0+-100 RESISTOR 2K 1% .125W F TC=0+-100 RESISTOR 150K 2% .125W F TC=0+-100 RESISTOR 5.11K 1% .125W F TC=0+-100 | 24546 24546 24546 24546 24546 | C4-1/8-T0-1271-F C4-1/8-T0-8871-F C4-1/8-T0-801-F C4-1/8-T0-1502-G C4-1/8-T0-5111-F |
| A9R72 A9R73 A9R74 A9R75 A9R76 | 0757-0283 0698-4202 0757-0976 0757-0453 0757-0438 | 6 1 4 2 3 | 1 | RESISTOR 2K 1% .125W F TC=0+-100 RESISTOR 8.87K 1% .125W F TC=0+-100 RESISTOR 150K 2% .125W F TC=0+-100 RESISTOR 30.1K 1% .125W F TC=0+-100 RESISTOR 5.11K 1% .125W F TC=0+-100 | 24546 24546 24546 24546 24546 | C4-1/8-T0-2001-F C4-1/8-T0-8871-F C4-1/8-T0-1502-G C4-1/8-T0-5012-F C4-1/8-T0-5111-F |
| A9R77 A9R78 A9R79 A9R81 A9R82 | 0757-0438 0683-1025 0757-0434 0698-3437 0698-3437 | 3 9 9 2 2 | 1 3 | RESISTOR 5,11K 1% 125W F TC=0+-100 RESISTOR 1K 5% ,25W FC TC=-400/+600 RESISTOR 3,65K 1% ,125W F TC=0+-100 RESISTOR 133 1% ,125W F TC=0+-100 RESISTOR 133 1% ,125W F TC=0++100 | 24546 01121 24546 24546 24546 | C4-1/8-T0-5111-F C81025 C4-1/8-T0-3651-F C4-1/8-T0-133R-F C4-1/8-T0-133R-F |
| A9R83 A9R84 A9R85 A9R86 A9R87 | 0698-3437 0757-0404 0757-0404 0757-0404 0757-0404 | 2 3 3 3 3 | 4 | RESISTOR 133 1% .125W F TC=0+-100 RESISTOR 130 1% .125W F TC=0+-100 | 24546 24546 24546 24546 24546 | C4-1/8-T0-133R-F C4-1/8-T0-131-F C4-1/8-T0-131-F C4-1/8-T0-131-F C4-1/8-T0-131-F |
| A9R88 A9R89 A9R91 A9R92 A9R93 | 0698-3446 0757-0438 0757-0161 0698-4441 0698-4020 | 3 9 0 | 1 2 1 | RESISTOR 363 1% .125W F TC=0+-100 RESISTOR 5.11K 1% .125W F TC=0+-100 RESISTOR 604 1% .125W F TC=0+-100 RESISTOR 3.74K 1% .125W F TC=0+-100 RESISTOR 9.53K 1% .125W F TC=0+-100 | 24546 24546 24546 24546 24546 | C4-1/8-T0-383R-F C4-1/8-T0-5111-F C4-1/8-T0-004R-F C4-1/8-T0-3741-F C4-1/8-T0-9531-F |
| A9R94 A9R95 A9R96 A9R97 A9R98 | 0757-0435 0757-0161 0757-0435 0757-0280 0698-4486 | 0 9 0 3 3 | 3 | RESISTOR 3.92K 1% .125W F TC=0+-100 RESISTOR 604 1% .125W F TC=0+-100 RESISTOR 3.92K 1% .125W F TC=0+-100 RESISTOR 1K 1% .125W F TC=0+-100 RESISTOR 24.9K 1% .125W F TC=0+-100 | 24546 24546 24546 24546 24546 | C4-1/8-T0-3921-F C4-1/8-T0-604R-F C4-1/8-T0-3921-F C4-1/8-T0-1001-F C4-1/8-T0-2492-F |
| A9R99 A9R101 A9R102 A9R103 A9R104 | 0757-0280 0698-4486 0757-0271 0757-0161 0757-0401 | 3 3 2 9 0 | 1 | RESISTOR 1K 1% 125W F TC=0+=100 RESISTOR 24.9K 1% 125W F TC=0+=100 RESISTOR 124K 1% 125W F TC=0+=100 RESISTOR 604 1% 125W F TC=0+=100 RESISTOR 100 1% 125W F TC=0+=100 | 24546 24546 24546 24546 24546 | C4-1/8-T0-1001-F C4-1/8-T0-2492-F C4-1/8-T0-1243-F C4-1/8-T0-804R-P C4-1/8-T0-801-F |
| A9R105 A9R106 A9R107 A9R108 A9R109 | 0757-0438 0757-0435 0698-3158 0683-1025 0757-0422 | 3 0 4 9 5 | 1 | RESISTOR 5.11K 1% .125W F TC=0+=100 RESISTOR 3.92K 1% .125W F TC=0+=100 RESISTOR 23.7K 1% .125W F TC=0+=100 RESISTOR 1K 5% .25W FC TC==400/+600 RESISTOR 909 1% .125W F TC=0+=100 | 24546 24546 24546 01121 24546 | C4-1/8-T0-5111-F C4-1/8-T0-3921-F C4-1/8-T0-2372-F C8-1/8-T0-909R-F |
| A9R111 A9R112 A9R113 A9R114 A9R115 | 0698-4441 0757-0413 0683-2045 0684-1041 0684-1031 | 0 4 5 1 9 | 1 1 | RESISTOR 3.74K 1% .125W F TC=0+=100 RESISTOR 392 1% .125W F TC=0+=100 RESISTOR 200K 5% .25W FC TC==800/+900 RESISTOR 100K 10% .25W FC TC==400/+800 RESISTOR 10K .25W FC TC=400/+700 | 24546 24546 01121 01121 | C4-1/8-T0-3741-F C4-1/8-T0-392R-F C82045 C81041 C81031 |
| A9R116 A9R117 A9R118 A9R119 A9R121 | 0698-3557 0683-2045 0684-1041 0684-1031 0698-3153 | 7 5 1 9 | 1 | RESISTOR 806 1x .125W F TC=0+-100 RESISTOR 200K 5x .25W FC TC=-800/+900 RESISTOR 100K 10x .25W FC TC=-400/+900 RESISTOR 10K 10x .25W FC TC=-400/+700 RESISTOR 3.85K 1x .125W F TC=0+-100 | 24546 01121 01121 01121 24546 | C4-1/8-T0-806R-F C82045 C81041 C81031 C4-1/8-T0-3831-F |
| A9R122 A9R123 A9R124 A9R125 A9R126 | 0684-1011 0684-4701 0683-1025 0757-0438 0683-4715 | 5 6 9 3 0 | 1 2 | RESISTOR 100 10% .25W FC TC==400/+500 RESISTOR 47 10% .25W FC TC==400/+500 RESISTOR 1K 5% .25W FC TC==400/+600 RESISTOR 5.11K 1% .125W F TC=0+-100 RESISTOR 470 5% .25W FC TC==400/+600 | 01121 01121 01121 24546 01121 | CB1011 CB4701 CB1025 C4-1/8-T0-5111-F CB4715 |
| A9R127 A9R128 A9R129 A9R131 A9R132 | 0757-0438 | 3 9 6 3 0 | 2 | RESISTOR 75K 1% .125W F TC=0+=100 RESISTOR 1K 5% .25W FC TC==400/+600 RESISTOR 47 10% .25W FC TC==400/+500 RESISTOR 5.11K 1% .125W F TC=0+=100 RESISTOR 470 5% .25W FC TC==400/+600 | 24546 01121 01121 24546 01121 | C4-1/8-T0-7502-F C81025 C84701 C4-1/8-T0-5111-F C84715 |
| A9R133 | 0757-0462 | 3 | | RESISTOR 75K 1% .125W F TC=0+=100 | 24546 | C4=1/8-T0=7502=F |
| A982 | 3100-2738 | 2 | 1 | SWITCH-ROTARY 1.031 STRUT CTR SPCG; 8 | 28480 | 3100-2738 |
| A98W1 | 03580-61905 | 5 | 1 | SWITCH ASSEMBLY | 28480 | 03560-61905 |
| A 9 U 2 | | 5 | 1 | IC OP AMP GP DUAL 14-DIP-C IC MODULATOR TO-100 | 07263 04713 | UA739DC MC1496G |
| | | | | | | |

Table 6-3. Replaceable Parts (Cont'd).

| Reference Designation | HP Part Number | C | Qty | able 6-3. Replaceable Parts (Cont' Description | Mfr | Mfr Part Number |
|--|---|-----------------------|--------------|---|---|--|
| Designation | Number | Н | | | Code | |
| 41141 | 03580-66531 | 3 | 1 | BOARD ASSEMBLY-HIGH VOLTAGE POWER SUPPLY | 28480 | 03560-66531 |
| 41141C1 41141C2 41141C3 41141C4 41141C5 | 0150-0050 0150-0012 0150-0012 0160-3008 0160-3008 | 9 3 3 4 4 | 1 3 | CAPACITOR-FXD 1000PF +80-20% 1KVDC CER CAPACITOR-FXD .01UF +-20% 1KVDC CER CAPACITOR-FXD .01UF +-20% 1KVDC CER CAPACITOR-FXD 4700PF +-20% 4KVDC CER CAPACITOR-FXD 4700PF +-20% 4KVDC CER | 28480 56289 56289 28480 28480 | 0150=0050 C023A102J103M938 C023A102J103M938 0160=3008 0160=3008 |
| A1141C6 | 0160-3008 | 4 | | CAPACITOR-FXD 4700PF +-20% 4KVDC CER | 28480 | 0160-3008 |
| A11A1CR1 A11A1CR2 A11A1CR3 A11A1CR4 A11A1CR5 | 1901-0033 1901-0033 1901-0033 1901-0033 1901-0341 | 5 5 5 5 | 7 | DIODE-GEN PRP 180V 200MA DO-7 DIODE-GEN PRP 180V 200MA DO-7 DIODE-GEN PRP 180V 200MA DO-7 DIODE-GEN PRP 180V 200MA DO-7 DIODE-GEN PRP 180V 200MA DO-7 DIODE-HV RECT 7.5KV 10MA 250NS | 28480 28480 28480 28480 28480 | 1901=0033 1901=0033 1901=0033 1901=0033 1901=0341 |
| A11A1R1 A11A1R2 A11A1R4 A11A1R5 | 0757-0465 2100-3358 0683-1535 0683-3345 | 6 3 6 0 | 14 1 1 | RESISTOR 100K 1% .125W F TC=0+=100 RESISTOR=TRWR 1W 20% C SIDE=ADJ 1=TRN RESISTOR 15K 5% .25W FC TC==400/+800 RESISTOR 330K 5% .25W FC TC==800/+900 | 24546 28480 01121 01121 | C4-1/8-T0-1003-F 2100-3358 C81535 C83345 |
| 41145 | 03580-66532 | 4 | 1 | BOARD ASSEMBLY-HIGH VOLTAGE POWER SUPPLY (DOES NOT INCLUDE ATTAZTI) | 28480 | 03580-66532 |
| A11A2C1 A11A2C2 A11A2C3 A11A2C4 A11A2C5 | 0160-3859 0160-3859 0150-0012 0160-3007 0160-2544 | 3 3 3 1 | 2 1 1 | CAPACITOR-FXD 560PF +-20% 6KVDC CER CAPACITOR-FXD 560PF +-20% 6KVDC CER CAPACITOR-FXD .01UF +-20% 1KVDC CER CAPACITOR-FXD 4700PF20% 4KVDC CER CAPACITOR-FXD 270PF +-10% 1KVDC CER | 28480 28480 56289 28480 28480 | 0160-3859 0160-3859 C02341023103M838 0160-3007 0160-2544 |
| A1142CR1 A1142CR2 A1142CR3 A1142CR4 A1142CR5 | 1902-3237 1902-3428 1902-3428 1901-0033 1901-0033 | 6 7 7 2 2 | 2 | DIODE-ZNR 20V 5% DO-35 PDE,44 TC=+,073% DIODE-ZNR 100V 5% DO-7 PDE,44 TC=+,083% DIODE-ZNR 100V 5% DO-7 PDE,44 TC=+,083% DIODE-GEN PRP 180V 200MA DO-7 DIODE-GEN PRP 180V 200MA DO-7 | 28480 28480 28480 28480 | 1902-3237 1902-3428 1902-3428 1901-0033 |
| A1145CR6 | 1901-0033 | s | | DIODE-GEN PRP 180V 200MA DO-7 | 28480 | 1901-0033 |
| 41142K1 41142R2 41142R3 41142K4 41142K4 | 0683-4725 0683-1065 0683-1055 0683-4725 0687-2751 | 2 7 5 2 8 | 2 1 1 | RESISTOR 4.7k 5% .25m FC TC==400/+700 RESISTOR 10M 5% .25m FC TC==900/+1100 RESISTOR 1M 5% .25m FC TC==800/+900 RESISTOR 4.7k 5% .25m FC TC==400/+700 RESISTOR 2.7M 10% .5m CC TC=0+1000 | 01121 01121 01121 01121 01121 | C84725 C81085 C81055 C84725 E82751 |
| A11A2R6 A11A2T1 | 0698-8427 9100-3440 | 0 | 1 | RESISTOR 29M 10% 1W C TC=0+-250 HV TRANSFORMER | 28480 | 0698-8427 |
| 413 | 03580-66513 | 1 | 1 | BOARD ASSEMBLY-REFLECTION | 28480 | 03580-66513 |
| A13C1 A13C2 A13C3 | 0160-0168 0180-0291 0180-0291 | 1 3 3 | 5 1 | CAPACITOR-FXD .1UF +-10% 200VDC POLYE CAPACITOR-FXD 1UF+-10% 35VDC TA CAPACITOR-FXD 1UF+-10% 35VDC TA | 28480 56289 56289 | 0160-0168 150D105x9035A2 150D105x9035A2 |
| A13CR1 A13CR2 A13CR3 A13CR4 | 1901-0040 1901-0040 1901-0040 1901-0040 | 1 1 1 1 | 12 | DIODE-SHITCHING 30V 50MA 2NS DO-35 DIODE-SHITCHING 30V 50MA 2NS DO-35 DIODE-SHITCHING 30V 50MA 2NS DO-35 DIODE-SHITCHING 30V 50MA 2NS DO-35 | 28480 28480 28480 28480 | 1901=0040 1901=0040 1901=0040 1901=0040 |
| A1301 A1302 A1303 A1304 A1305 | 1854-0474 1854-0474 1854-0474 1854-0474 1854-0071 | 4 4 7 | 8 5 | TRANSISTOR NPN SI PD=310MW FT=100MHZ TRANSISTOR NPN SI PD=300MW FT=200MHZ | 04713 04713 04713 04713 28480 | 2N5551 2N5551 2N5551 2N5551 1854-0071 |
| A1336 A1337 A1338 A1339 A13311 | 1854-0071 1854-0474 1854-0474 1854-0474 1854-0474 | 7 4 4 4 4 4 | | TRÂNSISTOR NPN SI PD#300MM FT#200MHZ TRÂNSISTOR NPN SI PD#310MM FT#100MHZ | 26480 04713 04713 04713 04713 | 1854-0071 285551 285551 285551 285551 |
| 413912 413913 | 1854-0071 1854-0071 | 7 | | TRANSISTOR NPN SI PD=300MW FT=200MHZ Transistor npn si PD=300MW FT=200MHZ | 28480 28480 | 1854-0071 1854-0071 |
| A1391 A1382 A1383 A1384 A1385 | 2100-0558 2100-3252 2100-3253 2100-3252 2100-0558 | 9 6 7 6 9 | 2 2 1 | RESISTOR-TRMR 26K 10% C TOP-ADJ 1-TRN RESISTOR-TRMR 5K 10% C TOP-ADJ 1-TRN RESISTOR-TRMR 50K 10% C TOP-ADJ 1-TRN RESISTOR-TRMR 5K 10% C TOP-ADJ 1-TRN RESISTOR-TRMR 20K 10% C TOP-ADJ 1-TRN | 28480 28480 28480 28480 28480 | 2100-0558 2100-3252 2100-3253 2100-3253 2100-0558 |
| A13R6 A13R7 A13R8 A13R8 A13R9 | 0757-0469 0757-0469 0757-0465 0757-0440 0757-0442 | 0 0 6 7 9 | 8 5 14 | RESISTOR 150K 1% .125W F TC=0+-100 RESISTOR 150K 1% .125W F TC=0+-100 RESISTOR 100K 1% .125W F TC=0+-100 RESISTOR 7.5K 1% .125W F TC=0+-100 RESISTOR 10K 1% .125W F TC=0+-100 | 24546 24546 24546 24546 | C4=1/8-T0=1503=F C4=1/8-T0=1503=F C4=1/8-T0=1003=F C4=1/8-T0=7501=F C4=1/8-T0=1502=F |
| A13R12 A13A13 A13R14 A13R15 A13R16 | 0757-0440 0757-0469 0757-0430 0757-0429 0757-0469 | 7 0 5 2 0 | 1 1 | RESISTOR 7.5K 1x .125W F TC=0+-100 RESISTOR 150K 1x .125W F TC=0+-100 RESISTOR 2.21K 1x .125W F TC=0+-100 RESISTOR 1.82K 1x .125W F TC=0+-100 RESISTOR 1.50K 1x .125W F TC=0+-100 | 24546 24546 24546 24546 | C4=1/8-T0=7501=F C4=1/8-T0=1503=F C4=1/8-T0=2211=F C4=1/8-T0=1821=F C4=1/8-T0=1503=F |

Table 6-3. Replaceable Parts (Cont'd).

| Reference Designation | HP Part Number | C | Qty | Description | Mfr Code | Mfr Part Number |
|---|---|-----------------------|-----------------------|--|--|--|
| A13917 A13918 A13919 A13921 A13922 | 0698-4481 0698-4435 0698-4435 0757-0469 0757-0469 | 8 2 2 0 0 | 2 | RESISTOR 16.5K 1% .125W F TC=0+=100 RESISTOR 2.49K 1% .125W F TC=0+=100 RESISTOR 2.49K 1% .125W F TC=0+=100 RESISTOR 150K 1% .125W F TC=0+=100 RESISTOR 150K 1% .125W F TC=0+=100 | 24546 24546 24546 24546 | C4-1/8-T0-1652-F C4-1/8-T0-2491-F C4-1/8-T0-2491-F C4-1/8-T0-1503-F C4-1/8-T0-1503-F |
| A13825 A13824 A13825 A13826 A13827 | 0757-0465 0757-0440 0757-0442 0757-0440 0757-0469 | 6 7 9 7 0 | | RESISTOR 100K 1% .125W F TC=0+=100 RESISTOR 7.5K 1% .125W F TC=0+=100 RESISTOR 10K 1% .125W F TC=0+=100 RESISTOR 7.5K 1% .125W F TC=0+=100 RESISTOR 150K 1% .125W F TC=0+=100 | 24546 24546 24546 24546 | C4-1/8-T0-1003-F C4-1/8-T0-7501-F C4-1/8-T0-1002-F C4-1/8-T0-501-F C4-1/8-T0-1503-F |
| 413928 413929 413931 413932 413933 | 0757-0469 0757-0449 0698-3484 0698-4481 0698-4435 | 8 9 9 0 | 3 1 | RESISTOR 150K 1% .125W F TC=0+=100 RESISTOR 20K 1% .125W F TC=0+=100 RESISTOR 6.65K 1% .125W F TC=0+=100 RESISTOR 16.5K 1% .125W F TC=0+=100 RESISTOR 2.49K 1% .125W F TC=0+=100 | 24546 24546 24546 24546 | C4-1/8-T0-1503-F C4-1/8-T0-2002-F C4-1/8-T0-0651-F C4-1/8-T0-1652-F C4-1/8-T0-2491-F |
| A 1 3 8 3 4 A 1 3 8 3 5 A 1 3 8 3 6 A 1 3 8 3 6 A 1 3 8 3 8 * • • • • • • • • • • • • • • • • • • | 0698-4435 0757-0467 0757-0467 0757-0461 0757-0401 2100-3210 3101-1162 1251-3378 03580-66534 | 288 | 3 1 1 | RESISTOR 2.49K 1% ,125W F TC=0+-100 RESISTOR 121K 1% ,125W F TC=0+-100 RESISTOR 121K 1% ,125W F TC=0+-100 RESISTOR 101 % .125W F RESISTOR 100 1% .125W F RESISTOR 106 K .125W F RESISTOR 10F N .125W F SWITCH-SLIPE SPOT NS CONNECTOR 10-PIN M POST TYPE BOARD ASSEMBLY-BANDWIDTH SWITCH | 24546 24546 24546 28480 28480 28480 28480 28480 28480 28480 | C4-1/8-T0-2491-F C4-1/8-T0-1213-F C4-1/8-T0-1213-F 0757-0401 0757-0401 2100-3210 3101-1162 1251-3378 03580-66534 |
| A14C4 A14C5 A14C6 A14C7 A14CR | 0180-0197 0180-0373 0180-1735 0180-2050 0189-1701 | 8 2 2 6 2 | 1 1 1 1 | CAPACITOR=FXD 2.2UF+=10% 20VDC TA CAPACITOR=FXD .68UF+=10% 35VDC TA CAPACITOR=FXD .22UF+=10% 35VDC TA CAPACITOR=FXD .082UF+=10% 35VDC TA CAPACITOR=FXD .68UF+=20% 6VDC TA | 56289 56289 56289 56289 56289 | 150P225X9020A2 150D684X9035A2 150D224X9035A2 150D823X9035A2 150D865X0006A2 |
| 41459 | 0160-0162 | 5 | 1 | CAPACITOR-FXD .022UF +-10% 200VDC POLYE | 28480 | 0160-0162 |
| A14J1 | 1251-0561 | 2 | 5 | CONNECTOR 34-PIN F POST TYPE | 28480 | 1251-0561 |
| A1431 | 1855-0081 | 1 | 3 | TRANSISTOR J=FET N=CHAN D=MODE SI | 01295 | 2N5245 |
| A1 422 A1 423 A1 423 A1 424 A1 425 | 0698-3453 0698-4488 0698-3558 0698-3519 0698-3228 | 2 5 8 1 9 | 1 1 1 2 6 | RESISTOR 196K 1% .125W F TC#0+=100 RESISTOR 26.7K 1% .125W F TC#0+=100 RESISTOR 4.02K 1% .125W F TC#0+=100 RESISTOR 12.4K 1% .125W F TC#0+=100 RESISTOR 49.9K 1% .125W F TC#0+=100 | 24546 24546 24546 24546 28480 | C4-1/8-T0-1963-F C4-1/8-T0-2672-F C4-1/8-T0-4021-F C4-1/8-T0-1242-F 0698-3228 |
| 41426 41427 41428 41429 414210 | 0757-0473 0684-1051 0684-2251 0684-1041 0684-3941 | 6 3 7 1 4 | 1 1 1 1 | RESISTOR 221% 1% .125m F TC=0+-100 RESISTOR 1M 10% .25m FC TC=-800/+900 RESISTOR 2.2M 10% .25m FC TC=-900/+1100 RESISTOR 100% 10% .25m FC TC=-400/+800 RESISTOR 390% 10% .25m FC TC=-800/+900 | 24546 01121 01121 01121 | C4-1/8-T0-2213-F C81051 C82251 C81041 C83941 |
| A14R11 A14R12 A14R13 A14R14 A14R15 | 0698-5102 0698-4443 0757-0454 0698-4506 0698-3459 | 2 3 8 8 | 1 1 1 1 | RESISTOR 1.2M 10% .25% FC TC=-900/+1100 RESISTOR 4.53% 1% .125% F TC=0+-100 RESISTOR 33.2% 1% .125% F TC=0+-100 RESISTOR 73.2% 1% .125% F TC=0+-100 RESISTOR 383% 1% .125% F TC=0+-100 | 01121 24546 24546 24546 28480 | C81251 C4-1/8-T0-4531-F C4-1/8-T0-3322-F C4-1/8-T0-7322-F 0698-3459 |
| A14816 A14817 A14818 A14819 A14819 A14820 | 0698-4524 0757-0442 0698-3441 0698-4427 0698-4511 | 0 9 8 2 5 | 3 1 1 3 | RESISTOR 174K 1% 125W F TC=0+=100 RESISTOR 10K 1% 125W F TC=0+=100 RESISTOR 215 1% 125W F TC=0+=100 RESISTOR 1.65K 1% .125W F TC=0+=100 RESISTOR 86.6K 1% .125W F TC=0+=100 | 24546 24546 24546 24546 | C4=1/8=T0=1743=F C4=1/8=T0=1002=F C4=1/8=T0=215R=F C4=1/8=T0=1851=F C4=1/8=T0=8662=F |
| A14R21 A14R22 A14R31 A14R32 A14R33 | 0757-0446 0698-4511 0698-4500 | 5 3 5 2 5 | 2 1 2 | RESISTOR 43.2K 1% .125W F TC=0+=100 RESISTOR 15% 1% .125W F TC=0+=100 RESISTOR 86.6K 1% .125W F TC=0+=100 RESISTOR 57.6K 1% .125W F TC=0+=100 RESISTOR 43.2K 1% .125W F TC=0+=100 | 24546 24546 24546 24546 | C4-1/8-T0-4322-F C4-1/8-T0-1502-F C4-1/8-T0-8662-F C4-1/8-T0-5762-F C4-1/8-T0-4322-F |
| A14R34 A14R35 A14R36 A14R37 A14R38 | 0698-3455 0757-0468 0698-7802 | 3 4 9 3 3 | 1 2 1 | RESISTOR 34.8K 1% .125M F TC=0+-100 RESISTOR 261K 1% .125M F TC=0+-100 RESISTOR 130K 1% .125M F TC=0+-100 RESISTOR 523K 1% .125M F TC=0+-100 RESISTOR 52.3K 1% .125M F TC=0+-100 | 28480 24546 24546 28480 24546 | 0757-0123 C4-1/8-T0-2613-F C4-1/8-T0-1303-F 0698-7802 C4-1/8-T0-5232-F |
| A 1 4 R 4 4 4 4 4 4 4 4 4 4 4 4 4 4 4 4 4 | 0698-3215 0698-3228 0698-3279 | 44900 | 1 1 2 | RESISTOR 64.9K 1% .125W F TC=0+-100 RESISTOR 499K 1% .125W F TC=0+-100 RESISTOR 49.9K 1% .125W F TC=0+-100 RESISTOR 4.99K 1% .125W F TC=0+-100 RESISTOR 1.74K 1% .125W F TC=0+-100 | 24546 28480 24546 24546 | C4-1/8-T0-6492-F 0698-3215 0698-3228 C4-1/8-T0-4991-F C4-1/8-T0-1743-F |
| 414R60 | 0757-0427 | ٥ | 1 | RESISTOR 1.5K 1% .125W F TC=0+-100 | 24546 | C4-1/8-T0-1501-F |
| A 1 4 5 2 | 03580=61901 03580=61907 | 1 7 | 1 1 | SWITCH ASSEMBLY SWITCH ASSEMBLY | 28480 28480 | 03580-61901 03580-61907 |
| | | | | | | |

Table 6-3. Replaceable Parts (Cont'd).

| -06535 0000 0040 0485 0485 0485 0485 0561 0561 0561 0561 0615 0780 | 7 10675 22 1112 32195 44126 8649 | 1 | BOARD ASSEMBLY-SWEEP SWITCH DIODE-SWITCHING 30V SOMA 2NS DO-35 DIODE-SCHOTTKY DO-35 LED-VISIBLE LUM-INT=IMCD IF=20MA-MAX LED-VISIBLE LUM-INT=IMCD IF=20MA-MAX LED-VISIBLE LUM-INT=00UCD IF=30MA-MAX CONNECTOR 34-PIN F POST TYPE CABLE ASSEMBLY TRANSISTOR NPN SI PD=350MN FT=300MHZ TRANSISTOR J=FET N=CMAN D=MODE SI TRANSISTOR 121 1X .125M F TC=0+=100 RESISTOR 301 1X .125M F TC=0+=100 RESISTOR 1.21K 1X .125M F TC=0+=100 RESISTOR 1.21K 1X .125M F TC=0+=100 RESISTOR 3.01K 1X .125M F TC=0+=100 | 28480 28480 28480 28480 28480 28480 28480 28480 28480 24546 24546 24546 24546 | 03580-66535 1901-0040 MSCM-1001 5082-4684 5082-4584 5082-4984 1251-0561 03580-61613 2N3904 2N5245 2N5245 2N5245 055245 0698-7802 C4-1/8-70-301R-F C4-1/8-70-301R-F C4-1/8-70-301R-F |
|---|----------------------------------|---|--|--|---|
| 0441 0487 0487 0485 0485 0561 0561 0081 0036 0036 0036 003 04410 0161 277 34440 24440 24440 4440 4440 4440 4440 4 | 0675 22 1112 32195 44126 864 | 1 1 1 5 5 1 1 1 2 2 2 1 1 | DIODE-SCHOTTKY DO-35 LED-VISIBLE LUM-INTBIMCD IFB20MA-MAX LED-VISIBLE LUM-INTBIMCD IFB20MA-MAX LED-VISIBLE LUM-INTBOOULD IFB30MA-MAX CONNECTOR 34-PIN F POST TYPE CABLE ASSEMBLY TRANSISTOR NPN SI PDB350MH FTB300MHZ TRANSISTOR J-FET N-CHAN D-MODE SI TRANSISTOR J-FET N-CHAN D-MOD | 28480 28480 28480 28480 28480 04713 01295 01295 014940 28480 24546 | HSCH=1001 5082-4584 5082-4584 1251-0561 03580-61613 2N3904 2N5245 1853-0036 0698-7802 C4-1/8-T0-121R-F C4-1/8-T0-301R-F |
| -01013 0215 0081 | 2 1112 32195 44126 864 | 1 5 1 1 1 2 2 2 1 | CABLE ASSEMBLY TRANSISTOR NPN SI PD=350MW FT=300MMZ TRANSISTOR J=FET N=CMAN D=MODE SI TRANSISTOR J=FET N=CMAN D=MODE SI TRANSISTOR PNP SI PD=310MW FT=250MMZ RESISTOR 523K 1% 125W F TC=0+=100 RESISTOR 121 1% 125W F TC=0+=100 RESISTOR 301 1% 125W F TC=0+=100 RESISTOR 604 1% 125W F TC=0+=100 RESISTOR 1.21K 1% 125W F TC=0+=100 RESISTOR 1.21K 1% 1.25W F TC=0+=100 RESISTOR 3.01K 1% 1.25W F TC=0+=100 | 28480 04713 01295 01295 28480 28484 24546 24546 | 03580-61613 2N3904 2N5245 2N5245 1853-0036 0698-7802 C4-1/8-T0-121R-F C4-1/8-T0-301R-F |
| 0081 10086 10086 10086 10081 1 | 32195 44126 864 | 1 1 1 2 2 2 1 | TRANSISTOR J=FET N=CHAN D=MODE SI TRANSISTOR J=FET N=CHAN D=MODE SI TRANSISTOR PNP SI PDB310MW FTE250MHZ RESISTOR 523K 1% .125W F TC=0+=100 RESISTOR 121 1% .125W F TC=0+=100 RESISTOR 301 1% .125W F TC=0+=100 RESISTOR 604 1% .125W F TC=0+=100 RESISTOR 1,21K 1% .125W F TC=0+=100 RESISTOR 3,01K 1% .125W F TC=0+=100 | 01295 01295 28480 28480 24546 24546 | 2N5245 2N5245 1853-0036 0698-7802 C4-1/8-70-121R-F C4-1/8-70-301R-F |
| 1403 1410 1274 1273 1444 1457 1467 1469 1499 | 2195 44126 864 | 1 1 2 2 2 1 | RESISTOR 121 1% .125W F TC=0+=100 RESISTOR 301 1% .125W F TC=0+=100 RESISTOR 604 1% .125W F TC=0+=100 RESISTOR 1.21K 1% .125W F TC=0+=100 RESISTOR 3.01K 1% .125W F TC=0+=100 | 24546 24546 24546 | C4-1/8-T0-121R-F C4-1/8-T0-301R-F |
| 497 444 453 572 467 499 442 442 442 442 | 4 1 2 6 8 6 | 2 2 1 | | 24340 | C4-1/8-T0-1213-F |
| 499 442 444 572 572 | 6 4 | | RESISTOR 6.04K 1% .125W F TC=0+-100 RESISTOR 12.1K 1% .125W F TC=0+-100 RESISTOR 30.1K 1% .125W F TC=0+-100 RESISTOR 60.4K 1% .125W F TC=0+-100 | 24546 24546 24546 24546 24546 | C4-1/8-T0-3011-F C4-1/8-T0-604R-F C4-1/8-T0-1212-F C4-1/8-T0-3012-F C4-1/8-T0-8042-F |
| 572 | 1 | 1 | RESISTOR 121K 1% .125W F TC=0+=100 RESISTOR 40.2K 1% .125W F TC=0+=100 RESISTOR 6.04K 1% .125W F TC=0+=100 RESISTOR 10K 1% .125W F TC=0+=100 RESISTOR 12.1K 1% .125W F TC=0+=100 | 24546 24546 24546 24546 24546 | C4-1/8-T0-1213-F C4-1/8-T0-4022-F C4-1/8-T0-604R-F C4-1/8-T0-102-F C4-1/8-T0-1212-F |
| 519 | 1 9 0 0 | 1 | RESISTOR 12,5K .5% .125W F TC=0+-100 RESISTOR 12,5K .5% .125W F TC=0+-100 RESISTOR 10K 1% .125W F TC=0+-100 RESISTOR-VAR CONTROL CC 10K 10% LIN RESISTOR 12,4K 1% .125W F TC=0+-100 | 24546 24546 24546 28480 24546 | C4-1/8-T0-1252-D C4-1/8-T0-1252-D C4-1/8-T0-1002-F 2100-06-8 C4-1/8-T0-1242-F |
| 758 580 573 292 581 | 6 0 1 3 | 1 1 1 1 | RESISTOR 12,5% _5% _125W F TC=0+=50 RESISTOR 25K _5% _125W F TC=0+=100 RESISTOR 50K _5% _125W F TC=0+=100 RESISTOR 125K _5% _125W F TC=0+=100 RESISTOR 250K _5% _125W F TC=0+=100 | 24546 24546 24546 28480 19701 | NC4-1/8-T2-1252-D C4-1/8-T0-2502-D C4-1/8-T0-5002-D 0688-6292 MF4C1/8-T0-2503-D |
| 486 489 351 524 455 | 1 6 0 0 4 | 1 1 1 | RESISTOR 750K 1% .125W F TC=0+-100 RESISTOR 26K 1% .125W F TC=0+-100 RESISTOR 3.3M 10% .25W F TC=-900/+1100 RESISTOR 174K 1% .125W F TC=00+-100 RESISTOR 261K 1% .125W F TC=0+-100 | 28480 24546 01121 24546 24546 | 0757-0486 C4-1/8-T0-2802-F C83351 C4-1/8-T0-1743-F C4-1/8-T0-2613-F |
| 500 511 442 442 160 | 2 5 9 9 8 | 1 | RESISTOR 57.6K 1% .125W F TC=0+-100 RESISTOR 86.6K 1% .125W F TC=0+-100 RESISTOR 10K 1% .125W F TC=0+-100 RESISTOR 10K 1% .125W F TC=0+-100 RESISTOR 31.6K 1% .125W F TC=0+-100 | 24546 24546 24546 24546 | C4-1/8-T0-5762-F C4-1/8-T0-8662-F C4-1/8-T0-1002-F C4-1/8-T0-1002-F C4-1/8-T0-3162-F |
| 15 32 16 442 557 | 9 7 | 1 1 | RESISTOR 100K 5% .125W F RESISTOR 1M 1% .25W F RESISTOR 383 1% .125W F RESISTOR 30% 1% .125W F TC=0+-100 RESISTOR 806 1% .125W F TC=0+-100 | 28480 28480 28480 24546 24546 | 0683-1045 0698-7332 0698-3446 C4-1/6-T0-1002-F C4-1/8-T0-806R-F |
| | 5 8 | 1 1 | RESISTOR 499 1% .125# F TC=0+=100 RESISTOR 3,48K 1% .125# F TC=0+=100 | 24546 24546 | C4-1/8-T0-499R-F C4-1/8-T0-3481-F |
| , | 3 8 7 | 1 1 1 | SWITCH ASSEMBLY-SPAN SWITCH ASSEMBLY-MODE SWITCH-SL DPDT MINTR .5A 125VAC/DC | 28480 28480 28480 | 03580-61903 03580-61908 3101-0199 |
| 043 | 4 | 3 | IC OP AMP GP TO=99 IC OP AMP GP TO=99 | 01928 01928 | CA3077 CA3077 |
| 6536 | 8 | 1 | BOARD ASSEMBLY_FCM | 28460 | 03580-66536 |
| 104 571 | 7 8 | 2 7 1 | CAPACITOR=FXD 200UF+75=10% 164DC AL CAPACITOR=FXD 200UF+75=10% 164DC AL CAPACITOR=FXD .1UF +80=20% 504DC CER CAPACITOR=FXD .047UF +=5% 804DC POLYE | 56289 56289 28480 28480 | 30D207G016DF2 30D207G016DF2 0160-4571 0160-2672 |
| | 0 | 1 1 1 | CAPACITOR-FXD 60UF+-20X 6VDC TA CAPACITOR-FXD 50UF+75-10X 16VDC AL CAPACITOR-FXD 1000PF +-20X 50VDC CER CAPACITOR-FXD 1UF +80-20X 50VDC CER | 56289 56289 28480 28480 | 150De06x0006B2 30D506G016CB2 0160-4532 0160-4571 |
| 100 | 3 536 4 1 1 2 | 536 8 7 7 1 8 6 9 9 0 1 1 | 536 8 1 4 7 2 4 7 7 1 8 7 2 6 1 | 536 6 1 BOARD ABSEMBLY_FCM 4 7 2 CAPACITOR=FXD 200UF+75=10% 16VDC AL CAPACITOR=FXD 200UF+75=10% 16VDC AL CAPACITOR=FXD .1UF +80=20% 50VDC CER CAPACITOR=FXD .047UF +=5% 80VDC POLYE 6 9 1 CAPACITOR=FXD 60UF+=20% 6VDC TA 9 0 1 CAPACITOR=FXD 50UF+=10VDC AL CAPACITOR=FXD 50UF+=10VDC AL CAPACITOR=FXD 10VDC FX 50VDC CER | 536 8 1 BOARD A8SEMBLY-FCM 28480 4 7 2 CAPACITOR-FXD 200UF+75-10X 16VDC AL 56289 4 7 CAPACITOR-FXD 200UF+75-10X 16VDC AL 56289 1 8 7 CAPACITOR-FXD 1UF +80-20X 50VDC CER 28480 2 6 1 CAPACITOR-FXD 047UF +5X 80VDC POLYE 28480 6 9 1 CAPACITOR-FXD 60UF+-20X 6VDC TA 56289 9 0 1 CAPACITOR-FXD 50UF+-20X 6VDC TA 56289 2 1 1 CAPACITOR-FXD 1000PF +-20X 50VDC CER 28480 |

Revised: September 1987

Table 6-3. Replaceable Parts (Cont'd).

| Table 6-3. Replaceable Parts (Cont'd). | | | | | | | | |
|--|--|---------------------|------------------|---|---|--|--|--|
| Reference Designation | HP Part Number | C D | Qty | Description | Mfr Code | Mfr Part Number | | |
| A16C43 A16C44 A16C45 A16CR2 A16CR3 A16CR4 #16CR6 A16CR7 A16CR7 A16CR3 A16CR31 A16CR32 | 0160-4571 0160-4571 1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 1901-0040 | 5 1 1 1 1 1 1 1 1 3 | 1 | CAPACITOR-FXD .1UF +80-20% 50VDC CER CAPACITOR-FXD .1UF +80-20% 50VDC CAPACITOR-FXD .1UF +80-20% 50VDC DIODE-SMITCHING 30V 50MA 2NS D0-35 DIODE-ZNR 1N825 6.2V 5X D0-7 PD0.4W | 28480 28480 28480 28480 28480 28480 28480 28480 28480 28480 28480 | 0160-4571 0160-4571 10160-4571 101-0040 101-0040 101-0040 101-0040 101-0040 101-0040 101-0040 101-0040 101-0040 101-0040 101-0040 | | |
| A10L1 A10L2 A10J2 A10J2 A10J3 A10J4 | 9100-1644 9100-0541 1855-0386 1854-0071 1853-0036 1853-0036 | 3797222 | 1 1 1 | INDUCTORRF-CH-MLD 330UH 5% .2D%.45LG INDUCTORRF-CH-MLD 250UH 10% .250%,5LG TRANSISTOR J-FET 200392 N-CHAN D-MODE TRANSISTOR NPN 3I PD=310MW FT=250MHZ TRANSISTOR PNP 3I PD=310MW FT=250MHZ | 28480 28480 04713 28480 28480 28480 28480 | 9100-1644 9100-0541 2N4392 1854-0071 1853-0036 1853-0036 | | |
| 416R2 A16R3 A16R4 416R5 A16R6 | 0757-0472 0757-0449 0757-0273 0757-0465 0757-0465 | 5 6 4 6 6 | 1 | RESISTOR 200K 1X .125M F TC=0+-100 RESISTOR 20K 1X .125M F TC=0+-100 RESISTOR 3.01K 1X .125M F TC=0+-100 RESISTOR 100K 1X .125M F TC=0+-100 RESISTOR 100K 1X .125M F TC=0+-100 | 24546 24546 24546 24546 | C4-1/8-T0-2303-F C4-1/8-T0-2002-F C4-1/8-T0-3011-F C4-1/8-T0-1003-F C4-1/8-T0-1003-F | | |
| A1687 A1648 A1649 A16410 A16411 | 0757-0463 0698-3493 0757-0457 0698-3228 0757-0465 | 4 0 6 9 6 | 1 | RESISTOR 82.5K 1% .125M F TC=0+=100 RESISTOR 4.12K 1% .125M F TC=0+=100 RESISTOR 47.5K 1% .125M F TC=0+=100 RESISTOR 49.9K 1% .125M F TC=0+=100 RESISTOR 100K 1% .125M F TC=0+=100 | 24546 24546 24546 24546 24546 | C4-1/8-T0-8252-F C4-1/8-T0-4121-F C4-1/8-T0-4752-F 0696-3228 C4-1/8-T0-1003-F | | |
| A16812 A16813 A16814 A16815 A16816 | 0757-0463 0757-0465 0683-1555 0757-0465 0698-3157 | 0 6 3 | 1 | RESISTOR 82.5% 1% .125% F TC=0+=100 RESISTOR 100K 1% .125% F TC=0+=100 RESISTOR 1.5% 5% .25% FC TC==900/+1100 RESISTOR 100K 1% .125% F TC=0+=100 RESISTOR 19.6K 1% .125% F TC=0+=100 | 24546 24546 01121 24546 24546 | C4-1/8-T0-8252-F C4-1/8-T0-1003-F C81555 C4-1/8-T0-1003-F C4-1/8-T0-1962-F | | |
| A16818 A16821 | 0683-2235 0698-3557 0757-0465 | 5 7 6 | 1 | RESISTOR 22K 5% .25W FC TC=-400/+800 RESISTOR 806 1% .125W F TC=0+-100 RESISTOR 100K 1% .125W F TC=0+-100 | 01121 24546 24546 | C82235 C4-1/8-T0-806R-F C4-1/8-T0-1003-F | | |
| A 1 6 7 2 2 A 1 6 7 2 3 A 1 6 7 2 4 A 1 6 8 2 5 A 1 6 8 3 0 A 1 6 9 3 1 A 1 6 9 3 2 | 0757-0465 0698-3228 0757-0465 0757-0442 0683-1535 0757-0440 0757-0449 | 7 6 6 | 9 | RESISTOR 100K 1% 125W F TC=0+=100 RESISTOR 49,9K 1% 125W F TC=0+=100 RESISTOR 100K 1% 125W F TC=0+=100 RESISTOR 15K 5% 125W F RESISTOR 7.5K 1% 125W F TC=0+=100 RESISTOR 20K 1% 125W F TC=0+=100 RESISTOR 475 1% 125W F TC=0+=100 RESISTOR 475 1% 125W F TC=0+=100 | 24546 28480 24546 28480 28480 24546 24546 24546 | C4=1/8-T0=1003-F 0698-3228 C4=1/8-T0=1003-F 0757-0442 0683-1535 C4=1/8-T0=7501-F C4=1/8-T0=2002-F C4=1/8-T0=475R-F | | |
| A16R33 A16R34 A16R35 A16R36 A16R40 | 0698-3382 0698-5673 0698-3279 0757-0407 0698-3228 | 6 2 0 6 9 | 1 1 | RESISTOR 5,49K 1% .125W F TC=0+-100 RESISTOR 3,9K 1% .125W F TC=0+-25 RESISTOR 4,99K 1% .125W F TC=0+-100 RESISTOR 200 1% .125W F TC=0+-100 RESISTOR 49,9K 1% .125W F TC=0+-100 | 24546 28480 24546 24546 28480 | C4=1/8=T0=5491=F 0698=5673 C4=1/8=T0=4991=F C4=1/8=T0=201=F 0698=3228 | | |
| A16941 A16942 A16843 A16845 A16850 | 0698-7332 0757-0199 0698-7332 0698-4542 0698-3228 | 3 4 2 9 | 1 | RESISTOR 1M 1X .125W F TC=0+=100 RESISTOR 21.5K 1X .125W F TC=0+=100 RESISTOR 1M 1X .125W F TC=0+=100 RESISTOR 453K 1X .125W F TC=0+=100 RESISTOR 49.9K 1X .125W F TC=0+=100 | 28480 24546 28480 28480 28480 | 0698-7332 C4-1/8-T0-2152-F 0698-7332 0698-4542 0698-3228 | | |
| A16452 A16853 A16854 | 2100-3352 0698-7332 0757-0442 | 7 | 1 | RESISTOR 10K 1% _125W F TC=0+-100 RESISTOR=TRMR 1K 10% C SIDE-ADJ 1-TRN RESISTOR 1M 1% .125W F RESISTOR 10K 1% .125W F | 24546 28480 28480 28480 | C4=1/8=T0=1002=F 2100=3352 0698-7332 0757-0442 | | |
| A16U1 A16U2 A16U8 A16U9 | 1826-0043 1826-0759 1826-0304 1826-0561 | 0 0 | 1 2 1 2 | IC OP AMP GP TO-99 IC COMPARATOR GP GUAD 14-DIP-P IC OP AMP TO-99 IC OP AMP OPO2CJ | 01928 01295 27014 27014 | CA307T IC339 LF355H LF355H | | |
| 416J1 | 03560-61612 | 4 | 1 | CABLE ASSEMBLY CABLE ASSEMBLY CONNECTOR 20-PIN M POST TYPE | 28480 28480 28480 | 03580-61612 03580-61612 1251-5240 | | |
| | | | | | | | | |

Table 6-3. Replaceable Parts (Cont'd),

| | Table 6-3. Replaceable Parts (Cont'd). | | | | | | | | |
|---|---|-----------------------|------------------|---|--|--|--|--|--|
| Reference Designation | HP Part Number | C D | Qty | Description | Mfr Code | Mfr Part Number | | | |
| Α18 Δ1 | 03581-66518 | | 1 | BOARD ASSY: INPUT, BALANCED (FOR OPTION 002 ONLY) | 28480 | 03580-66518 | | | |
| A18C1 A18C4 A18C5 A18C6 | 0180-0091 0180-0091 0160-2206 0140-0204 | | 2 | C:FXD 10UF +50 -10% 100 VDC AL C:FXD 10UF +50 -10% 100 VDC AL C:FXD 160PF 5% 300 VDCW C:FXD 47PF 5% 500 VDCW | 56289 56289 28480 72136 | 300106F100DC2 300106F100DC2 0160-2206 DM15E470J0500WY1CR | | | |
| A18J1 A18J2 | 1251-2969 1251-3638 | | 1 | CONN:PHONO, SINGLE JACK CONN:POST TYPE | 27264 27264 | 15-24-0501 09-65-1061 | | | |
| A18R1 A18R2 A18R3# | 0698-4882 0698-5874 0757-0284 | | 1 1 1 | R:FXD 976 OHM 1% .5 W F TUBULAR R:FXD 639 OHM 1% .5 W F TUBULAR R:FXD 150 OHM 1% .125 W F TUBULAR FACTORY SELECTED PART | 24546 24546 24546 | NA6 NA6 C4-1/8-TO-151-F | | | |
| A18R4 A18R5 A18T1 | 0757-0472 0698-4308 9100-1460 | | 1 | F:FXD 200K 1% .125 W F TUBULAR F:FXD 16.9K 1% .125 W F TUBULAR TRANSFORMER AUDIO | 24546 16299 28480 | C4-1/8-TO-2003-F C4-1/8-TO 1692-F 9100-1460 | | | |
| A20 | 0960-0505 | | 1 | POWER INPUT MODULE | 28480 | 0960-0444 | | | |
| A33 | 03580-66533 | 5 | 1 | BOARD ASSEMBLY-DISPLAY, FCM | 28480 | 03580-66533 | | | |
| A33C1 A33C2 A33C4 A33C5 | 0160-4571 0160-5104 0160-2205 0160-4571 | 8 5 1 8 | 1 1 | CAPACITOR-FXD 1.UF +80 -20% 50 VDC CER CAPACITOR-FXD 3.9PF 5% CAPACITOR-FXD 120PF +-5% 300 VDC MICA CAPACITOR-FXD .2UF +80 -20% 50 VDC CER | 28480 28480 28480 28480 | 0160-4571 0160-5104 0160-2205 0160-4571 | | | |
| A33CR1 | 1901-0841 | 0 | 1 | DIO-SI .05A 30V | 28480 | 1901-0841 | | | |
| A33L1 A33Q1 | 9100-1650 1853-0036 | 1 2 | 1 | INDUCTOR RF-CH-MLD 680 UH 5% .2DX.45LG TRANSISTOR PNP SI PD=310 MW FT=250 MHZ | 28480 28480 | 9100-1650 1853-0036 | | | |
| A33R1 A33R2 A33R3 A33R4 A33R5 | 0698-4517 0757-0465 0757-0465 0757-0401 0757-0435 | 1 6 6 0 | 1 3 1 | RESISTOR 127K 1% .125 W F TC=0+-100 RESISTOR 100K 1% .125 W F TC=0+-100 RESISTOR 100K 1% .125 W F TC=0+-100 RESISTOR 100 1% .125 W F TC=0+-100 RESISTOR 3.92K 1% .125 W F TC=0+-100 | 24546 24546 24546 24546 24546 | C4-1/8-TO-1273-F C4-1/8-TO-1003-F C4-1/8-TO-1003-F C4-1/8-TO-101-F C4-1/8-TO-3921-F | | | |
| A33R6 A33R7 A33R8 A33R9 A33R10 | 0757-0401 0698-3557 0698-3149 0757-0401 0757-0442 | 0 7 3 1 9 | 1 | RESISTOR 100 1% .125 W F TC=0+-100 RESISTOR 806 1% .125 W F TC=0+-100 RESISTOR 255K 1% .125 W F TC=0+-100 RESISTOR 100 1% .125 W F TC=0+-100 RESISTOR 10K 1% .125 W F TC=0+-100 | 24546 24546 24546 24546 24546 24546 | C4-1/8-TO-101-F C4-1/8-TO-806R-F C4-1/8-TO-2553-F C4-1/8-TO-101-F C4-1/8-TO-1002-F | | | |
| A33R11 A33R12 A33R13 A33R14 A33R15 | 0757-0465 0757-0442 0757-0415 1757-0415 0757-0415 | 6 9 6 6 6 | | RESISTOR 100K 1% .125 W F TC=0+-100 RESISTOR 10K 1% .125 W F TC=0+-100 RESISTOR 475 1% .125 W F TC=0+-100 RESISTOR 475 1% .125 W F TC=0+-100 RESISTOR 475 1% .125 W F TC=0+-100 | 24546 24546 24546 24546 24546 | C4-1/8-TO-1003-F C4-1/8-TO-1002-F C4-1/8-TO-475R-F C4-1/8-TO-475R-F C4-1/8-TO-475R-F | | | |
| A33R16 A33R17- A33R18 A33R19 A33R20 | 0757-0415 0757-0415 0757-0415 0757-0415 0757-0415 | 6 6 6 6 | | RESISTOR 475 1% .125 W F TC=0+-100 RESISTOR 475 1% .125 W F TC=0+-100 | 24546 24546 24546 24546 24546 | C4-1/8-TO-475R-F C4-1/8-TO-475R-F C4-1/8-TO-475R-F C4-1/8-TO-475R-F C4-1/8-TO-475R-F C4-1/8-TO-475R-F | | | |
| A33R21 A33R22 A33R24 | 0757-0442 0757-0442 2100-3889 | 9 9 5 | 1 | RESISTOR 10K 1% .125 W F TC=0+-100 | 24546 24546 28480 | C4-1/8-TO-1002-F C4-1/8-TO-1002-F 2100-3889 | | | |
| A33U1 A33U2 A33U3 A33U4 A33U5 | 1820-1963 1820-2310 1820-1413 1858-0047 1820-0938 | 7 0 2 5 4 | 2 1 1 1 | IC CNTR PMOS DECD UP/DOWN SYNCHRO IC DCDR CMOS BCD-TO-7-SEG 4-TO-7-LINE TRANSISTOR ARRAY 16-PIN PLSTC DIP | 01928 50088 04713 13606 0192B | CB4013BAE MK50399N MC14511BCP ULN-2003A CD4027AE | | | |
| A33U6 A33U7 A33U8 A33U9 | 1820-0935 1820-1963 1826-0026 1820-1408 | 1 7 3 5 | 1 1 1 | IC FF CMOS D-TYPE POS-EDGE-TRIG DUAL IC COMPARATOR PRCN TO-99 | 0192B 0192B 01295 0192B | CD4020AE CD4013BAE LM311L CD40738F | | | |
| A33W1 | 03580-61611 | ٥ | 1 | CABLE ASSEMBLY | 28480 | 03580-61611 | | | |
| A40 | 03580-66540 | | 1 | DISPLAY ADJUSTMENT ASSY | | | | | |
| A40R1 A40R2 | 2100-3212 | | 2 | RESISTOR-VAR 200 10% | 32977 | 3386P-Y46-201 | | | |
| A40R3 | 2100-3212 2100-3210 | | 1 | RESISTOR-VAR 200 10% RESISTOR-VAR 10K 10% | | 3386P-Y46-201 3386P-Y46-103 | | | |
| A40W1 | 03582-61640 | | 1 | CABLE ASSY A13/A40 | | 03580-61640 | | | |
| | | | | | | | | | |

ALSEE BACKDATING.

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See introduction to this section for ordering information *Indicates factory selected value

Table 6-3. Replaceable Parts (Cont'd).

| Reference Designation | HP Part Number | CD | Qty | Description | Mfr Code | Mfr Part Number |
|---|--|----|---------------------------------|--|---|---|
| | | | | CHASSIS MOUNTED COMPONENTS | | |
| | 03580-69508 | | | BATTERY PACK KIT | 28480 | 03580-69508 |
| BT1 THRU BT4 BT5 | 1420-0203 1420-0202 03580-04108 10101B 0624-0410 0160-2050 | | 4 1 2 1 4 | BATTERY PACK (4 CELLS) (OPT 001 ONLY) BATTERY PACK (4 CELLS CENTER TAP) (OPT 001 ONLY) PLASTIC BATTERY END GUARD COVER:PROTECTIVE FRONT (OPT 001 ONLY) SCR: TPG 6-19 (OPT 001 ONLY) | 05397 05397 28480 28480 28480 | Y-6114 Y-5505 03580-04108 10101B |
| DS1 | 2140-0380 1450-0153 1450-0157 | | 1 1 1 | C: FXD 10 UF 30 V 10% LAMP: INCAND (POWER) LAMP HOLDER (FOR DS1) | 56289 17537 08717 | 127P1069R3S4 86 102SR |
| DS2 THRU DS4 | 1990-0450 5040-7626 1450-0404 | | 3 3 | LENS (FOR DS1) DIODE: LIGHT EMITTING CLAMP LED (FOR DS2-4) LENS-PILOT LIGHT | 08717 28480 28480 | 102xx-w 1990-0450 5040-7626 |
| F1 J1 J2 J3, J4 J5 THRU J10 J11, J12 | 2110-0012 1510-0084 1510-0087 1510-0076 1250-0083 1510-0076 0340-0732 2190-0027 2950-0006 | | 1 1 2 6 2 | FUSE: 0.5 A 250 V NB BINDING POST: J-GRAY/RED BINDING POST: J-GRAY/BLK BINDING POST: J-GRAY CONN: BNC BINDING POST: J-GRAY (OPT 002 ONLY) INSULATOR: BINDING POST WASHER-INTERNAL LOCK NUT-HEX | 75915 28480 28480 28480 02660 28480 28480 78189 73734 | 312.500 1510-0084 1510-0087 1510-0076 31-221-1020 1510-0076 0340-0732 1914-00 9000 |
| K1 L1 | 0490-0499 01200-6601 | | 1 1 | RELAY:SPDT 2A 12 VDC PEN LIST COIL: TRACE ALIGN | 12300 28480 | RS5D-12VDC 01200-44703 |
| R1 R2 R3 R4 | 2100-0573 2100-0572 2100-0571 2100-1714 | | 1 1 1 | R: VAR LINEAR 200K OHM (INTENSITY) 20% 1/2W R: VAR C COMP 100K OHM (ADAPTIVE SWEEP-INCLUDES S1) R: VAR 5M OHM (FOCUS) 20% R: VAR C COMP LINEAR 1K OHM 20% 1/2W (CAL 10 KHZ) | 01121 12697 12697 01121 | WA4N040S204MZ 381 381 7YPE W |
| R5 R6 R7 R24A/B S1 S2 S3 S4 S5 | 2100-2843 2100-0564 2100-0574 2100-3889 2100-0572 03580-01901 3101-0548 3101-0199 3101-0199 | | 1 1 1 1 1 1 | R: VAR COMP LINEAR 5K OHM 10% 1/2W (LEVEL) R: VAR, 100 K 20 R: VAR 10 TURN 5 K - 10% R: VAR 10K/10K SWITCH: SPST (P/O R2) SWITCH: PUSHBUTTON (DISPLAY) SWITCH: PUSHBUTTON (AMPLITUDE MODE) SWITCH: SLIDE DPDT (DBV/DBM) 0.5A 125V SWITCH: SLIDE DPDT (EXT REF/NORMAL) | 28480 28480 28480 28480 12697 28480 28480 79727 79727 | 2100-2843 2100-0564 2100-0574 2100-3889 381 03580-01901 3101-0548 6126-0012 6126-0012 |
| S6 S7 T1 T2 | 3101-0575 3101-0199 9100-3425 9100-3883 | | 1 1 1 1 | SWITCH: SLIDE (BAL,BRIDGED,TERMINATED)(OPT 002 ONLY) SWTICH: SLIDE TRANSFORMER: POWER TRANSFORMER: OUTPUT (BALANCED TRACKING OSC OUT) (OPT 002 ONLY) | 79727 28480 28480 28480 | G168S-0000 3101-0199 9100-3425 9100-3883 |
| V1 | 5083-1871 | | 1 | TUBE: CATHODE RAY | 28480 | 5083-1871 |
| W1 W2 W3 | 8120-1348 03580-61606 03580-61604 | | 1 1 1 | CORD: POWER, DETACHABLE CABLE ASSY: POWER CABLE ASSY: DIGITAL STORAGE | 70903 28480 28480 | KHS-7041 03580-61606 03580-61604 |
| w4 W5 W6 | 03580-61603 03580-61602 03580-61608 03580-61602 | | 1 1 1 1 | CABLE ASSY: POT (INCLUDES FOCUS POT, R3) CABLE ASSY: DBV/DBM SWITCH CABLE ASSY: CRT CABLE ASSY: INPUT (OPT 002 ONLY) | 28480 28480 28480 28480 | 03580-61603 03580-61608 03580-61602 |
| XA1 | 1200-0037 | | 1 | SOCKET: CRT | 72825 | 97097 |
| | 5020-0476 03580-04102 03580-0608 03580-04104 03580-04103 01200-44701 1390-0084 1390-0084 1390-0088 03580-60121 5060-0548 5040-5862 5040-5861 | | 1 1 2 1 1 1 | MISCELLANEOUS MECHANICAL PARTS BEZEL: CRT (METAL) COVER: BOTTOM COVER: STORM NEST COVER: SIDE RAIL COVER: STOP CRT NECK-CLAMP FASTENER-FANEL: RECEPTACLE, QUARTER TURN FASTENER-PANEL: SCREW, QUARTER TURN FASTENER-PANEL: RETAINER (FOR SCREW) DECK: MAIN FACE PLACE: CRT (BLUE) FOOT: REAR PANEL BASE: FOOT CAP: END | 28480 28480 28480 28480 28480 28480 94222 28480 28480 28480 28480 | 5020-0476 03580-04102 03580-0608 03580-04104 03580-04103 01200-44701 82-47-101-15 1390-0339 1390-0088 03580-60121 5060-0548 5040-5862 5040-5862 |
| | 03580-20013 03580-20014 1510-0038 7120-4609 03580-23706 1440-0103 5040-0508 | | 1 1 1 1 2 2 1 | FRAME: FRONT FRAME: REAR BINDING POST-SINGLE WARNING LABEL FRAME SIDE RAIL HANDLE: STRAP LIGHT SHIELD: CRT (PLASTIC) | 28480 28480 28480 28480 28480 28480 28480 28480 | 03580-20001 03580-20012 1510-0038 7120-4609 03580-23702 1440-0103 5040-0508 |

Table 6-3. Replaceable Parts (Cont'd).

| Number 03580-04104 5040-7042 03580-24706 03580-26001 3050-0456 5040-0508 03580-00211 03580-00211 03580-00212 03580-00205 1460-1341 03580-09301 0370-10005 0370-2182 0370-2188 0370-2188 | CD | 2 4 4 4 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 1 | Description MISCELLANEOUS MECHANICAL PARTS (CONT'D) COVER: SIDE RAIL CAP: END (FOR HANDLE) RETAINER (FOR HANDLE) SCREW (FOR HANDLE) WASHER (FOR HANDLE) LIGHT SHIELD: CRT (PLASTIC) PANEL: FROUENCY CONTROL MODULE PANEL: FROMT (INSTRUMENT) STANDARD 3580A OPTION 002 PANEL: REAP STANDARD 3580A OPTION 002 STAND: TILT BEZEL: LED | 28480 28480 28480 28480 28480 28480 28480 28480 28480 28480 | Mfr Part Number 03580-04104 5040-7042 03580-24706 03580-24706 03580-26001 5808-16-15 5040-0508 03580-00203 03580-00201 03580-00204 03580-00212 03580-00205 |
|--|---|---|---|---|--|
| 5040-7042 03580-24706 03580-26001 3050-0456 5040-0508 03580-00209 03580-00211 \(\triangle \) 03580-00212 03580-00212 03580-00205 1460-1341 03580-09301 0370-10005 0370-2182 0370-2188 0370-2188 | | 4 4 4 1 1 1 1 | COVER: SIDE RAIL CAP: END (FOR HANDLE) RETAINER (FOR HANDLE) SCREW (FOR HANDLE) WASHER (FOR HANDLE) LIGHT SHIELD: CRT (PLASTIC) PANEL: FREQUENCY CONTROL MODULE PANEL: FRONT (INSTRUMENT) STANDARD 3580A OPTION 002 PANEL: REAP STANDARD 3580A OPTION 002 STANDARD 3580A OPTION 002 | 28480 28480 28480 86928 28480 28480 28480 28480 28480 | 5040-7042 03580-24706 03580-26001 5808-16-15 5040-0508 03580-00203 03580-00201 03580-00204 |
| 0370-3034 0370-2185 0370-2185 0370-2185 0370-1005 0370-1005 0370-2006 0370-2994 0370-2184 0370-2184 0370-2189 0370-2189 0370-0906 0370-0934 0370-0934 0370-0914 3030-0007 0350-0137 0350-0135 0350-0138 03580-24305 3101-0199 1450-0404 03580-06011 1990-0450 2100-0564 2100-3809 3101-0199 1990-0819 03580-60209 | | 1 1 1 1 1 1 1 | KNOBS ADAPTIVE SWEEP AMPLITUDE REF LEVEL BANDWIDTH DISPLAY SMOOTHING FOCUS COARSE FREQUENCY FREQUENCY SPAN INPUT SENSITIVITY DECAL INTENSITY MANUAL FINE FREQ POWER SWEEP MODE SWEEP MODE SWEEP TIME CONCENTRIC KNOB (SMOOTHING) VERNIER FINE FREQUENCY PUSHBUTTON-BASE PUSHBUTTON-BASE PUSHBUTTON-BEZEL SET SCREW LABEL: PUSHBUTTON, 1 DB LABEL: PUSHBUTTON, 10 DB LABEL: PUSHBUTTON, PLAIN PLATE-FRONT (SUB-PANEL) SWITCH-SLIDE LENS CAP FCM SHIELD DIODE-LED R: VAR, 100 K 20° R: VAR, 5 K SWITCH SLIDE LED BANK PANEL: FCM W/BEZEL RETROFIT KIT FOR FCM (MECH. TO ELECT.) | 28480 | 1460-1341 03580-09301 0370-1005 0370-2181 0370-2188 0370-1005 0370-3034 0370-2185 03580-67401 7120-3115 0370-1005 0370-2187 0370-2187 0370-2187 0370-2189 0370-0906 0370-2189 0370-0914 0350-0137 0350-0136 0350-0135 0350-0135 0350-0135 0350-0138 03580-0450 1450-0404 03580-0450 12100-0564 2100-0564 2100-0199 |
| 03580-00216 03580-01218 03580-23707 1440-0159 | | 1 2 2 | RACK MOUNT KIT (OPTION 003) FRONT PANEL (OPT 003) SIDE BRACKET (OPT 003) SIDE RAIL (OPT 003) OVAL HANDLE (OPT 003) | | |
| 000000000000000000000000000000000000000 | 0370-1005 0370-3034 0370-3034 0370-3034 0370-3034 03580-67401 0370-2185 0370-3006 0370-2994 0370-2187 0370-2187 0370-2188 0370-2188 0370-2188 0370-2188 0370-2188 0370-2096 0370-2994 0370-0914 0370-0934 0370-0934 0370-0914 0300-0007 0350-0135 0350-0135 0350-0135 0350-0135 0350-0135 0350-0138 03580-0137 03580-0139 | 0370-1005 0370-3034 0370-3034 0370-3034 0370-2185 03580-67401 1720-4008 0370-2994 0370-2187 0370-2187 0370-2184 0370-2184 0370-2189 0370-3006 0370-2189 0370-3006 0370-0914 0030-0007 0350-0137 0350-0136 0350-0135 0350-0138 03580-24305 101-0199 990-0450 101-0199 990-0819 03580-60209 | 1370-1005 1370-3034 1370-3185 13580-67401 17120-4008 1370-2187 1370-2187 1370-2187 1370-2189 1370-3006 1370-994 1370-9016 1370-9016 1370-0914 1030-0007 1350-0136 1350-0136 1350-0138 1350-0138 1350-0138 1350-0138 1350-0138 1350-0138 13580-0139 1101-0199 1450-0404 1580-0404 100-3809 1101-0199 1990-0450 1101-0199 1990-0450 1101-0199 1990-0819 13580-60209 13580-60209 13580-00216 13580-01218 23580-23707 2440-0159 2 | 1370-1005 1 | FOCUS TOURS TOUR |

AFOR S/N 1312A-00365 AND BELOW: ORDER 03580-00201 (STD) OR 03580-00204 (OPT. 002). See introduction to this section for ordering information *Indicates factory selected value

Revised: September 1987

SECTION VII TROUBLESHOOTING AND CIRCUIT DIAGRAMS

7-1. INTRODUCTION.

7-2. This section of the manual contains troubleshooting information and circuit diagrams for the Model 3580A Spectrum Analyzer. Included are troubleshooting information, information on factory selected components, functional block diagrams, schematic diagrams and component location diagrams.

7-3. TROUBLESHOOTING AND PREVENTIVE MAINTENANCE.

7-4. General Troubleshooting Procedures.

- 7-5. Troubleshooting information for the 3580A can be found in the functional block diagrams and circuit diagrams at the end of Section VII. An extensive set of notes, waveforms, and tables has been provided to help narrow the problems down from the functional block, to a board, and finally to a component.
- 7-6. Use the Overall Functional Block Diagram (Figure 7-1) to narrow the 3580A problem down into one of the four major functional blocks:
 - 1) Input Section
 - 2) Frequency and Sweep Section
 - 3) IF Section
 - 4) Display Section.

This diagram gives a good overall look at the 3580A operation. Once the diagram is understood, the failure symptoms alone may be adequate to lead you to the proper block. Other times, the output signals from the 3580A will suffice. For instance, the RECORDER X-AXIS and Y-AXIS outputs give an indication of proper instrument operation up to, but not including, the A7 Logic Board. The TRACKING OSC OUTPUT indicates if the Frequency and Sweep Section is working properly.

7-7. If the external control signals and front panel failure symptoms are not adequate to localize a problem to a particular block, remove the 3580A outer covers and check the appropriate input and output lines of each block. This will localize the problem to a block. The Analog Block Diagram (Figure 7-2), circuit schematics and associated notes can then be used to isolate the problem to the component.

7-8. A2 Board VTO Troubleshooting.

7-9. The A2 VTO is part of a complex feedback loop. If the VTO circuitry is not working properly, the feedback loop can be broken by applying approximately - 1.6 V dc to A2TP4. A 0 to +9 V dc signal supplied to the VTO ERROR AMP on the RED jumper lead to the A2 board should then cause the oscillator frequency to vary from 1.0 to 1.5 MHz (0 to 50 kHz Input Frequency). This signal can then be followed around the feedback loop to find the faulty components. Use the waveforms supplied with the A2 board to aid in this process.

7-10. A3 Board Troubleshooting.

7-11. This part of Section VII contains test procedures for the digital control circuitry of the A3 Sweep Board (Schematic 4). If the previous troubleshooting procedures indicate problems with the normal or adaptive sweep circuitry, perform these test procedures.

a. Position the 3580A front panel controls to:

| SWEEP MODE R | E | 3 | , |
|--------------|---|---|---|
|--------------|---|---|---|

Short A3TP1 to the gray jumper wire connected near the center of the A3 board (Don't remove the gray jumper).

- b. Adjust A3R54 (INTEGRATOR BALANCE) to verify that the output of the Ramp Integrator (A3TP1) can be adjusted from a positive to negative dc voltage. Readjust A3R54 for 0.000 volts ± .001 volts.
- c. Measure Vsg on the dual FET, A3Q1. Both FET's should have $Vsg \le 3 V dc$.
- d. Set switch S1 to the test position (UP position). Verify that CLOCK OUTPUT (A3U8 pin 11) is a TTL HIGH (≥ 2.0 V dc). Return S1 to the normal position.
- e. Remove the clock test jumper between Q18 and S1. Reposition:

| SWEEP MODERES |
|---------------|
|---------------|

f. Connect a logic clip to A3U5. Turn the 3580A POWER switch OFF then back to ON. The instrument should come up in state 000 or 100, where the C, B, and A state outputs are located on pins 13, 14 and 15 respectively of A3U5. If the instrument comes up in state 000, clock it to state 100 by momentarily switching A3S1 into, and then out of the test position. (This process will be called "clocking S1" from now on.)

g. Reposition the following front panel controls:

ADAPTIVE SWEEP OFF
RESOLUTION BANDWIDTH ... 100 Hz
FREQ. SPAN/DIV ... 2 KHz
SWEEP TIME/DIV ... 1 SEC/DIV

- h. Check the following:
 - 1. Collector of A3Q4: 10 volts ± .1 volts
 - 2. Collector of A3Q16: 0.0 volts ± .1 volts
 - 3. A3U5 pin 5: TTL LOW (as measured by logic clip).
 - 4. A3U5 pins 2, 3, 4, 6 and 9: TTL HIGH (as measured by logic clip).
 - 5. A3TP2: -.25 volts $\pm .02$ volts.
 - 6. A3TP3: $\pm .175$ volts $\pm .02$ volts.
 - 7. A3U8 pin 6: TTL HIGH (> 2.0 volts).
- i. Manually "clock" S1 once and verify that the state does not change from 100.
- j. Short A3TP3 to A3TP4. Verify that the voltage at A3TP11 can be changed from a negative to positive voltage by rotating A3R14. Readjust A3R14 so the voltage at TP11 is at the 0 V transition point. (In some cases it will alternate between positive and negative.)
- k. Check for proper source voltage on A3Q14. $(.1 < V_s < +4)$.
 - 1. Readjust A3R14* fully CCW. Reposition:

m. (L)RESP (A3U7 pin 5) should be a TTL HIGH. Verify that any one of the following will cause (L)RESP to go LOW.

If (L)RESP doesn't function properly, check the A8 board.

- n. In the following tests, the proper next state qualifiers are set up and the control logic is manually stepped to the next state by "clocking" S1 once. In each case the control logic should go to the next state only when all qualifiers are met and S1 is clocked.
- o. If the control logic fails to clock to the proper state, reset the logic to state 000 or 001 by selecting:

and momentarily turning the POWER switch OFF and then back to ON. Use Table 7-1 to reclock the control logic up to that state which will not go to the proper next state after clocking S1. Then recheck all the next state qualifiers, as given in Table 7-1 and test for proper inputs to the state flip—flops (U6 and U7). The J and K inputs to these flip—flops should correspond to the change the flip—flop will make on the next clock pulse. For instance, if a flip—flop's Q-output is to change from a 0 to a 1, its J input should be high. Likewise, if it is to change from a 1 to a 0, the K input should be high. If it is to stay at 1, the K input should be a 0. If it is to stay at 0, the J input should be a 0.

Table 7-1. Conditions for Single Stepping A3 Logic.

(Initial Setup: [Gray Jumper - TP1], [TP3 - TP4], A3R14* fully CCW, ADAPTIVE SWEEP - OFF, 100 Hz Bandwidth, 2 kHz/DIV, .1 SEC/DIV, RESET.)

| Present State | Next State | Conditions to go to next State | Next State Qualifiers |
|---------------|--------------|--------------------------------|--|
| CBA 0 000 | CBA 1 100 | SWEEP MODE: RESET | (L)SING - HIGH and (H)GEW - HIGH or (L)RESET - LOW |
| 1 100 | 2 101 | SWEEP MODE: SING | (H)DLYO - HIGH (L)RESET - HIGH |
| 2 101 | 3 111 | ADAP. SWEEP: CW | (L)RESP - HIGH |
| 3 111 | 4 110 | R14: CW* | CCMP - HIGH (L)RESP - HIGH |
| 4 110 | 5 010 | ADAP. SWEEP: CCW | (L)RESET LOW |
| 5 010 | 6 011 | R14: CCW* | CCMP - LOW |
| 6 011 | 7 001 | (Clock after delay) | (H)DLYO - HIGH |
| 7 001 | 2 101 | R14: CW* | CCMP - HIGH |

^{*}If A3R14 has a black casing, set it opposite to the setting given.

See Table 7-2. Notice also that the J and K inputs are not directly accessible on U7. All the inputs to each of the input AND gates must be high before there is a corresponding HIGH level given to one of the internal J or K inputs of the flip—flop.

Table 7-2. Excitation Table for J - K Flip—Flop.

| Q _t (Before Clock) | Q _t + 1 (After Clock) | J | К |
|-------------------------------------|--|------------|------------|
| 0 | 1 | 1 | don't care |
| 1 | 0 | don't care | 1 |
| 1 | 1 | don't care | 0 |
| 0 | 0 | 0 | don't care |

p. Reposition (Only those controls printed in BOLD require a change from the previous tests.)

| ADAPTIVE SWEEP | OFF |
|----------------------|----------|
| RESOLUTION BANDWIDTH | . 100 Hz |
| FREQ. SPAN/DIV | 2 KHz |
| SWEEP TIME/DIV | 0.1 SEC |
| SWEEP MODE | RESET |
| A3R14 fu | |

- q. State 100
 - 1. Clock S1, observe no change of state.
 - 2. Check the voltage at A3TP5 and A3TP6, it should be 0 V dc \pm .1 V.
 - 3. Short the A3 gray jumper to A3TP1 and short A3TP3 to A3TP4 if not already done.
 - 4. Reposition:

SWEEP MODE SING

- 5. Clock S1, and the logic should go to State 101.
- r. State 101
 - 1. Clock S1 and observe no change of state.
 - 2. Check for the following levels:

Collector A3Q4 : < - 8 V dc Collector A3Q16: 0 V dc \pm .1 V

A3TP5: < -7 V dcA3TP6: 0 V dc ± .1 V

A3TP8: TTL LOW (< .8 Vdc)

3. Reposition:

ADAPTIVE SWEEPCW

- 4. Clock S1 once, and the logic should go to State 111.
- s. State 111

- 1. Adjust A3R14 fully CCW*.
- 2. Clock S1 and observe no change of state.
- 3. Check for the following levels:

Collector A3Q16: $-9.9 \text{ V} \pm .1 \text{ V}$

A3TP5: < -7 V

A3TP8: TTL LOW (< .8 V)

4. Reposition:

ADAPTIVE SWEEPCCW

Clock S1 and observe no change of state.

- 5. Adjust A3R14 fully CCW. Clock S1 and observe no change of state.
- 6. Reposition:

ADAPTIVE SWEEPCW*

Adjust A3R14 fully CCW*. Clock S1 and observe no change of state.

- 7. Adjust A3R14 fully CW*. Clock S1, and the logic should go to state 110.
- t. State 110
 - 1. Clock S1 and observe no change of state.
 - 2. Remove the test lead between A3TP3 and A3TP4.

 The voltage at A3TP3 should be -.25 V
 ± .1 V.

 Replace the jumper.
 - 3. Check the following levels:

A3TP6: < -6 V

A3TP5: 0 V dc ± .1 V

Collector of A3Q16: 0 V. ± .1 V.

4. Adjust R14 fully CCW*.

Clock S1 and observe no change of state.

5. Reposition:

ADAPTIVE SWEEPCCW

Adjust R14 fully CW*.

- 6. Clock S1, and the logic should go to state 010.
- u. State 010
 - 1. Clock S1 and observe no change of state.
 - 2. Check for the following levels:

A3TP6: > +6 V A3TP5: 0 V ± .1 V A3TP8: +9.5 V ± .5 V

Collector A3Q16: $-9.9 \text{ V} \pm .1 \text{ V}$

3. Reposition:

RESOLUTION BANDWIDTH 1 Hz

- Connect an oscilloscope to the collector of A3Q11. Wait 5 seconds. The voltage should be a TTL HIGH (≥ 2 V dc).
- 5. Adjust A3R14 fully CCW*.
- 6. While watching the oscilloscope clock S1. The oscilloscope should indicate a TTL LOW (< .8 V) for a few seconds and then return HIGH. The logic state should be 011.</p>

v. State 011

1. Check the following levels:

A3TP5: $0 \text{ V dc} \pm .1 \text{ V}$ A3TP6: $0 \text{ V dc} \pm .1 \text{ V}$

Collector of A3Q16: $-9.9 \text{ V} \pm .1 \text{ V}$ A3TP8: TTL LOW (< .8 V)

2. Reposition:

RESOLUTION BANDWIDTH ... 100 Hz

- 3. Clock S1 and the control logic should go to state 001.
- w. State 001
 - 1. Clock S1 and observe no change of state.
 - 2. Check the following levels:

Collector A3Q16: - 9.9 V ± .1 V

A3TP5: <-7 V

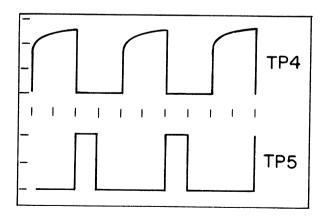
A3TP8: TTL LOW (< .8 V)

- 3. Adjust A3R14 fully CW*. Clock S1, and the control logic should go to State 101.
- x. Adjust R14 so that the voltage at A3TP11 is at the transition between a plus and minus voltage.
- y. Remove all test leads and replace the clock jumper. The 3580A should sweep normally. The penlift relay should "click" in single sweep mode and the output of the A3 RAMP GENERATOR (A3TP1) should be +5 volts nominal for a front panel display indication at the right graticule. If the LOG SWEEP mode will not work, see the A3 schematic notes.

7-12. A7 Board.

- 7-13. The A7 Board (03580-66507) is available as a rebuilt exchange board (03580-69507) through your local -hp-Sales and Service Office. Many times, however, the board can be repaired without purchasing an exchange board. The following procedure will aid in determining whether the A7 board or the analog circuits preceeding the A7 board are at fault.
- a. Connect the 3580A X-AXIS output on the rear panel to the X deflection EXT INPUT of an oscilloscope. A scope with variable persistance works best but is not absolutely necessary. Connect the 3580A Y-AXIS output to the vertical input of the scope. This procedure effectively half splits the 3580A for troubleshooting purposes.
 - 1. If the signal seen of the scope is correct and the signal seen on the 3580A display is incorrect then the problem is in the A7, A8, or A13 boards. If the signal seen on the scope is incorrect DO NOT troubleshoot the A7 board until repairs are made to preceeding circuitry. (See Functional Block Diagram in the Operating and Service Manual.)
 - 2. If the scope presentation is good but the 3580A display is incorrect, check A7TP1. If the presentation is bad there then troubleshoot the A7 board, otherwise troubleshoot A8 or A13.
- 7-14. Troubleshooting the A7 board.
- a. Check A7Q2, Q4, Q6, Q8, and Q9. If these parts are P/N 1853-0098 replace all 5 of them with P/N 1853-0010. The new type is much more reliable and is being used in all instruments with serial numbers above 1415A01276.
- b. Check A7TP4 and A7TP5. They should look similar to the figure shown below.

Horiz 2 μ sec/div Vert .2 V/div (with 10:1 probe)



The frequency must be 55 K - 70 kHz! If the frequency is off check A8TP9. The clock frequency is determined by the A8 board.

- c. Clean the A7 board connector with alcohol and see if this eliminates the problem.
- d. Flex the board slightly. Occasionally the mounting screws on the A7 board apply pressure in such a way as to intermittantly open traces.
- e. If random glitches appear on the display try holding in the CLEAR WRITE button. If the glitches are still present probably a RAM is bad. Short pin 12 of each RAM to ground one at a time. When the glitches disappear replace that RAM.
- f. Depress the STORE button, and then release STORE (depress again). If the display appears to shift one or more centimeters then replace A7U58.
- g. If an unnatural phenomenon appears repetitively at the same location on the 3580A display (may appear intermittantly) try paralleling the black 99.25 K resistors with a 50 K resistor one at a time. When the 50 K resistor creates an anomaly at the same point, replace the paralleled 99.25 K resistor.
- h. Verify that the CLOCK (A7 pin 8, waveform 2) is present. Also, verify that all the internal clocks are operating as indicated by the clock waveforms supplied with the A7 schematic. If these are working properly, check the Y-AXIS A to D and X-AXIS A to D and output D to A for proper operation.
- 7-15. The two A to D converters are basically counters which count up or down until their digital output is equal to the analog input. The digital output is fed back around to the input via a D to A converter. This feedback signal is then compared with the input signal to control the count of the A to D. By verifying that the feedback signal of the A to D converter is approximately equal to the input signal, the converters can be tested. This feedback signal is available at TP2 and TP3 of the X-AXIS A to D and Y-AXIS A to D respectively. Use MANUAL SWEEP mode when checking these converters.
- 7-16. The output D to A converter (U53, U61 to U63, and associated resistors) should also be checked for proper operation. It is basically a summing device which converts the digital output from the memory into currents proportional to their digital value. U53 sums these currents into an analog signal present at A7TP1. By using a small FREQ. SPAN/DIV (5 Hz) and a wide bandwidth (300 Hz), the memory can be loaded with a constant value so that the input to the D to A is a constant. Use A4TP4 to determine the input signal level to the memory of the A7 board, and test for proper output.
- 7-17. As a last test, verify that the U56 and U57 binary counters are receiving a clock pulse at pin 15, and that they are counting.
- 7-18. If these tests fail, it is probably best to exchange your board for a rebuilt exchange board (03580-69507).

This board is available through your local -hp- Sales and Service Office. Exchange credit will be given if you return your original 03580-66507 or 03580-69507 board. Please remember the A7 board uses CMOS integrated circuits extensively and proper handling is important. DO NOT return A7 boards in a plastic bag.

7-19. High Voltage Power Supply.

- 7-20. The A11A1 and A11A2 High Voltage Power Supply boards operate in conjunction with the feedback control circuitry on the A8 board to produce the regulated high voltage for the CRT. One winding of the high voltage transformer (A11A2T1) is further used to produce the + 158 V dc supply for the Deflection Amplifiers. The + 158 V dc regulator is located on the A8 board.
- 7-21. The high voltage transformer is driven by the high voltage oscillator consisting of A8Q21 and associated circuitry. Oscillation is sustained by positive ac feedback from a tertiary winding on the transformer to the base of A8Q21. Note that the 55 kHz to 65 kHz signal from the collector of A8Q21 serves as the primary clock for the Sweep Generator (A3) and Digital Storage (A7) boards.
- 7-22. The high voltage output level is determined by the drive level of the high voltage oscillator. This is controlled by dc feedback from the CRT cathode supply. The feedback voltage is fed through divider resistors A11A2 R5 and R6 and applied to the A8 board (A8J1) through a flying red lead. To prevent damage to the high voltage supply, a safety interlock disables the high voltage oscillator when the feedback lead is unplugged from A8J1. On the A8 board, the feedback voltage is processed by control amplifiers Q23 and Q22 and applied to the base of A8Q21 through the tertiary feedback winding of the high voltage transformer.
- 7-23. The voltage at the cathode of the CRT (CRT pin 2) is normally about 2,900 V and is not critical. Note, however, that the intensity grid voltage (CRT pin 3) cannot be more than 30 or 40 volts more negative than the cathode voltage. If it is, the display will be blanked.

WARNING

Do not attempt to measure the difference between the cathode and intensity grid with a floating voltmeter. Measure the absolute voltage at each point with a high voltage probe and then calculate the difference. These voltages can cause serious injury or even death if proper care is not taken.

7-24. The A11A1 and A11A2 boards have dangerous voltages which make troubleshooting both hazardous and difficult. Generally, the safest and most efficient approach is to remove all power from the 3580A and check these boards with an ohmmeter. Note that A11A1 CR1 and CR2

each contain many diodes in series and their forward resistance (as measured with -hp- Model 412A) can be as high as 50 megohms while their leakage (reverse) resistance will generally be about 100 megohms. The primary windings of the high voltage transformer and the CRT heater windings have a dc resistance of only a few ohms. The other two secondary windings have dc resistances of 100 to 200 ohms.

7-25. Crystal Replacement.

7-26. If it is found that the A5 filters or A2 crystal oscillator need a new crystal, the crystal cannot be exchanged individually but must be exchanged as a matched set of crystals and resistors. For this reason, the 03580-69505 exchange assembly, and 03580-69515 replacement assemblies are available. These assemblies consist of:

| Item | Qty | Description |
|------|-----|--|
| 1 | 1 | A5 IF Filter Board, 03580-66505 (Exchange Ass'y, 03580-69505, contains a rebuilt A5 Board; Replacement Ass'y, 03580-69515 contains a new A5 Board). |
| 2 | 1 | 0410-0480 Crystal Set (This is a matched set of six crystals. Five of the crystals are already part of Item 1; the sixth cyrstal is for the A2 Tracking Oscillator). |
| 3 | 1 | A resistor matched to the sixth crystal supplied by Item 2. |

7-27. If you need a new crystal, order the exchange or replacement assembly through your local -hp- Sales and Service Office. Exchange credit can only be given if you return both your old 03580-66505 board and the appropriate crystal and matching resistor from the A2 board. Always use care when removing these crystals, as undue stress on the leads can damage the glass encapsulation.

NOTE

This 03580-69505 exchange assembly is intended as an aid in crystal replacement. It is not intended to be used in place of repairing other components on the A5 board (03580-66505). The 03580-69515 replacement assembly is provided for those who want to purchase a new A5 Assembly and do not wish to use the exchange program.

7-28. CRT Replacement.

WARNING

Use care when handling the CRT. Undue stress can cause dangerous implosion of the tube.

When shipping the CRT, follow the shipping instructions outlined in the Cathode Ray Tube Warranty information at the beginning of this manual.

- 7-29. If it is determined that the CRT needs replacement, fill out the Cathode-Ray Tube Failure Report supplied at the beginning of this manual. To remove the CRT, use the following procedure:
 - a. Remove the front panel bezel (black hood).
- b. Remove the metal support and plastic lens (under bezel).
- c. Remove the rear protective CRT cover (on rear panel).
 - d. Remove the CRT rear tube socket.
 - e. Remove the bottom instrument cover.
- f. Through a hole in the left side, at the rear of the instrument, unscrew the CRT neck clamp using a long shaft screwdriver.
- g. Slide the CRT out. This may require moderate force. On instruments which have been used extensively, it may be necessary to cut the white CRT mounting tape to separate the CRT from the CRT tube shield. This tape is located on the top and bottom of the CRT, one inch to the rear of the CRT face.

Send the CRT and Failure Report to your local -hp- Sales and Service Office.

NOTES

- 1. If the CRT Mounting Tape is cut, replace it with a new mounting tape -hp- Part No. 0460-1115.
- 2. When reinstalling the CRT, push the CRT slightly forward while tightening the CRT neck clamp. This secures the plastic lens in front of the CRT

7-30. Battery Replacement (Option 001 only).

7-31. Each of the five battery sticks can be replaced individually. Do not attempt to replace individual cells within a battery stick. When ordering a new battery stick, order either the center tapped stick (-hp- Part No. 1420-0203) or the regular stick (-hp- Part No. 1420-0202).



Do not remove the individual battery sticks until the entire battery pack has been removed from the instrument. The battery pack can be removed by disconnecting the battery plug (P1) and removing the four screws holding the pack to the side of the instrument chassis. The individual battery sticks may short out against the sides of the instrument if the entire battery pack is not first removed.

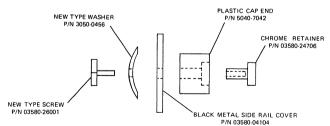
7-32. To determine which battery stick is faulty, place the 3580A on CHARGE for 14 hours and then run the 3580A on battery power until the undervoltage relays shut the battery power off. (Good batteries will run for 5 hours without a recharge). Measure the voltage across each battery stick. The nominal voltage should be approximately 5 volts per stick. Test for the stick which is lower in voltage than the other battery sticks. A bad stick will differ from the other battery sticks by .5 or more volts.

7-33. The normal warranty period on batteries is 90 days. Proper operation implys that the battery, operated under normal temperatures and load, will charge from a state of complete discharge in 14 hours, and will then power the instrument for 5 hours of continuous and normal use.

7-34. Cleaning and Lubricating Rotary Switches.

7-35. Faulty switches can cause intermittent performance, spurious responses, noise, and many other annoying problems. Tests have shown that the typical operating life of a switch is 25,000 operations or more. With proper cleaning and lubrication, this life may be extended to as much as 100,000 or more operations. Freon TF cleaner (-hp- Part No. 8500-0232) is available for cleaning switches. Electrotube 2G (-hp- Part No. 5060-6086) is available for lubricating high impedance switches. Electrotube 2A (-hp- Part No. 6040-0300) is available for lubricating low impedance switches. Follow the instructions given with these cleaners, -hp- Service Note M45B (available from your local -hp-Sales and Service Office) also gives detailed information on how to use these cleaners.

7-36. Repairing Handles. (For S/N 1415A00975 and below) Anytime a loose or broken handle is repaired the new type screw should be used. The illustration indicates the assembly order. The screw which attaches the L shaped plastic piece remains the same.



In order to repair both handles on one instrument the following new parts are needed.

4 ea screw P 4 ea washer P

P/N 03580-26001 P/N 3050-0456

7-37. FACTORY SELECTED COMPONENTS.

7-38. Certain components within the 3580A are individually selected at the factory to compensate for slightly varying circuit parameters. These components are identified by an asterisk (*) in the parts list and schematic diagrams.

Table 7-3. Factory Selected Components.

| | T Tackery Sele | tea components. |
|--|--|---|
| Component | Function | Value Range |
| A3R88* | Controls Sweep Time/Div. Increasing A3R88* increases sweep time. Decreasing A3R88* decreases sweep time. | 3 13.7 kΩ ± 1% 1/8 W $1 15.4 kΩ ± 1% 1/8 W$ |
| A5R1 * A5R26 * A5R48 * A5R71 * A5R73 * | Assures equal gain within all crystal stages. Selected by value of the crystal resistance: Crystal Resistance $<91~\mu$ $91~\Omega~to~110~\Omega$ $111~\Omega~to~130~\Omega$ $131~\Omega~to~150~\Omega$ $151~\Omega~to~170~\Omega$ $171~\Omega~to~190~\Omega$ $191~\Omega~to~210~\Omega$ | |
| A6R41* | Adjusts +10 V power supply to 10 V ± .050 V. Increasing A6R41* increases the voltage. Decreasing A6R41* decreases the voltage. | 243 Ω to 1.96 kΩ 1/8 W typical: 1.05 kΩ |
| A7R109* | Adjusts positive pulse width at A7TP5 to 1.0 to 1.4 μsec (Rev. A) or 2.0 to 2.4 μsec (Rev. B). Increasing A7R109* increases pulse width. Decreasing A7R109* decreases pulse width. | |
| A7R111* | Adjusts positive pulse at A7TP4 to 3.5 to 3.9 µsec. Increasing A7R111* increases the pulse width. Decreasing A7R111* decreases the pulse width. | 24.9 kΩ to 41.2 kΩ 1/8 W typical: 32.4 kΩ |
| A11A1R2* | Gives proper intensity limit adjustment. | 100 kΩ or 1 MΩ typical: 100 kΩ |
| A18R3* | | A O Ω B 51.1 Ω ± 1% 1/8 W C 100 Ω ± 1% 1/8 W D 150 Ω ± 1% 1/8 W E 182 Ω ± 1% 1/8 W F 221 Ω ± 1% 1/8 W G 267 Ω ± 1% 1/8 W H 332 Ω ± 1% 1/8 W I 392 Ω ± 1% 1/8 W J 475 Ω ± 1% 1/8 W K 562 Ω ± 1% 1/8 W |
| A34R46 | END SET" control. | 11 k Ω ± 1% 1/8 W 12.1 k Ω ± 1% 1/8 W 13.3 k Ω ± 1% 1/8 W 14.7 k Ω ± 1% 1/8 W 15.4 k Ω ± 1% 1/8 W |

A typical value is given for each. Table 7-3 is a list of the factory selected components, functions, and value ranges. A detailed description of selecting A3R88* is given in Paragraph 7-39. The other components will usually not require reselection. (The crystal padding resistors are factory selected and cannot be selected in the field. See Crystal Replacement, Paragraph 7-19).

7-39. A3R88* should be reselected if the frequency ramp integrating capacitor (C1) is changed (See Schematic 4). To

select A3R88*, select the following front panel control settings:

| ADAPTIVE SWEEP | | | | | | | | | OFF | 7 |
|----------------|--|------|--|--|--|--|--|---|-----|---|
| SWEEP TIME/DIV | | | | | | | | 1 | SEC | , |
| SWEEP MODE | | | | | | | | | REF |) |

Measure the time interval between the negative and positive voltage transition at A3TP5 with an electronic counter. For the -hp- 5326A Counter, the controls should be:

Sample Rate:

Fast

Function:

T.I. A to B

Multiplier:

.1 sec. Slope -

Channel A:

Slope -D.C. Atten X1

Level: set to trigger on negative

edge of pulse.

Channel B:

Slope + D.C.

Atten X1

Level: set to trigger on positive

edge of pulse.

BNC Input:

Com

The time interval should be 10.4 to 10.6 sec. The other sweep times can be easily tested at this time. The time interval should be $10.5 \times SWEEP TIME/DIV (\pm 5\%)$.

7-40. SCHEMATIC DIAGRAMS.

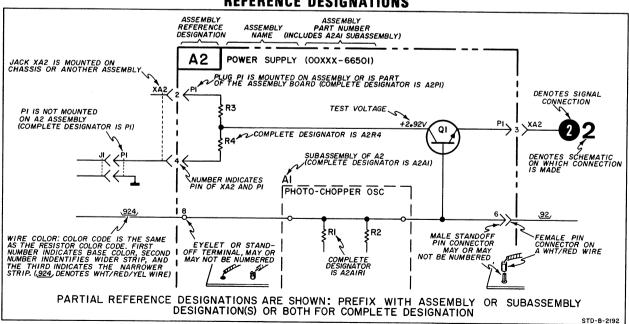
7-41. The schematic diagrams, Figure 7-3 through 7-12 show the detailed circuits of the Model 3580A. Each

schematic is assigned a numerical callout (1 through 10) which is used for referencing. The schematics are arranged to provide as much signal continuity as possible and assemblies do not necessarily appear in the order of their reference designations. Refer to Table 7-4 for a complete cross reference listing. Refer to the General Schematic Notes for further information concerning the schematic diagrams.

Table 7-4. Assembly Cross Reference.

| Assembly Number | Assembly Title | Schematic Number |
|------------------------|--------------------------|---------------------|
| A2 (03581-66502) | VTO Tracking Oscillator | 6 |
| A3 (03580-66503) | Main and Log Sweep | 4 |
| A4 (03581-66504) | Detector | 3 |
| A5 (03580-66505) | IF Filter | 2 |
| A6 (03580-66506) | Low Voltage Power Supply | 9 |
| A7 (03580-66507) | | _ |
| or | | |
| (03580-69507) | Digital Storage | 7 |
| A8 (03580-66508) | Control Board | 8 |
| A9 (03580-66509) | | |
| (Standard) | Input Circuits | 1 |
| (03580-66519) | | |
| (Option 002) | | |
| A10 (03580-66510 | Connector Board | 7 |
| A11A1 (03580-66531) | High Voltage | 8 |
| A11A2 (03580-6653·2) | • | - |
| and | | |
| (03580-66537) | HV Transformer | 8 |
| A13 (03580-66513) | Deflection Amp, | 8 |
| A14 (03580-66514) | Bandwidth/Sweep Time | 5 |
| A15 (03580-66515) | Freq Span/Sweep Mode | 5 |
| A16 (03580-66516) | Combining Board | 5 |
| A18 (03581-66518, Opt. | | |
| 002 only) | Balanced Input | 1 |

REFERENCE DESIGNATIONS



GENERAL SCHEMATIC NOTES

- 1. PARTIAL REFERENCE DESIGNATIONS ARE SHOWN. PREFIX WITH ASSEMBLY OR SUBASSEMBLY DESIGNATION(S) OR BOTH FOR COMPLETE DESIGNATION.
- 2. COMPONENT VALUES ARE SHOWN AS FOLLOWS UN-LESS OTHERWISE NOTED.

RESISTANCE IN OHMS
CAPACITANCE IN MICROFARADS
INDUCTANCE IN MILLIHENRYS

3. DENOTES EARTH GROUND.
USED FOR TERMINALS WITH NO LESS THAN A
NO. 18 GAUGE WIRE CONNECTED BETWEEN
TERMINAL AND EARTH GROUND TERMINAL OR
AC POWER RECEPTACLE.

4. DENOTES FRAME GROUND.
USED FOR TERMINALS WHICH ARE PERMANENTLY CONNECTED WITHIN APPROXIMATELY
0.1 OHM OF EARTH GROUND.

5. DENOTES GROUND ON PRINTED CIRCUIT ASSEMBLY. (PERMANENTLY CONNECTED TO FRAME GROUND).

6. — DENOTES ASSEMBLY.

7. DENOTES MAIN SIGNAL PATH.

DENOTES FEEDBACK

10. DENOTES FRONT PANEL MARKING.

11. DENOTES REAR PANEL MARKING.

12. DENOTES SCREWDRIVER ADJUST.

13. *A AVERAGE VALUE SHOWN, OPTIMUM VALUE SE-LECTED AT FACTORY. THE VALUE OF THESE COMPONENTS MAY VARY FROM ONE INSTRU-MENT TO ANOTHER. THE METHOD OF SELECTING THESE COMPONENTS IS DESCRIBED IN SECTION V OF THIS MANUAL.

14. DENOTES SECOND APPEARANCE OF A CONNECTOR PIN.

15. \924 \int DENOTES WIRE COLOR: COLOR CODE SAME AS RESISTOR COLOR CODE. FIRST NUMBER IDENTIFIES BASE COLOR, SECOND NUMBER IDENTIFIES WIDER STRIP, THIRD NUMBER IDENTIFIES NARROWER STRIP. (e.g. \924 = WHITE, RED, YELLOW.)

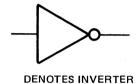
17. ALL RELAYS ARE SHOWN DEENERGIZED.

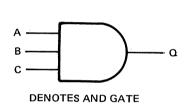
- 18. WAVEFORMS AND AC VOLTAGE MEASUREMENTS WERE MADE WITH RESPECT TO CHASSIS GROUND USING AN OSCILLOSCOPE WITH A 10:1 DIVIDER PROBE (10 MEGOHM, 10 pF). THE VOLTAGE LEVELS SHOWN ON THE WAVEFORMS ARE ACTUAL VOLTAGE LEVELS AND ARE NOT TO BE CONFUSED WITH OSCILLOSCOPE SETTING. THE VOLTAGE LEVELS SHOWN ARE NOMINAL AND MAY VARY FROM ONE INSTRUMENT TO ANOTHER. A VARIATION OF ± 10 % IN MEASUREMENTS SHOULD BE ALLOWED.
- 19. DC VOLTAGE LEVELS WERE MEASURED WITH RESPECT TO CIRCUIT GROUND USING A VTVM WITH 10 MEGOHM

INPUT IMPEDANCE. THE VOLTAGE LEVELS SHOWN ARE NOMINAL AND MAY VARY FROM ONE INSTRUMENT TO ANOTHER DUE TO CHANGE IN TRANSISTOR CHARACTERISTICS. A VARIATION OF ± 10 % SHOULD BE ALLOWED.



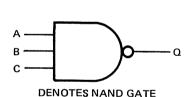
DENOTES BUFFER





| 0 | 0 | 0 | 0 |
|---|---|---|---|
| 0 | 0 | 1 | 0 |
| 0 | 1 | 0 | 0 |
| 0 | 1 | 1 | 0 |
| 1 | 0 | 0 | 0 |
| 1 | 0 | 1 | 0 |
| 1 | 1 | 0 | 0 |
| 4 | | 4 | |

ABCQ

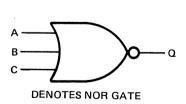


| Α | В | С | Q |
|---|---|---|---|
| 0 | 0 | 0 | 1 |
| 0 | 0 | 1 | 1 |
| 0 | 1 | 0 | 1 |
| 0 | 1 | 1 | 1 |
| 1 | 0 | 0 | 1 |
| 1 | 0 | 1 | 1 |
| 1 | 1 | 0 | 1 |

1 1 1 0

ABCQ

0 0 0 1



| U | U | U | • | |
|---|---|---|---|--|
| 0 | 0 | 1 | 0 | |
| 0 | 1 | 0 | 0 | |
| 0 | 1 | 1 | 0 | |
| 1 | 0 | 0 | 0 | |
| 1 | 0 | 1 | 0 | |
| 1 | 1 | 0 | 0 | |
| 1 | 1 | 1 | 0 | |



| | D | u | |
|---|---|---|--|
| 0 | 0 | 0 | |
| 0 | 1 | 1 | |
| 1 | 0 | 1 | |
| 1 | 1 | 0 | |
| 1 | 0 | 1 | |

DENOTES EXCLUSIVE OR GATE

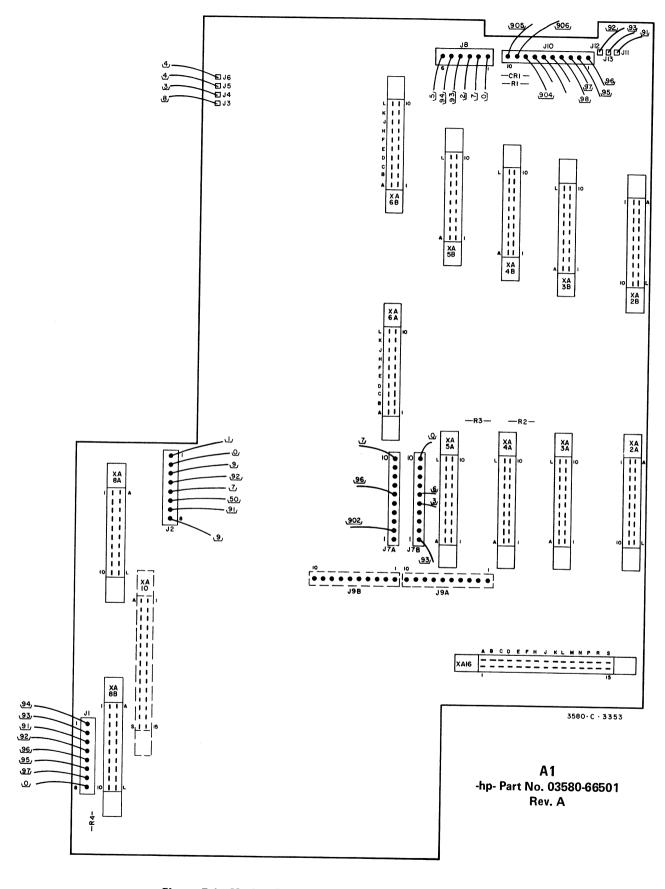


Figure 7-1. Mother Board (A1) Component Location Diagram.

GENERAL WAVEFORM INFORMATION

3580A Control Settings

These waveforms were made with the 3580A INPUT SENSITIVITY switch in the CAL position. The FREQUENCY and SWEEP controls are in the MANUAL mode, 300 Hz Bandwidth, and adjusted to read the 10 KHz harmonic of the CAL signal. The values given are those which would be observed for a full scale screen display of this 10 KHz signal. Set the dBv/LIN - dBm 600 Ω switch to dBv/LIN

(For Option 002 set the dBm $900~\Omega/\text{LIN}$ - dBm $600~\Omega$ switch to dBm $900~\Omega$. Waveform Ω is of slightly less magnitude for Option 002 instruments when using the LOG mode.)

⑫

These waveforms were made with the 3580A front panel controls adjusted as follows:

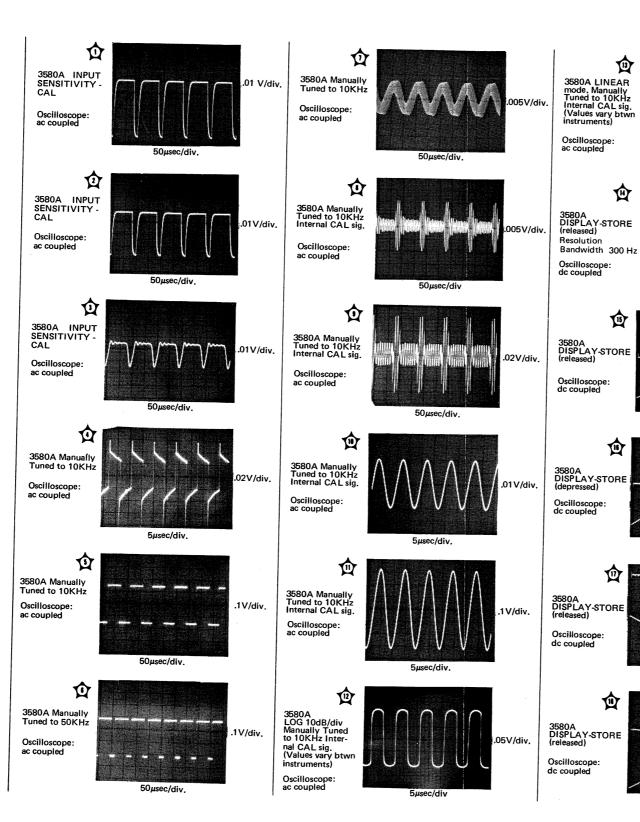
| ADAPTIVE SWEEP | OF |
|----------------------|--------------------|
| DISPLAY BLA | NK STORE released |
| | STORE as indicated |
| AMPLITUDE MODE | LOC 10 dp./plv |
| AMPLITUDE REF LEVEL | ALOG TO UBV/DIV |
| dBv/LIN - dBm 600 Ω | INDRIVIAL |
| INPLIT SENSITIVITY | aBV/LIN |
| INPUT SENSITIVITY | CAL |
| VERNIER (Amplitude) | |
| EDECLIENCY | (Fully CW) |
| FREQUENCY | 10,0 kHz |
| START - CTR | CTR |
| RESOLUTION BANDWIDTH | |
| | |
| FREQ. SPAN/DIV | 0.5 KHz |
| | I SEC: |
| SWEEP MODE | |
| 24 | |

Option 002: Set dBm 900 Ω/LIN - dBm 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL.

The front panel ZERO CAL was adjusted to give a display of the 10 KHz CAL signal in the center of the screen. It is easiest to set this adjustment in the MANUAL mode, and then switch to the REPetitive sweep mode to measure the waveforms.

Oscilloscope Settings

All waveforms were recorded using a 10:1 divider probe on the oscilloscope inputs. The vertical amplitude sensitivity is the actual amplifier setting and does not include the X10 multiplier introduced by the probe. All dual traces were made with the oscilloscope in the chopped mode and triggered by Channel B.



3580A LINEAR

mode, Manually Tuned to 10KHz Internal CAL sig. (Values vary btwi

够

⑩

⑫

⑩

Chan B

.2 V/div

←ov

Chan A

.1 V/div

Chan A

Chan B .2V/div

Chan A .2V/div

٠0٧

Chan A 2V/div

Chan B 2V/div

Chan A 2V/div

Chan B 2V/div

-0V

⊢0V

⊢0V

⊷nv

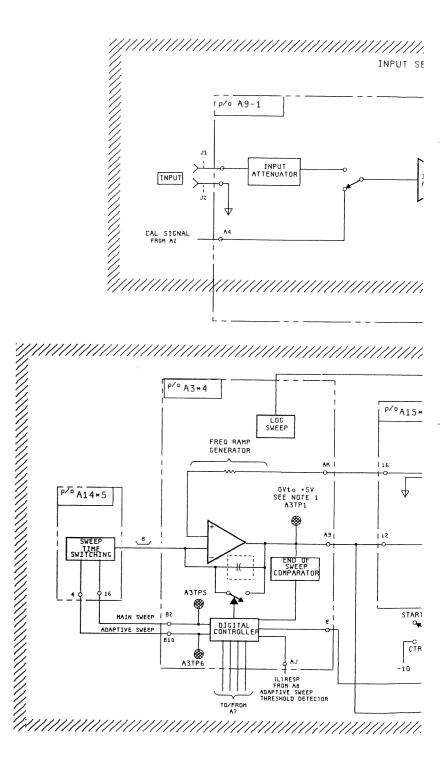
5msec/div

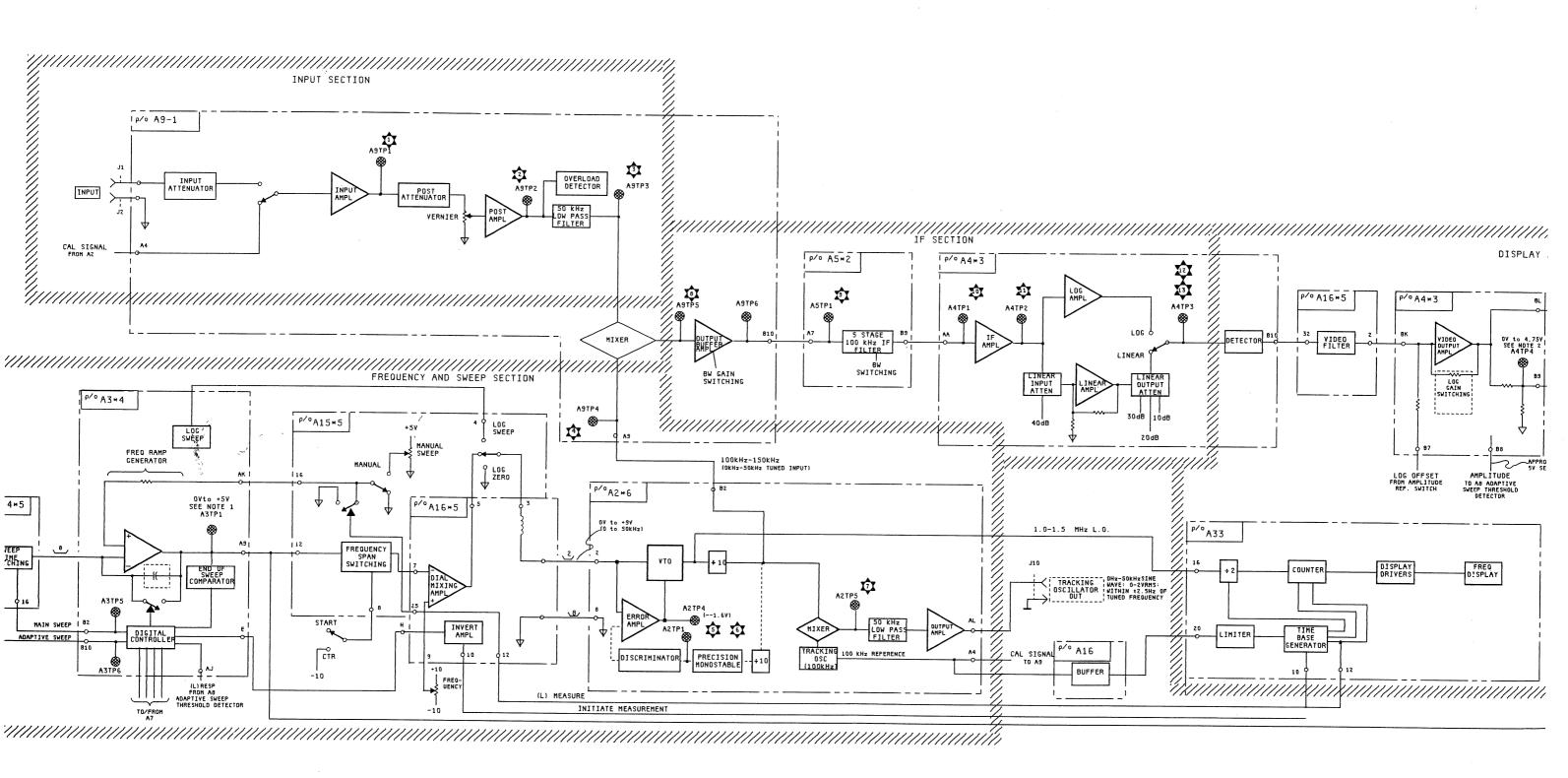
2msec/div

2msec/div

2msec/div

Oscilloscope





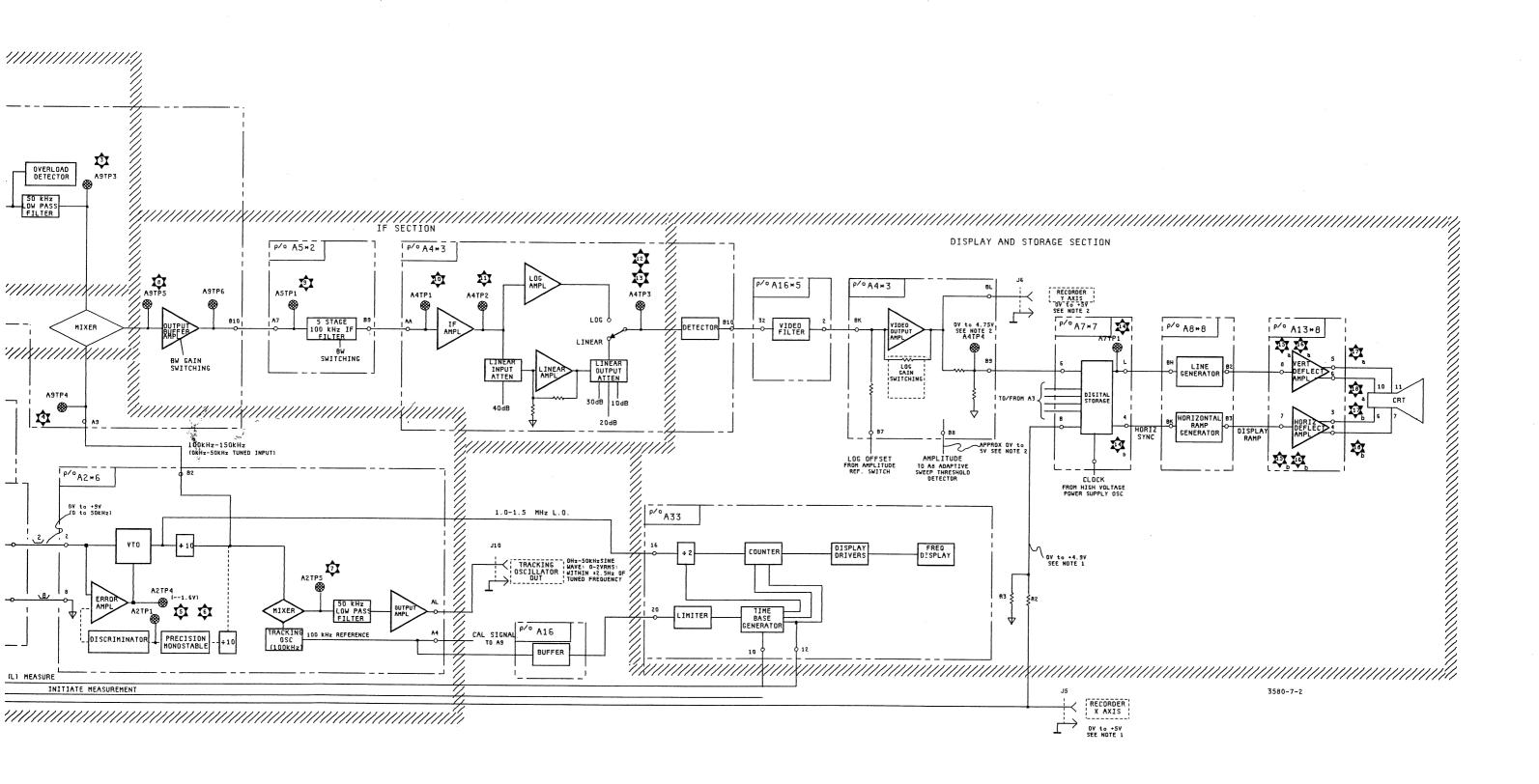


Figure 7-2. Functional Block Diagram.

Table 1. Input Circuit Amplitude Levels and Gains for Full Scale Sine Wave Inputs. (Amplitude Ref Level - Normal). Note: All voltages in RMS.

| | , | | | | | | | | | | |
|--|--|--|--|---|--|---|--|--|--|---|---|
| INPUT SENSITIVITY | + 30 dB/20 V | + 20 dB/10 V | + 10 dB/2 V | 0 dB/1 V | - 10 dB/0.2 V | 20 dB/0.1 V | 30 dB/20 mV | 40 dB/10 mV | - 50 dB/2 mV | - 60 dB/1 mV | - 70 dB/0.2 m |
| Maximum Input Log (dBm 600 Ω) Log (dBV) Linear Log (dBm 900 Ω)* | 24.5 V 31.6 V 20.0 V 30.0 V | 7.75 V 10.00 V 10.00 V 9.49 V | 2.45 V 3.16 V 2.00 V 3.00 V | .775 V 1.000 V 1.000 V .949 V | .245 V .316 V .200 V .300 V | 77.5 mV .1 V .1 V 94.9 mV | 24.5 mV 31.6 mV 20.0 mV 30:0 mV | 7.75 mV 10.00 mV 10.00 mV 9.49 mV | 2.45 mV 3.16 mV 2.00 mV 3.00 mV | .775 mV 1.000 mV 1.000 mV .949 mV | .245 mV .316 mV .200 mV |
| Input Attenuator Input Attenuator Out (Gate of A9Q1) Log (dBm 600 \Omega) Log (dBV) Linear Log (dBm 900 \Omega)* | - 60. dB 24.5 mV 31.6 mV 20.0 mV 30.0 mV | 40. dB 77.5 mV .1 V .1 V 94.9 mV | 40. dB 24.5 mV 31.6 mV 20.0 mV 30.0 mV | - 20. dB 77.5 mV .1 V .1 V 94.9 mV | - 20. dB 24.5 mV 31.6 mV 20.0 mV 30.0 mV | 0 dB 77.5 mV .1 V .1 V 94.9 mV | 0 dB 24.5 mV 31.6 mV 20.0 mV 30.0 mV | 0 dB 77.5 mV 10.0 mV 10.0 mV 9.49 mV | 0 dB 2.45 mV 3.16 mV 2.00 mV 3.00 mV | 0 dB .775 mV 1.000 mV 1.000 mV .949 mV | 0 dB .245 mV .316 mV .200 mV .300 mV |
| Input Amp Gain Input Amp Out (A9TP1) Log (dBm 600 Ω) Log (dBV) Linear Log (dBm 900 Ω)* | 1.8 dB 30.1 mV 38.9 mV 24.6 mV 36.9 mV | 1.8 dB 95.6 mV .123 V .123 V .117 V | 1.8 dB 30.1 mV 38.9 mV 24.6 mV 36.9 mV | 1.8 dB 95.6 mV .123 V .123 V .117 V | 1.8 dB 30.1 mV 38.9 mV 24.6 mV 36.9 mV | 1.8 dB 95.6 mV .123 V .123 V .117 V | 1.8 dB 30.1 mV 38.9 mV 24.6 mV 36.9 mV | 1.8 dB 9.56 mV 12.3 mV 12.3 mV 11.7 mV | 1.8 dB 3.01 mV 3.89 mV 2.46 mV 3.69 mV | 21.8 dB 9.56 mV 12.3 mV 12.3 mV 11.7 mV | 21.8 dB 3.01 mV 3.89 mV 2.46 mV 3.69 mV |
| Post Attenuation Log (dBm 600 Ω) Log (dBV) Linear Log (dBm 900 Ω)* Post Attenuator Out (Base A9Q6) Log Linear | 2.8 dB 5.0 dB 5.2 dB 4.6 dB | - 12.8 dB - 15.0 dB - 15.0 dB - 14.6 dB | 2.8 dB - 5.0 dB - 5.2 dB - 4.6 dB | - 12.8 dB 15.0 dB 15.0 dB 14.6 dB | 2.8 dB 5.0 dB 5.2 dB 4.6 dB | 12.8 dB - 15.0 dB - 15.0 dB 14.6 dB | 2.8 dB - 5.0 dB 5.2 dB 4.6 dB | 12.8 dB 15.0 dB 15.0 dB 14.6 dB | 2.8 dB 5.0 dB 5.2 dB 4.6 dB | - 12.8 dB - 15.0 dB - 15.0 dB - 14.6 dB | - 2.8 dB - 5.0 dB - 5.2 dB - 4.6 dB |
| Post Amp Gain Post Amp Out (A9TP2) Log Linear | 13.5 mV 13.2 dB 100. mV 61.9 mV | 13.2 dB 100. mV 100. mV | 13.5 mV 13.2 dB 100. mV 61.9 mV | 21.8 mV 13.2 dB 100. mV 100. mV | 13.5 mV 13.2 dB 100. mV 61.9 mV | 21.8 mV 13.2 dB 100. mV 100. mV | 13.5 mV 13.2 dB 100. mV 61.9 mV | 2.18 mV 33.2 dB 100. mV 100. mV | 1.35 mV 33.2 dB 100. mV 61.9 mV | 2.18 mV 33.2 dB 100. mV 100. mV | 1.35 mV 33.2 dB 100. mV 61.9 mV |
| Low Pass Filter Out (A9TP3) Log Linear | 50. mV 30.8 mV | 50. mV 50. niV | 50. mV 30.8 mV | 50. mV 50. mV | 50. mV 30.8 mV | 50. mV 50. mV | 50. mV 30.8 mV | 50. mV 50. mV | 50. mV 30.8 mV | 50. mV 50. mV | 50. mV 30.8 mV |
| Total Gain Log (dBm 600 Ω) Log (dBV) Linear Log (dBm 900 Ω) • | 47.8 dB 50. dB 50.2 dB 49.6 dB | - 37.8 dB - 40. dB - 40. dB - 39.6 dB | - 27.8 d8 30. d8 30.2 d8 29.6 d8 | - 17.8 dB - 20. dB - 20. dB - 19.6 dB | 7.8 d8 10. d8 10.2 d8 9.6 d8 | + 2.2 dB 0 dB 0 dB + 0.4 dB | + 12.2 dB + 10. dB + 9.8 dB + 10.4 dB | + 20. dB + 20. dB + | + 30. dB + 29.8 dB | +42.2 dB +40. dB +40. dB +40.4 dB | + 52.2 dB + 50. dB + 49.8 dB + 50.4 dB |

*Option 002 only.

Table 2. Approximate IF Level Changes with Bandwidth.
(Full Scale Sine Wave Input, LOG Mode, Manually Tuned to Input Frequency)

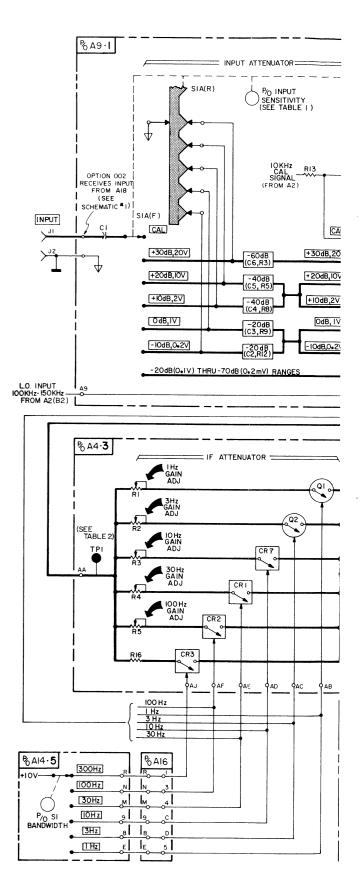
| Bandwidth | I.F. Input A9TP6, A5TP1 | I.F. Output A5 pin B9, A4TP1 |
|-----------|----------------------------|---------------------------------|
| 300 Hz | 640 mV p-p | 420 mV p-p |
| 100 Hz | 640 mV p-p | 420 mV p-p |
| 30 Hz | 325 mV p-p | 180 mV p-p |
| 10 Hz | 325 mV p-p | 140 mV p-p |
| 3 Hz | 110 mV p-p | 60 mV p-p |
| 1 Hz | 110 mV p-p | 50 mV p-p |
| | 1 | l . |

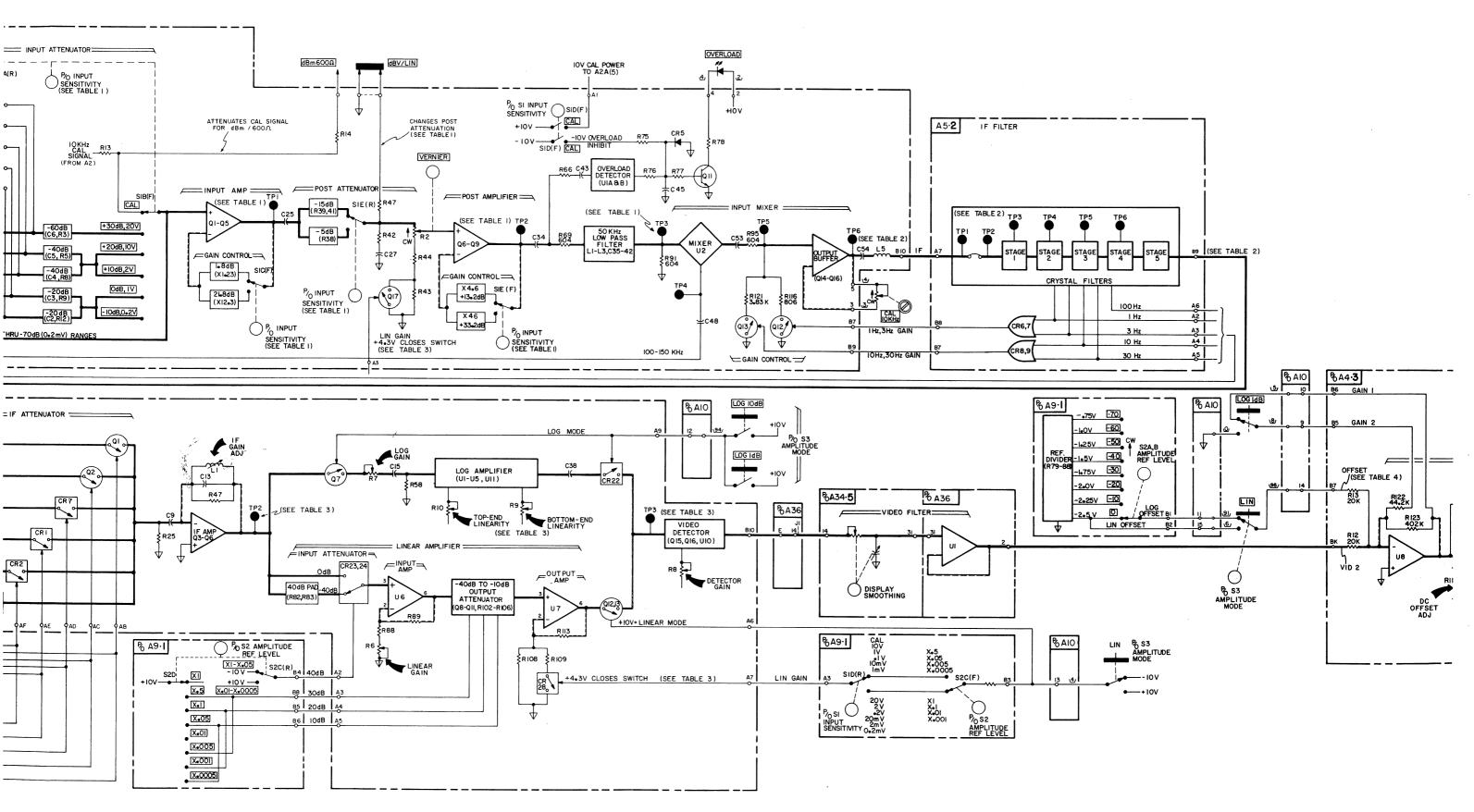
| Full Scale Input | MAX INPUT (INPUT SENSITIVITY Switch) | AMPLITUDE REF LEVEL | A4TP2 | Input Attenuation | Input Amp Gain | A4U6 pin 6 | Output Attenuation | A4 pin 7, A9 pin 3 (LIN GAIN) | Output Amp Gain | A4TP3 (Appr. Value) |
|---|--------------------------------------|--|--|---|--|--|--|--|--|-------------------------------------|
| 1 V rms .2 V rms .1 V rms .02 V rms .01 V rms 2 mV p-p 1 mV p-p .2 mV p-p | 1 V 1 V 1 V 1 V 1 V | X1 X.5 X.1 X.05 X.01 X.005 X.001 X.0005 | 6.7 V p-p 1.32 V p-p .67 V p-p .132 V p-p .067 V p-p | - 40 dB - 40 dB - 40 dB - 40 dB - 40 dB - 0 dB - 0 dB - 0 dB | + 40 dB + 40 dB | 6.7 V p-p 1.32 V p-p .67 V p-p .132 V p-p 6.7 V p-p 1.32 V p-p .67 V p-p .132 V p-p | - 40 dB - 30 dB - 20 dB - 10 dB - 40 dB - 30 dB - 20 dB - 10 dB | 28 V dc + 4.3 V dc 28 V dc + 4.3 V dc 28 V dc + 4.3 V dc 28 V dc + 4.3 V dc | + 34 dB (X50) + 38.2 dB (X81) + 34 dB (X50) + 38.2 dB (X81) + 34 dB (X50) + 38.2 dB (X81) + 34 dB (X50) + 38.2 dB (X81) | 3.4 V p-p 3.4 V p-p 3.4 V p-p |

Table 3. Approximate Full Scale Levels in Linear Amplifier (LINEAR Mode, Full Scale Input, Manually Tuned to Input Frequency).

Table 4. Linear and Log Offsets.

| AMPLITUDE MODE | AMPLITUDE REF LEVEL | A4 pin B7 |
|-------------------|------------------------|-------------|
| LOG | 0 | - 2.50 V dc |
| LOG | - 10 | - 2.25 V dc |
| LOG | - 20 | - 2,00 V dc |
| LOG | - 30 | - 1.75 V dc |
| LOG | - 40 | - 1,50 V dc |
| LOG | - 50 | - 1.25 V dc |
| LOG | - 60 | - 1.00 V dc |
| LOG | - 70 | - 0.75 V dc |
| LINEAR | Any Setting | - 2.50 V dc |





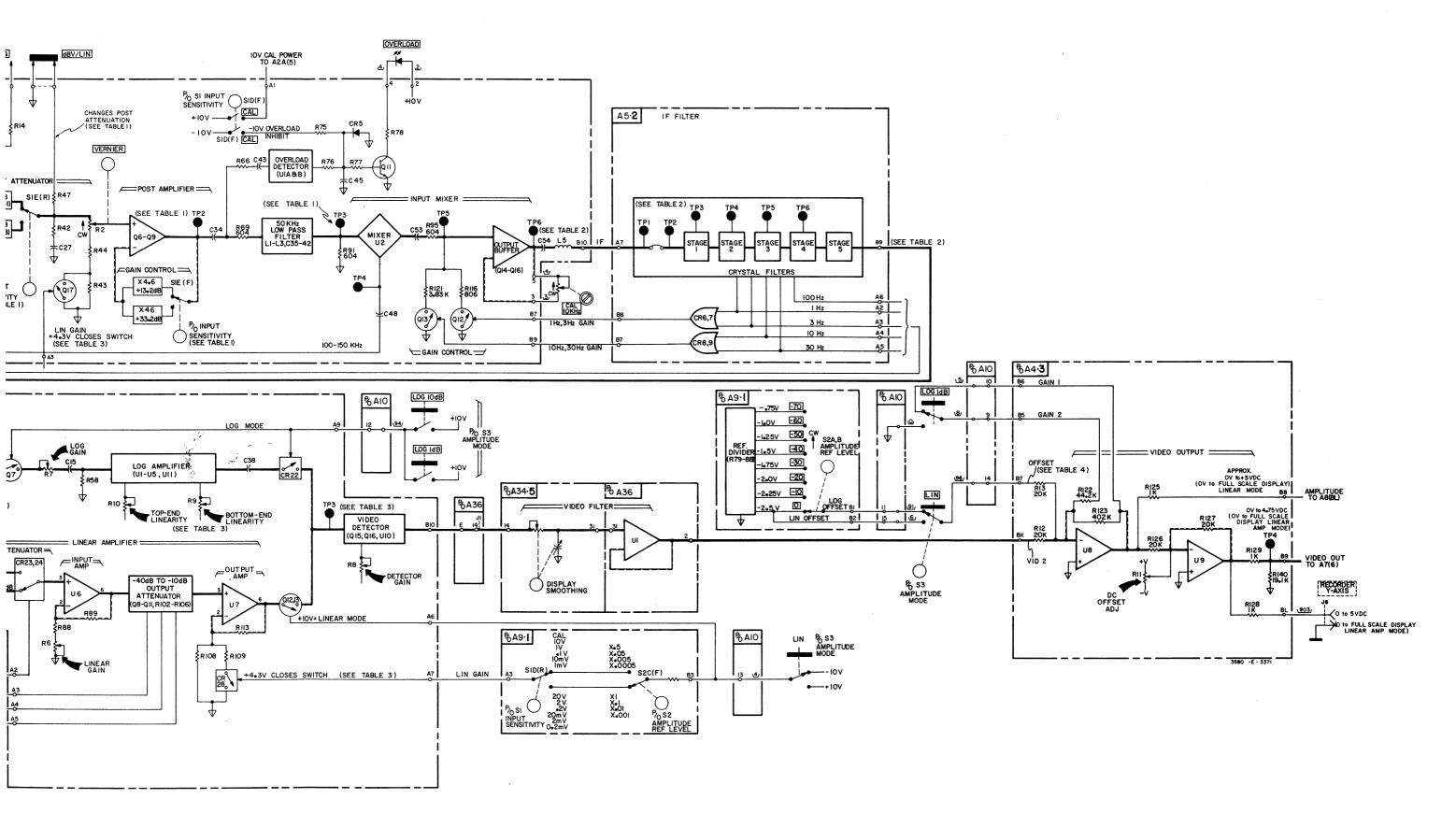


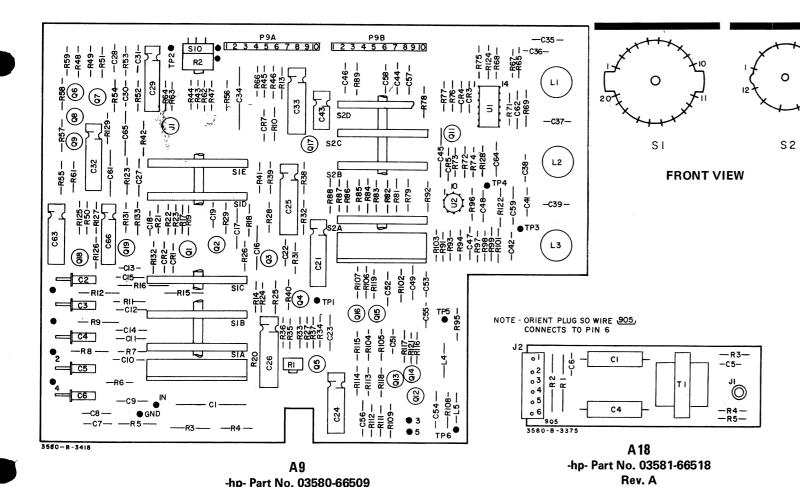
Figure 7-3. Amplitude Section Detailed Block Diagram.

| INPUT SENSITIVITY | + 30 dB/20 V | + 20 dB/10 V | + 10 dB/2 V | 0 dB/1 V | - 10 dB/0.2 V | 20 dB/0.1 V | 30 dB/20 mV | 40 dB/10 mV | - 50 dB/2 mV | - 60 dB/1 mV | - 70 dB/0.2 mV |
|---|--|--|---|---|--|---|--|--|--|---|---|
| Maximum Input Log (dBm 600 Ω) Log (dBV) Linear Log (dBm 900 Ω)* | 24.5 V 31.6 V 20.0 V 30.0 V | 7.75 V 10.00 V 10.00 V 9.49 V | 2.45 V 3.16 V 2.00 V 3.00 V | .775 V 1.000 V 1.000 V .949 V | .245 V .316 V .200 V .300 V | 77.5 mV .1 V .1 V 94.9 mV | 24.5 mV 31.6 mV 20.0 mV 30.0 mV | -7.75 mV 10.00 mV 10.00 mV 9.49 mV | 2.45 mV 3.16 mV 2.00 mV 3.00 mV | .775 mV 1.000 mV 1.000 mV 949 mV | .245 mV .316 mV .200 mV .300 mV |
| Input Attenuator Input Attenuator Out (Gate of A9Q1) Log (dBm 600 Ω) Log (dBV) Linear Log (dBm 900 Ω) * | - 60. dB 24.5 mV 31.6 mV 20.0 mV 30.0 mV | 40. dB 77.5 mV .1 V .1 V 94.9 mV | -40. dB 24.5 mV 31.6 mV 20.0 mV 30.0 mV | - 20. dB 77.5 mV .1 V .1 V 94.9 mV | 20. dB 24.5 mV 31.6 mV 20.0 mV 30.0 mV | 0 dB 77.5 mV .1 V .1 V 94.9 mV | 0 dB 24.5 mV 31.6 mV 20.0 mV 30.0 mV | 0 dB 77.5 mV 10.0 mV 10.0 mV 9.49 mV | 0 dB 2.45 mV 3.16 mV 2.00 mV 3.00 mV | 0 dB .775 mV 1.000 mV 1.000 mV .949 mV | 0 dB .245 mV .316 mV .200 mV .300 mV |
| Input Amp Gain Input Amp Out (A9TP1) Log (dBm 600 Ω) Log (dBV) Linear Log (dBm 900 Ω)* | 1.8 dB 30.1 mV 38.9 mV 24.6 mV 36.9 mV | 1.8 dB 95.6 mV .123 V .123 V .117 V | 1.8 dB 30.1 mV 38.9 mV 24.6 mV 36.9 mV | 1.8 dB 95.6 mV .123 V .123 V .117 V | 1.8 dB 30.1 mV 38.9 mV 24.6 mV 36.9 mV | 1.8 dB 95.6 mV .123 V .123 V .117 V | 1.8 dB 30.1 mV 38.9 mV 24.6 mV 36.9 mV | 1.8 dB 9.56 mV 12.3 mV 12.3 mV 11.7 mV | 1.8 dB 3.01 mV 3.89 mV 2.46 mV 3.69 mV | 21.8 dB 9.56 mV 12.3 mV 12.3 mV 11.7 mV | 21.8 dB 3.01 mV 3.89 mV 2.46 mV 3.69 mV |
| Post Attenuation Log (dBm 600 \(\delta \)) Log (dBV) Linear Log (dBm 900 \(\delta \)) * | 2.8 dB 5.0 dB 5.2 dB 4.6 dB | - 12.8 dB - 15.0 dB - 15.0 dB - 14.6 dB | - 2.8 dB - 5.0 dB - 5.2 dB - 4.6 dB | - 12.8 dB 15.0 dB 15.0 dB 14.6 dB | 2.8 dB 5.0 dB 5.2 dB 4.6 dB | 12.8 dB - 15.0 dB - 15.0 dB - 14.6 dB | 2.8 dB 5.0 dB 5.2 dB 4.6 dB | 12.8 dB 15.0 dB 15.0 dB 14.6 dB | 2.8 dB 5.0 dB 5.2 dB 4.6 dB | - 12.8 dB - 15.0 dB - 15.0 dB - 14.6 dB | - 2.8 dB - 5.0 dB - 5.2 dB - 4.6 dB |
| Post Attenuator Out (Base A9Q6) Log Linear | 21.8 mV 13.5 mV | 21.8 mV 21.8 mV | 21.8 mV 13.5 mV | 21.8 mV 21.8 mV | 21.8 mV 13.5 mV | 21.8 mV 21.8 mV | 21.8 mV 13.5 mV | 2.18 mV 2.18 mV | 2.18 mV 1.35 mV | 2.18 mV 2.18 mV | 2.18 mV 1.35 mV |
| Post Amp Gain Post Amp Out (A9TP2) Log Linear | 13.2 dB 100, mV 61.9 mV | 13.2 dB 100. mV 100. mV | 13.2 dB 100. mV 61.9 mV | 13.2 dB 100. mV 100. mV | 13.2 dB 100. mV 61.9 mV | 13.2 dB 100. mV 100. mV | 13.2 dB 100. mV 61.9 mV | 33.2 dB 100. mV 100. mV | 33.2 dB 100. mV 61.9 mV | 33.2 dB 100. mV 100. mV | 33.2 dB 100, mV 61.9 mV |
| Low Pass Filter Out (A9TP3) Log Linear | 50. mV 30.8 mV | 50. mV 50. nıV | 50. mV 30.8 mV | 50. mV 50. mV | 50. mV 30.8 mV | 50. mV 50. mV | 50. mV 30.8 mV | 50. mV 50. mV | 50. mV 30.8 mV | 50. mV 50. mV | 50. mV 30.8 mV |
| Total Gain $\label{eq:continuous} \begin{tabular}{ll} Log (dBm 600 \Omega) \\ Log (dBV) \\ Linear \\ Log (dBm 900 \Omega) \end{tabular} \begin{tabular}{ll} Log (dBm 900 & \Omega) \end{tabular} \begin{tabular}{l$ | 47.8 dB 50. dB 50.2 dB 49.6 dB | - 37.8 dB - 40. dB - 40. dB 39.6 dB | 27.8 dB 30. dB 30.2 dB 29.6 dB | - 17.8 dB - 20. dB - 20. dB - 19.6 dB | 7.8 dB 10. dB 10.2 dB 9.6 dB | + 2.2 dB 0 dB 0 dB + 0.4 dB | + 12.2 dB + 10. dB + 9.8 dB + 10.4 dB | + 22.2 dB + 20. dB + 20. dB + 20.4 dB | + 32.2 dB + 30. dB + 29.8 dB + 30.4 dB | + 42.2 dB + 40. dB + 40. dB + 40.4 dB | + 52.2 dB + 50. dB + 49.8 dB + 50.4 dB |

Table 1. Input Circuit Amplitude Levels and Gains for Full Scale Sine Wave Inputs.

(Amplitude Ref. Level — Normal).

NOTE: All voltages in RMS.



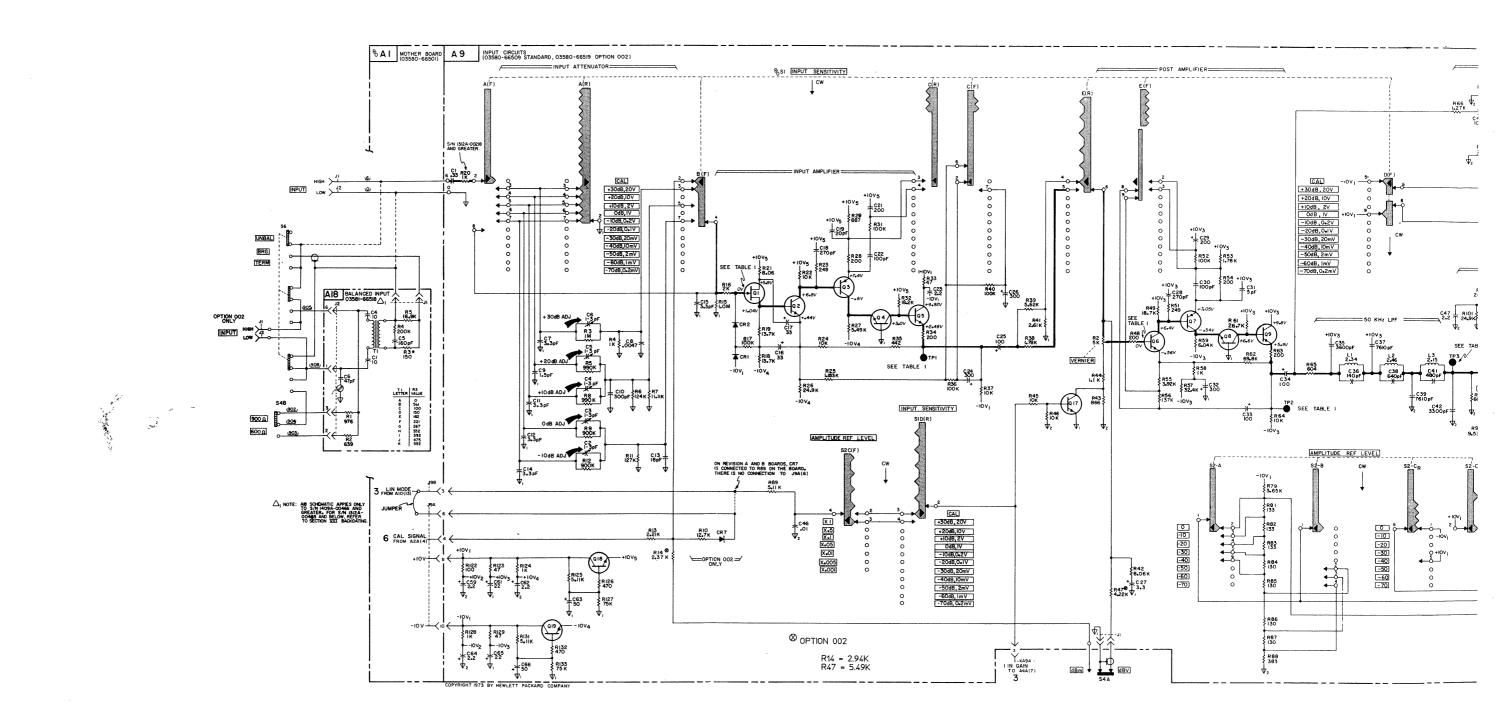
-hp- Part No. 03580-66509 Rev. C NOTE

CR7 and R10 appear on 03580-66519 only.

All dc voltages measured with a low capacitive, high resistance dc probe and voltmeter (-hp- Model 412A or 10 k Ω film resistor in series with the input of a high impedance voltmeter).

CAUTION

The A9 board must be clean handled.



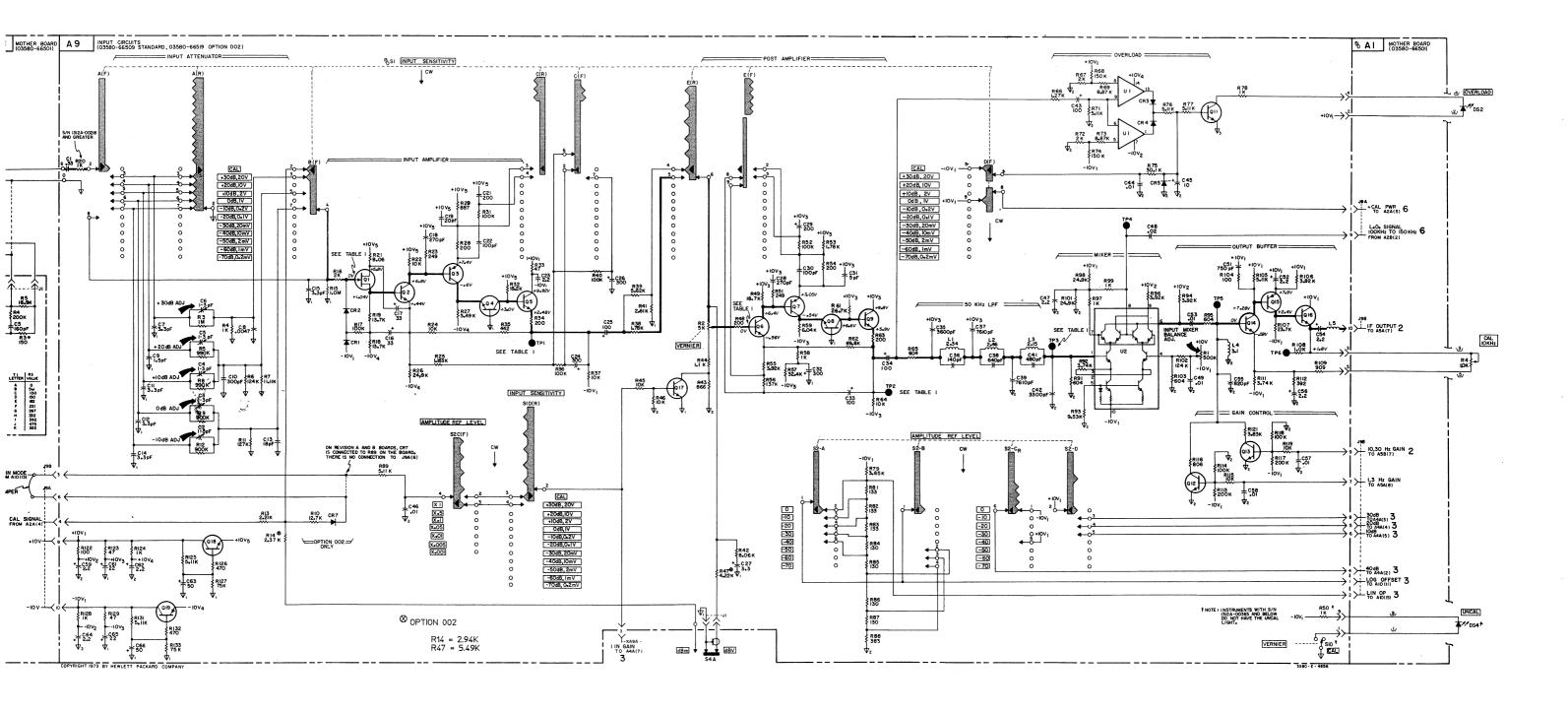


Figure 7-4. Input Assembly (A9) and Balanced Input Assembly (A18) Schematics and Component Location Diagrams.

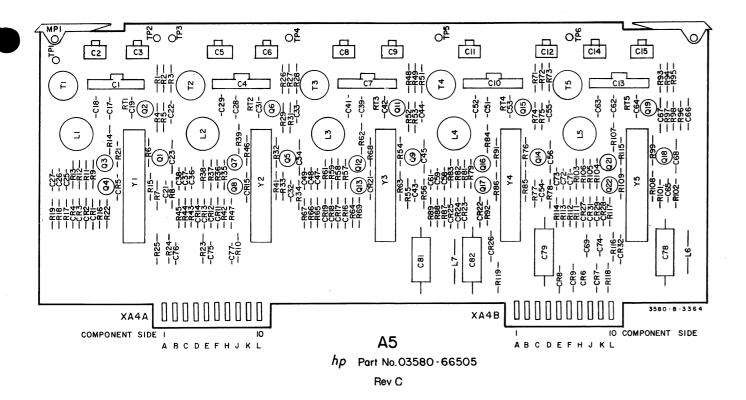


Table 1. I.F. Input Level Change With Bandwidth.

(Full Scale Sine Wave Input, Log Mode, Manually Tuned to Input Frequency)

| 47 | |
|-----------|------------|
| Bandwidth | A5TP1 |
| | Tap . |
| 300 Hz | 640 mV p-p |
| 100 Hz | 640 mV p-p |
| 30 Hz | 325 mV p-p |
| 10 Hz | 325 mV p-p |
| 3 Hz | 110 mV p-p |
| 1 Hz | 110 mV p-p |

Table 2. I.F. Output Level Change With Bandwidth.

(Full Scale Input, Log Mode, Manually Tuned to Input

| | · · · · · · · · · · · · · · · · · · · | | | | |
|-----------|---------------------------------------|--|--|--|--|
| Bandwidth | A5 pin B9 | | | | |
| 300 Hz | 420 mV p-p | | | | |
| 100 Hz | 410 mV p-p | | | | |
| 30 Hz | 180 mV p-p | | | | |
| 10 Hz | 140 mV p-p | | | | |
| 3 Hz | 60 mV p-p | | | | |
| 1 Hz | 50 mV p-p | | | | |
| | | | | | |

3580A Full Scale Input
Log Mode, 300 Hz Bandwidth Manually tuned to Input Frequency Oscilloscope:

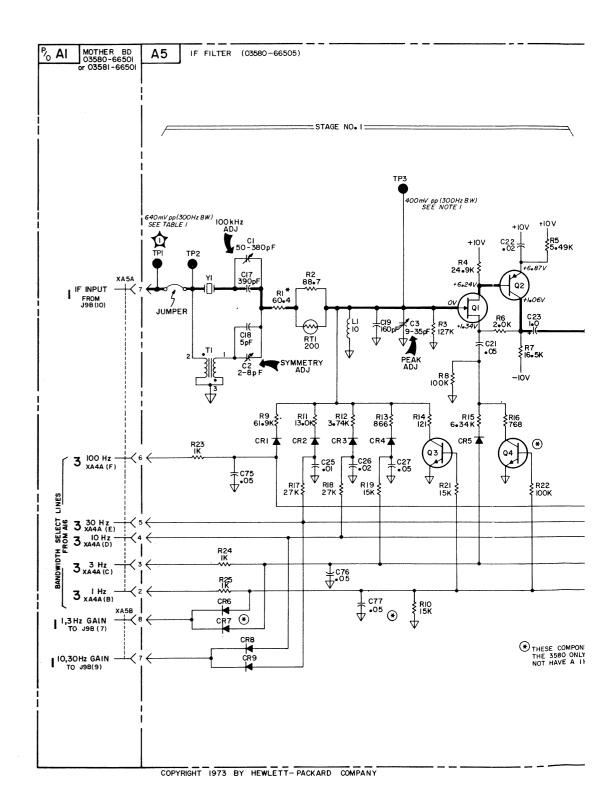


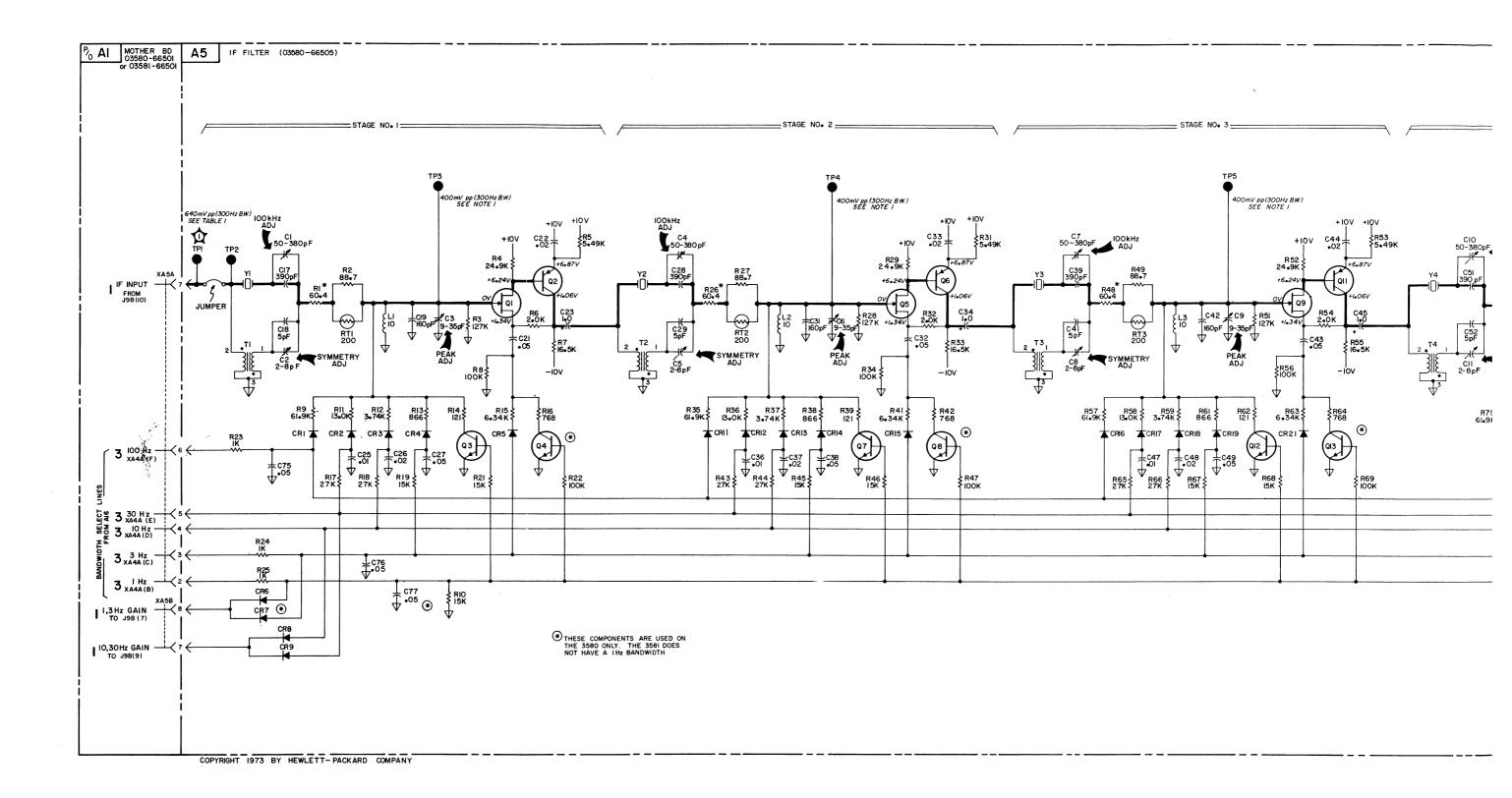
A 10:1 divider probe was used on the oscilloscope input. The vertical amplitude sensitivity is the actual amplifier setting and does not include the X10 multiplier introduced by the probe.

20μsec/div.

NOTE 1: AC voltage readings were taken with an oscilloscope. equipped with 10:1 divider probes. Some loading occurs during the measurement, so values are approximate. The values given correspond to a full scale input (LOG MODE, 300 Hz BANDWIDTH). The 3580A must be manually tuned to the input frequency. See Table 1 and Table 2 for level changes with Bandwidth.

NOTE 2: DC levels taken with a low capacitive, high resistance dc probe and voltmeter (-hp- 412A or 10 k Ω film resistor in series with the input of a high impedance voltmeter).





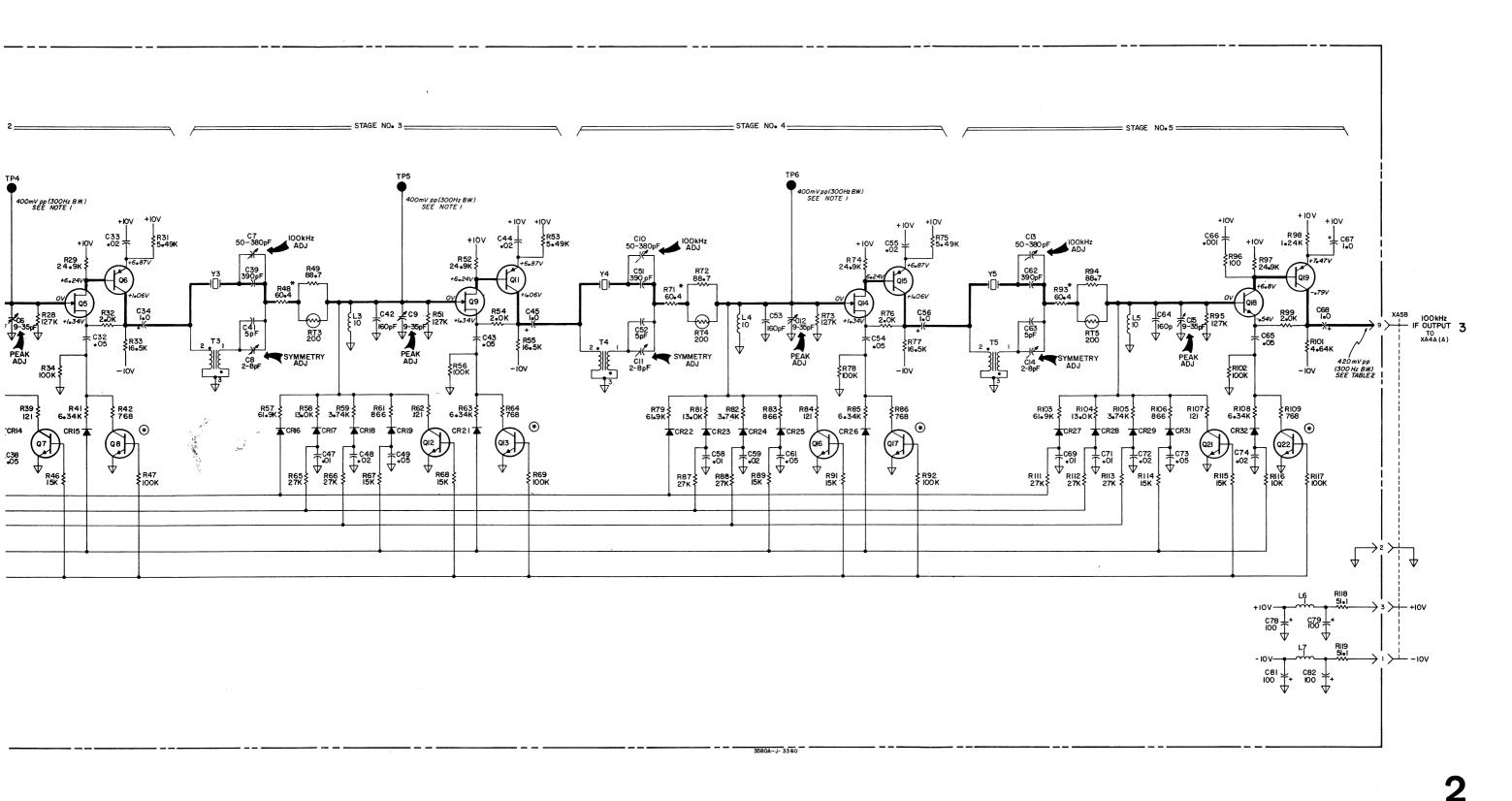


Figure 7-5. IF Filter Assembly (A5) Schematic and Component Location Diagram.

| Full Scale Input | MAX INPUT (INPUT SENSITIVITY Switch) | AMPLITUDE REF LEVEL | A4TP2 | Input Attenuation | Input Amp Gain | A4U6 pin 6 | Output Attenuation | A4 pin 7, A9 pin 3 (LIN GAIN) | Output Amp Gain | A4TP3 (Appr. Value) |
|---|--|--|--|--|--|--|--|--|--|---|
| 1 V rms .2 V rms .1 V rms .02 V rms .01 V rms 2 mV p-p 1 mV p-p .2 mV p-p | 1 V | X1 X.5 X.1 X.05 X.01 X.005 X.001 X.0005 | 6.7 V p-p 1.32 V p-p .67 V p-p .132 V p-p .067 V p-p | - 40 dB - 40 dB - 40 dB - 40 dB O dB O dB O dB | + 40 dB + 40 dB | 6.7 V p-p 1.32 V p-p .67 V p-p .132 V p-p 6.7 V p-p 1.32 V p-p .67 V p-p .132 V p-p | - 40 dB - 30 dB - 20 dB - 10 dB - 40 dB - 30 dB - 20 dB - 10 dB | 28 V dc + 4.3 V dc 28 V dc + 4.3 V dc 28 V dc + 4.3 V dc 28 V dc + 4.3 V dc | + 34 dB (X50) + 38.2 dB (X81) + 34 dB (X50) + 38.2 dB (X81) + 34 dB (X50) + 38.2 dB (X81) + 34 dB (X50) + 38.2 dB (X81) | 3.4 V p-p 3.4 V p-p 3.4 V p-p 3.4 V p-p 3.4 V p-p |

Table 1. Approximate Full Scale Levels in Linear Amplifier (LINEAR Mode, Full Scale Input, Manually Tuned to Input Frequency).

| AMPLITUDE MODE | AMPLITUDE REF LEVEL | A4 pin B7 |
|-------------------|------------------------|-------------|
| LOG | 0 | - 2.50 V dc |
| LOG | -10 | - 2.25 V dc |
| LOG | - 20 | - 2.00 V dc |
| LOG | - 30 | - 1.75 V dc |
| LOG | - 40 | - 1.50 V dc |
| LOG | - 50 | -1.25 V dc |
| LOG | - 60 | - 1.00 V dc |
| LOG | - 70 | - 0.75 V dc |
| LINEAR | Any Setting | - 2.50 V dc |

Table 2. Linear and Log Offsets.

| Bandwidth | A4TP1 |
|-----------|------------|
| 300 Hz | 420 mV p-p |
| 100 Hz | 420 mV p-p |
| 30 Hz | 180 mV p-p |
| 10 Hz | 140 mV p-p |
| 3 Hz | 60 mV p-p |
| 1 Hz | 50 mV p-p |

Table 3. A4 Input Level with Bandwidth Change.
(LOG Mode, Full Scale Input)

NOTE 1: The values given are for a full scale input, LOG Mode. The 3580A must be manually tuned to the input frequency.

NOTE 2: DC Levels taken in LOG Mode, Full Scale Input. Use a low capacitive, high resistance dc probe and voltmeter (-hp- Model 412A or 10 k Ω film resistor in series with the input of a high impedance voltmeter).

NOTE 3: R122B* is selected to provide a full scale output of 5 V \pm 50 mV at the Y AXIS output.

NOTE 4: The attenuation is not equal to the Amplitude Ref Level switch settings (see Table 1).

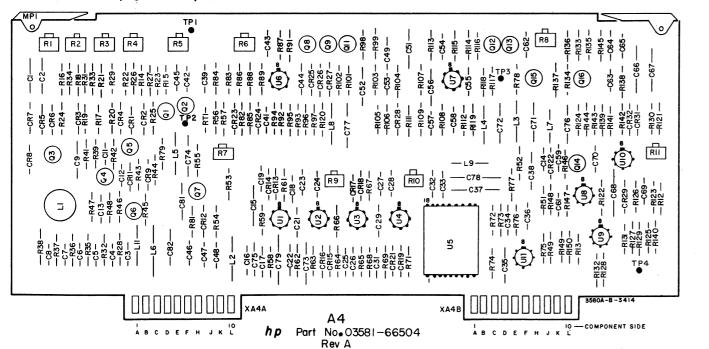
GENERAL WAVEFORM INFORMATION

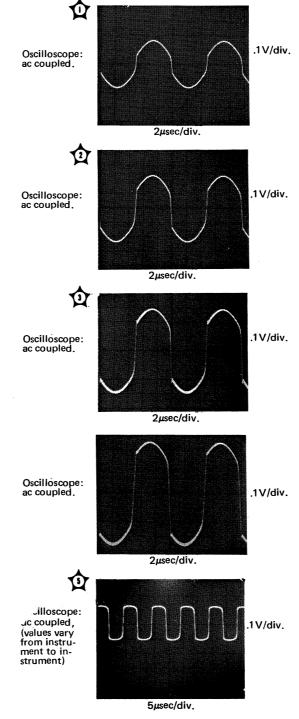
3580A Control Settings

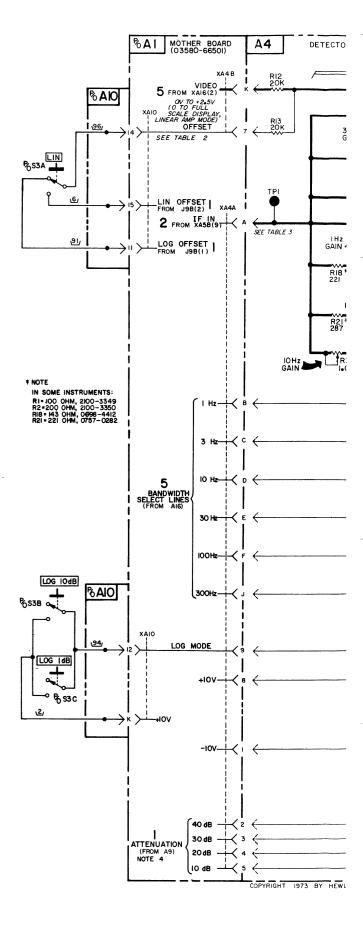
Manually tune the 3580A to the 10 KHz harmonic of the internal CAL signal. Values given are for a full scale display indication, LOG Mode.

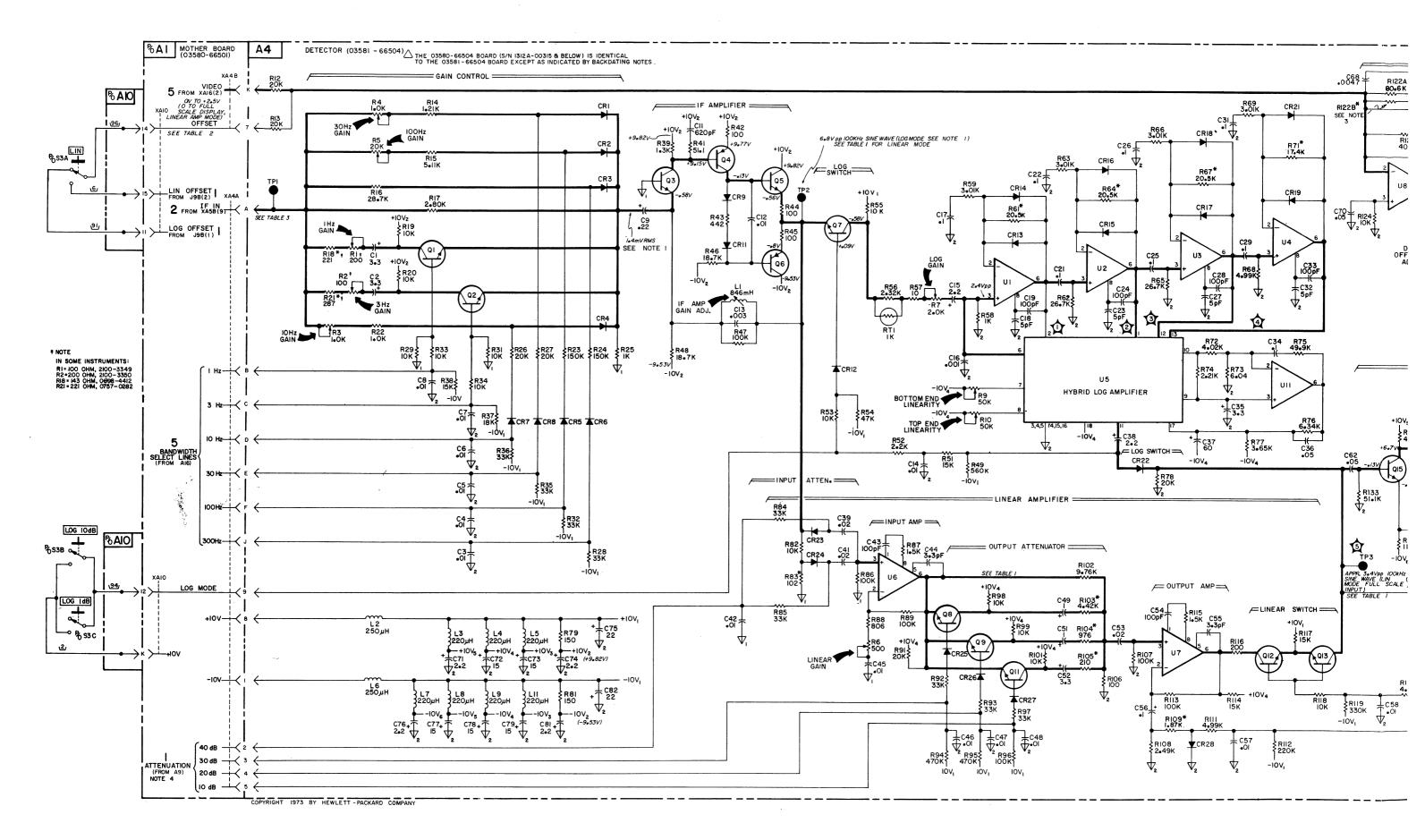
Oscilloscope Settings

All waveforms were recorded using a 10:1 divider probe on the oscilloscope inputs. The vertical amplitude sensitivity is the actual amplifier setting and does not include the X10 multiplier introduced by the probe.









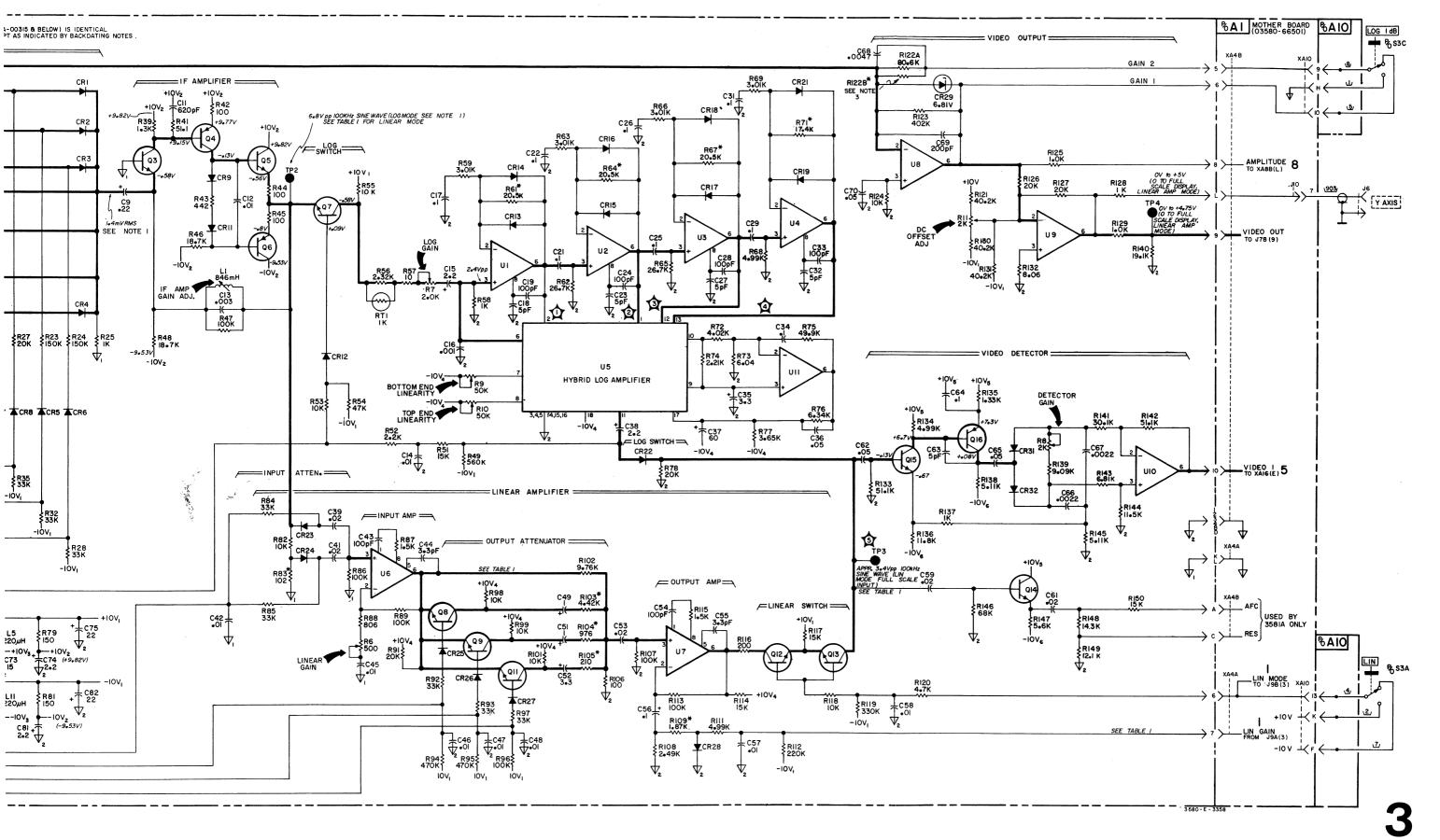
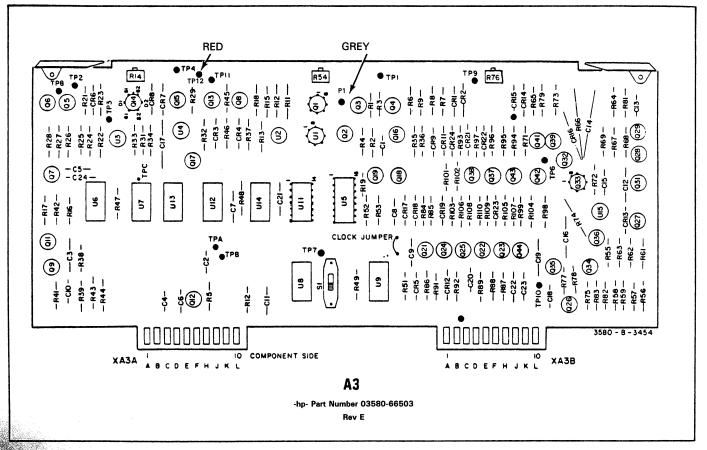


Figure 7-6. Detector Assembly (A4) Schematic and Component Location Diagram.
7-19/7-20



3580A

 ADAPTIVE SWEEP
 OFF

 DISPLAY
 All pushbuttons released

 FREQUENCY
 00.0 kHz

 START - CTR
 START

 DISPLAY SMOOTHING
 MIN

 RESOLUTION BANDWIDTH
 300 Hz

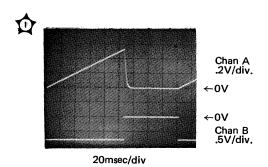
 FREQ. SPAN/DIV
 .5 KHz

 SWEEP TIME/DIV
 0.01 SEC

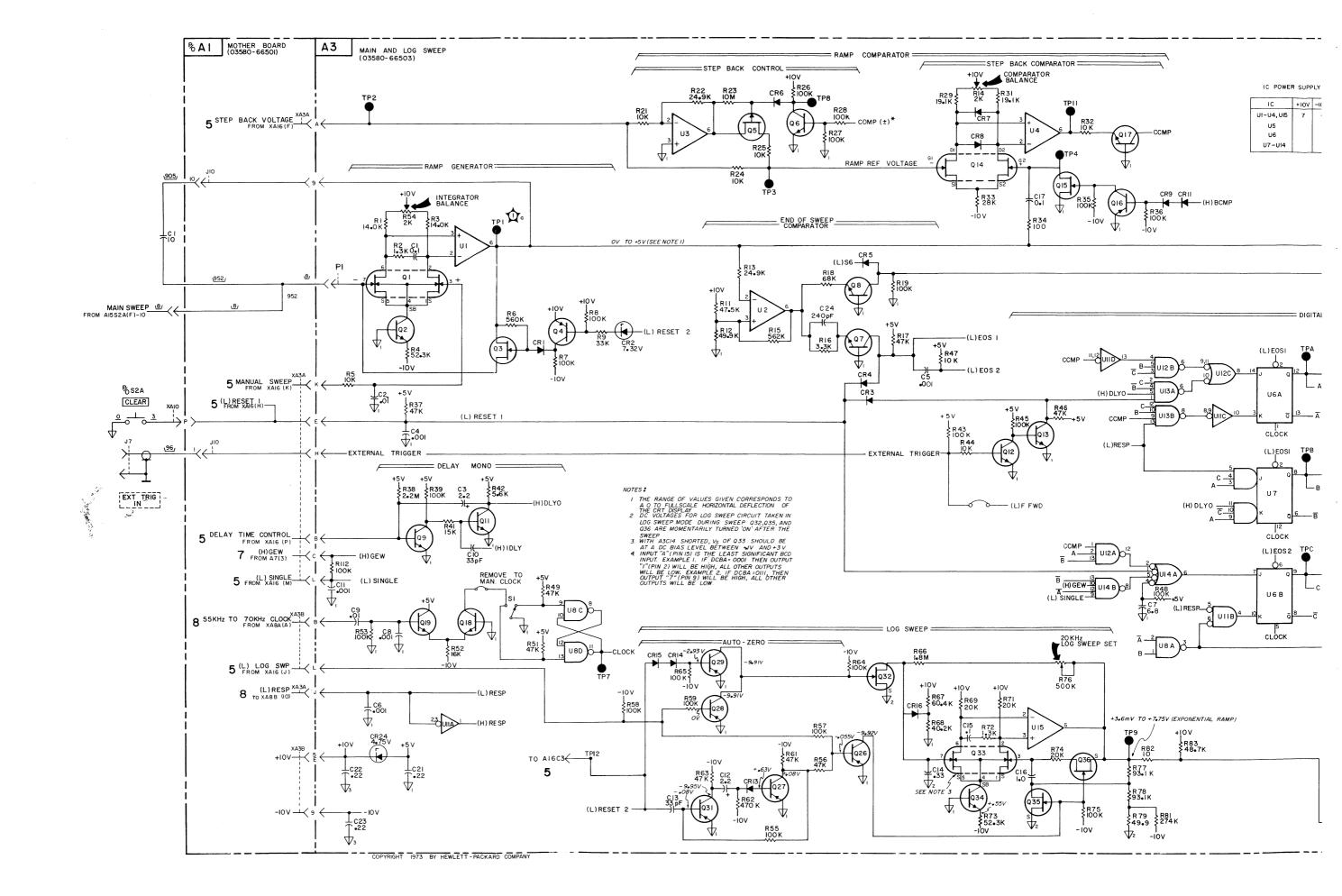
 SWEEP MODE
 REP

OSCILLOSCOPE

DC coupled, dual trace (chopped), triggered by Channel B.



A 10:1 divider probe was used on the oscilloscope input. The vertical amplitude sensitivity is the actual amplifier setting and does not reflect the X10 multiplier introduced by the probe.



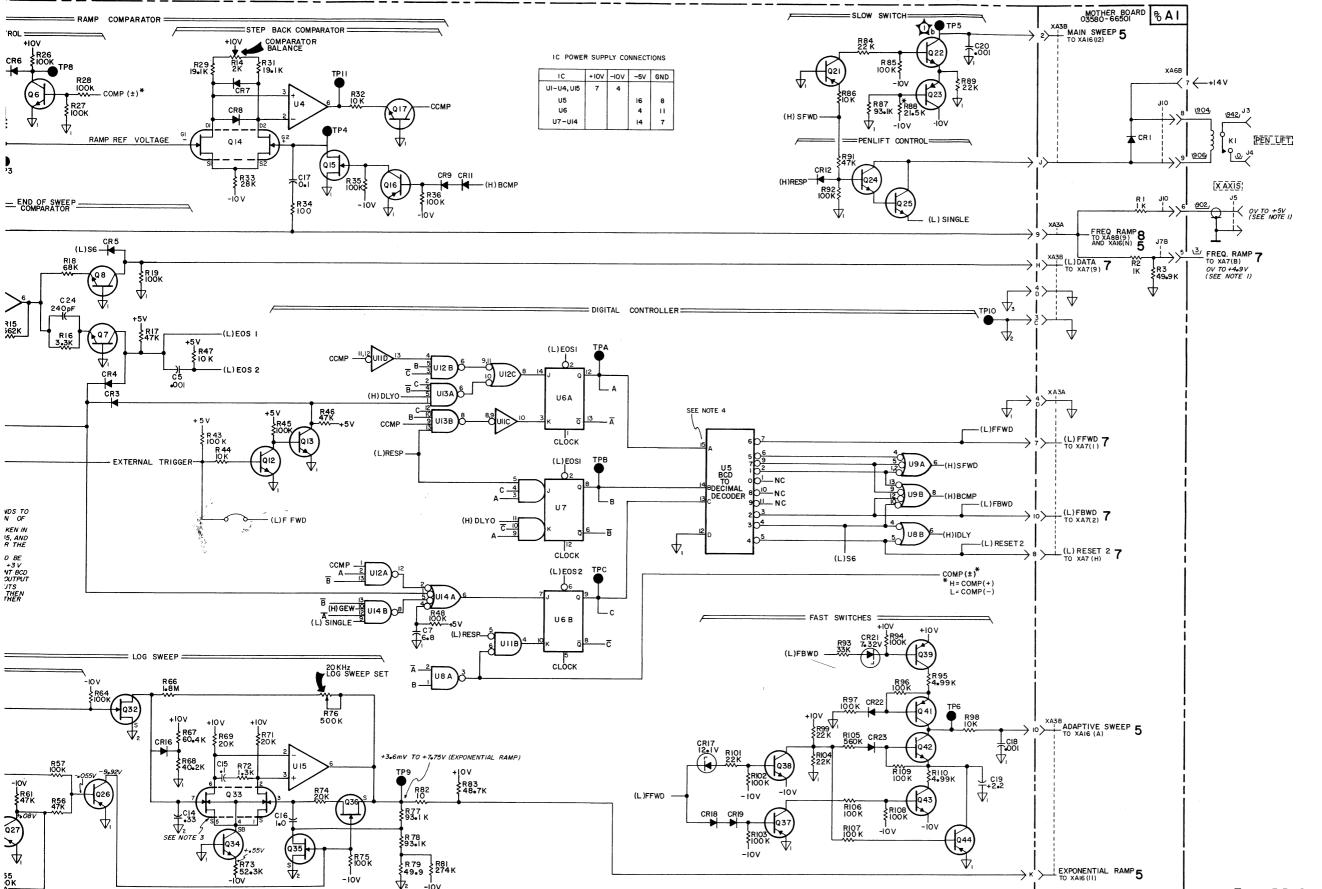
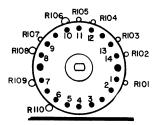


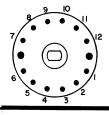
Figure 7-7. Sweep Generator (A3) Schematic and Component Location Diagram.

Revised: September 1987

RESISTORS MOUNTED BETWEEN WAFERS B & C ON SI.

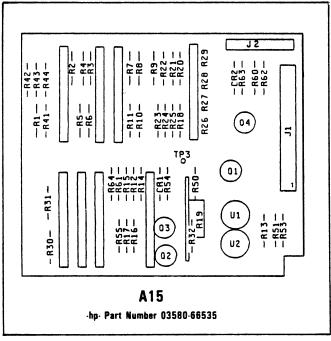


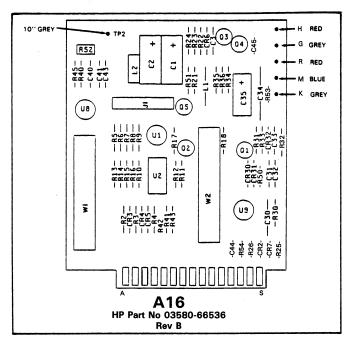
FRONT VIEW PIN POSITIONS FOR AI4SI

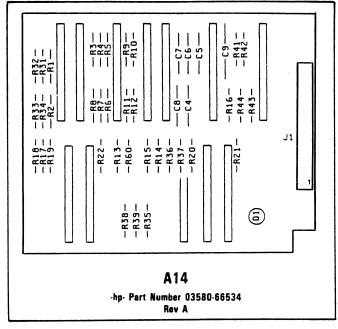


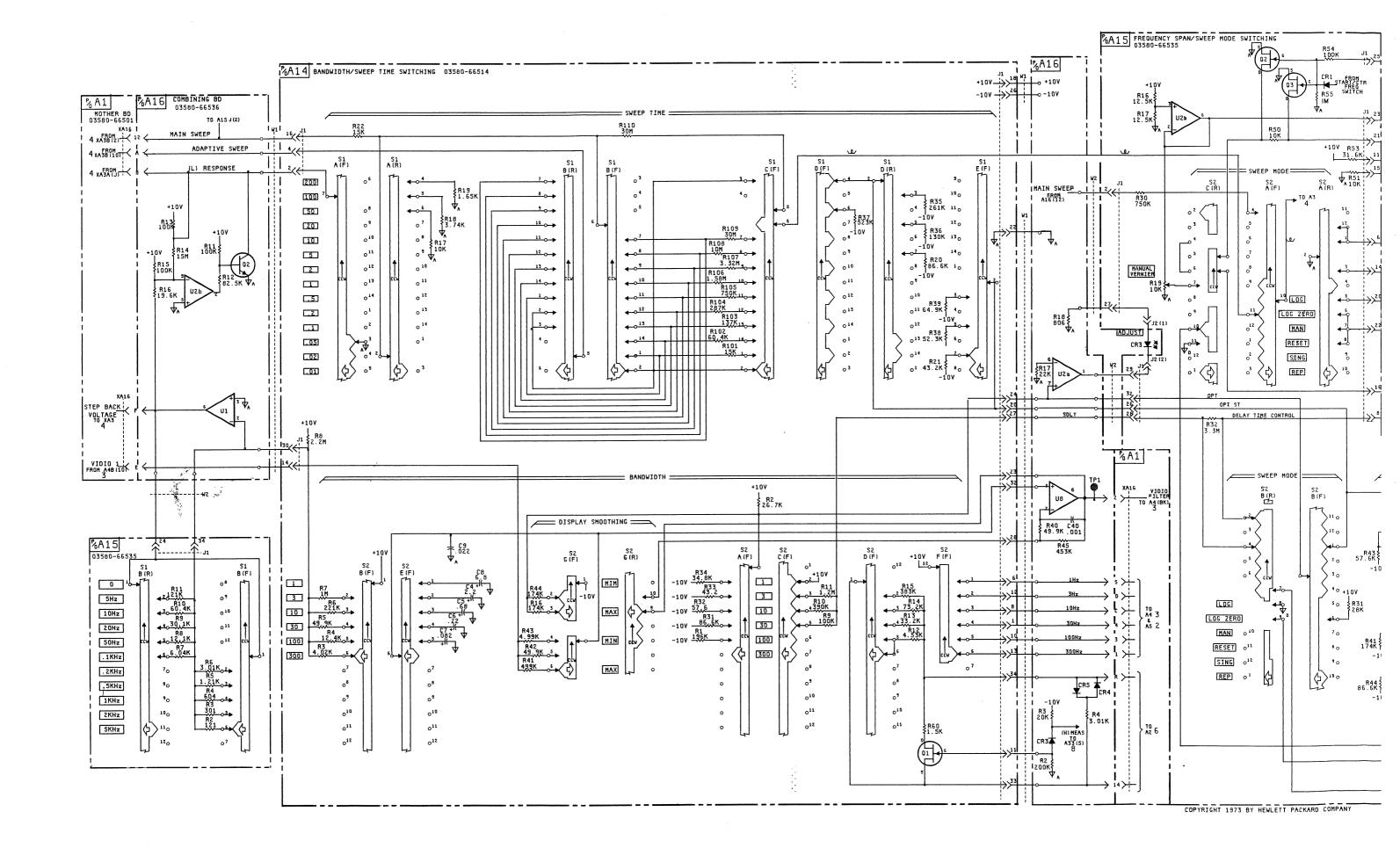
FRONT VIEW

PIN POSITIONS FOR AI4S2 AI5SI AI5S2









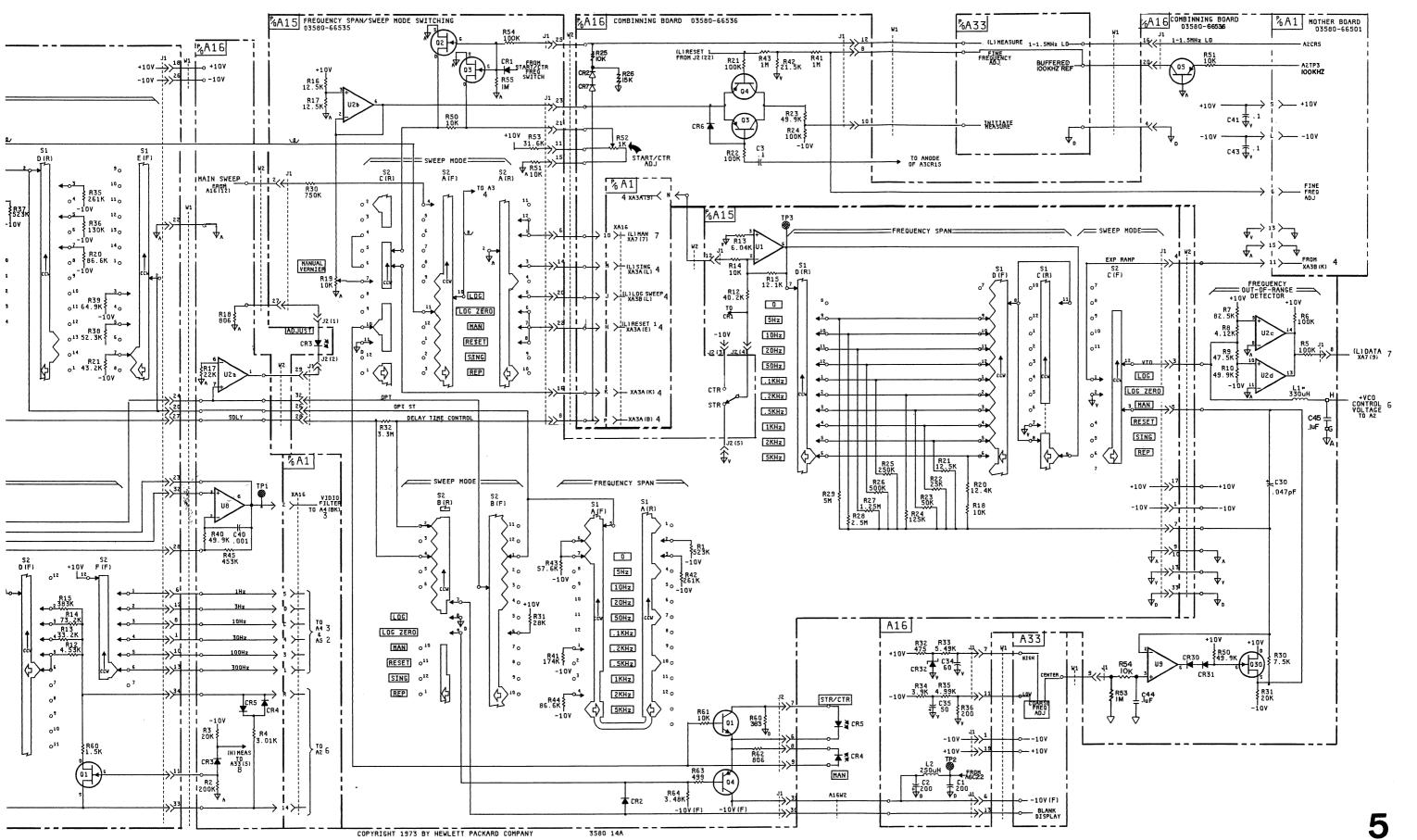
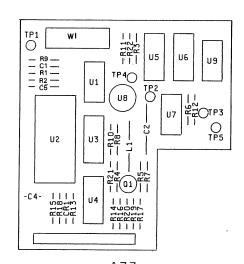


Figure 7-8. Frequency Control Circuits (A14, A15, A16) Schematic and Component Location Diagrams.



A33 H-P Part No 03580-66533 Rev A

A33 TIME BASE/COUNTER/DISPLAY 03580-66533 A16 -10V(F) R2 100K V₀ | R Vss | 7 | -10V | (L) COUNT TP3 **⊗** |7 -10V(F) 8 -10V(F)

COPYRIGHT 198

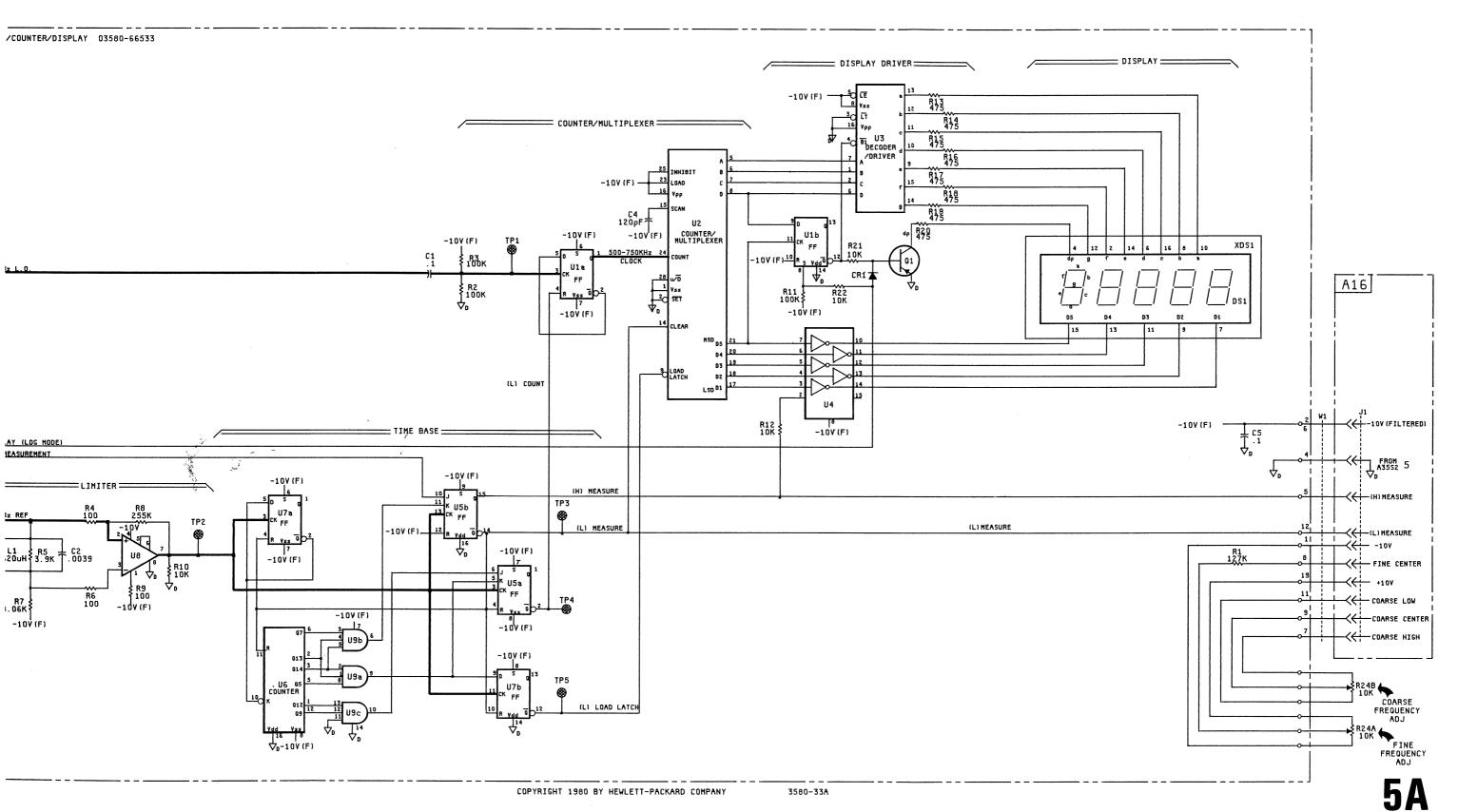
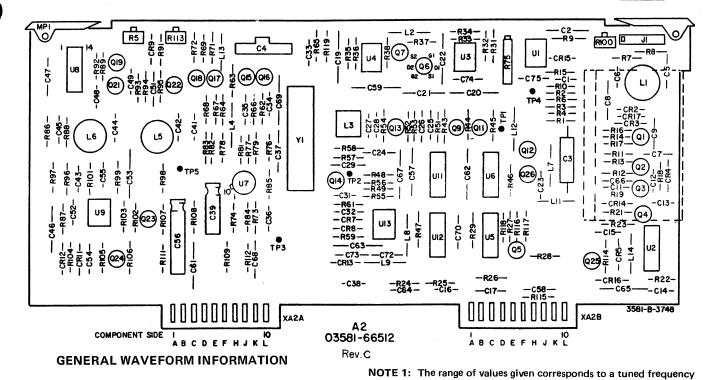


Figure 7-8(a). Frequency Counter and Display Circuits (A33)
Schematic and Component Location Diagrams.

Revised: September 1987
7-24(a)/7-24(b)



between 0 Hz and 50 kHz.

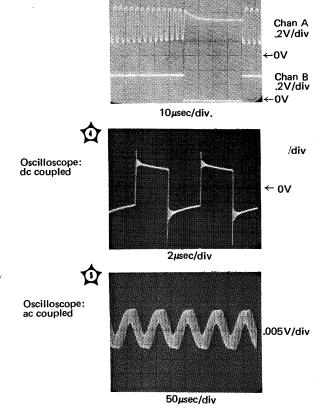
A2R65, See Paragraph 7-25,

3580A Control Settings

The 3580A is manually tuned to 10 kHz.

Oscilloscope Settings

All waveforms were recorded using a 10:1 divider probe on the oscilloscope inputs. The vertical amplitude sensitivity is the actual amplifier setting and does not include the X10 multiplier introduced by the probe, All dual traces were made with the oscilloscope in the chopped mode and triggered by Channel B.



NOTE 2: Replacement circuit boards do not contain A2Y1 and

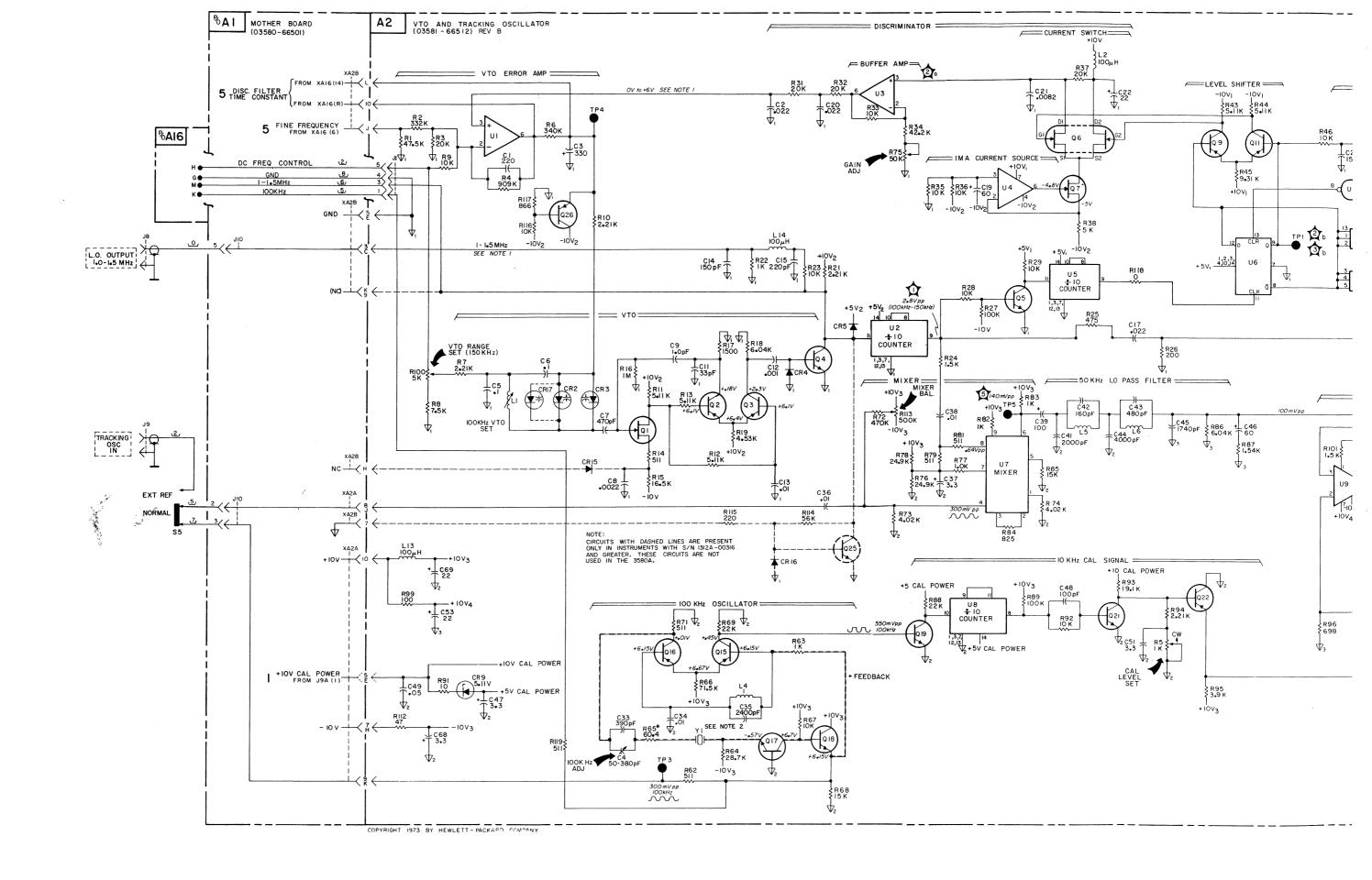
Oscilloscope:
ac coupled

2µsec/div

Chan A
1V/div
←0V

Chan B
2V/div

10 µsec/div



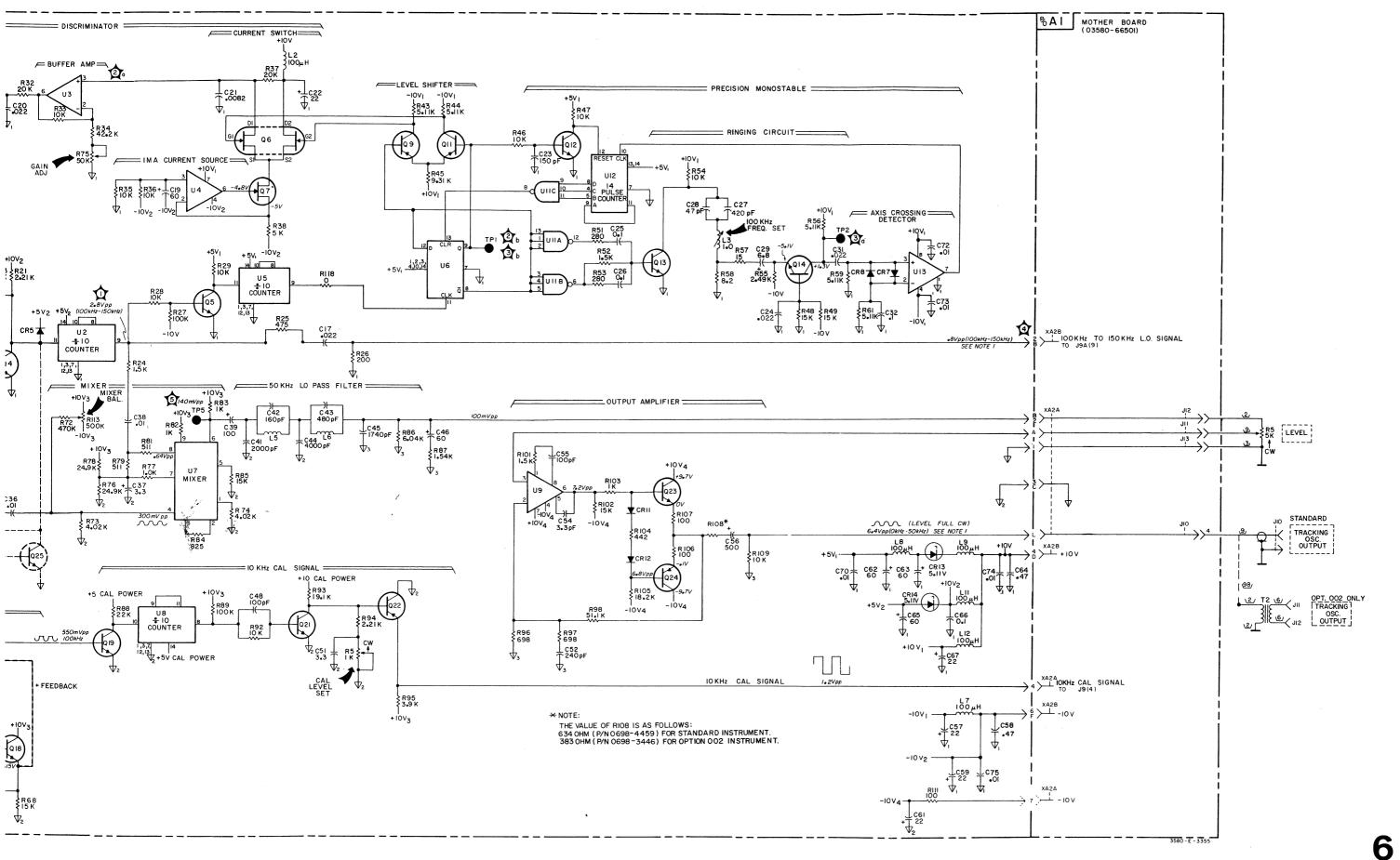
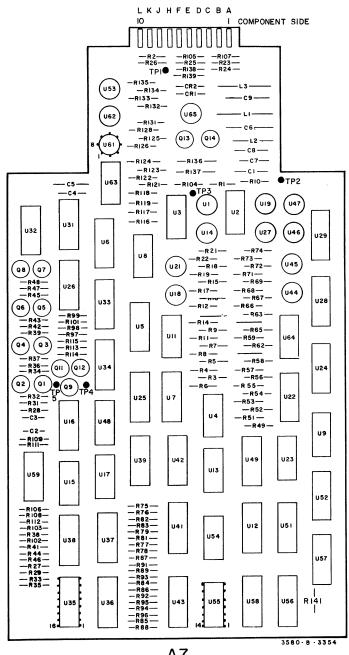


Figure 7-9. VTO and Tracking Oscillator Assembly (A2) Schematic and Component Location Diagram.



A7 hp Part No. 03580-66507 Rev. C

NOTE 1: For serial number 2030A04745 and lower see Δ 17. NOTE 2: For serial number 2030A4460 and lower see Δ 18.

GENERAL WAVEFORM INFORMATION

| 3580A |
|--|
| ADAPTIVE SWEEPOFF |
| DISPLAY All pushbuttons released |
| AMPLITUDE MODELOG 10 dBv/DIV |
| AMPLITUDE REF LEVELNORMAL |
| $dBv/LIN - dBm 600 \Omega \dots dBv/LIN$ |
| INPUT SENSITIVITYCAL |
| VERNIER (Amplitude)CAL |
| (Fully CW) |
| FREQUENCY10 kHz |
| START - CTRCTR |
| RESOLUTION BANDWIDTH 300 Hz |
| DISPLAY SMOOTHINGMIN |
| |

Option 002: Set dBm 900 Ω/LIN - dBm 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL.

 FREQ. SPAN/DIV
 0.5 KHz

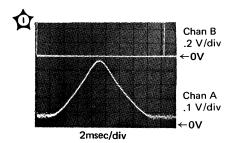
 SWEEP TIME/DIV
 .1 SEC

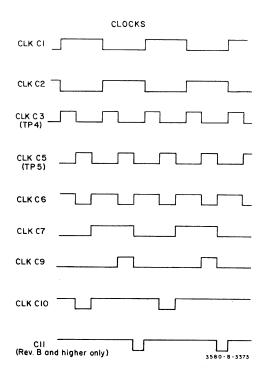
 SWEEP MODE
 ... REP

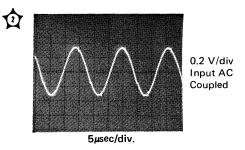
The front panel ZERO CAL and CAL 10 KHz were adjusted to give a full scale display of the 10 KHz CAL signal in the center of the screen. It is easiest to set this adjustment in the MANUAL mode, and then switch to the REPetitive sweep mode to measure the waveforms.

OSCILLOSCOPE

10:1 Probe, Dual Trace (Chopped), triggered by Channel B. The vertical amplitude sensitivity is the actual amplifier setting and does not include the X10 multiplier introduced by the probe.

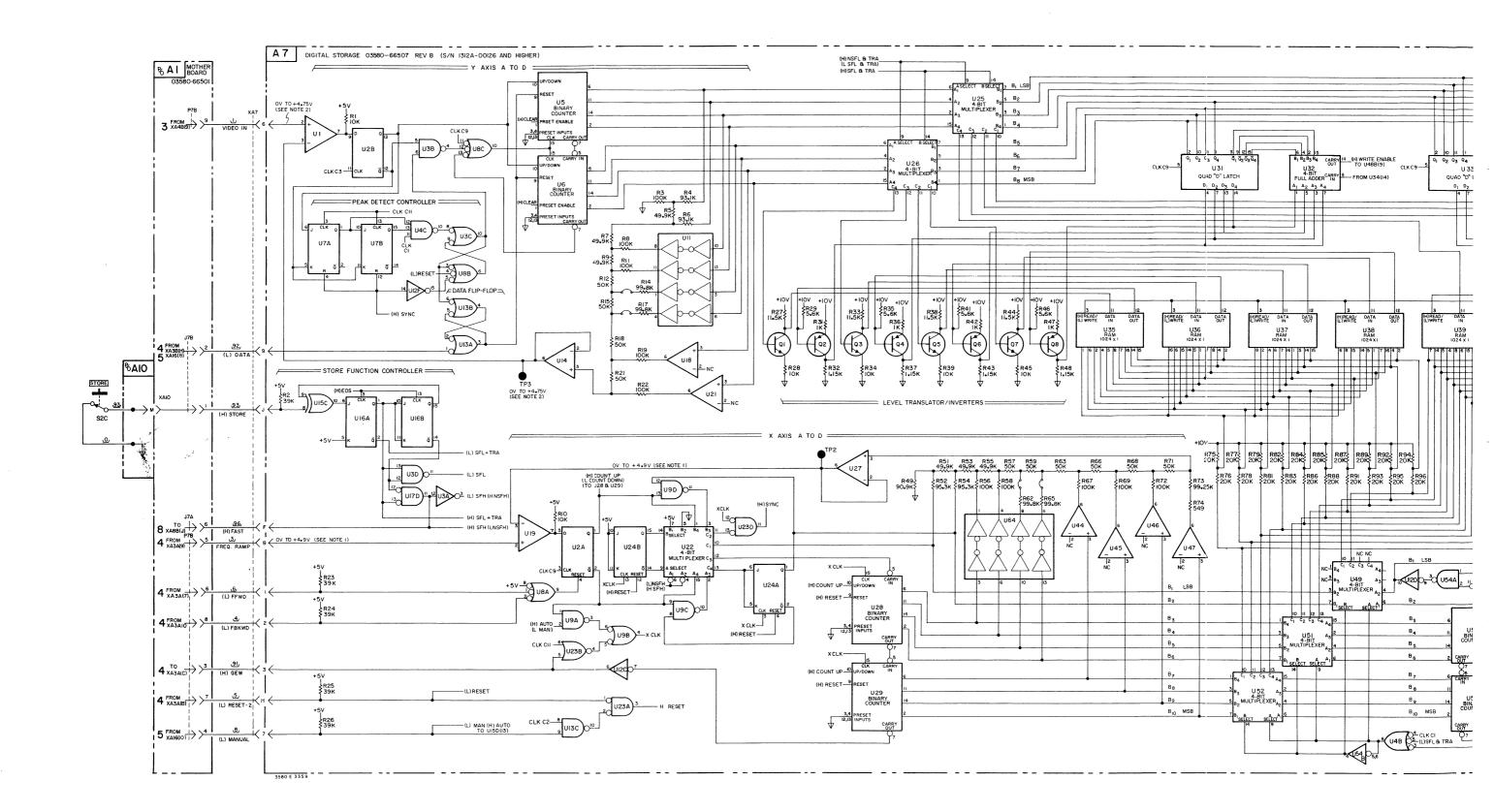






NOTE 1: The voltage range given is for a zero to full scale horizontal frequency sweep of the CRT display.

NOTE 2: The voltage range given is for a zero to full scale vertical deflection of the CRT display, LINEAR MODE. In LOG 10 dB/div Amplitude Mode, 0 volts cannot be obtained because of the - 80 dB noise floor.



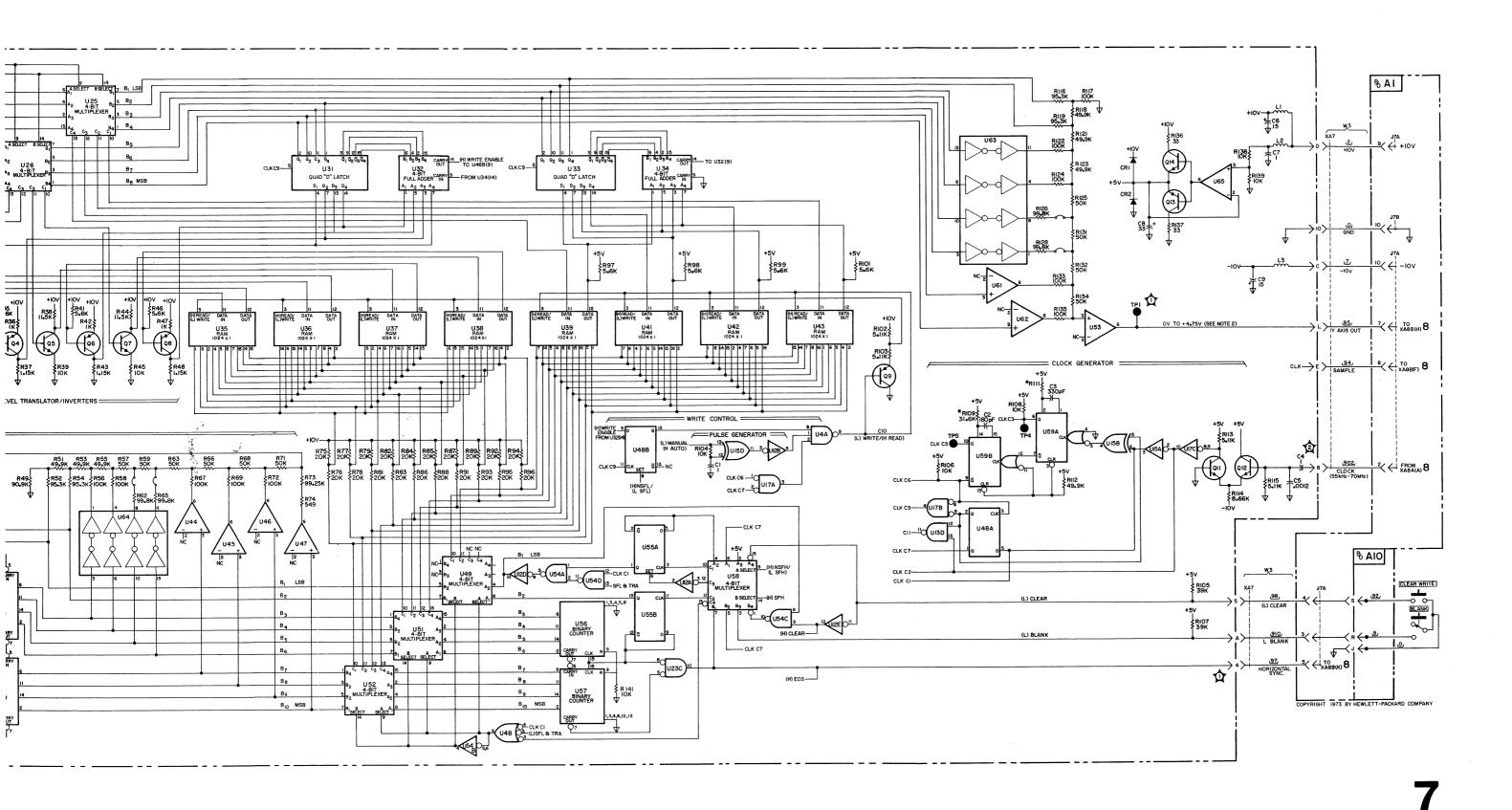
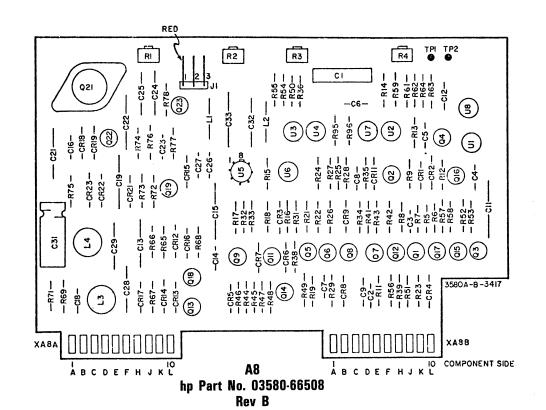
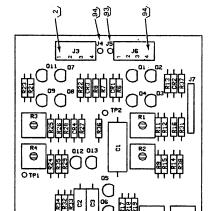


Figure 7-10. Digital Storage Assembly (A7) Schematic and Component Location Diagram.



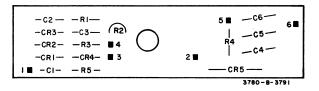
R1 R2 R3

A40 hp Part No. 03580-66540 Rev A

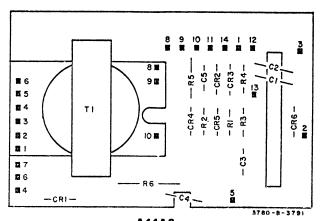


A13 hp Part No. 03580-66513 Rev B

NOTE: For serial numbers 2030A06029 and below. R37, R38, and R39 are used when the A13 Rev B board is used without an A40 board.



A11A1 -hp- Part No. 03580-66531 Rev. A



A11A2 -hp- Part No. 03580-66532 Rev. A

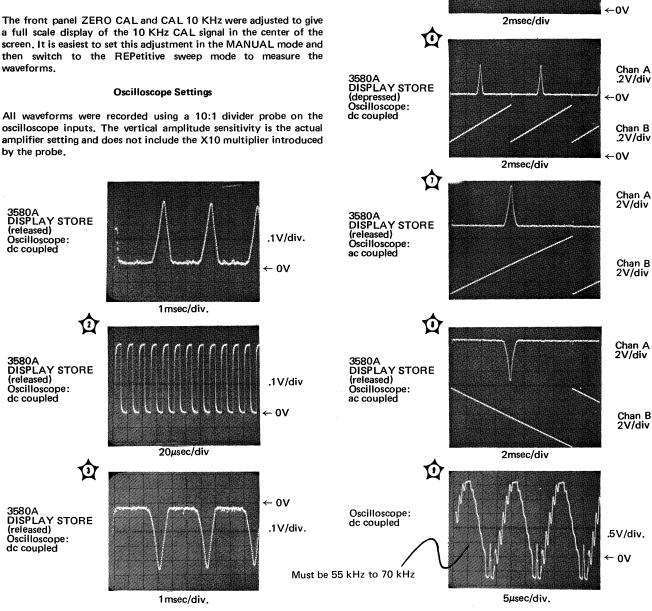
These waveforms were made with the 3580A front panel controls adjusted as follows:

| ADAPTIVE SWEEPOFF |
|--|
| DISPLAY BLANK STORE released |
| STORE as indicated |
| AMPLITUDE MODELOG 10 dBv/DIV |
| AMPLITUDE REF LEVELNORMAL |
| $dBv/LIN - dBm 600 \Omega \dots dBv/LIN$ |
| INPUT SENSITIVITYCAL |
| VERNIER (Amplitude) |
| (Fully CW) |
| FREQUENCY 10.0 kHz |
| START - CTRCTR |
| RESOLUTION BANDWIDTH |
| DISPLAY SMOOTHINGMIN |
| FREQ. SPAN/DIV |
| SWEEP TIME/DIV1 SEC |
| SWEEP MODEREF |

Option 002: Set dBm 900 Ω/LIN - dBm 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL.

a full scale display of the 10 KHz CAL signal in the center of the screen. It is easiest to set this adjustment in the MANUAL mode and then switch to the REPetitive sweep mode to measure the waveforms.

oscilloscope inputs. The vertical amplitude sensitivity is the actual amplifier setting and does not include the X10 multiplier introduced by the probe.



妏

⑫

.1V/div.

Chan A ,2V/div

Chan B .2V/div

←0V

← 0V

1msec/div.

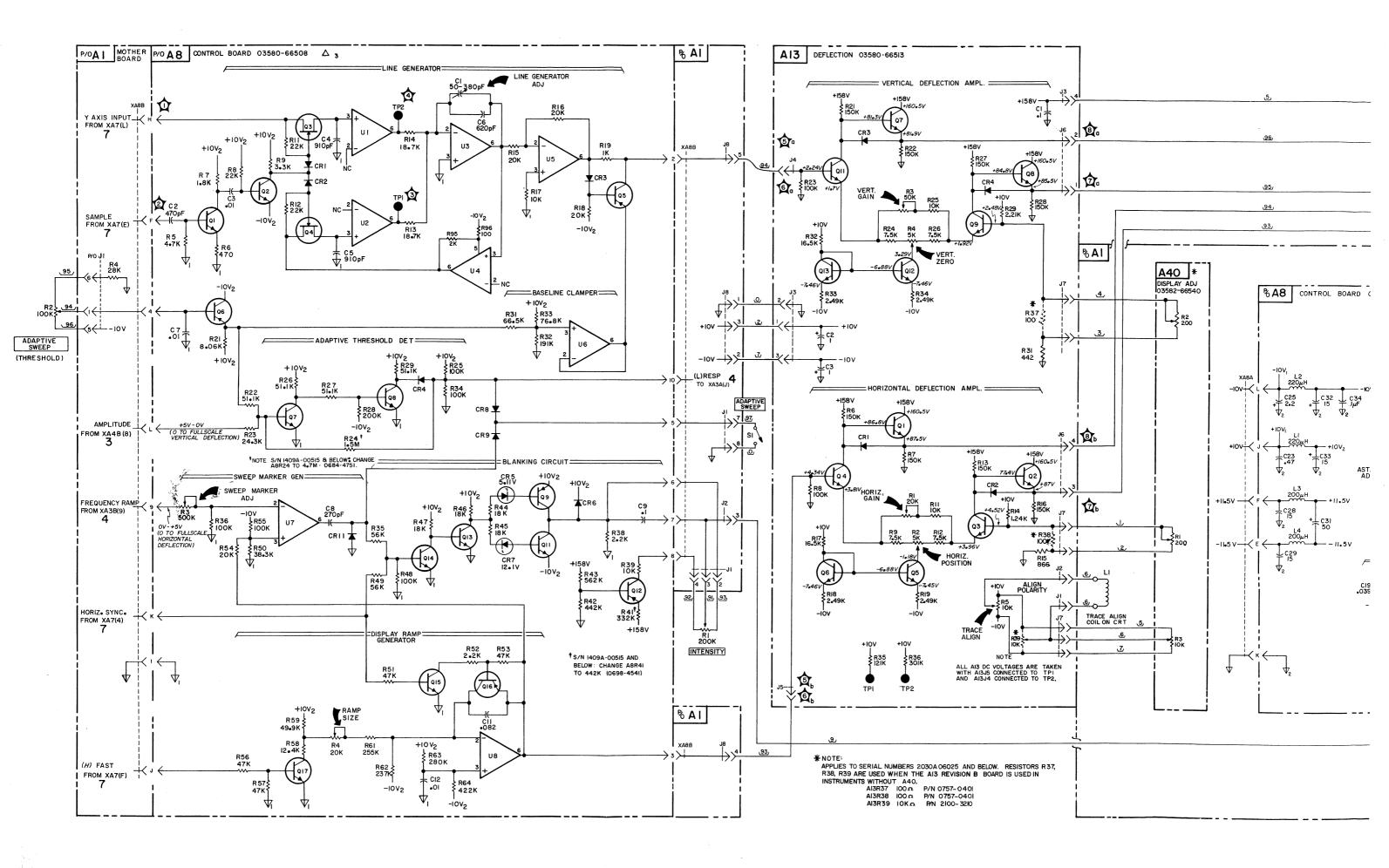
3580A

dc coupled

DISPLAY STORE (released) Oscilloscope:

3580A DISPLAY STORE (released) Oscilloscope:

dc coupled



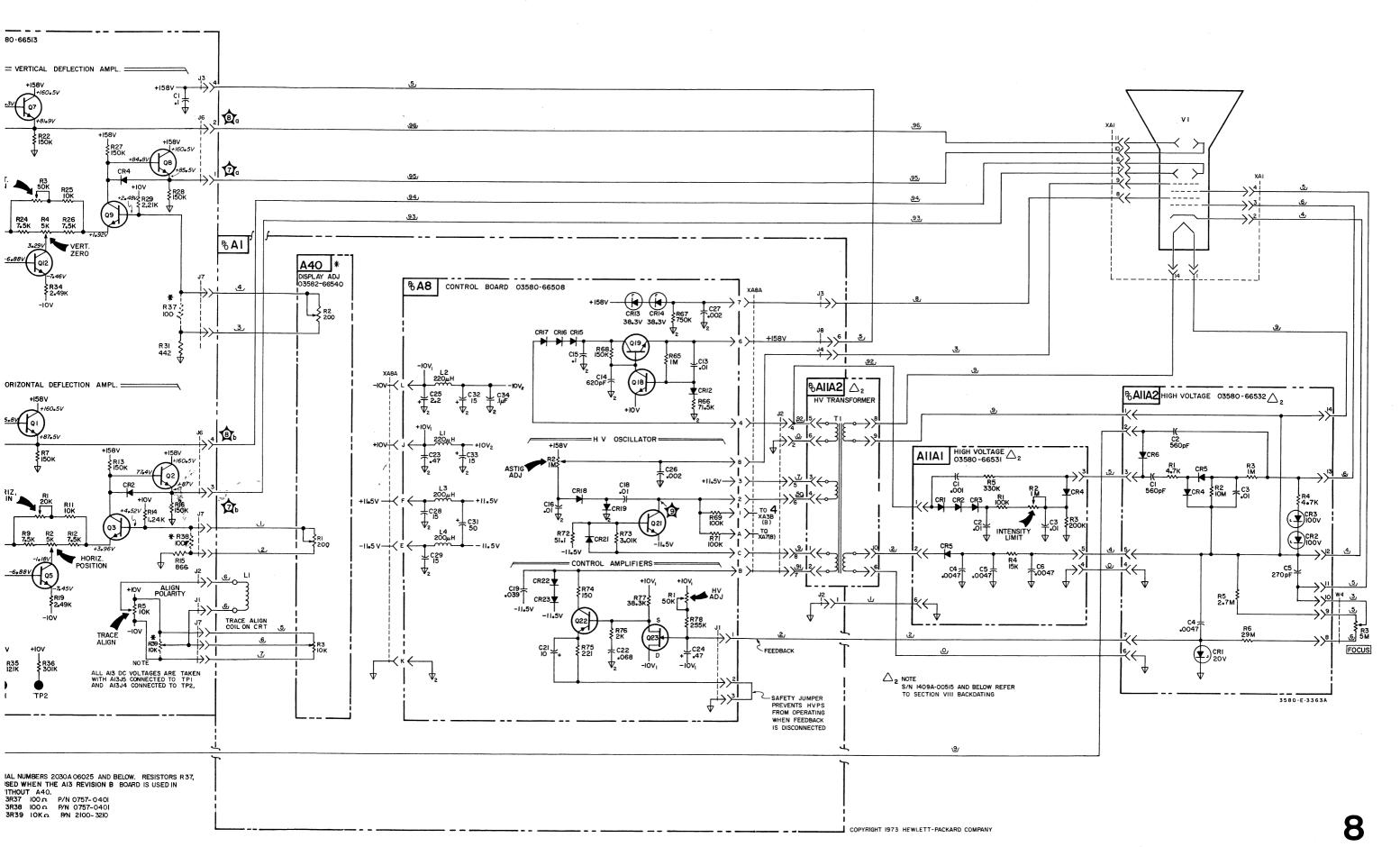
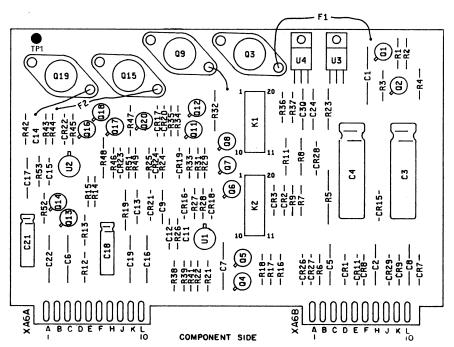


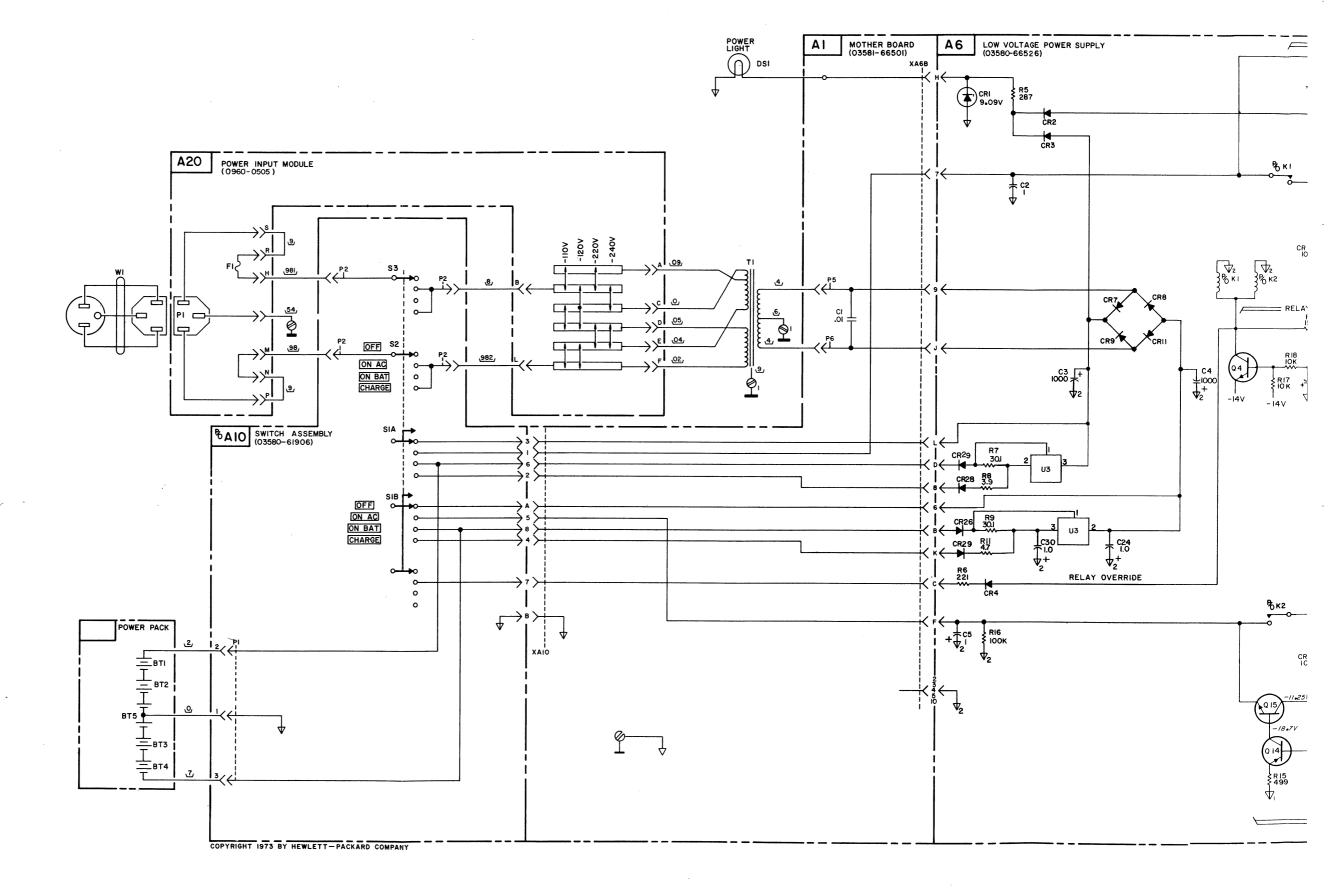
Figure 7-11. Control Assembly (A8), Deflection Amplifier (A13) and High Voltage Power Supply (A11) Schematics and Component Location Diagrams.



A6 hp Part No. 03580-66526 Rev C

*NOTE

 \pm 11.5 V Power Nominal \pm 14 V Power Nominal Standard Instrument \approx \pm 20 V Opt 001 (Line Powered) \approx \pm 18 V Opt 001 (Battery Powered) \approx \pm 12 V



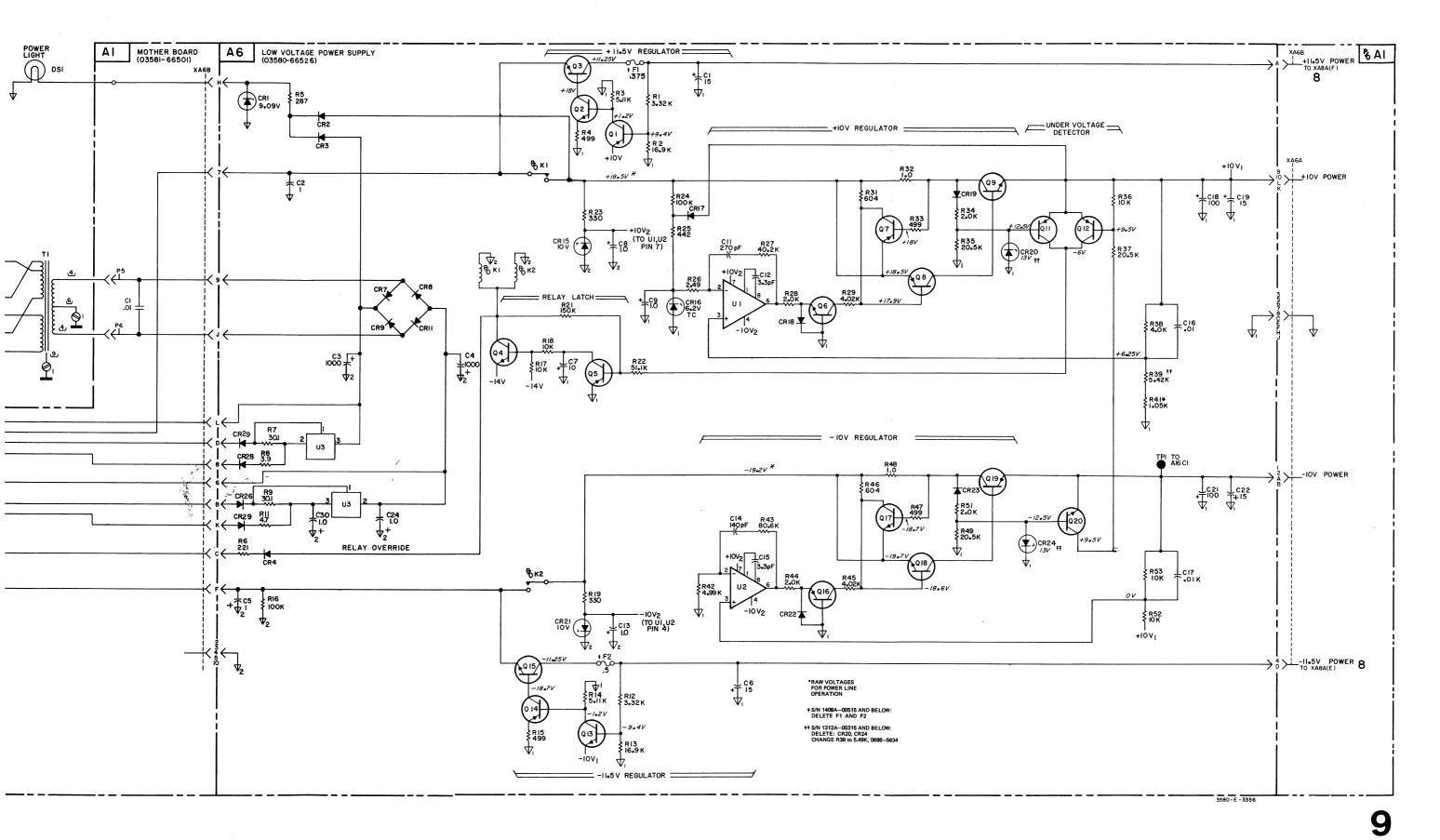


Figure 7-12. Low Voltage Power Supply (A6) Schematic and Component Location Diagram, and Power Input Module (A20) Schematic.

SECTION VIII BACKDATING

8-1. INTRODUCTION.

8-2. This section contains backdating changes which make this manual applicable to earlier instruments. Where possible, backdating changes have been integrated into the manual text, parts list and schematic diagrams. Changes that are too long or otherwise impractical to integrate into the manual are covered in this section. Backdating changes included in this section are referenced by a numbered delta (Δ_1) which appears in the text, parts list and schematic diagrams. The number indicates the number of the corresponding backdating change. Make all backdating changes that apply to your instrument.

CHANGE NO. Δ_1 : Applies to Option 002 instruments with serial number 1312A-00465 and below.

Table 1-1: Change Balanced Input Frequency Response specification to ± 0.5 dB, 300 Hz to 20 kHz.

Paragraph 3-188: Change CAUTION to read as follows:

ECAUTION 3

When using the balanced terminated input configuration, the differential input level must

not exceed $+27 \, dBm$ or $\pm 15 \, V$ dc. Exceeding these input levels will damage the input circuitry.

Figure 3-25(C): Change the Terminated input configuration as shown in Figure 8-1.

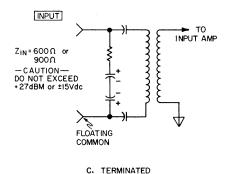


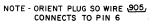
Figure 8-1. Balanced-Terminated Input Configuration.

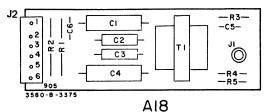
Table 5-13 (pp. 5-15): Delete 40 Hz from the Frequency Response Test.

Table 6-1: Change the Balanced Input Assembly (A18) parts list as follows:

| A18 | 03580-66518 | 1 | BOARD ASSY:INPUT, BALANCED (OPTION 002) | 28480 | 03580-66518 |
|---|-------------------------------------|---|--|----------------------------------|--|
| A18C1 | 0170-0042 | | C:FXD MY 0.33UF 5% 100VDCW | 99515 | E1-334D TYPE E120 |
| A18C2 | 0180-0228 | | C:FXD ELECT 22 UF 10% 15VDCW | 56289 | 1500226X901582-DYS |
| A18C3 | 0180-0228 | | C:FXD ELECT 22 UF 10% 15VDCW | 56289 | 1500226X901582-DYS |
| A18C4 | 0170-0042 | | C:FXD MY 0.33UF 5% 100VDCW | 99515 | E1-334D TYPE E120 |
| A18C5 | 0160-2206 | | C:FXD MICA 160 PF 5% | 28480 | 0160-2206 |
| A18C6 | 0140-0204 | | C:FXD MICA 47 PF 5% | 14655 | RDM15E470J5C |
| A18J1 | 1251-2969 | | CONN:PHONO, SINGLE JACK | 27264 | 15-24-0501 |
| A18J2 | 1251-3638 | | CONN:POST | 28480 | 1251-3638 |
| A18R1 A18R2 A18R3 * A18R3 A18R4 | 0757-0819 0698-4870 0757-0284 | 1 | R:FXD MET FLM 909 OHM 1% 1/2W R:FXD FLM 604 OHM 1.0% 1/2W R:FXD MET FLM 150 OHM 1% 1/8W FACTORY SELECTED PART R:FXD MET FLM 200K OHM 1% 1/8W | 28480 28480 28480 28480 | 0757-0819 0698-4870 0757-0284 0757-0472 |
| A18R5 | 0698-3245 | 1 | R:FXO MET FLM 20.5K OHM 1% 1/8W | 28480 | 0698-3245 |
| A18T1 | 9100-1460 | | TRANSFORMER AUDIO | 28480 | 9100-1460 |

Schematic No. 1: Change the Balanced Input Assembly (A18) schematic and component locator as shown in Figure 8-2.





hp Part No 03580-66518 Rev A

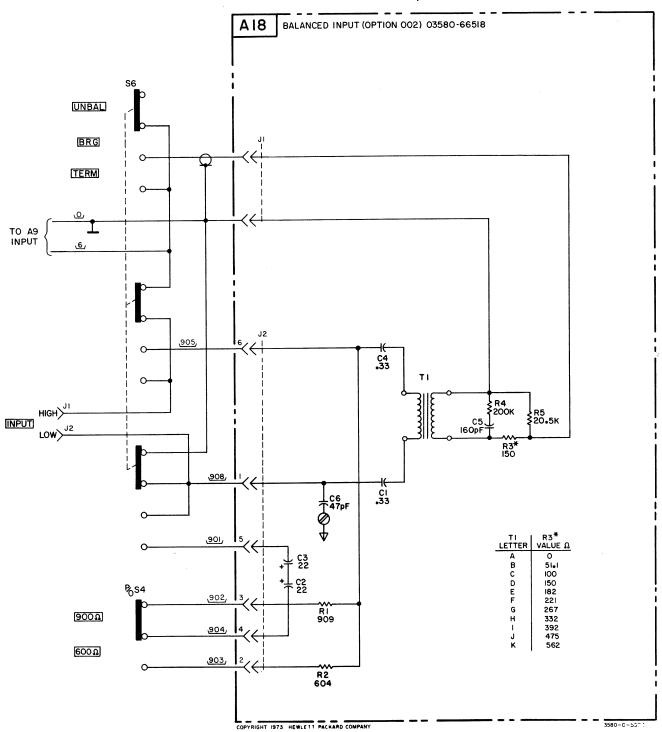


Figure 8-2. Balanced Input Assembly.

Model 3580A Section VIII

CHANGE NO. Δ_2 : Applies to instruments with the following serial numbers:

1312A-00399 and lower

1312A-00402

1312A-00403

1312A-00405

1312A-00408

1312A-00410

1312A-00413

1312A-00416 thru 1409A-00515

Table 6-1: Change the High Voltage Power Supply (A11) parts list as follows:

| All | 03580-64201 | 1 | POWER SUPPLY-HIGH VOLTAGE | 28480 | 03580-64201 |
|---------------|------------------------|-------|--|----------------|-------------------------|
| | 1251-3069 1251-3201 | 1 1 | CONNECTOR:PC 8 MALE CONTACT CONNECTOR:POST TYPE 3-CONTACT POSITION | 28480 27264 | 1251-3069 09-50-7031 |
| A11A1 | 0358066511 | 1 1 | PC ASSY:POWER SUPPLY 1, HIGH VOLTAGE | 28480 | 03580-66511 |
| Allaici | 0160-3007 | 5 | C:FXD CER 4700 PF 20% 4K VDCW | 72982 | 3888-024-Y5S0-472M |
| A11A1C2, C3 | 0160-3008 | 1 4 1 | C:FXD CER 4700 PF 20% 4K VDCW | 72982 | 3888-024-Y5SO-472M |
| A11A1C4 | 0160-3007 | · | C:FXD CER 4700 PF 20% 4K VDCW | 72982 | 3888-024-Y5SO-472M |
| A11A1C5 | 0160-3008 | | C:FXD CER 4700 PF 20% 4K VDCW | 72982 | 3888-024-Y5SO-472M |
| | | | | | |
| A11A1CR1, CR2 | 1901-0341 | 2 | DIODE:SI 7000 PIV 50MA | 28480 | 1901-0341 |
| Allalel | 2100-3359 | 1 1 | R:VAR CERMET 2 MEGOHM 20% TYPE VI 1/2W | 28480 | 2100-3359 |
| Allair2 * | 0687-1041 | 1 | R:FXD COMP 100K OHM 10% 1/2W | 01121 | EB 1041 |
| A11A2 | 03580-66512 | 1 | FACTORY SELECTED PART BOARD ASSY: POWER SUPPLY 2 - DOESN'T INCLUDE A11A2T1 | 28480 | 03580-66512 |
| A11A2C1 | 0160-3007 | | C:FXD CFR 4700 PF 20% 4K VOCW | 72982 | 3888-024-Y5S0-472M |
| A11A2C2 | 0160-3008 | | C:FXU CER 4700 PF 20% 4K VDCW | 72 9 82 | 3888-024-Y550-472M |
| A11A2C3 | 0160-3007 | | C:FXD CFR 4700 PF 20% 4K VDCW | 72982 | 3888-024-Y5SD-472M |
| A11A2C4 | 0160-3007 | | C:FXD CER 4700 PF 20% 4K VDCW | 72982 | 3888-024-Y5SO-472M |
| A11A2C5 | 0160-3007 | | C:FXD CER 4700 PF 20% 4K VDCW | 72982 | 3888-024-Y5SO-472M |
| A11A2C6 | 0160-2544 | 1 1 | C:FXD CER 270 PF 10% 1000 VDCW | 56289 | C0168102E271KS27-CDH |
| Alla2CR1 | 1902-3428 | 2 | DIDDE BREAKDOWN:SILICON 100V 5% | 28480 | 1902-3428 |
| Alla2CR2 | 1902-3428 | | DIODE BREAKDOWN: SILICON 100V 5% | 28480 | 1902-3428 |
| Alla2CR3 | 1902-3237 | 1 1 | DIODE BREAKDOWN: SILICON 20.0V 5% | 28480 | 1902-3237 |
| A11A2R1 | 0836-0001 | 1 | R:FXD CARBON 50 MEGDHM 10% 2W | 28480 | 0836-0001 |
| Allazr2 | 0687-1051 | 1 | R:FXD COMP 1 MEGOHM 10% 1/2W | 01121 | E8 1051 |
| A11A2R3 | 0687-1531 | 1 | R:FXD COMP 15K OHM 10% 1/2W | 01121 | EB 1531 |
| A11A2R4 | 0687-2221 | 1 1 | R:FXD COMP 2200 DHM 10% 1/2W | 01121 | F8 2221 |
| A11A2R5 | 0687-2751 | 1 1 | R:FXO COMP 2.7 MEGOHM 10% 1/2W | 01121 | FB 2751 |
| Allazre | 0698-8427 | 1 | R:FXD MET FLM 29 MFGOHM 10% 1.0W | 28480 | 0698-8427 |
| A11A2T1 | 9100-3263 | 1 | TRANSFORMER: H.V. (INCLUDES 03580-66517) | 28480 | 9100-3263 |

Schematic No. 8: Use the High Voltage Power Supply schematic (Figure 8-6) in place of the existing schematic.

CHANGE NO. Δ_3 : Applies to instruments with serial numbers 1415A00935 and below.

Table 6-1, Page 6-22. Delete A8R95 and A8R96 from the A8 assembly parts list.

Figure 7-11, Page 7-29/7-30. Change the A8 schematic as shown in Figure 8-3.

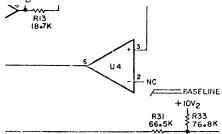


Figure 8-3. Control Board Circuit Change.

CHANGE NO. Δ_4 : Applies to instruments with serial numbers 1415AA00740 and below.

The new crystal used on the A2 board, Tracking Oscillator Assembly, differs in size from that used in the serial numbers listed above (see Figure 8-4 and 8-5). In order for the tie wrap to hold the new crystal, some new holes must be drilled in the A2 board.

Follow the Crystal Replacement procedure given in Section VII of the manual. While the A2 board crystal is removed, drill two holes in the A2 board about .120 inches (#31 drill bit) in diameter 1/4 inch above the existing tie wrap holes (see Figure 8-5). The new holes may now be used to secure the crystal to the board. The rest of the crystal replacement procedure is unchanged.

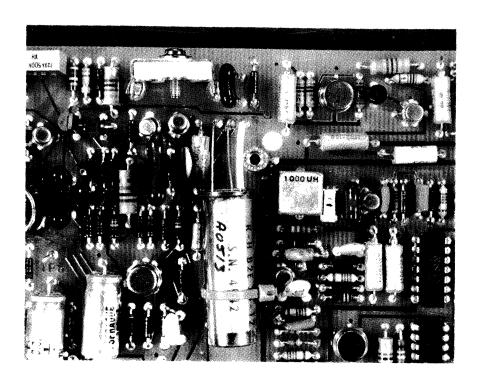


Figure 8-4. Old Style Crystal.

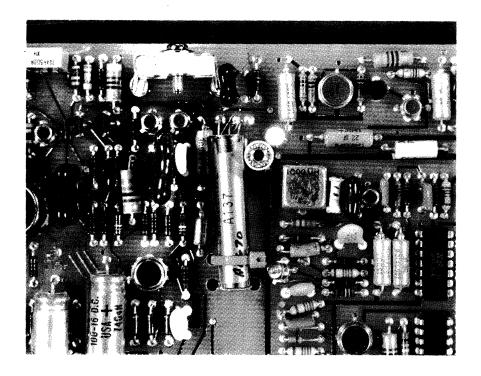
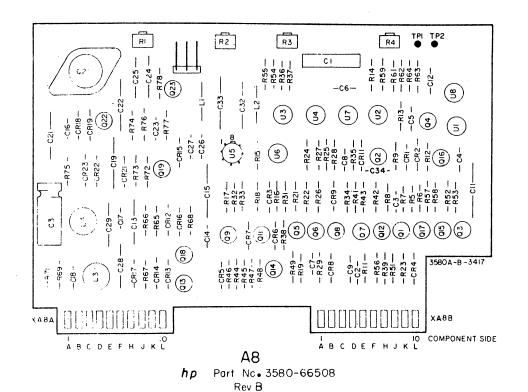
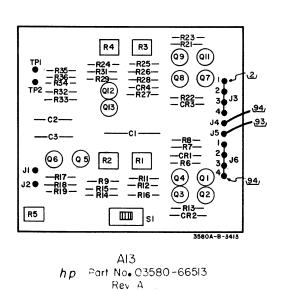
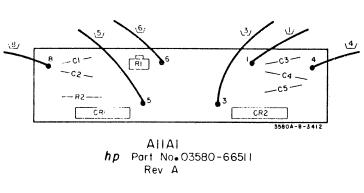
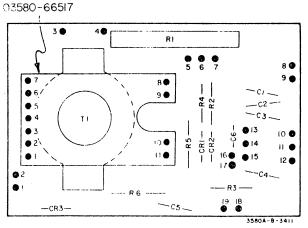


Figure 8-5. New Crystal on Modified A2 Board.









ATTA2

hp Part No. 03580-66512

Rev B

These waveforms were made with the 3580A front panel controls adjusted as follows:

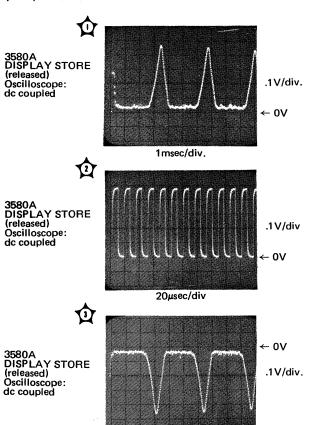
| ADAPTIVE SWEEP | OFF |
|----------------------------------|---------------------|
| DISPLAY BLA | NK STORE released |
| DIOI 27(1 | • |
| | STORE as indicated. |
| AMPLITUDE MODE | LOG 10 dBv/DIV |
| AMPLITUDE REF LEVEL | NORMAL |
| $dBv/LIN - dBm 600 \Omega \dots$ | dBv/LIN |
| INPUT SENSITIVITY | CAL |
| VERNIER (Amplitude) | |
| • • | (Fully CW) |
| FREQUENCY | 10.0 kHz |
| START - CTR | CTR |
| RESOLUTION BANDWIDTH | |
| DISPLAY SMOOTHING | |
| FREQ. SPAN/DIV | |
| | |
| SWEEP TIME/DIV | |
| SWEEP MODE | REP |
| | |

Option 002: Set dBm 900 Ω/LIN - dBm 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL.

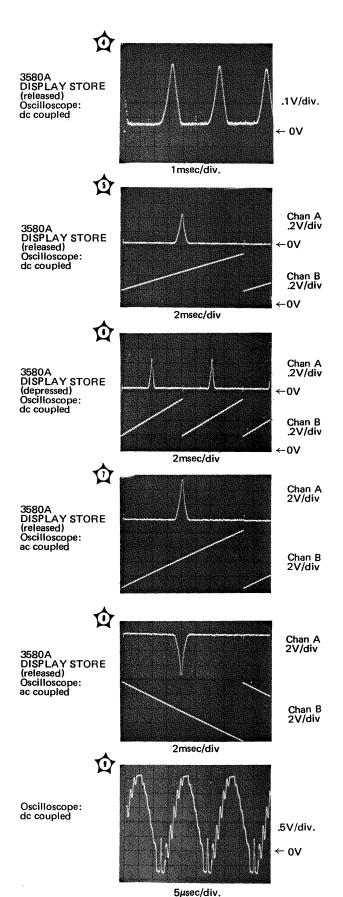
The front panel ZERO CAL and CAL 10 KHz were adjusted to give a full scale display of the 10 KHz CAL signal in the center of the screen. It is easiest to set this adjustment in the MANUAL mode and then switch to the REPetitive sweep mode to measure the waveforms.

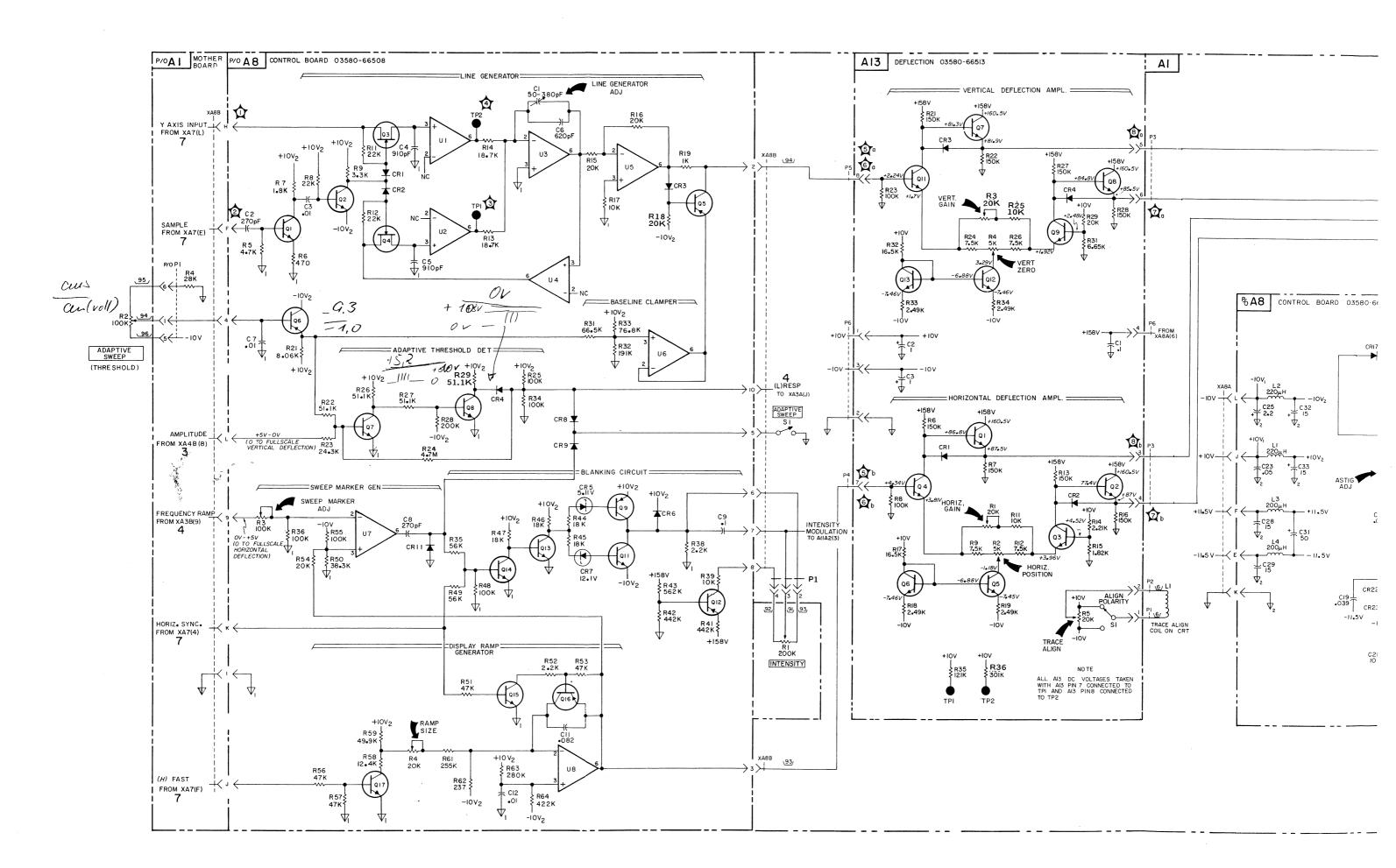
Oscilloscope Settings

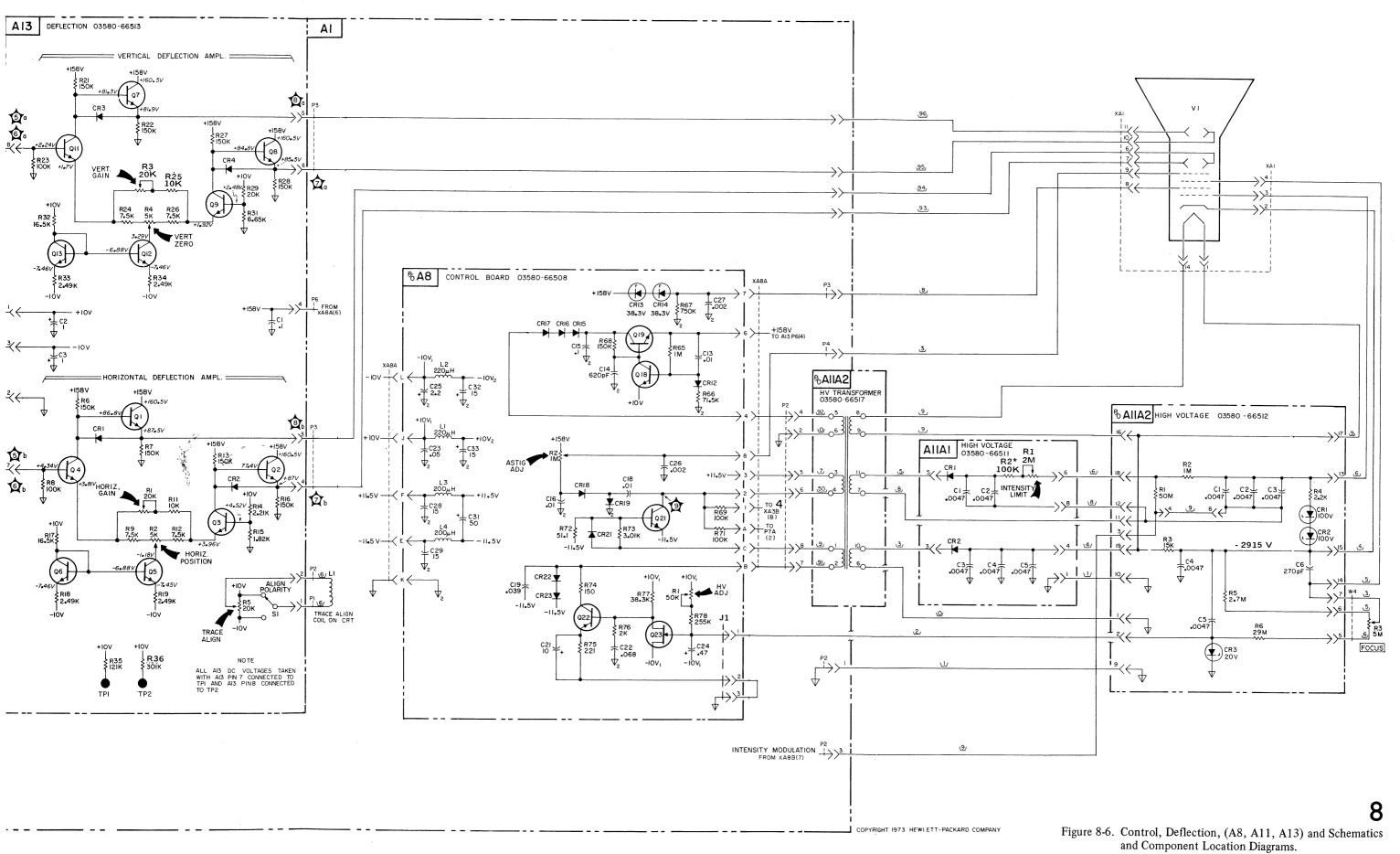
All waveforms were recorded using a 10:1 divider probe on the oscilloscope inputs. The vertical amplitude sensitivity is the actual amplifier setting and does not include the X10 multiplier introduced by the probe.



1msec/div.







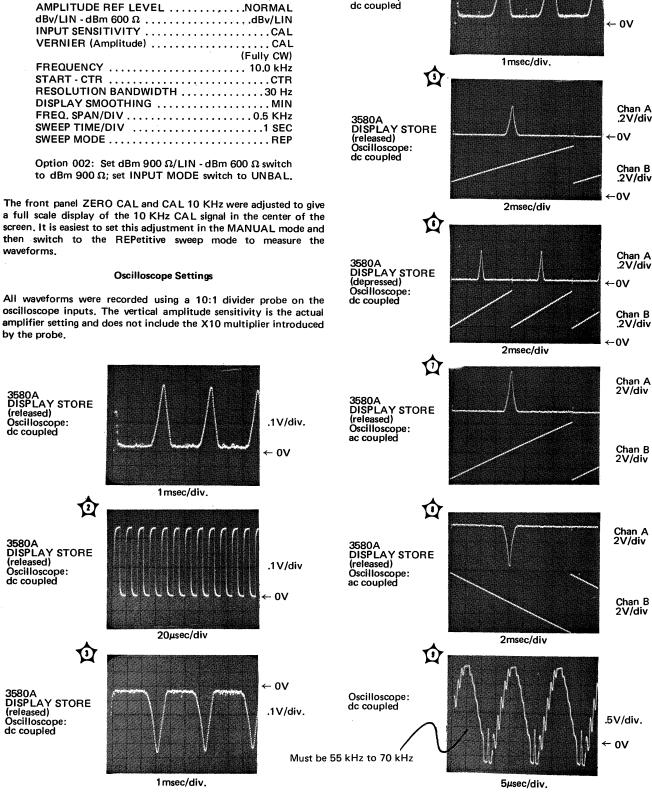
Location Diagrams. 8-7/8-8

These waveforms were made with the 3580A front panel controls adjusted as follows:

| ADAPTIVE SWEEPOFF |
|--|
| DISPLAY BLANK STORE released, |
| STORE as indicated. |
| AMPLITUDE MODELOG 10 dBv/DIV |
| AMPLITUDE REF LEVELNORMAL |
| $dBv/LIN - dBm 600 \Omega \dots dBv/LIN$ |
| INPUT SENSITIVITY |
| VERNIER (Amplitude) |
| (Fully CW) |
| FREQUENCY 10.0 kHz |
| START - CTRCTR |
| RESOLUTION BANDWIDTH30 Hz |
| DISPLAY SMOOTHINGMIN |
| FREQ. SPAN/DIV |
| SWEEP TIME/DIV1 SEC |
| SWEEP MODE REP |

a full scale display of the 10 KHz CAL signal in the center of the screen. It is easiest to set this adjustment in the MANUAL mode and then switch to the REPetitive sweep mode to measure the waveforms.

All waveforms were recorded using a 10:1 divider probe on the oscilloscope inputs. The vertical amplitude sensitivity is the actual amplifier setting and does not include the X10 multiplier introduced

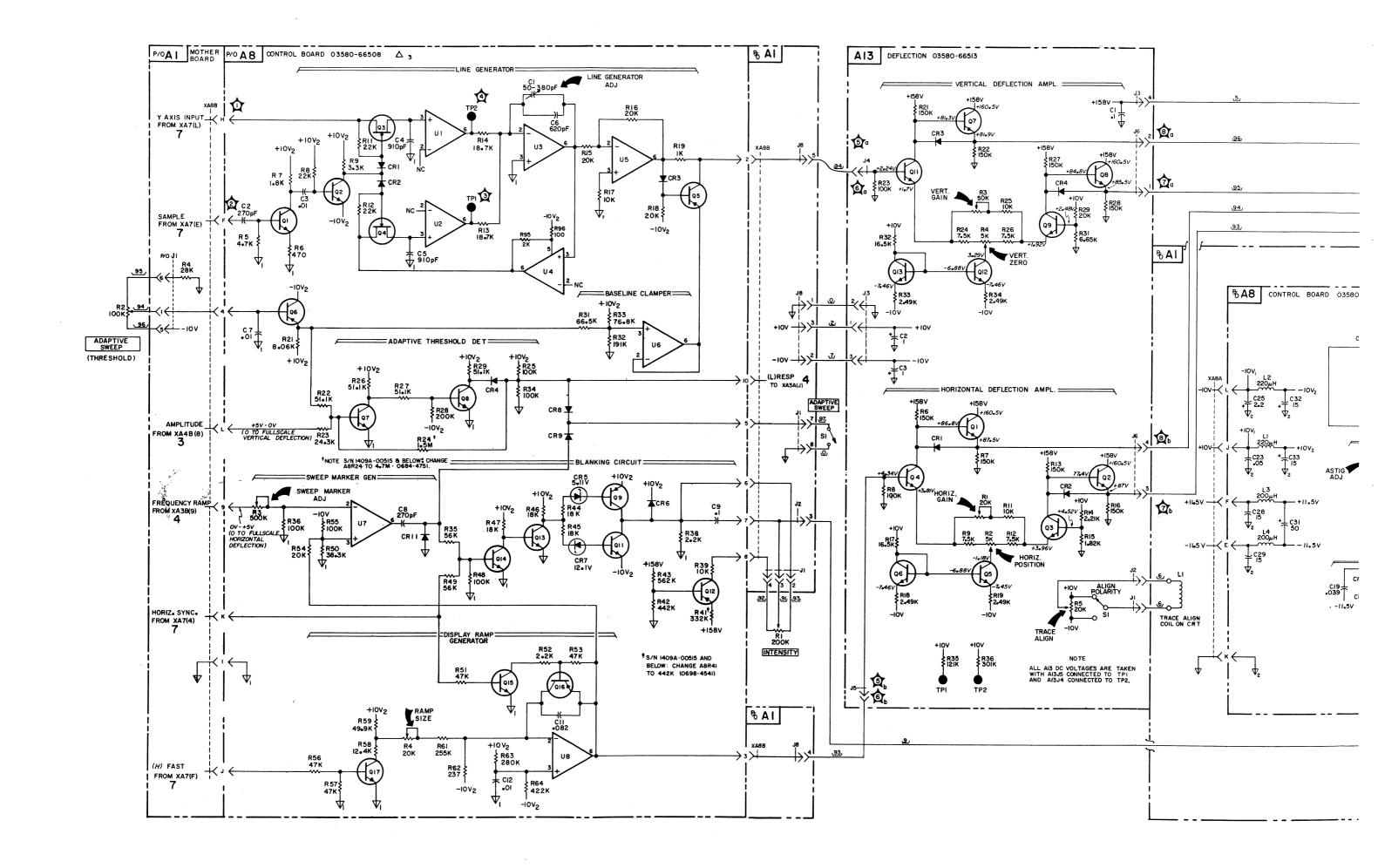


位

.1V/div.

3580A DISPLAY STORE (released)

Oscilloscope:



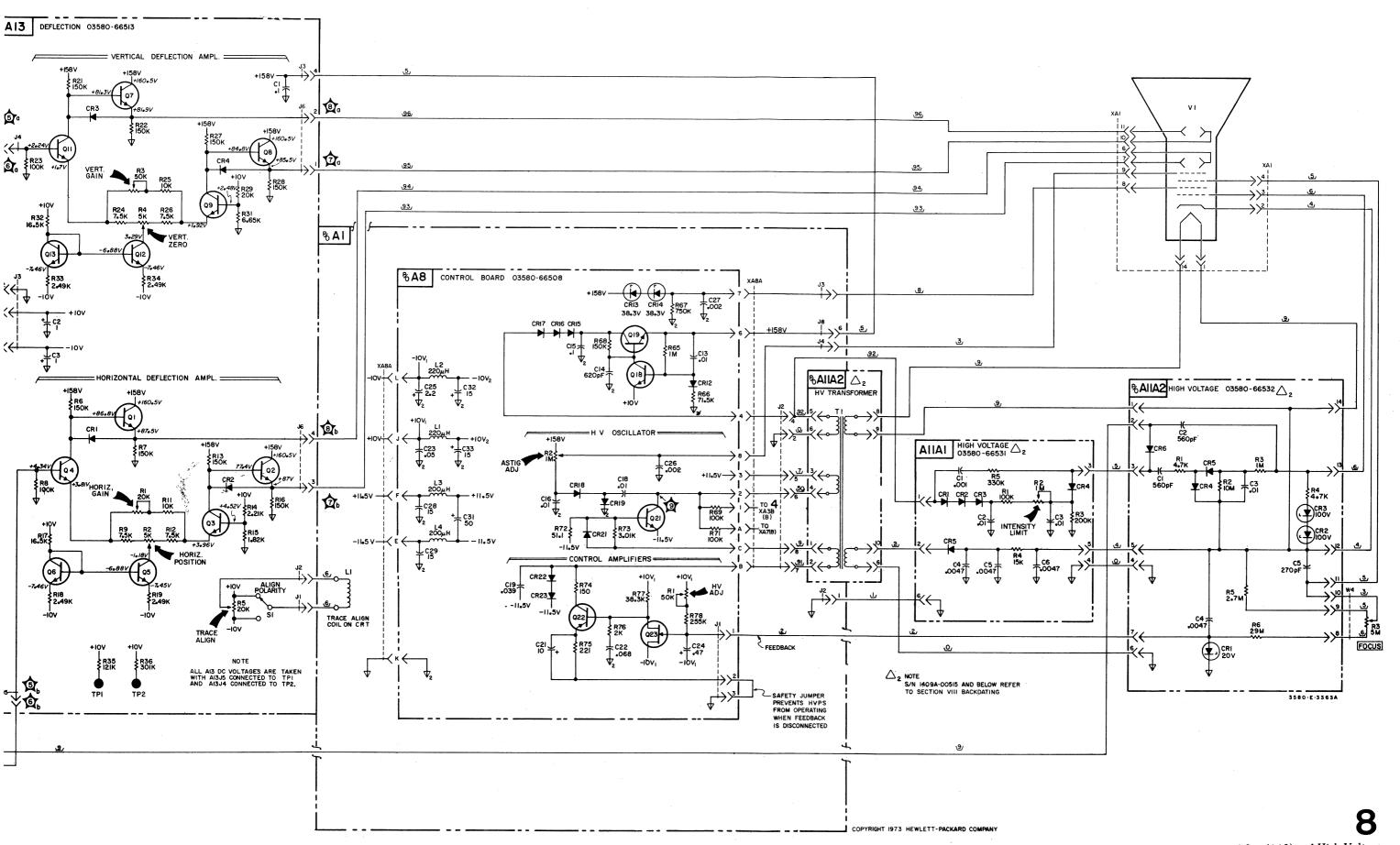


Figure 8-6a. Control Assembly (A8), Deflection Amplifier (A13) and High Voltage Power Supply (A11) Schematics and Component Location Diagrams.

CHANGE NO. Δ_5 : Applies to instruments with serial numbers 1415A02090 and below.

Page 6-5. Change the part numbers of the following:

| A2U2, 5, 8 | to | 1820-0600 |
|------------|----|-----------|
| A2U6 | to | 1820-0594 |
| A2U11 | to | 1820-0587 |

Page 6-7. Change the part numbers of the following:

| A3U5 | to | 1820-0777 |
|--------------|----|-----------|
| A3U6 | to | 1820-0595 |
| A3U7 | to | 1820-0594 |
| A3U8 | to | 1820-0583 |
| A3U9, 13, 14 | to | 1820-0588 |
| A3U11 | to | 1820-0584 |
| A3U12 | to | 1820-0587 |
| A3R52 | to | 0684-2231 |

Page 6-20. Change A7U15 to 1820-0947.

Page 7-21/7-22, Figure 7-7. Change the value of R52 on schematic 4 to 22 k ohms.

CHANGE NO. Δ_6 : Applies to instruments with serial number 1415A02010 and below.

Page 6-3. Change A3CR2, 3 to 0122-0059.

Page 6-16. Change A6F1, F2 to 2110-0343.

CHANGE NO. Δ_7 : Applies to instruments with serial number 1415A02050 and below.

Page 6-17. Delete 1400-0507 from the 03580-66506 Hardware List.

Page 6-35. Change part number 0370-2994 to 0370-2473.

CHANGE NO. Δ_8 : Applies to instruments with serial number 1415A02280 and below.

Page 6-11. Change the part numbers of the following:

| A4R71 | to | 0757-0434 |
|---------|----|-----------|
| A4R109 | to | 0698-4430 |
| A4R122B | to | 0757-0465 |

Page 7-19/7-20, Figure 7-6. Change the values of these components on Schematic 3:

| R71 | to | 174 k ohms |
|------|----|-------------|
| R109 | to | 1.87 k ohms |

CHANGE NO. Δ_9 : Applies to instruments with serial number 1415A02140 and below.

Page 6-33. Delete part number 7124-2308.

CHANGE NO. Δ_{10} : Applies to instruments with serial number 1415A01775 and below.

CHANGE NO. Δ_{11} : Applies to instruments having a serial number of 1415A03290 or lower.

Page 6-5. Change A3C24 to 0140-0149.

Page 6-6. Change A3R16 to 0684-1031.

Page 7-21/7-22, Figure 7-7. Change the following values on Schematic 4:

| C24 | to | 470 pF |
|-----|----|----------|
| R16 | to | 10 k ohm |

CHANGE NO. Δ_{12} : Applies to instruments having a serial number of 1415A03390 or lower.

Page 6-16. Add:

A6K1, 0490-0366, Sw-Reed A6K1, 0490-0515, Coil Assy A6K2, 0490-0366, Sw-Reed A6K2, 0490-0515, Coil Assy A6K1, 0490-1208, Reed Relay A6K2, 0490-1208, Reed Relay

CHANGE NO Δ_{13} : Applies to instruments having a serial number of 1415A03490 or lower.

Page 6-34. Under Misc. Mechanical Parts: Add 03580-23701 Rail-Rt. Side. Change the description of 03580-23702 to Frame: Left Side Rail, Quantity 1.

CHANGE NO. Δ_{14} : Applies to instruments having a serial number of 1415A03590 or lower.

Page 6-35. Change R7 to 2100-0574.

CHANGE NO. Δ_{15} : Applies to instruments having serial numbers of 1415A03440 and lower.

Page 6-16.

a. Delete the following parts:

A6CR26 A6CR27 A6CR28 A6CR29 A6C24 A6C30

b. Change the following parts:

| A6C3 | to | 0180-1943 |
|--------|----|-----------|
| A6C4 | to | 0180-1943 |
| A6C8 | to | 0180-0197 |
| A6C9 | to | 0180-0197 |
| A6C13 | to | 0180-0197 |
| A6CR7 | to | 1901-0045 |
| A6CR8 | to | 1901-0045 |
| A6CR9 | to | 1901-0045 |
| A6CR11 | to | 1901-0045 |
| A6F1 | to | 2110-0343 |
| A6F2 | to | 2110-0343 |
| | | |

Page 6-17.

- a. Change 03580-21102 to 03580-21101.
- b. Change the following parts:

| A6R7 | to | 0757-0799 |
|-------|----|-----------|
| A6R8 | to | 0766-0014 |
| A6R9 | to | 0757-0799 |
| A6R11 | to | 0766-0014 |
| A6R19 | to | 0757-0809 |
| A6R23 | to | 0757-0809 |

c. Change the part number of INSULATOR:

TRANSISTOR to 0340-0162

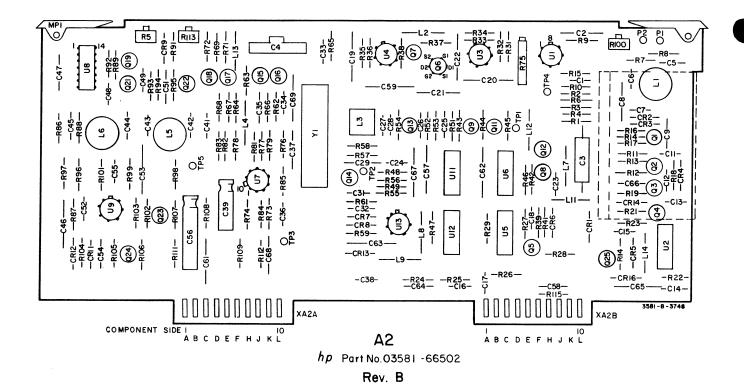
d. Delete the following parts:

A6U3 A6U4

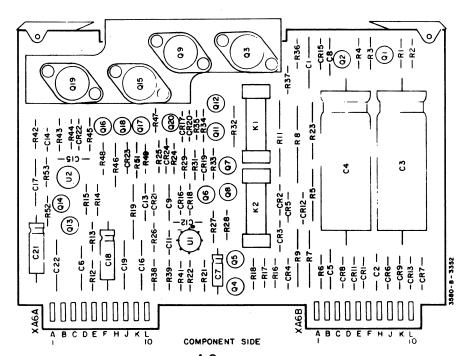
Page 7-15/7-16, Figure 7-4. Change J9A at "Cal Pwr to A2A" to J9B.

Page 7-23/7-24, Figure 7-8. On board A15 change R to L, L to R.

Page 7-25. Revise the component locator for the -hp-part number 03581-66502 as follows:



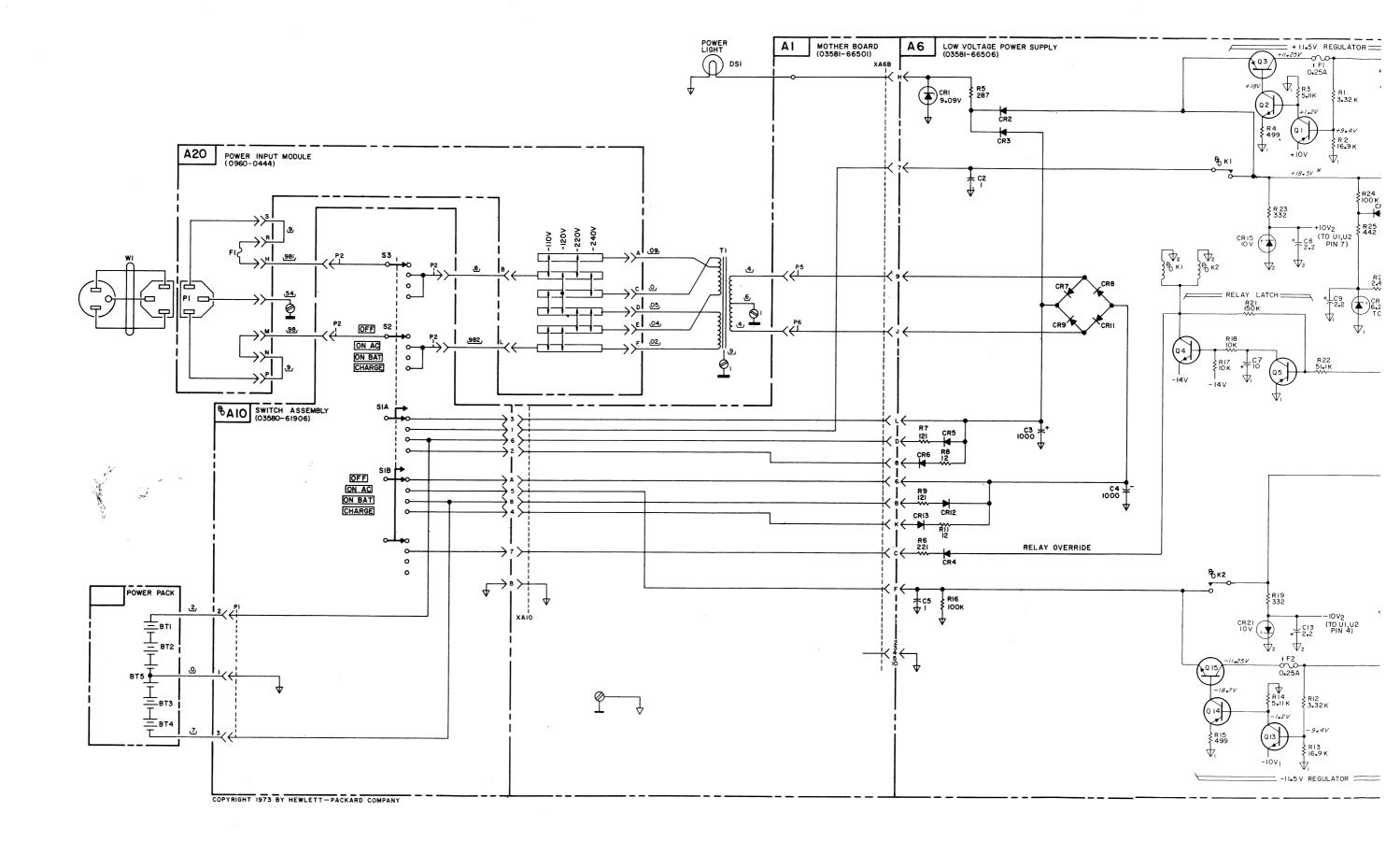
Page 7-31/7-32, Figure 7-12. Insert the following schematic drawing.

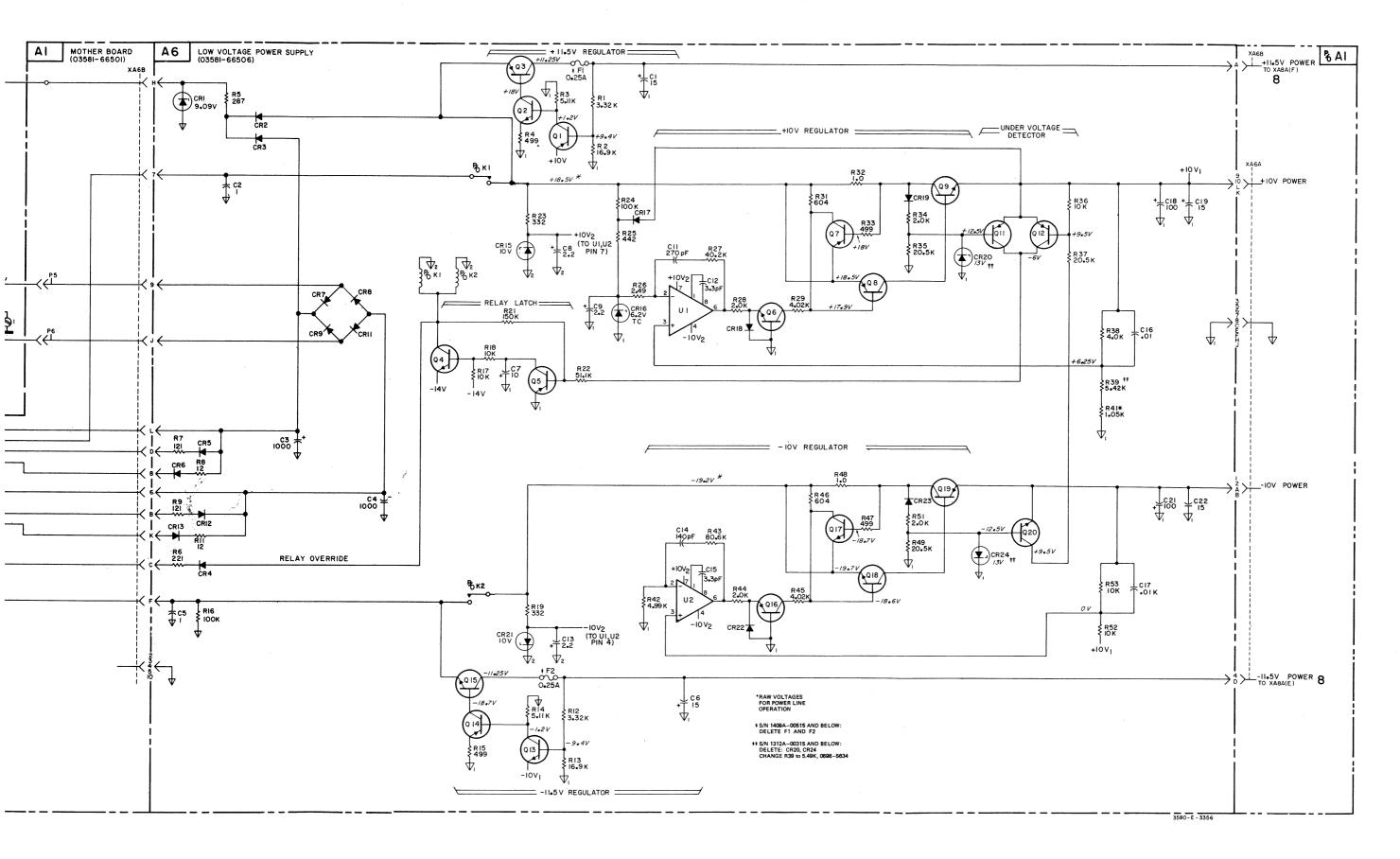


A 6 hp Part No. 03580-66506 Rev D

*NOTE

 \pm 11.5 V Power Nominal \pm 14 V Power Nominal Standard Instrument \approx \pm 20 V Opt 001 (Line Powered) \approx \pm 18 V Opt 001 (Battery Powered) \approx \pm 12 V

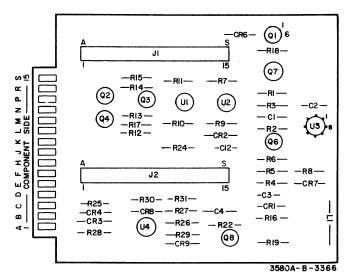




9

CHANGE NO. Δ_{16} . Applies to instruments having a serial number of 1415A04280 or lower.

Page 7-23/7-24, Figure 7-8. Change the component locator for the A16 board (-hp- part number 03580-66516) as follows:



AI6

hp Part No. 03580 – 66516 Rev A

Page 1-0, Table 1-1. Change the Frequency Dial Accuracy to \pm 100 Hz, 20°C to 30°C; \pm 300 Hz 0°C to 55°C.

Page 1-2, Table 1-2. The frequency control, dial, and stability should be as follows.

Frequency Control: The front panel FREQUENCY control tunes the frequency of the analyzer over the 0 Hz to 50 kHz range. The control can be used to set either the start or center frequency of linear or manual sweeps.

Coarse or Fine Tuning: Coarse tuning is selected by pushing the crank toward the front panel; fine tuning is selected by pulling the crank outward. In the coarse position, one revolution of the crank changes the frequency by approximately 2.7 kHz. In the fine position, one revolution of the crank changes the frequency by approximately 73 Hz.

Frequency Dial: Indicates start or center frequency in kHz.

Range: 00.0 kHz to approximately 50.8 kHz.

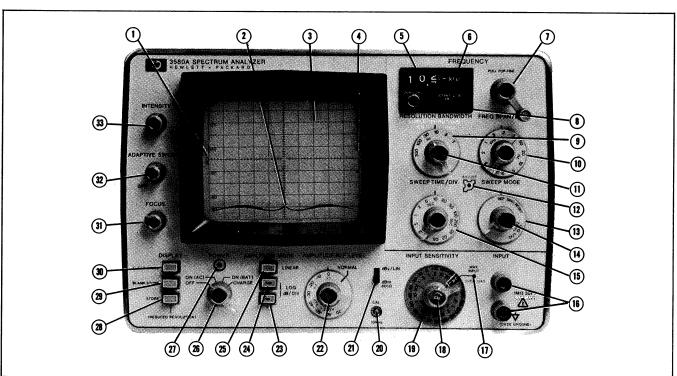
Resolution: 20 Hz (one minor division)

Typical Frequency Stability: ± 10 Hz/hr. after 1 hour; ± 5 Hz/°C

Page 3-2, Figure 3-1. Change items (5), (6), (7), and (8) to read as follows:

- FREQUENCY Dial: Indicates start or center frequency of linear or manual sweep. (Paragraph 3-99)
- § START/CTR Switch: When set to START position, FRE-QUENCY dial indicates start frequency of linear or manual sweep; when set to CTR position, FREQUENCY dial indicates center frequency of linear or manual sweeps. (Paragraph 3-100)
- FREQUENCY Control: Tunes frequency of instrument over 0 Hz to 50 Hz range. Is used to set start or center frequency of linear or manual sweeps. Push in for coarse tuning; pull out for fine tuning. (Paragraph 3-96)
- (8) ZERO CAL Potentiometer: Used to calibrate FREQUENCY dial for linear or manual sweeps and to set the correct starting point for log sweep. (Paragraph 3-102)

Page 3-2, Figure 3-1. The photograph of the front panel should appear as follows:



Page 3-15. Change paragraphs 3-97 through 3-102 to read as follows:

3-97. The front panel FREQUENCY control tunes the frequency of the analyzer over the 0 Hz to 50 kHz range. The control can be used to set either the start or center frequency of a linear sweep. The start or center frequency selected by the FREQUENCY control is indicated on the FREQUENCY dial.

3-98. The FREQUENCY control has two selectable drive ratios to permit coarse or fine tuning. Coarse tuning is selected by pushing the crank toward the front panel; fine tuning is selected by pulling the crank outward. In the coarse position, one revolution of the crank changes the FREQUENCY dial setting by approximately 2.7 kHz. In the fine position, one revolution of the crank changes the frequency by approximately 73 Hz.

3-99. Frequency Dial. The FREQUENCY dial indicates the start or center frequency in kHz. Dial settings range from 00.0 kHz to approximately 50.5 kHz. The frequency dial resolution is 20 Hz represented by one minor division on the frequency scale. When the instrument is properly calibrated (Paragraph 3-195), the frequency dial accuracy is:

- a. ± 100 Hz when the ambient temperature is within the range of 20°C (68°F) to 30°C (86°F).
- b. ± 300 Hz when the ambient temperature is within the range of 0°C (32°F) to 20°C (68°F) or 30°C (86°F) to 55°C (131°F).

3-100. Start/Center. With the START/CENTER slide switch in the START position, the FREQUENCY dial setting indicates the frequency represented by the first vertical line on the left-hand side of the display graticule. This is the "start frequency" or frequency at which the sweep begins. With the switch in the CENTER position, the FREQUENCY dial setting indicates the frequency represented by the center vertical line on the display graticule. This is the "center frequency" of the sweep.

3-101. When surveying a spectrum containing two or more signals, it is generally convenient to leave the START/CENTER switch in the START position. The FRE-QUENCY control can then be used to set the start frequency and the FREQUENCY SPAN control can be used to set the spectrum width or "end frequency". To observe one frequency component in a spectrum, set the START/CENTER switch to the CENTER position and set the FREQUENCY dial to the frequency of interest. The frequency of interest will appear in the center of the display. The width of the center frequency response can be adjusted by changing the FREQUENCY SPAN or BAND-WIDTH setting.

3-102. Zero Cal. Potentiometer. The purpose of the ZERO CAL potentiometer is to enable the operator to

compensate for slight variations in frequency dial accuracy that occur during warm-up or when the instrument is operated in an uncontrolled environment. The ZERO CAL potentiometer is also used in the Log Zero sweep mode to establish the correct starting point for the log sweep. Refer to Paragraph 3-195 for the Frequency Calibration Procedure.

Page 3-16. Delete the last two sentences of Paragraph 3-115.

Page 3-17. Change Paragraphs 3-119 through 3-124 to read as follows:

3-119. Reset Mode. When the Reset mode is selected, the sweep is reset to the left-hand side of the screen and the instrument remains at the *start* frequency determined by the FREQUENCY dial setting.

3-120. The Reset mode is used primarily for calibrating the FREQUENCY dial. In the Frequency Calibration Procedure (Paragraph 3-195), the Reset mode is selected and the FREQUENCY dial is set for a start frequency of 00.0 kHz. The ZERO CAL potentiometer is then adjusted so that the zero response peaks at 0 Hz on the display.

3-121. Manual Mode. In the Manual sweep mode, the electronic frequency sweep is disabled and frequency control is transferred to the MANUAL VERNIER potentiometer. By adjusting the MANUAL VERNIER, the frequency can be set anywhere within the selected spectrum. With the MANUAL VERNIER set fully counterclockwise, the CRT sweep is at the left-hand side of the screen and the instrument is tuned to the start frequency determined by the FREQUENCY setting. As the vernier is rotated in a clockwise direction, the frequency increases and the video information is written (and retained) on the CRT just as it is when using the electronic sweep.

3-122. The Manual sweep is useful for applications where it is necessary to precisely measure the frequency of a signal within the spectrum. For precise frequency measurements, an electronic counter is connected to the rear panel TRACKING OSC OUT or LO OUTPUT to monitor the frequency. Using a narrow bandwidth such as 10 Hz or 30 Hz, the MANUAL VERNIER is adjusted so that the CRT sweep is at the peak of the signal to be measured. If the TRACKING OSC OUT is used, the frequency of the signal can then be read directly from the counter. If the LO OUTPUT is used, the frequency must be calculated by dividing the counter reading by ten and subtracting 100 kHz (Paragraph 3-178).

NOTE

When the SWEEP MODE setting is changed from LOG ZERO to MAN or from RESET to MAN, the frequency sweep jumps from the start frequency to the frequency set by the MANUAL VERNIER. Conversely, when the SWEEP MODE is changed from MAN to LOG

ZERO or from MAN to RESET, the frequency sweep jumps from the frequency set by the MANUAL VERNIER to 0 Hz or to the start frequency. In either case, the rapid change in frequency will distort the trace being displayed on the CRT. If it is desirable to retain a specific trace when switching to or from the Manual mode, set the MANUAL VERNIER fully counterclockwise before changing the SWEEP MODE setting.

3-123. Log Zero Mode. The Log Zero mode is used to establish the correct starting frequency for the log sweep. When the Log Zero mode is selected, the sweep is reset to the left-hand side of the screen, the FREQUENCY and FREQUENCY SPAN controls are disabled and the start frequency is internally set to 0 Hz. To calibrate the log sweep, the front panel ZERO CAL potentiometer is adjusted to peak the zero response at the left-hand edge of the display graticule. Peaking the zero response at 0 Hz in the Log Zero mode nulls out any dc offsets in the frequency control circuit. This ensures that the log sweep will start at 20 Hz.

3-124. Log Sweep. When the Log sweep mode is selected, the following things take place:

- a. The FREQUENCY, FREQUENCY SPAN and SWEEP TIME controls are disabled and their settings do not effect the log sweep. The ZERO CAL potentiometer remains operative and, to ensure the proper starting point for the log sweep, must be adjusted for peak zero response in the Log Zero mode.
- b. The instrument sweeps logarithmically over the 20 Hz to 43 kHz frequency range. The log sweep is repetitive and the duration of each sweep is approximately 5 seconds.

NOTE

When the Log sweep mode is first selected or when the log sweep is initiated by external triggering, optimum frequency accuracy will not be obtained until 3 or 4 continuous sweeps have been made. This peculiarity of the Log sweep is caused by dielectric absorption (soak effect) in the integrating capacitor of the Log sweep generator.

Page 3-27. Change Paragraphs 3-195 through 3-198 to read the following:

3-195. Frequency Calibration Procedure.

3-196. The Frequency Calibration Procedure should be performed after warm-up each time the instrument is turned on. It should also be performed before and after using the log sweep.

- 3-197. For operation in the Repetitive, Single or Manual sweep mode, proceed as follows:
- a. Turn the instrument on as outlined in Paragraph 3-192.
 - b. Set the 3580A controls as follows:

A D A DOUGH OF CHURCH

| ADAPTIVE SWEEP | |
|-----------------------|-------------|
| DISPLAYSTORE and | BLANK STORE |
| | Released |
| AMPLITUDE MODE | |
| AMPLITUDE REF LEVEL . | NORMAL |
| dBv/LIN - dBm Switch | dBv/LIN |
| INPUT SENSITIVITY | CAL |
| VERNIER | CAL |
| | (Fully CW) |
| FREQUENCY | 00.0 kHz |
| START CTR | START |
| BANDWIDTH | |
| DISPLAY SMOOTHING | MIN |
| FREQ. SPAN/DIV | N/A |
| SWEEP TIME/DIV | N/A |
| SWEEP MODE | |
| | |

- c. Clear the display by pressing the CLEAR WRITE button.
- d. Adjust the front panel ZERO CAL potentiometer for peak zero response. (The zero response will appear on the first line on the left-hand side of the display graticule.)
 - e. Set the BANDWIDTH to 10 Hz. Repeat Step d.
- 3-198. For operation in the Log sweep mode, proceed as follows:
 - a. Set the 3580A controls as follows:

| ADAPTIVE SWEEP O | FF |
|-----------------------------|-----|
| DISPLAY STORE and BLANK STO | RE |
| Releas | sed |
| AMPLITUDE MODE LOG 10 dB/D | Ν |
| AMPLITUDE REF LEVELNORM. | |
| dBv/LIN - dBm SwitchdBv/L | |
| INPUT SENSITIVITY | |
| VERNIER | |
| (Fully C | (W) |
| FREQUENCY N | I/A |
| START CTR N | [/A |
| BANDWIDTH 30 | |
| DISPLAY SMOOTHING | IIN |
| FREQ. SPAN/DIV N | [/A |
| SWEEP TIME/DIV N | [/A |
| SWEEP MODE LOG ZEI | RO |
| | |

- b. Clear the display by pressing the CLEAR WRITE button.
- c. Adjust the front panel ZERO CAL potentiometer for peak zero response. (The zero response will appear on the first line on the left-hand side of the display graticule.)

Page 3-28.

- a. Paragraph 3-201, Step b. Change FREQUENCY to 10.0 kHz.
- b. Paragraph 3-201, Step e, should read as follows: "Pull out the Frequency control for fine tuning. Carefully adjust the FREQUENCY control for a peak 10 kHz response in the center of the display."
- c. Paragraph 3-202, Step b. Change FREQUENCY to 10.0 kHz.
- d. Paragraph 3-204, Step c. Change FREQUENCY to 00.0 kHz.

Page 3-29.

- a. Paragraph 3-206, Step a. The second sentence should read as follows: "Perform the Frequency Calibration Procedure (Paragraph 3-195) and the Amplitude Calibration Procedure (Paragraph 3-199).
- b. Paragraph 3-206, Step b. Change FREQUENCY to 00.0 kHz.

Page 3-30.

a. Paragraph 3-206, Step I. The second-to-the-last sentence should read as follows: "Set the FREQUEN-CY dial to 20.0 kHz."

Page 4-1. Paragraph 4-4 should read as follows:

4-4. Refer to the Simplified Block Diagram (Figure 4-1) for the following discussion.

The 3580A can be divided into four major sections:

- 1) Amplitude Section
- 2) Frequency and Sweep Section
- 3) Digital Storage Section
- 4) Display

Page 4-2. Delete the discussion of the Frequency Display Section beginning in Paragraph 4-27.

Page 4-24. Delete the discussion of the Frequency Counter and Display beginning in Paragraph 4-167.

Figure 4-1, Simplified Block Diagram, should appear as on Page 8-17/8-18.

UPDATE OF FCM

The 3580A has had two different types of frequency control modules or FCM. Original 3580A's had a mechanical FCM with gear driven parts while the new FCM is a digital design using a potentiometer.

Both the A2 and A16 boards have undergone revisions to account for the change from the mechanical to digital FCM. The major physical effect of the change in FCM has been the need for additional signal lines from the A2 to the A16 board. The older mechanical FCM required that two signals be brought by wires from the A2 board to the A16 board. This was accomplished by routing two wires from crimp posts on the A2 to similar posts on the A16.

When the instrument was revised for the new digital FCM, the lines required from the A2 to A16 board increased from two to four. On early versions of the digital FCM the four wires were routed to A16 from pins in different locations on the A2 board. In addition to units manufactured in this manner some old 3580A's have been retrofitted per Service Note 3580A-9.

The current A2 board (-hp-Part No. 03581-66512) has a five pin connector J1 which contains the wires which are routed to a mating connector on the cable to the A16 board.

Since there are several combinations of A2 and A16 boards which may exist the following outlines the procedure for ordering and installing either of the boards.

PROCEDURE:

The replacement of either the A2 or A16 board in the 3580A is affected by the current A2 board in the instrument. Identifying the A2 board currently in the instrument leads to the proper ordering and installation procedure.

Remove the A2 board from the instrument. Pay specific attention to the type and number of wired connections from the A2 to the A16 board.

If the A2 board in the instrument has two wires connecting it to the A16 board then go to Procedure A.

If the A2 board in the instrument has four wires connecting it to the A16 board and the board part number is 03581-66502, then then go to Procedure B.

If the A2 board has part number 03581-66512, then go to Procedure C.

PROCEDURE A:

In this case the instrument still has the mechanical FCM and has not been retrofitted. Follow the instructions given below depending upon the board needed:

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Replacing the A16 Board:

The A16 board for the old FCM is a direct replacement part but it has the old part number. Order - hp-Part No. 0358066516 and install directly in place of the A16 board currently in the instrument.

Replacing the A2 Board:

The new A1 board is compatable with the old FCM but is not a direct replacement. Since the new A1 board has the five pin connector, a cable must be ordered for connecting the A1 to the A16 board.

Order -hp- Part No. 03580-66512, A2 Board 03581-61613, Cable Assembly

- Remove both the A2 and A16 board from the instrument.
- 2. Desolder the red and grey wires from the A2 to A16 board and note their location.
- 3. Solder the red and grey wires to the corresponding lugs on the A16 board and reinstall it.
- 4. Remove crystal A2Y1 and resistor A2R65 from the old A2 board and reinstall in the new A2 board. These are a matched pair which must be reinstalled in the new A2.
- 5. Attach the connector from A16 to A2 and reinstall A2 in the instrument. Verify calibration of the instrument.

PROCEDURE B:

This procedure is used for 3580A's which have the new digital FCM either installed at the factory or as a field retrofit. The A2 board has lug jumper connectors. Follow the instructions below for the board being replaced.

Replacing The A16 Board:

When replacing the A16 board, the new A16 will have a connector soldered in place of the four loose wires on the old A16. The procedure to replace the A16 includes removal of this connector from the new A16.

Order -hp- Part No. 03580-66536, A16

- Remove the old A16 board from the 3580A being careful to note the position of the wires on the old board.
- Unsolder the five pin connector cables from the new A16 board. This will not be needed for installation.
- 3. Resolder the wires into position on the new A16 as shown in figure xx.

4. Re-install the new A16 board and calibrate the 3580A.

Replacing The A2 Board:

When replacing the A2 board there is a requirement for the cable assembly which connects the A2 board to the A16 board. The installation will include removal of the A16 board for installation of the cable assembly.

Order -hp- Part No. 03580-66512, A2 Order -hp- Part No. 03580-61615, Cable

- 1. Remove the A16 board from the instrument noting the location of all wires.
- 2. Solder the cable supplied to the A16 as shown in figure xx and reconnect the other wires to the A16.
- 3. Install the A16 board back in the instrument.
- 4. Remove crystal A2Y1 and resistor A2R65 from the old A2 board. These are a matched pair and must be reinstalled in the new A2 board.
- 5. Insert the new A2 board in the 3580A and connect the cable from A16 to the connector in the upper right hand corner of A2.
- 6. Re-assemble and calibrate the 3580A.

PROCEDURE C:

The A2 board in this case has the five pin connector J1 in the upper right hand corner for the new digital FCM. The procedure for replacing either the A16 or the A2 is as follows:

If the A16 board needs to be replaced then order Part No. 03580-66536.

If the A2 board needs to be replaced then order Part No. 03580-66512

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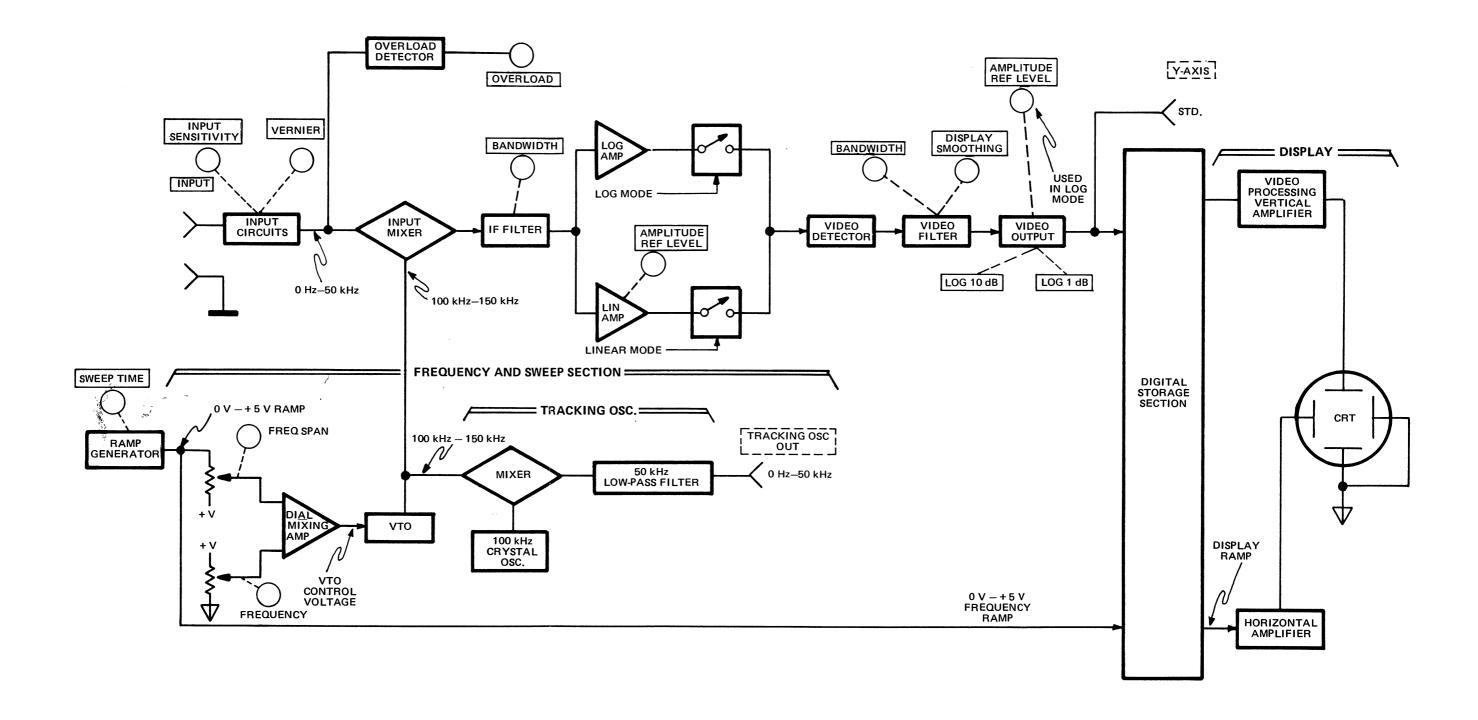


Figure 4-1. Simplified Block Diagram. 8-17/8-18

Pages 5-1 through 5-15. The following paragraphs in the Performance Tests should appear as follows:

5-9. Frequency Tests.

5-10. These tests verify part of the Frequency Characteristic Specifications listed in Table 1-1. If, for any reason, the instrument will not pass these tests, perform the Sweep Alignment and Dial Calibration (Paragraph 5-63) of the Adjustment Procedures.

5-11. Range and Frequency Dial Accuracy Test.

a. Position the following front panel controls:

| ADAPTIVE SWEEPOFF |
|--|
| DISPLAY All pushbuttons released |
| AMPLITUDE MODE LOG 10 dBv/DIV |
| AMPLITUDE REF LEVELNORMAL |
| $dBv/LIN - dBm 600 \Omega \dots dBv/LIN$ |
| INPUT SENSITIVITYCAL |
| VERNIER (Amplitude)CAL |
| (Fully CW) |
| FREQUENCY 00.0 kHz |
| START - CTRSTART |
| RESOLUTION BANDWIDTH 30 Hz |
| DISPLAY SMOOTHINGMIN |
| FREQ. SPAN/DIV 0 Hz |
| SWEEP TIME/DIV 0.2 SEC |
| SWEEP MODE REP |
| |

Option 002: Set dBm 900 Ω/LIN - dBm 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL.

b. Adjust the front panel ZERO CAL control for a maximum display indication. Readjust the FREQUENCY control for 10, 20, 30, 40 and 50 kHz. A peak response should occur for each of these settings (± .1 kHz) to verify the Range Specifications and Frequency Dial Accuracy Specifications (20°C to 30°C) given in Table 1-1.

NOTE

As the frequency of the peak response is approached, pull out the knob for easier tuning.

5-12. Display Accuracy Tests.

a. Reposition the following front panel controls.

| FREQUENCY | 00.0 kHz |
|----------------------|----------|
| RESOLUTION BANDWIDTH | 300 Hz |
| FREQ. SPAN/DIV | 5 KHz |

b. The 10 kHz CAL signal and its harmonics should be repetitively swept and appear on the display as shown by Figure 5-1. The separation between the Zero Response and 50 kHz harmonic should be 10 major divisions \pm 1 minor division. The separation between any two adjacent responses should be 2 major divisions \pm .2 minor divisions.

Momentarily push and release DISPLAY - STORE, watching the display to verify that the STORE and NON-STORE traces appear in the same position.

5-13. Sweep Tests.

5-14. These tests verify the Sweep Characteristics Specifications given in Table 1-1. If the instrument fails the Frequency Span Tests (Paragraph 5-15), perform the Sweep Alignment and Dial Calibration (Paragraph 5-63) of the Adjustment Procedures. If it fails only the Log Sweep Test (Paragraph 5-16), perform only the Log Sweep Adjustments (Paragraph 5-67) of the Sweep Alignment and Dial Calibration. All sweep time calibration is done with a factory selected resistor. If the instrument will not pass the Sweep Time Tests (Paragraph 5-17), refer to Section VII for additional information.

Equipment Required:

Electronic Counter (-hp- Model 5326A)

5-15. Frequency Span Test.

a. Position the following front panel controls: (Only those controls printed in **BOLD** require a change from the previous test.)

| ADAPTIVE SWEEPOFF |
|--|
| DISPLAY All pushbuttons released |
| AMPLITUDE MODE LOG 10 dBv/DIV |
| AMPLITUDE REF LEVELNORMAL |
| $dBv/LIN - dBm 600 \Omega \dots dBv/LIN$ |
| INPUT SENSITIVITY 0 dBV |
| VERNIER (Amplitude)CAL |
| (Fully CW) |
| FREQUENCY 00.0 kHz |
| START - CTRSTART |
| RESOLUTION BANDWIDTH 300 Hz |
| DISPLAY SMOOTHINGMIN |
| FREQ. SPAN/DIV 5 Hz |
| SWEEP TIME/DIV 0.2 SEC |
| SWEEP MODE |
| |

Option 002: Set dBm 900 Ω/LIN - dBm 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL.

b. Adjust MANUAL VERNIER full CCW.

c. Set the electronic counter to the frequency mode and adjust the time base/multiplier for a measurement of 1 MHz with 6 digits of resolution (1000.00 kHz). Adjust for maximum input sensitivity and either a zero trigger level or Preset. For the -hp- 5326A Counter, the controls should be set to:

Sample Rate:

Fast

Function:

Freq. A 10⁶

Multiplier: Channel A

Slope +

AC

Atten: X1

Level: Preset

BNC Input:

Sep.

- d. Connect the counter Channel A input to the L.O. OUTPUT terminal on the back panel of the 3580A.
- e. Adjust the ZERO CAL for approximately a 1000.00 kHz reading on the counter. Adjust the FRE-QUENCY dial (pulled out for fine tuning) for a 1000.00 kHz indication on the counter.
- f. Adjust MANUAL VERNIER full CW. The counter indication should be 1000.50 kHz ± .01 kHz.
- g. Readjust MANUAL VERNIER full CCW. Reposition FREQ. SPAN/DIV 10 Hz.
- h. Readjust the FREQUENCY dial (pulled out for fine tuning) for a 1000.00 kHz indication on the counter.
- i. Adjust MANUAL VERNIER full CW. The counter indication should be $1001.00 \text{ kHz} \pm .02 \text{ kHz}$.
- j. Continue this procedure for the remaining FREQ. SPAN/DIV settings. Refer to Table 5-2 for the proper tolerances.

Table 5-2. Frequency Span Test.

| | COUNTE | R READING |
|----------------|----------------------------|---------------------------|
| FREQ. SPAN/DIV | MANUAL VERNIER FULL CCW | MANUAL VERNIER FULL CW |
| 5 Hz | 1000.00 kHz | 1000.50 kHz ± .01 kHz |
| 10 Hz | 1000.00 kHz | 1001.00 kHz ± .02 kHz |
| 20 Hz | 1000.00 kHz | 1002.00 kHz ± .04 kH: |
| 50 Hz | 1000.00 kHz | 1005.00 kHz ± .10 kHz |
| .1 kHz | 1000.00 kHz | 1010.00 kHz ± .20 kHz |
| .2 kHz | 1000.00 kHz | 1020.00 kHz ± .40 kHz |
| .5 kHz | 1000.00 kHz | 1050.00 kHz ± 1.00 kH |
| 1 kHz | 1000.00 kHz | 1100.00 kHz ± 2.00 kH |
| 2 kHz | 1000.00 kHz | 1200.00 kHz ± 4.00 kH; |
| 5 kHz | | |
| (checked in | | |
| Para 5-16) | | 1 |

5-16. Log Sweep Test.

a. Reposition the controls as follows:

| INPUT SENSITIVITY | CAL |
|----------------------|--------|
| RESOLUTION BANDWIDTH | 30 Hz |
| SWEEP MODE | C ZERO |

- b. Momentarily press DISPLAY CLEAR WRITE.
- c. Adjust the ZERO CAL control for a maximum indication on the leftmost display graticule.

d. Reposition the controls as follows:

| RESOLUTION BANDWIDTH | 300 Hz |
|----------------------|--------|
| SWEEP MODE | LOG |

Allow time for three complete sweeps.

e. Verify that the 20 kHz harmonic of the internal CAL signal falls on the proper graticule (± 1 minor division). If the instrument will not pass this test, but passes all previous tests, perform only the Log Sweep Adjustments (Paragraph 5-67) of the Adjustment Procedures.

5-20. Bandwidth Switching Accuracy Tests.

a. Position the following front panel controls: (Only those controls printed in **BOLD** require a change from the previous tests.)

| ADAPTIVE SWEEP | OFF |
|----------------------------------|------------|
| DISPLAY All pushbutte | |
| AMPLITUDE MODE LOG | 1 dBv/DIV |
| AMPLITUDE REF LEVEL | |
| $dBv/LIN - dBm 600 \Omega \dots$ | dBv/LIN |
| INPUT SENSITIVITY | 0 dBV |
| VERNIER (Amplitude) | CAL |
| | (Fully CW) |
| FREQUENCY | |
| START - CTR | CTR |
| RESOLUTION BANDWIDTH | |
| DISPLAY SMOOTHING | |
| FREQUENCY SPAN/DIV | 5 Hz |
| SWEEP TIME/DIV | . 0.1 SEC |
| SWEEP MODE | .MANUAL |
| | |

Option 002: Set dBm 900 Ω/LIN - dBm 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL.

b. Connect a properly terminated frequency synthesizer to the 3580A INPUT and adjust the source for a 10 kHz, 0 dBV output level (0 dBm 900 Ω for instruments with Option 002).

NOTE

See Table 5-3 for the proper level to use with your source. See Figure 5-2 for the proper hookup with an -hp- 3320B Frequency Synthesizer.

- c. By alternately pressing and releasing DISPLAY-CLEAR WRITE while adjusting MANUAL VERNIER, center the display indication (a narrow spike). Adjust ZERO CAL for a peak display of this spike.
- d. Adjust the front panel ZERO CAL for a peak indication.

Model 3580A Backdating

5-24. Amplitude Reference Level Tests (Log Mode).

NOTE

If the instrument fails this test, see Section VII for troubleshooting information. There are no adjustments for this specification.

a. Reposition the following front panel controls:

AMPLITUDE MODELOG 10 dB/DIV AMPLITUDE REF LEVELNORMAL INPUT SENSITIVITY – 10 dBV

- b. Connect the digital multimeter (DC mode, 100 volt range) to the Y AXIS output of the 3580A.
- c. Adjust the signal source for a -70 dB V output (-70 dBm 900 Ω for Option 002). (See Table 5-3 for proper levels.) Adjust the MANUAL VERNIER and ZERO CAL for a peak display. Adjust VERNIER (Amplitude) for a 1.50 volt \pm .01 volt reading on the multimeter.
- d. Adjust the AMPLITUDE REF LEVEL switch to the settings given in Table 5-7. Check for the proper multimeter reading.

NOTE

MANUAL VERNIER may have to be readjusted to insure a peak display indication.

Table 5-7. Amplitude Ref. Level Tests (Log Mode).

| INPUT LEV STANDARD INSTRUMENT | EL (10 KHz) OPTION 002 900 Ω | AMPLITUDE REF. LEVEL | MULTIMETER READING |
|-------------------------------------|------------------------------------|-------------------------|-----------------------|
| - 70 dBv | - 70 dBm | - 10 dB | 2.00 V ± .02 V |
| - 70 dBv | - 70 dBm | - 20 dB | 2.50 V ± .02 V |
| - 70 dBv | - 70 dBm | - 30 dB | 3.00 V ± .03 V |
| - 70 dBv | - 70 dBm | - 40 dB | 3.50 V ± .03 V |
| - 70 dBv | - 70 dBm | - 50 dB | 4.00 V ± .04 V |
| - 70 dBv | - 70 dBm | - 60 dB | 4.50 V ± .04 V |
| - 70 dBv | - 70 dBm | - 70 dB | 5.00 V ± .05 V |

e. Disconnect the multimeter from the 3580A.

5-26. Frequency Response Tests.

a. Reposition the following front panel controls:

| AMPLITUDE MODE LOG 10 dE | 3/DIV |
|-----------------------------|-------|
| AMPLITUDE REF LEVEL | 30 dB |
| INPUT SENSITIVITY | |
| (according to white marker) | 0 dB |
| RESOLUTION RANDWIDTH | 3 Hz |

- b. Adjust the signal source for a 10 kHz 0 dBv output (0 dBm 900 Ω for Option 002). (See Table 5-3).
- c. By alternately pressing and releasing DISPLAY-CLEAR WRITE while adjusting MANUAL VERNIER, center the display indication (a narrow spike). Adjust ZERO CAL for a peak display of this spike.

d. Reposition the following front panel control:

AMPLITUDE MODE LOG 1 dB/DIV

- e. Readjust MANUAL VERNIER for a peak display.
- f. Adjust VERNIER (Amplitude) for a -1 dB display (1 dB/div).
- g. Adjust the signal source to the frequencies given in Table 5-10 (refer to Table 5-3) for an INPUT SEN-SITIVITY of 0 dB. At each frequency, adjust the FRE-QUENCY dial to that of the source. Then, slowly adjust the ZERO CAL for a peak display indication. Momentarily press DISPLAY-CLEAR WRITE. Check for proper level as given in Table 5-10. Note: The display is calibrated 1 dB per major division.
- h. Repeat Steps e through g for an INPUT SENSITI-VITY and source levels of -10 dB, -20 dB, -30 dB and -40 dB (according to white marker and with a -30 dB AMPLITUDE REF LEVEL). Consult Table 5-3 and Table 5-10 for the proper input level and frequencies to use. At the start of each new INPUT SENSITIVITY, always recalibrate the instrument at 10 KHz with CAL 10 KHz.

5-27. Internal Calibrator Test.

VEDNIED (A. 11, 1.)

a. Reposition the following front panel controls:

| VERNIER (Amplitude) | CAL |
|----------------------|----------|
| AMPLITUDE REF LEVEL | NORMAL |
| INPUT SENSITIVITY | 20 dB |
| FREQUENCY | 00.0 kHz |
| START - CTR | START |
| RESOLUTION BANDWIDTH | 300 Hz |
| FREQ. SPAN/DIV | . 5 kHz |
| SWEEP TIME/DIV | 0.2 SEC |
| SWEEP MODE | REP |
| | |

- b. Adjust the signal source for a 10 KHz 20 dBv (- $20~dBm~900~\Omega$ if Option 002) output. (See Table 5-3 for proper level.)
- c. Adjust the ZERO CAL for a display response on the 10 KHz graticule (2 major divisions from left graticule). (After each trial adjustment, allow 2 seconds for the next sweep before verifying the accuracy of the adjustment.)
- d. Adjust the CAL 10 KHz for a full scale 0 dB display. (After each trial adjustment, allow 2 seconds for the next sweep before verifying the accuracy of the adjustment.)
 - e. Reposition the following front panel control:

| INPUT SENSITIVITY | H | NP | 'U | T | SE | N | ١S | Γ | LI. | V | ľ | ΤY | | | | | | | | | | | | | | | \mathbf{C}_{i} | A | 1 | ĺ |
|-------------------|---|----|----|---|----|---|----|----------|-----|---|---|----|--|--|--|--|--|--|--|--|--|--|--|--|--|--|------------------|---|---|---|
|-------------------|---|----|----|---|----|---|----|----------|-----|---|---|----|--|--|--|--|--|--|--|--|--|--|--|--|--|--|------------------|---|---|---|

f. Verify that the 10 KHz harmonic of the CAL signal appears 2 major divisions from left graticule with a full scale 0 dB level (\pm .15 dB). (1 dB = 1 major division.)

5-28. Bandwidth Tests.

5-29. This test verifies the bandwidth specifications of Table 1-1. If the instrument will not pass this test, perform the IF Filter Alignment (Paragraph 5-70) of the Adjustment Procedures.

Equipment Required:

Frequency Synthesizer (-hp-Model 3320B, 50 ohms) 50 Ohm Termination (-hp-11048C)

a. Position the following front panel controls. (Only those controls printed in BOLD require a change from the previous test.)

| ADAPTIVE SWEEP OFF |
|--|
| DISPLAY All pushbuttons released |
| AMPLITUDE MODE LOG 10 dBv/DIV |
| AMPLITUDE REF LEVELNORMAL |
| $dBv/LIN - dBm 600 \Omega \dots dBv/LIN$ |
| INPUT SENSITIVITY 20 dB |
| VERNIER (Amplitude) CAL |
| (Fully CW) |
| FREQUENCY 10.0 KHz |
| START - CTRCTR |
| RESOLUTION BANDWIDTH 300 Hz |
| DISPLAY SMOOTHINGMIN |
| FREQUENCY SPAN/DIV 50 Hz |
| SWEEP TIME/DIV 0.2 SEC |
| SWEEP MODE |
| |

Option 002: Set dBm 900 Ω/LIN - dBm 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL.

- b. By alternately pressing and releasing DISPLAY-CLEAR WRITE while adjusting MANUAL VERNIER, center the display indication (a narrow spike).
- c. Connect a properly terminated frequency synthesizer to the input of the 3580A. Adjust the synthesizer for a 10 kHz 20 dBv signal (- 20 dBm 900 Ω for Option 002) output. (See Table 5-3.) Momentarily press DISPLAY CLEAR WRITE.
- d. Adjust the FREQUENCY dial (pulled out for fine tuning) for a peak display indication.
 - e. Reposition the following front panel controls:

AMPLITUDE MODE LOG 1 dB/DIV

f. Readjust the FREQUENCY dial (fine tune position) for a peak display indication of the 10 KHz input. Adjust CAL 10 KHz for a full scale 0 dB display, if not already so adjusted.

- g. Slowly rotate MANUAL VERNIER CW until the display dot has dropped 3 dB in amplitude. (Remember, the display is calibrated 1 dB/DIV). This is the upper 3 dB point of the filter.
- h. Momentarily press DISPLAY-CLEAR WRITE. Slowly increase the frequency of the source. The dot will move to a full scale display and then down to the lower 3 dB point of the filter.
- i. Note the frequency of the source at this lower 3 dB point ___ Hz. This frequency, less the original 10 KHz start frequency, is the 3 dB bandwidth of the 300 Hz filter. It should be 300 Hz ± 45 Hz.
- j. Repeat Steps f through i for the 100 Hz, 30 Hz and 10 Hz filters. See Table 5-11 for the start frequency of the source, FREQUENCY dial setting, RESOLUTION BAND-WIDTH, FREQ. SPAN/DIV, and the test limits. At the start of each new bandwidth setting, always center the display with MANUAL VERNIER, and adjust the FREQUENCY dial, and CAL 10 KHz for a full scale, peak display at the appropriate start frequency. Then make the appropriate adjustments for the upper and lower 3 dB points.
- k. Using Table 5-12 and the same technique used for the 300 Hz, 100 Hz, 30 Hz, and 10 Hz Bandwidths, test the 60 dB Bandpass of the 3 Hz and 1 Hz filters. However, use

AMPLITUDE MODE LOG 10 dB/DIV

and measure the frequency difference between the 60 dB points. As before, always adjust the FREQUENCY dial and CAL 10 KHz for a peaked full scale display before attempting to measure the 60 dB bandwidths. If the display is noisy at the 60 dB points, use Display Smoothing......Max. Note: The display is now calibrated 10 dB/DIV.

5-32. Noise Level Tests.

- a. Connect the 1 $k\Omega$ resistor across the INPUT terminals of the 3580A. Disconnect all signal sources.
- b. Position the following front panel controls: (Only those controls printed in BOLD require a change from the previous tests).

| ADAPTIVE SWEEP OFF |
|--|
| DISPLAY All pushbuttons released |
| AMPLITUDE MODE LOG 10 dBv/DIV |
| AMPLITUDE REF LEVELNORMAL |
| $dBv/LIN - dBm 600 \Omega \dots dBv/LIN$ |
| INPUT SENSITIVITY 70 dB |
| VERNIER (Amplitude) CAL |
| (Fully CW) |
| FREQUENCY 00.0 kHz |
| START - CTRSTART |
| RESOLUTION BANDWIDTH 300 Hz |

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| DISPLAY SMOOTHING MAX FREQ. SPAN/DIV 5 KHz SWEEP TIME/DIV 5 SEC | Momentarily press the following control: DISPLAY | |
|---|--|--|
| SWEEP MODEMANUAL | o. The display indication should be less than - 150 dB | |
| Option 002: Set dBm 900 $\Omega/$ LIN - dBm 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL. | (8 major divisions down from top graticule). p. Readjust MANUAL VERNIER for a display indication at 100 Hz (1 major division from leftmost graticule). Momentarily press the following control: | |
| c. Adjust the MANUAL VERNIER full CCW. Adjust ZERO CAL for a peak display indication. | DISPLAY | |
| d. Adjust the MANUAL VERNIER for a display indication at 10 KHz (2 major divisions from left graticule). Momentarily press the following control: | q. The indication should be less than -143 dB (7.3 major divisions down from top graticule). | |
| DISPLAY CLEAR WRITE | r. Reposition the following controls: | |
| e. The display indication should always be less than -130 dB (6 major divisions down from top graticule, since | DISPLAY SMOOTHING MIN FREQ. SPAN/DIV 5 Hz | |
| Full Scale = -70 dB).f. Reposition the following front panel control: | s. Adjust MANUAL VERNIER full CCW. Momentarily press the following front panel control: | |
| RESOLUTION BANDWIDTH 30Hz | DISPLAY CLEAR WRITE | |
| g. Momentarily press the following control: | t. Adjust ZERO CAL for a peak response at the leftmost graticule. Reposition the following front panel control: | |
| DISPLAYCLEAR WRITE | DISPLAY SMOOTHINGMAX | |
| The display indication should be less than -140 dB (7 major divisions down from top graticule). | u. Adjust the MANUAL VERNIER for a display indica- | |
| h. Reposition the following control: | tion at 10 Hz (2 major divisions from leftmost graticule). Momentarily press the following control: | |
| FREQ. SPAN/DIV 0.1 KHz | DISPLAY | |
| i. Adjust MANUAL VERNIER full CCW. Adjust ZERO CAL for a peak display indication. | v. The display indication should be less than - 135 dB (6.5 major divisions down from top graticule). Remove the | |
| j. Adjust MANUAL VERNIER for a display indication at 100 Hz (1 major division from leftmost graticule). Momentarily press the following control: | 1 kΩ resistor from the input terminals.5-33. Noise Sideband Test. | |
| DISPLAYCLEAR WRITE | a. Reposition the following controls: INPUT SENSITIVITY | |
| k. The display indication should be less than - 132 dB (6.2 major divisions down from top graticule). | FREQUENCY 10.0 kHz START - CTR CTR DISPLAY SMOOTHING MIN | |
| 1. Adjust MANUAL VERNIER for a display indication of 1 KHz (far right graticule). Momentarily press the following control: | FREQ. SPAN/DIV | |
| DISPLAY | b. By alternately pressing and releasing DISPLAY - CLEAR WRITE while adjusting MANUAL VERNIER, center the display indication (a narrow spike). | |
| m. The display indication should be less than - 140 dB (7 major divisions down from top graticule). | c. Adjust the FREQUENCY dial (pulled out for fine tuning) for a peak display of this spike. | |
| n. Reposition the following control: | d. Reposition the following controls: | |
| RESOLUTION BANDWIDTH 1 Hz | SWEEP MODE SING | |

Backdating Model 3580A

e. After waiting for the sweep to be completed (100 sec.), verify that the noise on the display ± 10 Hz (± 2 major divisions) away from the 10 KHz CAL signal (in center of display) is at least 70 dB below the CAL signal.

5-34. Spurious Response Test.

NOTE

The removable of A16C31 and A16C32 eliminates the need to perform the Spurious Response test. For serial numbers 2030A05375 and below see Change No. D19.

a. Reposition the following controls:

| INPUT SENSITIVITY | 20 dB |
|----------------------|---------|
| FREQUENCY | 00.0 Hz |
| START-CTR | START |
| RESOLUTION BANDWIDTH | 30 Hz |
| FREQ. SPAN/DIV | 2 KHz |
| SWEEP TIME/DIV | |
| SWEEP MODE | |

b. Momentarily press:

DISPLAYCLEAR WRITE

- c. Adjust ZERO CAL for a peak display on the leftmost display graticule.

and momentarily press:

DISPLAYCLEAR WRITE

- e. Connect the frequency synthesizer (use proper output impedance needed for the bandpass filter) to the input of the bandpass filter. Adjust the filter for a 5 kHz center frequency and adjust the synthesizer for a 5 kHz output. (For a 50 ohm source and the White 2640 filter, connect a 550 Ω resistor (± 10%) in series between the filter and synthesizer. This gives the 600 Ω source impedance required by the White filter (See Figure 5-3).
- f. Connect the output of the filter to the input of the 3580A. Always terminate properly if required. (The White Model 2640 filter requires no output termination. See Figure 5-3).
- g. Adjust MANUAL VERNIER for a display indication at 5 kHz (2 1/2 major divisions from left graticule). Adjust the source level for a 20 dBv (full scale) input to the 3580A (For the White 2640 filter and a 50 Ω source, this corresponds to -16.99 dBm 50 Ω level on the source). Readjust MANUAL VERNIER for a peak display. Adjust CAL 10 KHz for a full scale display.

 - i. After waiting for one complete sweep (50 sec.) verify

that all responses other than the zero response are at least 80 dB below the 5 kHz response.

5-35. Line Related Spurious Test.

Specification: > 80 dB below input reference level or $-135 \text{ dBV } (0.18 \mu\text{V})$

- a. Disconnect the Synthesizer and Bandpass Filter from the 3580A Input. Turn off all unnecessary equipment located near the 3580A. This especially includes large current users such as soldering irons, blowers, moters, etc.
- b. Using a short piece of wire, connect a short across the 3580A INPUT terminals.
- c. Reposition the following controls:

| 70 dB | INPUT SENSITIVITY | |
|--------------|----------------------|--|
| 3 Hz | RESOLUTION BANDWIDTH | |
| 5 Hz | FREQ. SPAN/DIV | |
| MAN | SWEEP MODE | |
| centered | MANUAL VERNIER | |
| MAX | DISPLAY SMOOTHING | |
| CTR | START-CTR | |
| MA center | SWEEP MODE | |

d. Connect the LO OUTPUT (rear panel) to the input of an Electronic Counter (-hp-Model 5326A).

NOTE

If the power-line frequency is 50 Hz, substitute the following Counter readings for Steps e and f.

Step e and f: 1000.48 kHz to 1000.52 kHz Step e and f: 1000.98 kHz to 1001.02 kHz Step e and f: 1001.48 kHz to 1001.52 kHz

- e. With the FREQUENCY control pulled out for fine tuning, tune the 3580A frequency for a Counter reading between 1000.58 kHz and 1000.62 kHz.
- f. Press CLEAR WRITE, then slowly turn the MANUAL VERNIER to obtain a peak reading. The peak should be more than 70 dB below full scale (-135 dBV).
- g. Repeat Steps e and f substituting 1001.18 kHz to 1001.22 kHz, and 1001.78 kHz to 1001.82 kHz for the Counter readings.

NOTE

If the instrument fails this test double check that the input short is as small as possible; that all power line current is kept at a minimum, and that all covers are tightly secured on the 3580A.

5-36. IF Feedthru and Zero Beat Response Tests.

5-37. These tests verify the ability of the instrument to reject a 100 kHz signal at the input and also how well the

Zero Beat Response is suppressed. Proceed to the Mixer Balance Adjustments (Paragraph 5-81) of the Adjustment Procedures if the Zero Beat Response is too large. Proceed to Section VII for troubleshooting information if there is too much IF Feedthru.

Equipment Required:

Frequency Synthesizer (-hp- Model 3320B, 50 ohm)

- a. Reconnect the synthesizer to the 3580A. Do not terminate. Adjust the source for a 10 volt 100 kHz output (+ 26.99 dBm 50 ohms setting on 3320B and unterminated).
- b. Position the following front panel controls: (Only those controls printed in BOLD require a change from the previous test).

| ADAPTIVE SWEEP OFF DISPLAY All pushbuttons released AMPLITUDE MODE LOG 10 dBv/DIV AMPLITUDE REF LEVEL NORMAL dBv/LIN - dBm $600~\Omega$ dBv/LIN |
|---|
| INPUT SENSITIVITY + 20 dB |
| VERNIER (Amplitude) CAL (Fully CW) |
| FREQUENCY |
| FREQ. SPAN/DIV 20 Hz SWEEP TIME/DIV 5 SEC SWEEP MODE MANUAL |
| Option 002: Set dBm 900 Ω/LIN - dBm 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL. |

- c. Adjust MANUAL VERNIER for a response in the center of the screen. The display indication should be at least 70 dB below full scale to verify the IF Feedthru specification of Table 1-1. If the instrument fails this test, see Section VII for troubleshooting information.
- d. Disconnect the synthesizer. Reposition the following front panel controls:

| RESOLUTION BANDWIDTH | 300 Hz |
|----------------------|--------|
| FREQ. SPAN/DIV | 5 KHz |
| SWEEP MODE | RESET |

e. Momentarily press the following front panel control:

DISPLAY CLEAR WRITE

f. Adjust ZERO CAL for a maximum display indication on the left graticule. This display should be at least 30 dB (3 major divisions) below full scale to verify the Zero Beat Response specification of Table 1-1. If the instrument fails this test, go to the Mixer Balance Adjustments (Paragraph 5-81) of the Adjustment Procedures.

5-38. Input Impedance Tests.

5-39. These tests verify the Input Impedance characteristics of Table 1-2. Since there is no adjustment for this parameter, see Section VII for troubleshooting information if the instrument fails this test.

Equipment required:

 $1 \text{ M}\Omega \pm 1\%$ film resistor (-hp- Part No. 0757-0344)

a. Position the following front panel controls. (Only those controls printed in **BOLD** require a change from the previous tests.)

| DISPLAY All pushbuttons released |
|--|
| |
| AMPLITUDE MODELOG 10 dBv/DIV |
| AMPLITUDE REF LEVELNORMAL |
| $dBv/LIN - dBm 600 \Omega \dots dBv/LIN$ |
| INPUT SENSITIVITY 0 dB |
| VERNIER (Amplitude CAL |
| (Fully CW) |
| FREQUENCY 00.0 kHz |
| START - CTRSTART |
| DISPLAY SMOOTHINGMIN |
| RESOLUTION BANDWIDTH 10 Hz |
| FREQ. SPAN/DIV 1 KHz |
| SWEEP TIME/DIV 5 SEC |
| SWEEP MODERESET |

Option 002: Set dBm 900 Ω/LIN - dBm 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL.

- b. Adjust ZERO CAL for a peak display on the left graticule.
- c. Reposition the following front panel controls:

| AMPLITUDE MODE | LOG 1 dB/DIV |
|----------------|--------------|
| SWEEP MODE | |

d. Connect the rear panel TRACKING OSC OUT to the front INPUT terminals of the 3580A. Adjust the rear panel TRACKING OSC LEVEL control fully CW. Adjust MANUAL VERNIER for a 1 kHz display indication (1 major division from left graticule). Readjust the TRACKING OSC LEVEL control for a full scale 0 dB display. Momentarily press the following control:

DISPLAY CLEAR WRITE

e. Connect the 1 M Ω resistor in series between the TRACKING OSC OUT and front panel INPUT terminals. The display indication should drop 6 dB \pm .3 dB (6 major divisions \pm .3 major divisions) to verify an input impedance of 1 M Ω .

| f. R | eposition | the | following | front | panel | conti | rol | l: |
|------|-----------|-----|-----------|-------|-------|-------|-----|----|
|------|-----------|-----|-----------|-------|-------|-------|-----|----|

INPUT SENSITIVITY - 10 dB

g. Readjust the rear panel TRACKING OSC LEVEL control for a full scale display. Adjust MANUAL VERNIER for a display indication at 10 kHz (far right display graticule). DO NOT REMOVE 1 M Ω RESISTOR. Momentarily press the following front panel control:

- h. 1) Std. 3580A: The amplitude should drop 3 dB
 ± 1 dB, verifying that the input shunt capacitance is 30 pF, nominal.
 - 2) Option 002: The amplitude should drop 4 dB ± 1 dB, verifying that the input shunt capacitance is 40 pF, nominal.
- i. Disconnect the cable connected between the TRACK-ING OSC OUT and the front panel INPUT terminals.

5-40. Output Tests.

5-41. These tests verify the Output specifications of the 3580A listed in Table 1-1.

Equipment Required:

Electronic Counter (-hp- Model 5326A)
Digital Multimeter (-hp- Model 34740/34702)
Distortion Analyzer (-hp- Model 333A)

5-42. TRACKING OSC OUTPUT Tests.

a. Position the following front panel controls. (Only those controls printed in BOLD require a change from the previous tests).

| ADAPTIVE SWEEP | |
|----------------------------------|-----------------|
| DISPLAY All pushbut | tons released |
| AMPLITUDE MODELOG | 10 dBv/DIV |
| AMPLITUDE REF LEVEL | |
| $dBv/LIN - dBm 600 \Omega \dots$ | dBv/LIN |
| INPUT SENSITIVITY | $\dots + 20 dB$ |
| VERNIER (Amplitude) | CAL |
| | (Fully CW) |
| FREQUENCY | 00.0 kHz |
| START - CTR | START |
| RESOLUTION BANDWIDTH | 10 Hz |
| DISPLAY SMOOTHING | MIN |
| FREQ. SPAN/DIV | 5 KHz |
| SWEEP TIME/DIV | 5 SEC |
| SWEEP MODE | RESET |
| | |

Option 002: Set dBm 900 Ω /LIN - dBm 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL.

- b. Momentarily press DISPLAY CLEAR WRITE. Adjust the ZERO CAL for a peak display (on leftmost display graticule).
- c. Connect the multimeter (AC mode 100 volt range) and a 600 Ω resistor termination to the rear panel TRACKING OSC OUT. Adjust the FREQUENCY dial for 50Hz (300 Hz for Option 002). Adjust the rear panel TRACKING OSC LEVEL control forj a 1.00 volt reading on the multimeter.*
- d. Adjust the FREQUENCY control to 50.0 kHz (20.0 kHz for Option 002 instruments). Verify that the multimeter reads 1.00 volts \pm .06 volts.
 - e. Reposition the following front panel controls:

| AMPLITUDE MODE | LIN |
|----------------------|---------|
| INPUT SENSITIVITY | 2 V |
| FREQUENCY | 00.0 Hz |
| RESOLUTION BANDWIDTH | 30 Hz |
| SWEEP MODE | MANUAL |

f. Connect the rear panel TRACKING OSC OUT to the front panel INPUT terminals. Momentarily press the following control:

- g. By alternately pressing and releasing DISPLAY-CLEAR WRITE while adjusting MANUAL VERNIER, center the display indication (a narrow spike).
- h. Adjust the rear panel TRACKING OSC LEVEL control for a full scale 2 V display. Reposition the following front panel control:

RESOLUTION BANDWIDTH 3 Hz

- i. The display indication should drop no lower than 1 V (5 major divisions) to verify the frequency accuracy of the tracking oscillator. If the tracking oscillator frequency is out of tolerance, remove the top cover and adjust A2C4 for a peak display indication.
 - j. Reposition the following front panel control:

FREQ. SPAN/DIV 0.1 KHz

k. Adjust MANUAL VERNIER for a 1 KHz display indication (indication on far right display graticule). Momentarily press the following front panel control:

1. Connect the TRACKING OSC OUT to the INPUT of

^{*}For measurements below 50 Hz, use a low frequency Digital Voltmeter such as the -hp- Model 3480/3484 with true rms.

the distortion analyzer. Adjust the TRACKING OSC LEVEL control fully CW.

m. Reference the TRACKING OSC OUT to 0 dB on the distortion analyzer. (For the -hp- 333A Distortion Analyzer, position the following controls:

| FUNCTION | SET LEVEL |
|------------------|------------|
| METER RANGE | 0 dB |
| FREQUENCY RANGE | X100 |
| FREQUENCY | 10 (1 kHz) |
| HIGH PASS FILTER | |

Adjust the SENSITIVITY and VERNIER controls of the distortion analyzer for a 0 dB meter indication. Set the distortion analyzer FUNCTION switch to DISTORTION.)

- n. Measure the distortion in dB by nulling the distortion analyzer.
- o. Adjust the FREQUENCY and BALANCE controls for a meter null. Use automatic nulling if available.
- p. The total distortion indication should be at least 40 dB below the reference level. If it is not, perform the Mixer Balance Adjustments (Paragraph 5-81). Disconnect the distortion analyzer from the 3580A.

5-46. Common Mode Rejection Test.

a. Position the following front panel controls:

| ADAPTIVE SWEEP OFF |
|---|
| DISPLAY All pushbuttons released |
| AMPLITUDE MODE LOG 10 dBv/DIV |
| AMPLITUDE REF LEVELNORMAL |
| $dBm 900 \Omega/LIN - dBm 600 \Omega \dots$ |
| - dBm 900 Ω/LIN |
| INPUT SENSITIVITY 0 dB |
| VERNIER (Amplitude) |
| (Fully CW) |
| INPUT MODE BRDG |
| FREQUENCY 00.0 kHz |
| START-CTRSTART |
| RESOLUTION BANDWIDTH 3 Hz |
| DISPLAY SMOOTHING MIN |
| FREQ. SPAN/DIV 10 Hz |
| SWEEP TIME/DIV 5 SEC |
| SWEEP MODE |
| |

b. Disconnect all inputs to the 3580A. Momentarily press the following front panel control:

c. Adjust the ZERO CAL control for a peak display at the leftmost graticule of the CRT. Reposition the following front panel control:

- d. Adjust the frequency synthesizer for a 60 Hz, +5 dBm 900 Ω output (+ 17.55 dBm/50 ohms). Connect the synthesizer (properly terminated) to the INPUT of the 3580A.
- e. Slowly adjust MANUAL VERNIER to the 60 Hz signal which will appear as a peak on the sixth major division from the left. Momentarily press the following front panel control:

- f. Adjust the VERNIER (Amplitude) for a full scale 0 dB display.
- g. Disconnect the synthesizer from the 3580A and connect two 453 ohm resistors in series between the INPUT terminals. (See Figure 5-4)
- h. Connect the synthesizer to the junction of the two resistors and to the chassis on the rear panel as shown in Figure 5-4. (Do not change the synthesizer amplitude setting.)
- i. The display indication on the 3580A should be at least 70 dB below full scale (10 dB/DIV).

5-47. Frequency Response Test.

- a. Disconnect the resistors from the 3580A INPUT terminals and reconnect the synthesizer (properly terminated in 50 ohms). Adjust the source for a 0 dBm 900 Ω (+ 12.55 dBm 50 Ω) 10 kHz signal.
 - b. Reposition the following front panel controls:

| FREQUENCY | 10.0 kHz |
|---------------------|------------|
| START-CTR | CTR |
| VERNIER (Amplitude) | . Fully CW |

- c. By alternately pressing and releasing DISPLAY-CLEAR WRITE while adjusting MANUAL VERNIER, center the display indication (a narrow spike). Adjust the FREQUENCY dial (pulled out for fine tuning) for a peak display of the 10 kHz input signal.
 - d. Reposition the following front panel control:

AMPLITUDE MODE LOG 1 dB/DIV

- e. Readjust the FREQUENCY dial for a peak display indication. Adjust VERNIER (Amplitude) for a full scale 1 dB display indication (1 major division down from full scale).
- f. Adjust the frequency synthesizer and 3580A FRE-QUENCY dial to the frequencies given by Table 5-13. Always peak the display indication with the FREQUENCY dial and check for proper amplitude accuracy.

Pages 5-16 through 5-28. The following items in the Adjustment Procedures should appear as follows:

5-48. ADJUSTMENT PROCEDURE.

5-49. This portion of Section V contains complete Adjustment Procedures for the Model 3580A Spectrum Analyzer:

POWER SUPPLY CHECKS AND ADJUST-MENTS (Paragraph 5-53).

DISPLAY ADJUSTMENTS (Paragraph 5-68).

SWEEP ALIGNMENT AND DIAL CALIBRATION (Paragraph 5-63).

LINE GENERATOR ADJUSTMENTS (Paragraph 5-68).

I.F. FILTER ALIGNMENT (Paragraph 5-70).

AMPLITUDE CALIBRATION (Paragraph 5-74).

MIXER BALANCE ADJUSTMENTS (Paragraph 5-81).

ADAPTIVE SWEEP MARKER ADJUSTMENT (Paragraph 5-84).

5-62. Trace Alignment Adjustment.

a. Position the 3580A front panel controls as follows:

| ADAPTIVE SWEEP |
|--|
| DISPLAY All pushbuttons released |
| AMPLITUDE MODE LOG 10 dB/DIV |
| AMPLITUDE REF LEVELNORMAL |
| $dBv/LIN - dBm 600 \Omega \dots dBv/LIN$ |
| INPUT SENSITIVITY30 dB |
| VERNIER (Amplitude)CAL |
| (Fully CW) |
| FREQUENCY 00.0 Hz |
| START - CTRSTART |
| RESOLUTION BANDWIDTH 300 Hz |
| DISPLAY SMOOTHING MIN |
| FREQ. SPAN/DIV 0.2 KHz |
| SWEEP TIME/DIV 0.1 SEC |
| SWEEP MODE REP |
| |

Option 002: Set dBm 900 Ω/LIN - dBm 600 Ω switch to dBm 900 $\Omega;$ set INPUT MODE switch to UNBAL.

- b. Adjust the front panel ADAPTIVE SWEEP for a line in the middle of the display. Adjust the front panel FOCUS control for the narrowest and sharpest line.
- c. Adjust A13R5 (TRACE ALIGN) for a level trace. If unable to achieve this, switch A13S1 and readjust A13R5.

5-63. Sweep Alignment and Dial Calibration.

5-64. These adjustments calibrate the front panel FRE-QUENCY dial plus align the frequency sweep limits. They should be done if the Frequency Tests (Paragraph 5-9) or Sweep Tests (Paragraph 5-13) of the Performance Tests cannot be passed by the instrument. In addition, the adjustment should be made if the high voltage supply was previously adjusted.

5-65. Recommended Test Equipment.

Digital Multimeter (-hp- Model 34740A and 34702A plug-on)
Electronic Counter (-hp- Model 5326A)
Oscilloscope (-hp- Model 180A with 1801A and 1820A plug-ins)

5-66. Linear Sweep Adjustments.

a. Position the following front panel controls: (Only those controls printed in BOLD require a change from the previous adjustments).

| ADAPTIVE SWEEP OFF |
|--|
| DISPLAY All pushbuttons released |
| AMPLITUDE MODE LOG 10 dB/DIV |
| $dBv/LIN - dBm 600 \Omega \dots dBv/LIN$ |
| INPUT SENSITIVITY |
| VERNIER (Amplitude)CAL |
| (Fully CW) |
| FREQUENCY 00.0 kHz |
| START - CTRSTART |
| RESOLUTION BANDWIDTH 300 Hz |
| DISPLAY SMOOTHINGMIN |
| FREQ. SPAN/DIV 0 kHz |
| SWEEP TIME/DIV 0.1 SEC |
| SWEEP MODE LOG ZERO |

Option 002: Set dBm 900 Ω/LIN - dBm 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL.

- b. Connect the multimeter (DC mode, 1 volt range) to the wiper of the front panel ZERO CAL pot (center terminal of pot). Adjust the front panel ZERO CAL pot for a dc reading on the multimeter of 0 ± 100 mV.
- c. Set the counter to the Frequency Mode and adjust the time base/multiplier for a measurement of 1 MHz with 6 digits of resolution (1000.00 kHz). Adjust for maximum input sensitivity and either a zero trigger Level or Preset. For the -hp- Model 5326A Counter, the controls should be set to:

Sample Rate: Fast Function: Freq. Multiplier 10⁶ Channel A: Slope + DC

Atten: X1

BNC Input:

Sep.

Connect the counter Channel A Input to the LO Output terminal on the back of the 3580A, and adjust the Level control until the LO frequency is displayed.

- d. Remove the inner circuit board shield (covering A2-A5). Connect the multimeter (DC mode 100 volt range) to A2TP4.
- e. Adjust A2L3 (100 kHz FREQ. ADJ.) for a reading of 1000.00 kHz ± .1 kHz on the counter. (100 kHz FREQ. ADJ. can be reached through side of circuit board card nest.)
- f. Adjust A2L1 (100 kHz VCO ADJ.) for a voltage reading on the multimeter between -1.5 V and -1.7 V. Record the reading.
- g. Repeat Steps e and f as necessary to meet the frequency and voltage specifications.
- h. Set the SWEEP MODE control to MANUAL and turn the MANUAL VERNIER control fully counterclockwise (CCW).
- i. Adjust A14R27 (DIAL LOW END ADJ.) for a display of $1000.00 \text{ kHz} \pm .01 \text{ kHz}$ on the counter.
 - j. Reposition the following front panel control:

FREQ. SPAN/DIV 5 kHz

- k. Adjust A3R54 (INTEGRATOR BALANCE) for a display of 1000.00 kHz ± .01 kHz on the counter.
- 1. Position the front panel MANUAL VERNIER control fully clockwise (CW).
- m. Adjust A2R75 (BUFFER AMP GAIN ADJ.) for a display of 1500.00 kHz \pm .01 kHz on the counter.
- n. Adjust A2R100 (VCO RANGE SET) for a reading on the multimeter equal to that obtained in Step $f (\pm 10 \text{ mV})$.
- o. Repeat Steps m and n as necessary to meet the frequency and voltage specifications.
- p. Position the front panel MANUAL VERNIER fully CCW.
 - q. Reposition the following front panel control:

FREQUENCY 50.0 kHz

- r. Adjust A14R25 (DIAL HIGH END SET) for a display of $1500.00 \text{ kHz} \pm .01 \text{ kHz}$ on the counter.
 - s. Reposition the following front panel control:

FREQUENCY 00.0 kHz

- t. Readjust A14R27 (DIAL LOW END ADJ.) for a display of $1000.00 \text{ kHz} \pm .01 \text{ kHz}$ on the counter.
- u. Repeat Steps q through t as necessary to meet the frequency specifications.
- v. Adjust the front panel FREQUENCY dial for 0, 10, 20, 30, 40 and 50 kHz. The corresponding frequency counter reading should be 1 MHz, 1.1 MHz, 1.2 MHz etc. with a tolerance of \pm 1 kHz.
 - w. Reposition the following front panel controls::

| FREQUENCY | 0.0 kHz |
|----------------------|---------|
| RESOLUTION BANDWIDTH | 300 Hz |
| SWEEP TIME/DIV | . 2 SEC |
| SWEEP MODE | |

- x. Adjust A13R1 (HORIZONTAL GAIN ADJ.) and A13R2 (HORIZONTAL POSITION ADJ.) for a full 10 cm display. The 10 kHz signal and its harmonics should fall on the proper graticule marking \pm 1/2 minor divisions (2nd, 4th, 6th, 8th and 10th graticule from the left).
- y. Connect the input of the oscilloscope to A3TP11. Set the oscilloscope input to dc coupling. Connect a jumper between A3TP3 and A3TP4.
- z. Adjust the A3R14 (RAMP COMPARATOR BALANCE) so that the output of the ramp comparator (on scope) just changes states.
 - aa. Remove the jumpers from the A3 board.
 - ab. Reposition the following front panel control:

SWEEP TIME/DIV0.1 sec

ac. Alternately press and release the STORE pushbutton, adjusting A8R4 (RAMP SIZE ADJ.) so that the 40 kHz harmonic of the CAL signal falls on the same point for both the STORE and non-STORE display modes.

5-67. Log Sweep Adjustments.

a. Position the following front panel controls: (Only those controls printed in BOLD require a change from the previous adjustments).

| ADAPTIVE SWEEP | OFF |
|-------------------|--------------------------|
| | All pushbuttons released |
| AMPLITUDE MODE. | LOG 10 dB/DIV |
| AMPLITUDE REF LE | VELNORMAL |
| | dBv/LIN |
| INPUT SENSITIVITY | |
| VERNIER (Amplitu | de) CAL |
| • | (Fully CW) |

| FREQUENCY | 00.0 kHz |
|----------------------|----------|
| START - CTR | START |
| RESOLUTION BANDWIDTH | 300 Hz |
| DISPLAY SMOOTHING | |
| FREQ. SPAN/DIV | 5 kHz |
| SWEEP TIME/DIV | 0.5 SEC |
| SWEEP MODE | LOG ZERO |

Option 002: Set dBm 900 Ω/LIN - dBm 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL.

b. Momentarily push:

Adjust the front panel ZERO CAL pot for a peak at the left graticule. If the peak is off the screen, adjust A13R3 for an on screen indication.

c. Reposition the following front panel control:

SWEEP MODE LOG

d. Allow the 3580A to make three complete sweeps. Then adjust A3R76 (20 kHz LOG SWEEP ADJ.) so that the 20 kHz harmonic of the CAL signal falls on the 20 kHz LOG SWEEP graticule.

NOTE

After each adjustment of A3R76, wait for the 3580A to sweep through 20 kHz before attempting to readjust the setting.

5-68. Line Generator Adjustments.

- 5-69. This adjustment properly aligns the line generator circuitry. The adjustment is usually not necessary, but should be done if components in the high voltage power supply are changed, or if the display exhibits overshoot to abrupt level changes.
- a. Position the following front panel controls: (Only those controls printed in BOLD require a change from the previous adjustments).

| ADAPTIVE SWEEP | OFF |
|----------------------------------|-------------------------------|
| DISPLAY All pushbutto | ons released |
| AMPLITUDE MODELOG | |
| AMPLITUDE REF LEVEL | .NORMAL |
| $dBv/LIN - dBm 600 \Omega \dots$ | dBv/LIN |
| INPUT SENSITIVITY | CAL |
| VERNIER (Amplitude) | \dots CAL |
| | (Fully CW) |
| | |
| FREQUENCY | 10.0 kHz |
| FREQUENCY START - CTR | |
| | \dots CTR |
| START - CTR | CTR 300 Hz |
| START - CTR | CTR 300 HzMIN |
| START - CTR | CTR300 HzMIN0.2 KHz |
| START - CTR | CTR 300 HzMIN 0.2 KHz 0.1 SEC |

Option 002: Set dBm 900 Ω/LIN - dBm 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL.

- b. Adjust MANUAL VERNIER for a peak display signal. Note: The Amplitude VERNIER may have to be adjusted to keep the signal within the display limits.
 - c. Momentarily press:

DISPLAY CLEAR WRITE

5-72. Tracking Oscillator and Center Frequency Adjustments.

a. Position the following front panel controls: (Only those controls printed in BOLD require a change from the previous adjustments).

| ADAPTIVE SWEEP OFF |
|--|
| DISPLAY All pushbuttons released |
| AMPLITUDE MODELINEAR |
| AMPLITUDE REF LEVELNORMAL |
| $dBv/LIN - dBm 600 \Omega \dots dBv/LIN$ |
| INPUT SENSITIVITY + 20 dB |
| VERNIER (Amplitude)CAL |
| (Fully CW) |
| FREQUENCY 10.0 kHz |
| START - CTRSTART |
| DISPLAY SMOOTHING MIN |
| RESOLUTION BANDWIDTH 1 Hz |
| FREQ. SPAN/DIV 0.5 KHz |
| SWEEP TIME/DIV |
| SWEEP MODE |
| Option 002: Set dBm 900 Ω/LIN - dBm 600 Ω |
| switch to dBm 900 Ω; set INPUT MODE |
| switch to UNBAL. |

5-73. Symmetry Adjustments.

a. Position the following front panel controls: (Only those controls printed in BOLD require a change from the previous adjustments).

| ADAPTIVE SWEEP OFF |
|---|
| DISPLAY All pushbuttons released |
| AMPLITUDE MODE LOG 10 dB/DIV |
| AMPLITUDE REF LEVEL NORMAL |
| $dBv/LIN - dBm 600 \Omega \dots dBv/LIN$ |
| INPUT SENSITIVITY CAL |
| VERNIER (Amplitude)CAL |
| (Fully CW) |
| |
| FREQUENCY 10.0 kHz |
| START-CTRCTR |
| RESOLUTION BANDWIDTH 30 Hz |
| DISPLAY SMOOTHING MIN |
| FREQ. SPAN/DIV 0.5 KHz |
| SWEEP TIME/DIV 0.1 SEC |
| SWEEP MODE |
| |
| OPTION 002: Set dBm 900 Ω/LIN - dBm |
| 600Ω switch to dBm 900Ω ; set INPUT |
| MODE switch to UNBAL. |

- c. Fine tune the FREQUENCY dial for a maximum display indication.
- i. Adjust MANUAL VERNIER while pressing and releasing DISPLAY - CLEAR WRITE for a spike display indication in the center of the screen. Adjust the front panel FREQUENCY dial for a maximum display indication.

5-77. Linear and Log Gain Adjustments.

a. Position the following front panel controls: (Only those controls printed in **BOLD** require a change from the previous adjustments).

Option 002: Set dBm 900 Ω/LIN - dBm 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL.

5-78. Bandwidth Gain Switching Adjustments.

a. Position the following front panel controls: (Only those controls printed in BOLD require a change from the previous adjustments).

| ADAPTIVE SWEEP OFF | F |
|--|---|
| DISPLAY All pushbuttons released | d |
| AMPLITUDE MODE LINEAR | |
| AMPLITUDE REF LEVEL NORMAI | L |
| $dBv/LIN - dBm 600 \Omega \dots dBv/LIN$ | V |
| INPUT SENSITIVITY20 dl | В |
| VERNIER (Amplitude) CAl | L |
| (Fully CW | |
| FREQUENCY 10.0 kH | Z |
| START - CTR CTI | |
| DISPLAY SMOOTHINGMIN | V |
| RESOLUTION BANDWIDTH 300 H | Z |
| FREQ. SPAN/DIV 50 H | Z |
| SWEEP TIME/DIV 0.1 SEG | С |
| SWEEP MODE | L |
| | |

d. Adjust MANUAL VERNIER and the front panel ZERO CAL pot for a peak reading in the center of the display. Make the following full scale adjustments on the appropriate bandwidth setting.

NOTE

The ZERO CAL pot may have to be readjusted after each Bandwidth/Freq. Span setting for a peak reading in the center of the screen.

5-79. Frequency Response Adjustments.

a. Position the following front panel controls: (Only those controls printed in **BOLD** require a change from the previous adjustments.

| ADAPTIVE SWEEP OF | F |
|--|----------------|
| DISPLAY All pushbuttons release | d |
| AMPLITUDE MODE 1 dB/DI | V |
| AMPLITUDE REF LEVEL NORMA | |
| $dBv/LIN - dBm'600 \Omega \dots dBv/LIN$ | N |
| INPUT SENSITIVITY20 d | |
| VERNIER (Amplitude) CA | L |
| (Fully CW | ¹) |
| FREQUENCY 01.0 kH | |
| START - CTR | |
| RESOLUTION BANDWIDTH 300 H | |
| DISPLAY SMOOTHING | |
| FREQ. SPAN/DIV 0.2 kH | Z |
| SWEEP TIME/DIV 0.1 SEG | С |
| SWEEP MODE | |
| Option 002: Set dBm 900 Ω/LIN - dBm 600 Ω | |
| switch to dBm 900 Ω ; set INPUT MODE | |
| switch to UNBAL. | |

- b. Adjust the frequency synthesizer for 1 kHz output at 20 dBV (- 20 dBm 900 Ω for Option 002), and connect it to the 3580A INPUT (properly terminated). Adjust the front panel CAL 10 KHz for a full scale (0 dB) display.
 - c. Reposition the following front panel controls:

| FREQUENCY | 40.0 kHz |
|-------------------|----------|
| INPUT SENSITIVITY | - 10 dR |

5-84. Adaptive Sweep Marker Adjustment.

- 5-85. This adjustment properly positions the ADAPTIVE SWEEP marker. If the marker (blank spot on screen) does not appear at the same point on the display as new information being written onto the display, do this adjustment:
- a. Position the following front panel controls: (Only those controls printed in BOLD require a change from the previous adjustments).

| ADAPTIVE SWEEP | OFF |
|----------------------------------|-------------------|
| | (Fully CCW) |
| DISPLAY All pus | hbuttons released |
| AMPLITUDE MODE I | LOG 10 dBV/DIV |
| AMPLITUDE REF LEVEL . | NORMAL |
| $dBv/LIN - dBm 600 \Omega \dots$ | dBv/LIN |
| INPUT SENSITIVITY | CAL |

| VERNIER (Amplitude) | CAL |
|----------------------|----------|
| • • • | |
| FREQUENCY | 00.0 kHz |
| START-CTR | START |
| RESOLUTION BANDWIDTH | 300 Hz |
| DISPLAY SMOOTHING | |
| FREQ. SPAN/DIV | |
| SWEEP TIME/DIV | 1 SEC |
| SWEEP MODE | |

Option 002: Set dBm 900 Ω/LIN - dBm 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL.

- b. Adjust the MANUAL VERNIER control until the trace is at the peak of the 10 kHz signal.
- c. Momentarily press the DISPLAY-CLEAR WRITE button. A dot should remain at the top of the scope.
- d. Turn the ADAPTIVE SWEEP on and adjust A8R3 (SWEEP MARKER ADJ.) until the sweep marker (blank spot in trace) blanks out the dot at the top of the scope.

Page 5-27/5-28, Figure 5-8 should appear as on Page 8-33/8-34.

Performance Test Card. The following items in the Performance Test Card (following Page 5-27/5-28) should appear as follows:

FREQUENCY SPAN TESTS

| Frequency Span/Div. | Counter Reading (Manual Vernier Fully CW) | Test Limits | | |
|---|---|--|--|--|
| 5 Hz 10 Hz 20 Hz 50 Hz . 1 kHz . 2 kHz | kHz kHz kHz kHz kHz | 1000.50 kHz ± .01 kHz 1001.00 kHz ± .02 kHz 1002.00 kHz ± .04 kHz 1005.00 kHz ± .10 kHz 1010.00 kHz ± .20 kHz 1020.00 kHz ± .40 kHz | | |
| . 5 kHz 1 kHz 2 kHz | kHz kHz kHz | 1050.00 kHz ± 1.00 kHz 1100.00 kHz ± 2.00 kHz 1200.00 kHz ± 4.00 kHz | | |

RANGE AND FREQUENCY DIAL ACCURACY TESTS

| Ideal Frequency Dial Setting | Actual Setting for a Peak | Test Limits |
|------------------------------|---------------------------|-------------|
| 10 kHz | kHz | ± .1 kHz |
| 20 kHz | kHz | ± .1 kHz |
| 30 kHz | kHz | ± .1 kHz |
| 40 kHz | kHz | ± .1 kHz |
| 50 kHz | kHz | ± .1 kHz |

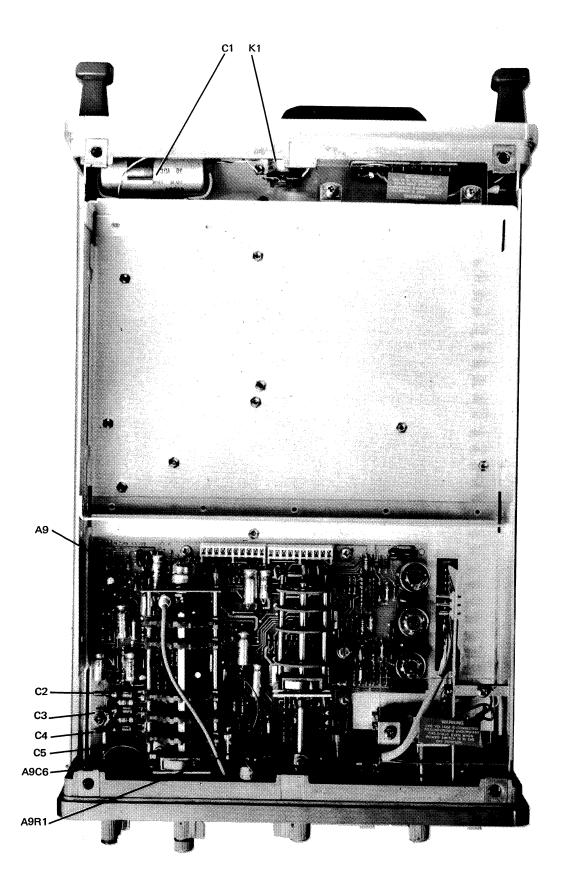
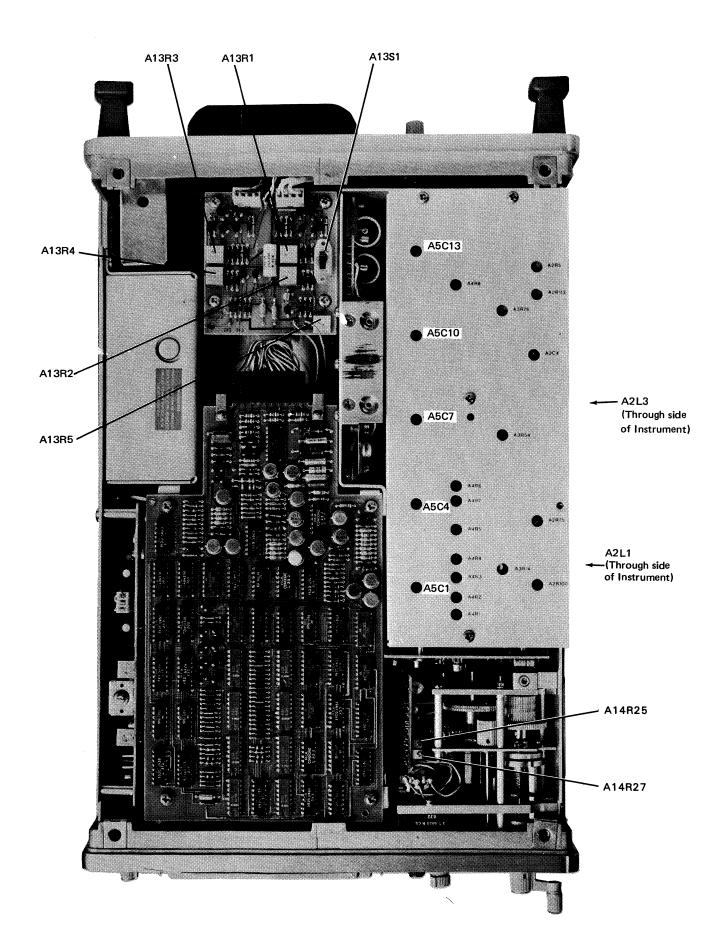


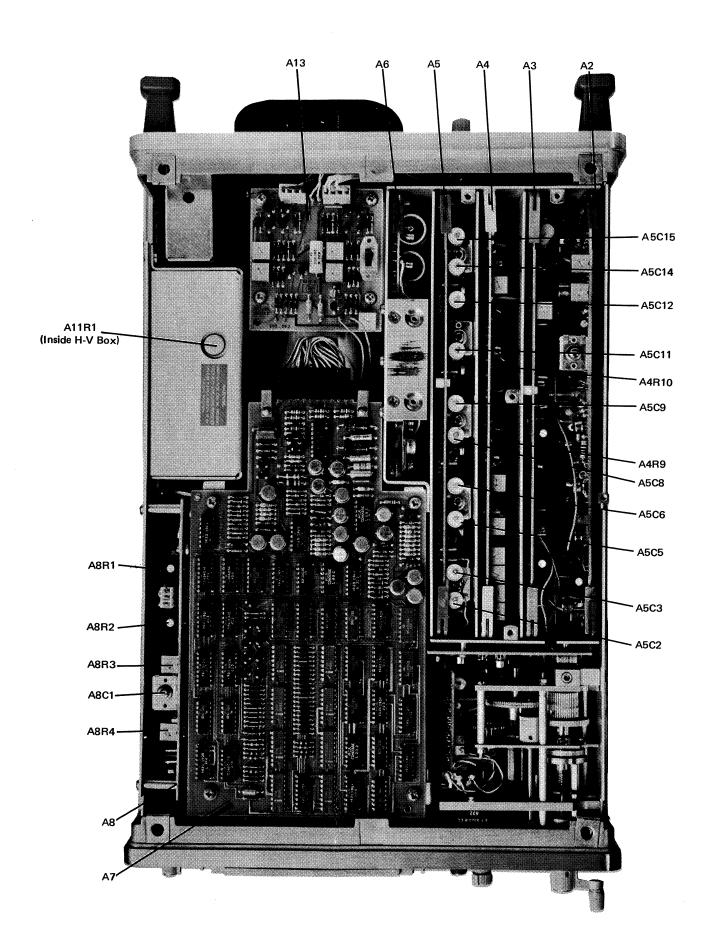
Figure 5-8. Test Point and Adjustment Locations. 8-33/8-34



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Page 6-31, 6-32, and 6-33 should be as follows:

| Reference Designation | HP Part Number | Qty | Description | Mfr Code | Mfr Part Number |
|--------------------------|-------------------------------------|-----|---|-------------------------|------------------------|
| | | | | | |
| A13R33 | 0698-4435 | | R:FXD FLM 2.49K OHM 1% 1/8W | 28480 | 0698-4435 |
| A13R34 | 0698-4435 | | R:FXD FLM 2.49K OHM 1% 1/8W | 28480 | 0698-4435 |
| A13R35 | 0757-0467 | . 2 | R:FXD MET FLM 121K OHM 1% 1/8W | 284 80 | 0757-0467 |
| A13R36 | 0757-0476 | 1 | R:FXD MET FLM 301K OHM 1% 1/8W | 284 80 | 0757-0476 |
| A13S1 | 3101-1162 | 1 | SWITCH: SLIDE MINIATURE, SPDT | 79727 | GF 124-0008 |
| A14 | 03580-66514 | 1 | BOARD ASSY:BANDWIDTH SWITCH | 28480 | 03580-66514 |
| A14C4 | 0180-0197 | ı | C:FXD ELECT 2.2 UF 10% 20VDCW | 56289 | 1500225X9020A2-DYS |
| A14C5 | 0180-0373 | | C:FXD ELECT 0.68 UF 10% 35VDCW | 56289 | 1500684X9035A2-DYS |
| A14C6 | 0180-1735 | 1 | C:FXD ELECT 0.22 UF 10% 35VDCW | 28480 | 0180-1735 |
| A14C7 | 0180-2050 | | C:FXD TANT. 0.082 UF 10% 35VDCW | 56289 | 1500823X9035A2-0YS |
| A14C8 | 0180-1701 | | C:FXD ELECT 6.8 UF 20% 6VDCW | 28480 | 0180-1701 |
| A14C9 | 0160-0162 | | C:FXD MY 0.022 UF 10% 200VDCW | 562 89 | 192P22392-PTS |
| A14C12 | 0180-0106 | | C:FXD ELECT 60 UF 20% 6VDCW | 28480 | 0180-0106 |
| A14C15 | 0180-0339 | | C:FXD AL ELECT 50 UF +75-10% 15VDCW | 56289 | 30D506G015C82-DSM |
| A14CR1 | 1902-0777 | _ | DIODE:BREAKDOWN 6.2V 5% | 04713 | 1N825 |
| A14R1 A14R2 A14R3 | 0698-3453 0698-4488 0698-3558 | 1 | R:FXD MET FLM 196K OHM 1% 1/8W R:FXD FLM 26.7K OHM 1% 1/8W | 28480 28480 | 0698-3453 0698-4488 |
| A14R4 A14R5 | 0698-3558 0698-3519 0698-3228 | | R:FXD MET FLM 4.02K OHM 1% 1/8W R:FXD MET FLM 12.4K OHM 1% 1/8W R:FXD MET FLM 49.9K OHM 1% 1/8W | 28480 28480 28480 | 0698-3558 0698-3519 |
| A14R6 | 0757-0473 | 1 | R:FXD MET FLM 221K OHM 1% 1/8W | 28480 | 0698-3228 0757-0473 |
| A14R7 A14R8 | 0684-1051 0684-2251 | • | R:FXD COMP 2-2 MEGOHM 1% 1/4W | 01121 01121 | CB 1051 CB 2251 |
| A14R9 A14R1C | 0684-1041 0684-3941 | 1 | R:FXD COMP 100K OHM 10% 1/4W R:FXD COMP 390K OHM 10% 1/4W | 01121 01121 01121 | CB 1041 CB 3941 |
| A14R11 | 0698-5102 | 1 | R:FXO COMP 1.2 MEGOHM 10% 1/4W | 01121 | CB 1251 |
| A14R12 | 0698-4443 | 1 | R:FXD FLM 4.53K OHM 1% 1/8W | 28480 | 0698-4443 |
| A14R13 | 0757-0454 | | R:FXD MFT FLM 33.2K OHM 1% 1/8W | 28480 | 0757-0454 |
| A14R14 | 0698-4506 | 1 | R:FXD FLM 73.2K OHM 1% 1/8W | 28480 | 0698-4506 |
| A14R15 | 0698-3459 | 1 | R:FXD MET FLM 383K OHM 1% 1/8W | 28480 | 0698-3459 |
| A14R16 | 0698-4524 | 3 | R:FXD FLM 174K OHM 1% 1/8W | 28480 | 0698-4524 |
| A14R17 | 0757-0442 | | R:FXD MET FLM 10.0K OHM 1% 1/8W | 28480 | 0757-0442 |
| A14R18 | 0698-4441 | 1 | R:FXD MET FLM 3.74K OHM 1% 1/8W | 28480 | 0698-4441 |
| A14R19 | 0698-4427 | | R:FXD FLM 1650 OHM 1% 1/8W | 28480 | 0698-4427 |
| A14R20 | 0698-4511 | 3 | R:FXD FLM 86.6K DHM 1.0% 1/8W | 28480 | 0698-4511 |
| A14R21 | 0757-0456 | 3 | R:FXD MET FLM 43.2K OHM 1% 1/8W | 28480 | 0757-0456 |
| A14R22 | 0757-0446 | | R:FXD MET FLM 15.0K OHM 1% 1/8W | 28480 | 0757-0446 |
| A14R23 | 0757-0415 | 1 | R:FXD MET FLM 475 OHM 1% 1/8W | 28480 | 0757-0415 |
| A14R24 | 0757-0407 | | R:FXD MET FLM 200 OHM 1% 1/8W | 28480 | 0757-0407 |
| A14R25 | 2100-3123 | 1 | R:VAR CERMET 500 OHM 10% TYPE P 3/4W | 28480 | 2100-3123 |
| A14R26 | 0698-5673 | 1 | R:FXD MET FLM 3.9K OHM 1% 1/8W | 28480 | 0698-5673 |
| A14R27 | 2100-3161 | 1 | R:VAR CERMET 20K OHM 10% TYPE P 3/4W | 28480 | 2100-3161 |
| A14R28 | 0698-3279 | | R:FXD MET FLM 4990 OHM 1% 1/8W | 28480 | 0698-3279 |
| A14R31 | 0698-4511 | | R:FXD FLM 86.6K OHM 1.0% 1/8W | 28480 | 0698-4511 |
| A14R32 | 0698-4500 | 2 | R:FXD FLM 57.6K OHM 1% 1/8W | 28480 | 0698 -4 500 |
| A14R33 | 0757-0456 | 1 | R:FXD MET FLM 43.2K OHM 1% 1/8W | 28480 | 0757-0456 |
| A14R34 | 0757-0123 | | R:FXD MET FLM 34.8K OHM 1% 1/8W | 28480 | 0757-0123 |
| A14R35 | 0698-3455 | 2 | R:FXD MET FLM 261K OHM 1% 1/8W | 28480 | 0698-3455 |
| A14R36 | 0757-0468 | | R:FXD FLM 130K OHM 1% 1/8W | 28480 | 0757-0468 |
| A14R37 | 0698-7802 | 2 | R:FXD FLM 523K OHM 1.0% 1/8W | 28480 | 0698-7802 |
| A14R38 | 0757-0272 | ı | R:FXD FLM 52.3K OHM 1% 1/8W | 28480 | 0757-0272 |
| A14R39 | 0698-4502 | | R:FXF FLM 64.9K OHM 1% 1/8W | 28480 | 0698-4502 |
| A14R40 | 0698-3228 | 1 | R:FXD MET FLM 49.9K OHM 1% 1/8W | 28480 | 0698-3228 |
| A14R41 | 0698-3215 | | R:FXD FLM 499K OHM 1.0% 1/8W | 28480 | 0698-3215 |
| A14R42 | 0698-3228 | | R:FXD MET FLM 49.9K OHM 1% 1/8W | 28480 | 0698-3228 |
| A14R43 | 0698-3228 | | R:FXD MET FLM 499.9K UHM 1% 1/8W | 28480 | 0698-3228 |
| A14R44 | 0698-4524 | 1 | R:FXD FLM 174K OHM 1% 1/8W | 28480 | 0698-4524 |
| A14R45 | 0698-4542 | | R:FXD FLM 453K OHM 1% 1/8W | 28480 | 0698-4542 |
| A14R46* | 0698-3540 | _ | R: FXD MET FLM 15.4 K OHM 1% 1/8W | 16299 | C4-1/8-TO-1542-F |
| A14R101 | 0757-0446 | | R: FXD MET FLM 15.0 K OHM 1% 1/8W | 28480 | 0757-0446 |
| A14R102 A14R103 | 0698-3572 0698-4518 | | R:FXD FLM 60.4K OHM 1% 1/8W R:FXD FLM 137K OHM 1% 1/8W | 28460 | 0698-3572 |
| A14R104 | 0698-4518 0698-3456 | 1 | R:FXD MET FLM 287K OHM 1% 1/8W | 28480 28480 | 0698-4518 0698-3456 |
| A14R105 | 0757-0486 | 1 | R:FXD MET FLM 750K OHM 1% 1/8W | 28480 | 0757-0486 |
| A14R106 | 0698-5904 | | R:FXD FLM 1.58 MEGOHM 1.0% 1/2W | 28480 | 0698-5904 |
| A14R107 | 0698-7094 | 1 | R:FXD MET FLM 3.32 MEGOHM 1% 1/4W | 28480 | 0698-7094 |
| A14R108 | 0698-7091 | 1 | R:FXD MET FLM 10 MEGOHM 1% 1/2W | 28480 | 0698-7091 |
| A14R109 | 0698-5675 | 2 | R:FXD MET FLM 30 MEGDHM 1% 1W | 28480 | 0698-5675 |
| A14R110 | 0698-5675 | | R:FXD MET FLM 30 MEGDHM 1% 1W | 28480 | 0698-5675 |
| A14S1 | 03580-61901 | 1 | SWITCH ASSY | 28480 | 03580-61901 |

| Reference Designation | HP Part Number | Qty | Description | Mfr Code | Mfr Part Number |
|--|---|-------------|---|---|--|
| | | | | | |
| A1451 A1451 A1452 A14U1 | 3100-2740 3100-2736 3100-2740 1826-0304 | 1 1 1 | SWITCH:ROTARY SWITCH:ROTARY SWITCH:BANDWIDTH IC LF 355 OP AMP | 28 4 80 28 4 80 28 4 80 28 4 80 27 0 1 4 | 3100-2740 3100-2736 3100-2740 LF355H |
| A15 | 03580-66515 | 1 | BOARD ASSY:SWEEP SWITCH | 28480 | 03580-66515 |
| A15R1 | 0698-7802 | | R:FXD FLM 523K OHM 1.0% 1/8W | 28480 | 0698-7802 |
| ` A15R2 A15R3 A15R4 A15R5 A15R6 | 0757-0403 0757-0410 0757-0161 0757-0274 0757-0273 | | R:FXD MET FLM 121 OHM 1% 1/8W R:FXD MET FLM 301 OHM 1% 1/8W R:FXD FLM 604 OHM 1% 1/8W R:FXD MET FLM 1-21K OHM 1% 1/8W R:FXD MET FLM 3.01K OHM 1% 1/8W | 284 80 28 480 28 480 28 480 28 480 | 0757-0403 0757-0410 0757-0161 0757-0274 0757-0273 |
| A15R7 A15R8 A15R9 A15R10 A15R11 | 0698-3497 0757-0444 0757-0453 0698-3572 0757-0467 | 2 | R:FXD FLM 6.04K OHM 1% 1/8W R:FXD MET FLM 12.1K OHM 1% 1/8W R:FXD MET FLM 30.1K OHM 1% 1/8W R:FXD FLM 60.4K OHM 1% 1/8W R:FXD MET FLM 121K OHM 1% 1/8W | 28480 28480 28480 28480 28480 | 0698-3497 0757-0444 0757-0453 0698-3572 0757-0467 |
| A15R12 A15R13 A15R14 A15R15 A15R16 A15R17 A15R18 A15R19 A15R20 A15R21 A15R21 | 0698-3499 0698-3497 0757-0442 0757-0444 0698-5572 0698-5572 0757-0442 2100-0668 0698-3519 0698-6758 0698-5580 | 2 | R:FXD FLM 40.2K OHM 1% 1/8W R:FXD FLM 6.04K OHM 1% 1/8W R:FXD MET FLM 10.0K OHM 1% 1/8W R:FXD MET FLM 12.1K OHM 1% 1/8W R:FXD FLM 12.5K OHM 0.5% 1/8W R:FXD FLM 12.5K OHM 0.5% 1/8W R:FXD MET FLM 10.0K OHM 1% 1/8W R:FXD MET FLM 10.0K OHM 1% 1/8W R:FXD MET FLM 12.4K OHM 1% 1/8W R:FXD MET FLM 12.4K OHM 1% 1/8W R:FXD FLM 22.5K OHM 0.5% 1/8W R:FXD FLM 25K OHM 0.5% 1/8W | 28480 28480 28480 28480 28480 28480 12697 28480 28480 28480 28480 | 0698-3499 0698-3497 0757-0442 0757-0444 0698-5572 0698-5572 381 0698-3519 0698-3519 0698-5758 |
| A15R23 A15R24 A15R25 A15R26 A15R27 | 0698-5573 0698-6292 0698-5581 0757-0015 0698-5916 | 1 1 1 | R:FXD FLM 50K OHM 0.5% 1/8W R:FXD FLM 125K OHM 0.5% 1/8W R:FXD FLM 250K OHM 0.5% 1/8W R:FXD MET FLM 500K OHM 1/2% 1/2W R:FXD MET FLM 1.25 MEGOHM 1.0% 1/2W | 28480 28480 28480 28480 28480 | 0698-5573 0698-6292 0698-5581 0757-0015 0698-5916 |
| A15R28 A15R29 A15R3C A15R31 A15R32 | 0698-5987 0698-3587 0757-0486 0698-4489 0684-3351 | 1 | R:FXD MET FLM 2.5 MEGGHM 1.0% 1/2W R:FXD MET FLM 5.00 MEGGHM 1% 1W R:FXD MET FLM 750K OHM 1% 1/8W R:FXD FLM 28K OHM 1% 1/8W R:FXD 3.3 MEGGHM 10% | 28480 28480 28480 28480 01121 | 0698-5987 0698-3587 0757-0486 0698-4489 CB 3351 |
| A15R41 A15R42 A15R43 A15R44 A15S1 | 0698-4524 0698-3455 0698-4500 0698-4511 03580-61903 | 1 | R:FXD FLM 174K OHM 1% 1/8W R:FXD MET FLM 261K OHM 1% 1/8W R:FXD FLM 57.6K OHM 1% 1/8W R:FXD FLM 86.6K OHM 1.0% 1/8W SWITCH ASSY:SPAN | 28480 28480 28480 28480 28480 | 0698-4524 0698-3455 0698-4500 0698-4511 03580-61903 |
| A15S1 A15S2 A15U1 A15U2 | 3100-2742 03580-61904 1826-0043 1826-0043 | 1 | SWITCH:ROTARY SWITCH ASSY:MODE IC:LINEAR OPERATIONAL AMPLIFIER IC:LINEAR OPERATIONAL AMPLIFIER | 28480 28480 28480 28480 | 3100-2742 03580-61904 1826-0043 1826-0043 |
| A16 | 03580-66516 | 1 | BOARD ASSY:FCM | 28480 | 03580-66516 |
| A16C1 | 0180-1743 | | C:FXD ELECT 0.1 UF 10% 35VDCW | 56289 | 1500.104X9035A2-DYS |
| A16C2 A16C3 A16C4 A16C12 A16CR1 | 0160-2207 0150-0093 0180-0376 0150-0093 1901-0040 | | C:FXD MICA 300 PF 5% C:FXD CER 0.01 UF +80-20% 100VDCW C:FXD ELECT 0.47 UF 10% 35VDCW C:FXD CER 0.01 UF +80-20% 100VDCW DINDE:SILICON 50 MA 30 WV | 28480 72982 56289 72982 07263 | 0160-2207 801-K800011 1500474X9035A2-DYS 801-K800011 FDG1088 |
| A16CR2 A16CR3 A16CR4 A16CR6 A16CR7 | 1901-0040 1901-0040 1901-0040 1902-0025 1901-0040 | | DIODE:SILICON 50 MA 30 MV DIODE:SILICON 50 MA 30 MV DIODE:SILICON 50 MA 30 MV DIODE:BREAKOOMN:10.0V 5% 400 MW DIODE:SILICON 50 MA 30 MV | 07263 07263 07263 28480 07263 | FDG1088 FDG1088 FDG1088 1902-0025 FDG1088 |
| A16CR8, CR9 A16J1, J2 A16L1 A16O1 A16O2 | 1901-0040 1251-2035 9100-1644 1854-0354 1853-0010 | 1 | DIODE:SILICON 50 MA 30 MY CONN:PC EDGE (2 × 15) 30 CONTACT COIL/CHOKE 330 UH 5% TSTR:SI MPN ISTR:SI PNP(SELECTED FROM 2N3251) | 07263 71785 28480 28480 28480 | FDG1088 252-15-30-300 9100-1644 1854-0354 1853-0010 |
| A1603 A1604 A1606 A1607 A1608 | 1854-0071 1853-0010 1854-0071 1854-0475 1855-0386 | 1 | TSTR:SI NPN(SELECTED FROM 2N3704) TSTR:SI PNP(SELECTED FROM 2N3251) TSTR:SI NPN(SELECTED FROM 2N3704) TSTR:SI NPN TSTR:SI NPN TSTR:FET N-CHANNEL | 28480 28480 28480 28480 80131 | 1854-0071 1853-0010 1854-0071 1854-0475 2N4392 |

| Reference Designation | | | Description | Mfr Code | Mfr Part Number | |
|---|---|-------------|--|--|---|--|
| | | | | | | |
| A16R1 A16R2 A16R3 A16R4 A16R5 | 0757-0270 0757-0270 0757-0426 0698-4499 0698-3162 | 2 1 2 | R:FXD MET FLM 249K OHM 1% 1/8W R:FXD MET FLM 249K OHM 1% 1/8W R:FXD FLM 1.3K OHM 1% 1/8W R:FXD FLM 54.9K OHM 1% 1/8W R:FXD MET FLM 46.4K OHM 1% 1/8W | 28480 28480 28480 28480 28480 | 0757-0270 0757-0270 0757-0426 0698-4499 0698-3162 | |
| 16R6 16R7 16R8 16R9 16R10 | 0698-4503 0684-2231 0757-0282 0684-1031 0698-3228 | | R:FXD FLM 66.5K OHM 1% 1/8W R:FXD COMP 22K OHM 10% 1/4W R:FXD MFT FLM 221 OHM 1% 1/8W R:FXD COMP 10K OHM 10% 1/4W R:FXD MET FLM 49.9K OHM 1% 1/8W | 28480 01121 28480 01121 28480 | 0698-4503 CB 2231 0757-0282 CB 1031 0698-3228 | |
| 16R11 16R12 16R13 16R14 16R15 | 0757-0456 0698-3228 0698-3228 0684-1041 0684-2251 | | R:FXD MET FLM 43.2K CHM 1% 1/8W R:FXD MET FLM 49.9K CHM 1% 1/8W R:FXD MET FLM 49.9K CHM 1% 1/8W R:FXD COMP 100K CHM 10% 1/4W R:FXD COMP 2.2 MEGOHM 10% 1/4W | 28480 28480 28480 01121 01121 | 0757-0456 0698-3228 0698-3228 CB 1041 CB 2251 | |
| 16R1 6 16R1 7 16R1 8 16R1 9 16R2 2 | 0757-0440 0757-0460 0698-3557 0698-3228 0684-2231 | | R:FXD MET FLM 7.50K OHM 1% 1/8W R:FXD MET FLM 61.9K OHM 1% 1/8W R:FXD FLM 806 OHM 1% 1/8W R:FXD MET FLM 49.9K OHM 1% 1/8W R:FXD COMP 22K OHM 10% 1/4W | 28480 28480 28480 28480 01121 | 0757-0440 0757-0460 0698-3557 0698-3228 CB 2231 | |
| 16R24 16R25 16R2 <i>6</i> 16R27 16R28 | 0757-0479 0757-0273 0698-3162 0698-3228 0757-0463 | 1 | R:FXD MET FLM 392K OHN 1% 1/8W R:FXD MET FLM 3.01K OHM 1% 1/8W R:FXD MET FLM 46.4K OHM 1% 1/8W R:FXD MET FLM 49.9K OHM 1% 1/8W R:FXD MET FLM 82.5K OHM 1% 1/8W | 28 4 8 0 28 4 8 0 28 4 8 0 28 4 8 0 28 4 8 0 | 0757-0479 0757-0273 0698-3162 0698-3228 0757-0463 | |
| 16R29 16R3C 16R31 16U1 16U2 | 0698-3557 0684-6831 0684-1041 1826-0043 1826-0043 | | R:FXD FLM 806 OHM 1% 1/8W R:FXD COMP 68K OHM 10% 1/4W R:FXD COMP 100K OHM 10% 1/4W IC:LINEAR OPERATIONAL AMPLIFIER IC:LINEAR OPERATIONAL AMPLIFIER | 28480 01121 01121 28480 28480 | 0698-3557 C8 6831 CB 1041 1826-0043 1826-0043 | |
| 16U3 16U4 17 | 1820-0223 1826-0111 | 1 | INTEGRATED CIRCUIT:OPERATIONAL AMPL. IC NOT ASSIGNED | 28480 04713 | 1820-0223 MC1458C | |
| λ18 Δ ₁ | 03581-66518 | 1 | BOARD ASSY: INPUT, BALANCED (FOR OPTION 002 ONLY) | 28480 | 0358066518 | |
| .18C1 .18C4 .18C5 .18C6 | 0180-0091 0180-0091 0160-2206 0140-0204 | 2 1 | C:FXD 10UF+50 -10% 100VDC AL C:FXD 10UF+50 -10% 100VDC AL C:FXD 160PF 5% 300VDCW C:FXD 47PF 5% 500VDCW | 56289 56289 28480 72136 | 30D106F100DC2 30D106F100DC2 0160-2206 DM15E470J0500WV1CR | |
| .18J1 .18J2 | 1251–2969 1251–3638 | 1 | CONN:PHONO, SINGLE JACK CONN:POST TYPE | 27264 27264 | 15-24-0501 09-65-1061 | |
| 18R1 18R2 18R3* | 0698-4882 0698-5874 07570284 | 1 1 1 | R:FXD 976 OHM 1% .5 W F TUBULAR R:FXD 639 OHM 1% .5W F TUBULAR R:FXD 150 OHM 1% .125W F TUBULAR FACTORY SELECTED PART | 24546 24546 24546 | NA6 NA6 C4-1/8-TO-151-F | |
| 18R4 18R5 | 0757-0472 0698-4308 | 1 | R:FXD 200K 1% .125W F TUBULAR R:FXD 16.9K 1% .125W F TUBULAR | 24546 16299 | C4-1/8-TO-2003-F C4-1/8-TO-1692-F | |
| 18T1 | 9100—1460 | 1 | TRANSFORMER AUDIO | 28480 | 91001460 | |
| 19 | | | NOT ASSIGNED | | | |
| 20 | 0960-0444 | 1 | POWER INPUT MODULE | 28480 | 0960-0444 | |
| | | | | | | |
| | | | | | | |

Page 6-34. Delete the parts list for the A33 board.

Pages 6-35, 6-36 should be as follows:

| Reference Designation | HP Part Number | Qty | Description | Mfr Code | Mfr Part Number | |
|--------------------------------------|---|---------------------------------|---|---|--|--|
| | 03580-04104 5040-7042 03580-24706 03580-26001 3050-0456 5040-0508 03580-00203 | 2 4 4 4 4 1 1 | MISCELLANEOUS MECHANICAL PARTS (CONT'D) COVER:SIDE RAIL CAP:END (FOR HANDLE) RETAINER (FOR HANDLE) SCREW (FOR HANDLE) WASHER (FOR HANDLE) LIGHT SHIELD:CRT (PLASTIC) PANEL:FREQUENCY CONTROL MODULE | 28480 28480 28480 28480 86928 28480 28480 | 03580-04104 5040-7042 03580-24706 03580-26001 5808-16-15 5040-0508 03580-00203 | |
| | 0358000211 Δ 0358000214 Δ | 1 | PANEL:FRONT STANDARD 3580A OPTION 002 PANEL:REAR | 28480 28480 | 03580-00201 03580-00204 | |
| | 03580-00212 03580-00205 | 1 | STANDARD 3580A OPTION 002 | 28480 28480 | 03580-00202 03580-00205 | |
| | 1460—1341 5060—7440 | 1 | STAND:TILT WINDOW:FREQUENCY KNOBS | 28480 28480 | 14601341 50607440 | |
| | 0370—1005 0370—2182 0370—2186 0370—2188 0370—1005 | | ADAPTIVE SWEEP AMPLITUDE REF LEVEL BANDWIDTH DISPLAY SMOOTHING FOCUS | 28480 28480 28480 28480 28480 28480 | 0370-1005 0370-2182 0370-2186 0370-2188 0370-1005 | |
| | 0370-1115 0370-2185 03580-67401 7120-4008 0370-1005 | | FREQUENCY (CRANK) FREQUENCY SPAN INPUT SENSITIVITY DECAL INTENSITY | 28480 28480 28480 28480 28480 | 0370-1115 0370-2185 03580-67401 7120-3115 0370-1005 | |
| | 0370-2188 0370-2473 0370-2187 0370-2184 0370-2188 | | MANUAL VERNIER POWER SWEEP MODE SWEEP TIME CONCENTRIC KNOB | 28480 28480 28480 28480 28480 | 0370-2188 0370-2473 0370-2187 0370-2184 0370-2188 | |
| | 0370-2189 0370-1019 0370-0906 0370-0934 0370-0914 | 6 6 6 | VERNIER ZERO CAL PUSHBUTTON—BASE PUSHBUTTON—CAP PUSHBUTTON—BEZEL | 28480 28480 28480 28480 28480 | 0370-2189 0370-1019 0370-0906 0370-0934 0370-0914 | |
| | 03500137 03500136 03500135 03500138 | 1 1 1 3 | LABEL:PUSHBUTTON, 1 DB LABEL:PUSHBUTTON, 10 DB LABEL:PUSHBUTTON, LIN LABEL:PUSHBUTTON, PLAIN | 28480 28480 28480 28480 | 0350-0137 0350-0136 0350-0135 0350-0138 | |
| | | | MECHANICAL PARTS (SEE FIGURE 6-1) | | | |
| MP1 MP2 MP3 MP4 MP5 | 1140-0059 03580-24302 03580-24303 1430-0777 03580-24304 | 1 1 1 1 | COUNTER:MECH PLATE:COUNTER PLATE:POT GEAR:SPUR PLATE:REAR | 28480 28480 28480 28480 28480 | 1140-0059 03580-24302 03580-24303 1430-0777 03580-24304 | |
| MP6 MP7 MP8 MP9 MP10 | 03580-20801 1430-0778 1430-0775 03580-24704 1460-0563 | 1 1 1 4 1 | HSG:DETENT GEAR:SPUR GEAR:SPUR HSG:SPACER SPRING:CLUTCH | 28480 28480 28480 28480 28480 | 03580-20801 1430-0778 1430-0775 03580-24704 1460-0563 | |
| MP11 MP12 MP13 MP14 MP15 | 03580-24705 03580-23704 03580-21401 03580-21204 3050-0587 | 2 1 2 1 | SPACER:RATIO DRIVE SHAFT:COUNTER RATIO DRIVE ADAPTER:CLUTCH WASH:NEOPRENE | 28480 28480 28480 28480 28480 | 03580-24705 03580-23704 03580-21401 03580-21204 3050-0587 | |
| MP16 MP17 MP18 MP19 MP20 | 03580-23703 1430-0713 03580-22402 03580-24301 03580-24702 | 1 1 1 1 3 | SHAFT:RATIO DRIVE GEAR:MITER GEAR:BEVEL, MOD PLATE:FRONT (THIS INCLUDES SWITCH, 3101–0199) SPACER:HSG | 28480 28480 28480 28480 28480 | 03580-23703 1430-0713 03580-22402 03580-24301 03580-24702 | |
| MP21 MP22 MP23 MP24 MP25 | 03580-01216 5040-7532 03580-24703 03580-62401 03580-22401 | 1 1 2 1 | PLATE:CLUTCH CLUTCH SPACER:HSG GEAR:ANTI-BACKLASH GEAR:STOP-MOD | 28480 28480 28480 28480 28480 | 03580-01216 5040-7532 03580-24703 03580-62401 03580-22401 | |
| MP26 MP27 | 00692-247 03580—23705 | 1 1 | GEAR:STOP SHAFT:LIMIT | 28480 28480 | 00692-247 03580—23705 | |
| R6 R7 | 2100-0564 2100-0574 | 1 1 | R:VAR, 100 K 20° R:VAR 10 TURN 5 K – 10% | 28480 28480 | 2100-0564 2100-0574 | |
| S7 | 31010199 | 1 | SWITCH SLIDE | 28480 | 3101-0199 | |

Δ For S/N 1312A-00365 and below: order 03580-00201 (Std) or 03580-00204 (Opt. 002).

Model 3580A Backdating

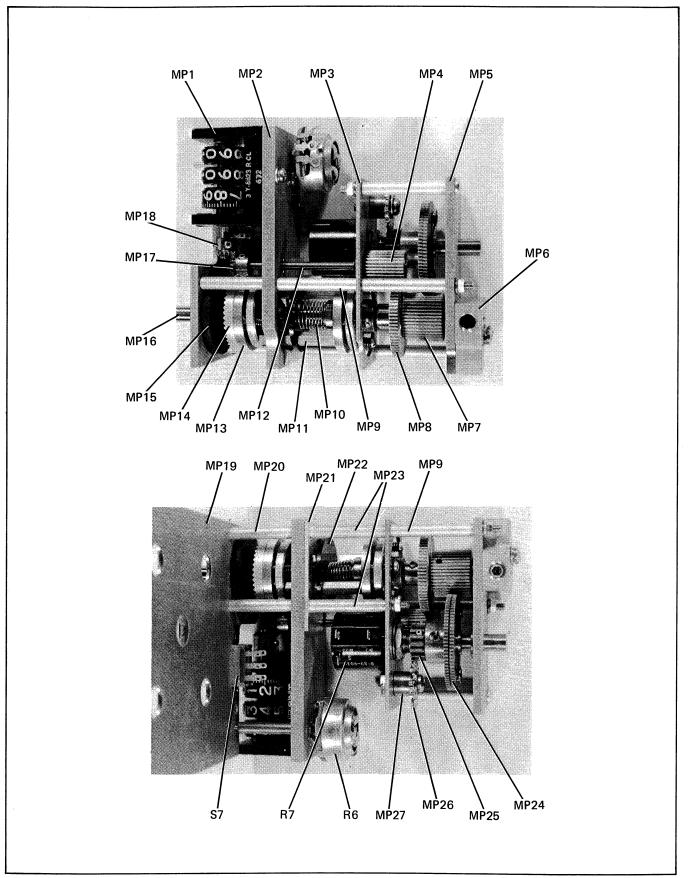


Figure 6-1. Frequency Control Component Locator.

GENERAL WAVEFORM INFORMATION

3580A Control Settings



These waveforms were made with the 3580A INPUT SENSITIVITY switch in the CAL position. The FREQUENCY and SWEEP controls are in the MANUAL mode, 300 Hz Bandwidth, and adjusted to read the 10 KHz harmonic of the CAL signal. The values given are those which would be observed for a full scale screen display of this $10\,\mathrm{KHz}$ signal. Set the dBv/LIN -dBm $600\,\Omega$ switch to dBv/LIN mode.

(For Option 002 set the dBm $900~\Omega/\text{LIN}$ - dBm $600~\Omega$ switch to dBm $900~\Omega$. Waveform Ω is of slightly less magnitude for Option 002 instruments when using the LOG mode.)



These waveforms were made with the 3580A front panel controls adjusted as follows:

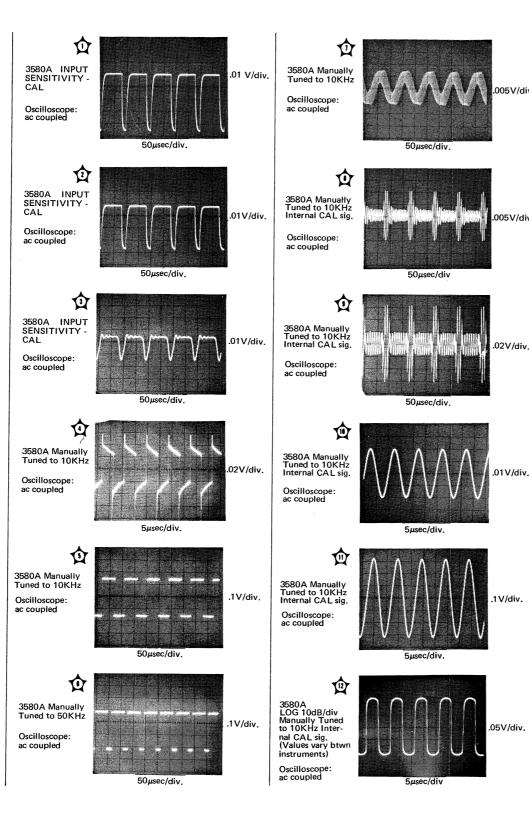
| ADAPTIVE SWEEPOFF |
|--|
| DISPLAY BLANK STORE released, |
| STORE as indicated |
| |
| AMPLITUDE MODE LOG 10 dBv/DIV |
| AMPLITUDE REF LEVELNORMAL |
| $dBv/LIN - dBm 600 \Omega \dots dBv/LIN$ |
| INPUT SENSITIVITYCAL |
| VERNIER (Amplitude) CAL |
| (Fully CW) |
| FREQUENCY 10.0 kHz |
| START-CTRCTR |
| RESOLUTION BANDWIDTH 30 Hz |
| DISPLAY SMOOTHING, |
| FREQ. SPAN/DIV |
| SWEEP TIME/DIV |
| SWEEP MODE |
| % |

Option 002: Set dBm 900 Ω /LIN - dBm 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL.

The front panel ZERO CAL was adjusted to give a display of the 10 KHz CAL signal in the center of the screen, It is easiest to set this adjustment in the MANUAL mode, and then switch to the REPetitive sweep mode to measure the waveforms.

Oscilloscope Settings

All waveforms were recorded using a 10:1 divider probe on the oscilloscope inputs. The vertical amplitude sensitivity is the actual amplifier setting and does not include the X10 multiplier introduced by the probe. All dual traces were made with the oscilloscope in the chopped mode and triggered by Channel B.





Chan B

.2 V/div

~ov

Chan A

.1 V/div ←0V

> Chan A .2V/div

Chan A

Chan B

Chan A 2V/div

Chan B 2V/div

Chan A 2V/div

Chan B 2V/div

⊢0V

⊬0V

⊬0V

-0V

–0V

5msec/div

2msec/div

2msec/div

3580A LINEAR mode, Manually Tuned to 10KHz

Internal CAL sig. (Values vary btwo

3580A DISPLAY-STORE (released)

Resolution Bandwidth 300 Hz

3580A DISPLAY-STORE (released)

Oscilloscope: dc coupled

3580A DISPLAY-STORE (depressed)

3580A DISPLAY-STORE (released)

3580A DISPLAY-STORE

Oscilloscope:

dc coupled

⑩

Oscilloscope:

dc coupled

Oscilloscope:

dc coupled

Oscilloscope dc coupled

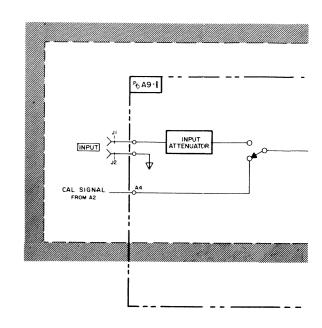
⑩

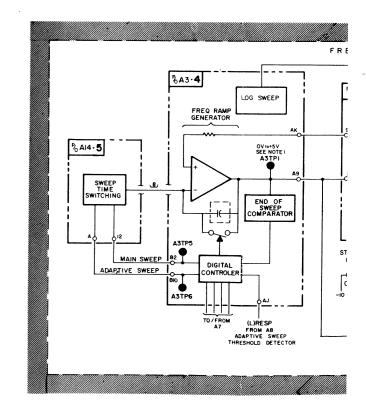
15

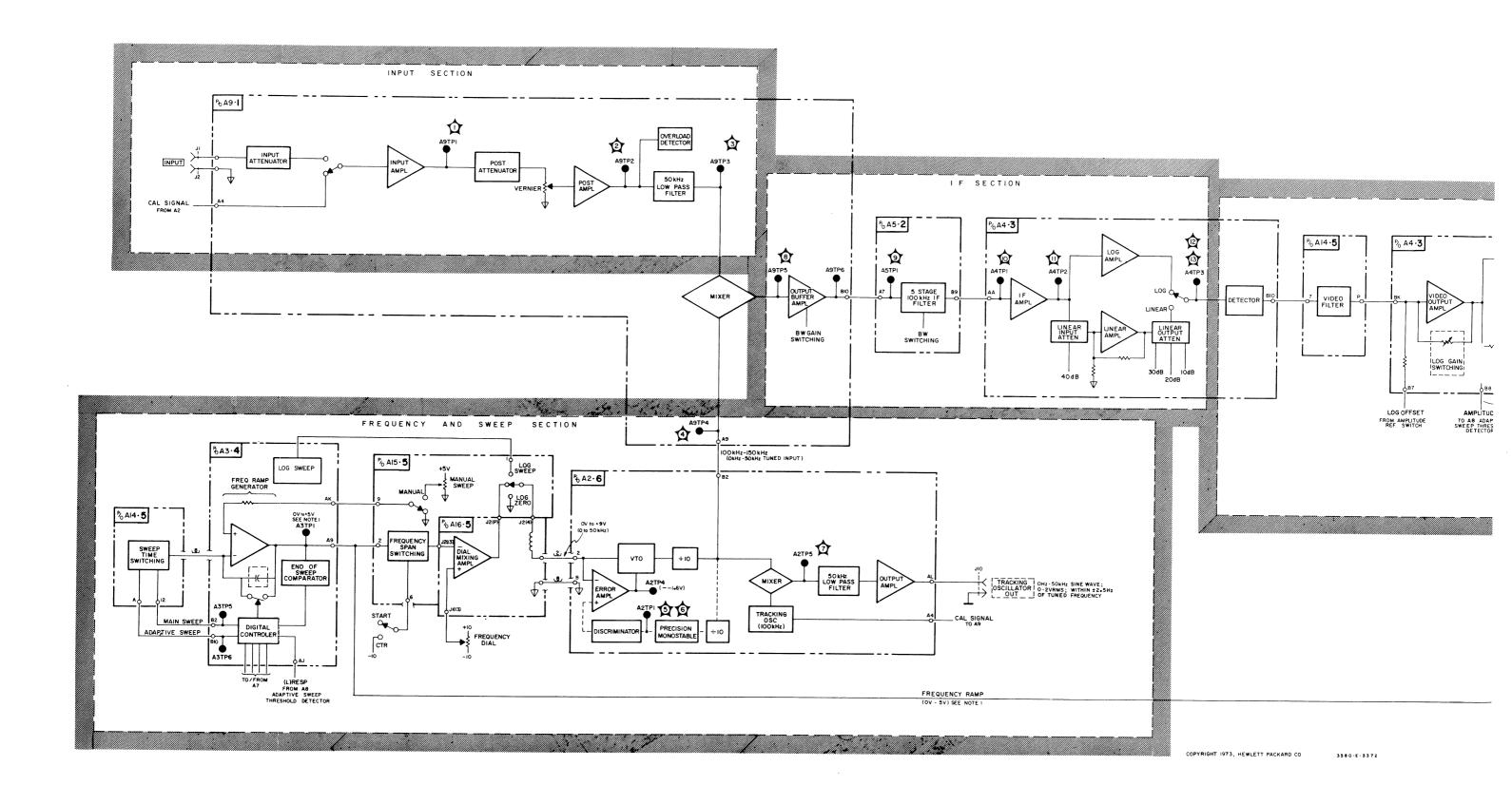
ᅇ

Oscilloscope:

ac coupled







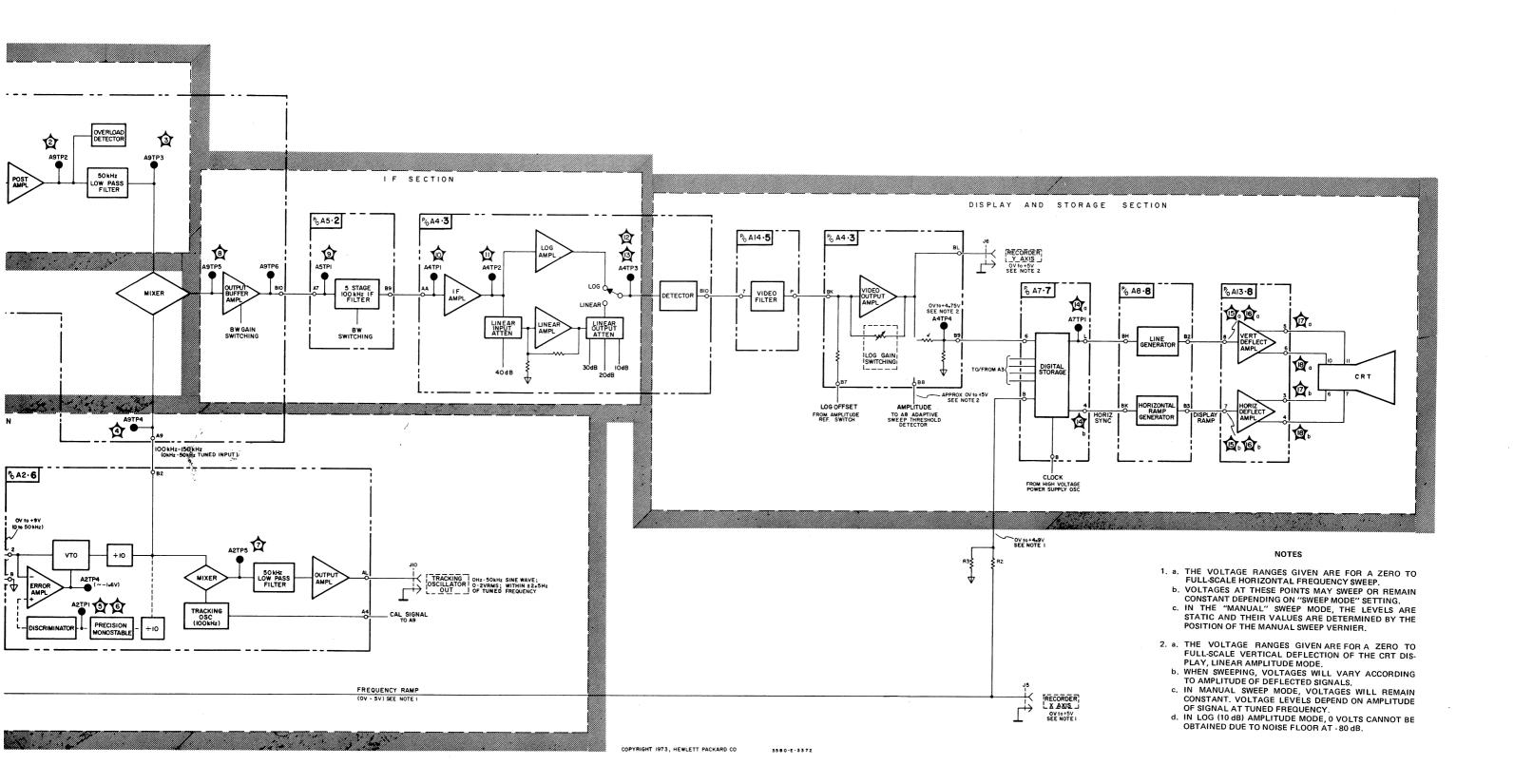


Figure 7-2. Functional Block Diagram. 8-41/8-42

Table 1. Input Circuit Amplitude Levels and Gains for Full Scale Sine Wave Inputs. (Amplitude Ref Level - Normal). Note: All voltages in RMS.

| INPUT SENSITIVITY | + 30 dB/20 V | + 20 dB/10 V | + 10 dB/2 V | 0 dB/1 V | - 10 dB/0.2 V | 20 dB/0.1 V | 30 dB/20 mV | 40 dB/10 mV | 50 dB/2 mV | - 60 dB/1 mV | - 70 dB/0.2 mV |
|--|--|---|--|--|--|---|--|--|--|---|---|
| Maximum Input Log (dBm 600 Ω) Log (dBV) Linear Log (dBm 900 Ω) * | 24.5 V 31.6 V 20.0 V 30.0 V | 7.75 V 10.00 V 10.00 V 9.49 V | 2.45 V 3.16 V 2.00 V 3.00 V | 1.000 V 1.000 V | .245 V .316 V .200 V .300 V | 77.5 mV .1 V .1 V 94.9 mV | 24.5 mV 31.6 mV 20.0 mV 30.0 mV | 7.75 mV 10.00 mV 10.00 mV 9.49 mV | 2.45 mV 3.16 mV 2.00 mV 3.00 mV | .775 mV 1.000 mV 1.000 mV .949 mV | .245 mV .316 mV .200 mV |
| Input Attenuator Input Attenuator Out (Gate of A9Q1) Log (dBm 600 Ω) Log (dBV) Linear Log (dBm 900 Ω)* | - 60. dB 24.5 mV 31.6 mV 20.0 mV 30.0 mV | 40. dB 77.5 mV .1 V .1 V 94.9 mV | 40. dB 24.5 mV 31.6 mV 20.0 mV 30.0 mV | 77.5 mV .1 V | 20. dB 24.5 mV 31.6 mV 20.0 mV 30.0 mV | 0 dB 77.5 mV .1 V .1 V 94.9 mV | 0 dB 24.5 mV 31.6 mV 20.0 mV 30.0 mV | 0 dB 77.5 mV 10.0 mV 10.0 mV 9.49 mV | 0 dB 2.45 mV 3.16 mV 2.00 mV 3.00 mV | 0 dB .775 mV 1.000 mV 1.000 mV .949 mV | 0 dB .245 mV .316 mV .200 mV .300 mV |
| Input Amp Gain Input Amp Out (A9TP1) Log (dBm 600 Ω) Log (dBV) Linear Log (dBm 900 Ω)* | 1.8 dB 30.1 mV 38.9 mV 24.6 mV 36.9 mV | 1.8 dB 95.6 mV .123 V .123 V .117 V | 1.8 dB 30.1 mV 38.9 mV 24.6 mV 36.9 mV | 95.6 mV .123 V .123 V | 1.8 dB 30.1 mV 38.9 mV 24.6 mV 36.9 mV | 1.8 dB 95.6 mV .123 V .123 V .117 V | 1.8 dB 30.1 mV 38.9 mV 24.6 mV 36.9 mV | 1.8 dB 9.56 mV 12.3 mV 12.3 mV 11.7 mV | 1.8 dB 3.01 mV 3.89 mV 2.46 mV 3.69 mV | 21.8 dB 9.56 mV 12.3 mV 12.3 mV 11.7 mV | 21.8 dB 3.01 mV 3.89 mV 2.46 mV 3.69 mV |
| Post Attenuation Log (dBm 600 Ω) Log (dBV) Linear Log (dBm 900 Ω) • | 2.8 dB 5.0 dB 5.2 dB 4.6 dB | - 12.8 dB 15.0 dB 15.0 dB 14.6 dB | - 2.8 dB - 5.0 dB - 5.2 dB - 4.6 dB | 15.0 dB 15.0 dB | 2.8 dB - 5.0 dB 5.2 dB - 4.6 dB | 12.8 dB - 15.0 dB 15.0 dB 14.6 dB | - 2.8 dB - 5.0 dB - 5.2 dB 4.6 dB | 12.8 dB 15.0 dB 15.0 dB 14.6 dB | 2.8 dB 5.0 dB 5.2 dB 4.6 dB | - 12.8 dB - 15.0 dB - 15.0 dB - 14.6 dB | - 2.8 dB - 5.0 dB - 5.2 dB - 4.6 dB |
| Post Attenuator Out (Base A9Q6) Log Linear | 21.8 mV 13.5 mV | 21.8 mV 21.8 mV | 21.8 mV 13.5 mV | 21.8 mV 21.8 mV | 21.8 mV 13.5 mV | 21.8 mV 21.8 mV | 21.8 mV 13.5 mV | 2.18 mV 2.18 mV | 2.18 mV 1.35 mV | 2,18 mV 2,18 mV | 2.18 mV 1.35 mV |
| Post Amp Gain Post Amp Out (A9TP2) Log Linear | 13.2 dB 100. mV 61.9 mV | 13.2 dB 100. mV 100. mV | 13.2 dB 100. mV 61.9 mV | | 13.2 dB 100. mV 61.9 mV | 13.2 dB 100. mV 100. mV | 13.2 dB 100. mV 61.9 mV | 33.2 dB 100. mV 100. mV | 33.2 dB 100. mV 61.9 mV | 33.2 dB 100. mV 100. mV | 33.2 dB 100. mV 61.9 mV |
| Low Pass Filter Out (A9TP3) Log Linear | 50. mV 30.8 mV | 50. mV 50. niV | 50. mV 30.8 mV | 50. mV 50. mV | 50. mV 30.8 mV | 50. mV 50. mV | 50. mV 30.8 mV | 50. mV 50. mV | 50. mV 30.8 mV | 50. mV 50. mV | 50. mV 30.8 mV |
| Total Gain Log (dBm 600 Ω) Log (dBV) Linear Log (dBm 900 Ω) • | ·· 47.8 dB ·· 50. dB ·· 50.2 dB ·· 49.6 dB | - 37.8 dB - 40. dB - 40. dB 39.6 dB | - 27.8 dB 30. dB 30.2 dB 29.6 dB | 17.8 dB 20. dB 20. dB 19.6 dB | 7.8 dB - 10. dB - 10.2 dB 9.6 dB | + 2.2 dB 0 dB 0 dB + 0.4 dB | + 12.2 dB + 10. dB + 9.8 dB + 10.4 dB | + 22.2 dB + 20. dB + 20. dB + 20.4 dB | + 32.2 dB + 30. dB + 29.8 dB + 30.4 dB | + 42.2 dB + 40. dB + 40. dB + 40.4 dB | + 52.2 dB + 50. dB + 49.8 dB + 50.4 dB |

*Option 002 only.

Table 2. Approximate IF Level Changes with Bandwidth.
(Full Scale Sine Wave Input, LOG Mode, Manually Tuned to Input Frequency)

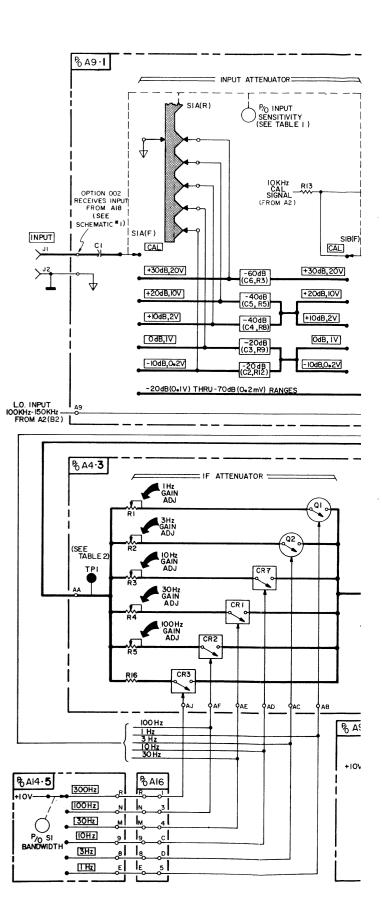
| Bandwidth | I.F. Input A9TP6, A5TP1 | I.F. Output A5 pin B9, A4TP1 |
|-----------|----------------------------|---------------------------------|
| 300 Hz | 640 mV p-p | 420 mV p-p |
| 100 Hz | 640 mV p-p | 420 mV p-p |
| 30 Hz | 325 mV p-p | 180 mV p-p |
| 10 Hz | 325 mV p-p | 140 mV p-p |
| 3 Hz | 110 mV p-p | 60 mV p-p |
| 1 Hz | 110 mV p-p | 50 mV p-p |

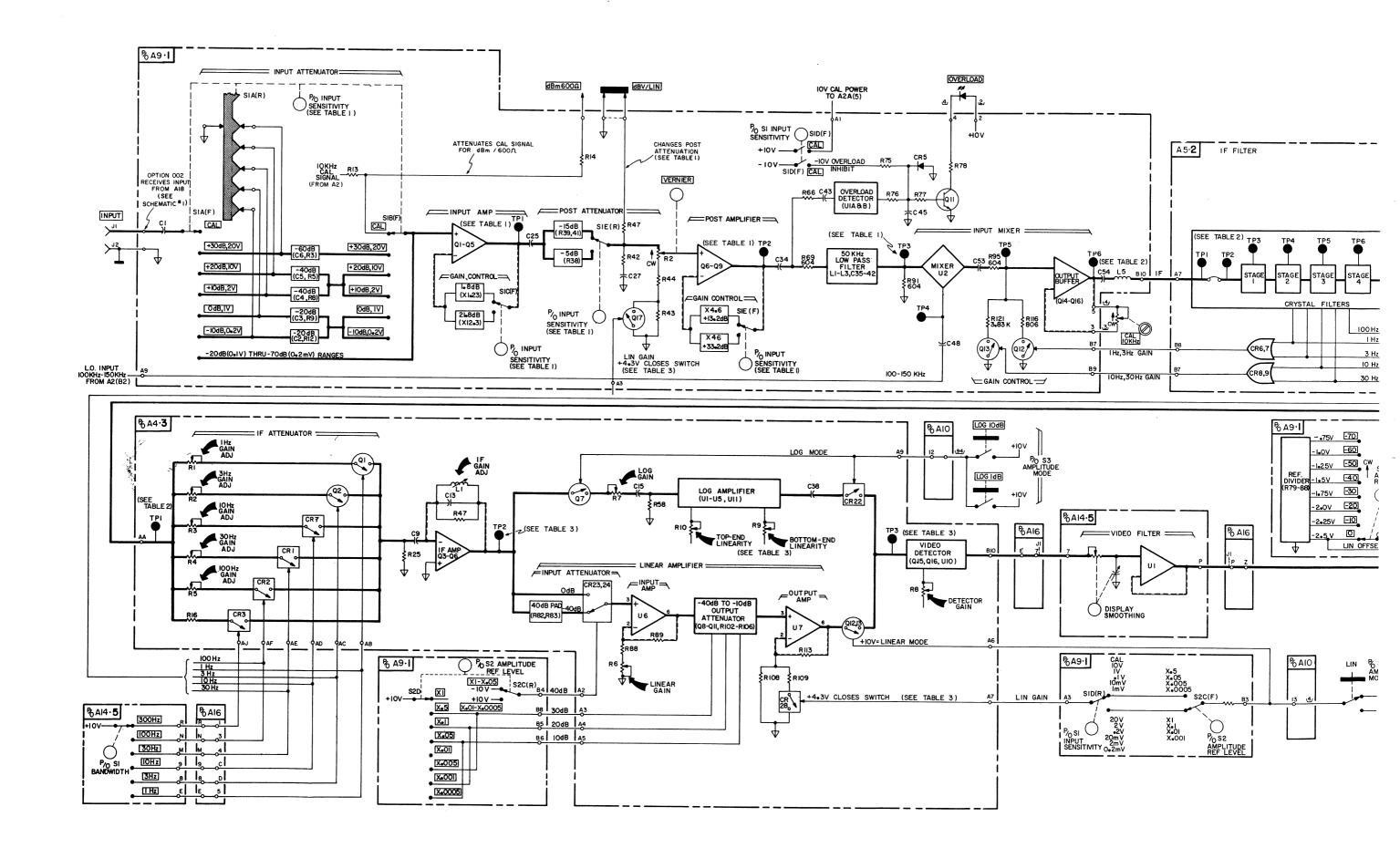
| Full Scale Input | MAX INPUT (INPUT SENSITIVITY Switch) | AMPLITUDE REF LEVEL | A4TP2 | Input Attenuation | Input Amp Gain | A4U6 pin 6 | Output Attenuation | A4 pin 7, A9 pin 3 (LIN GAIN) | Output Amp Gain | A4TP3 (Appr. Value) |
|---|--|--|--|---|--|--|--|--|--|--|
| 1 V rms .2 V rms .1 V rms .02 V rms .01 V rms 2 mV p-p 1 mV p-p .2 mV p-p | 1 V | X1 X.5 X.1 X.05 X.01 X.005 X.001 X.0005 | 6.7 V p-p 1.32 V p-p .67 V p-p .132 V p-p .067 V p-p | - 40 dB - 40 dB - 40 dB - 40 dB - 40 dB - 0 dB - 0 dB - 0 dB | + 40 dB + 40 dB | 6.7 V p-p 1.32 V p-p .67 V p-p .132 V p-p 6.7 V p-p 1.32 V p-p .67 V p-p .132 V p-p | - 40 dB - 30 dB - 20 dB - 10 dB - 40 dB - 30 dB - 20 dB - 10 dB | 28 V dc + 4.3 V dc 28 V dc + 4.3 V dc 28 V dc + 4.3 V dc 28 V dc + 4.3 V dc + 4.3 V dc | + 34 dB (X50) + 38.2 dB (X81) + 34 dB (X50) + 38.2 dB (X81) + 34 dB (X50) + 38.2 dB (X81) + 34 dB (X50) + 38.2 dB (X81) | 3.4 V p-p 3.4 V p-p |

Table 3. Approximate Full Scale Levels in Linear Amplifier (LINEAR Mode, Full Scale Input, Manually Tuned to Input Frequency).

Table 4. Linear and Log Offsets.

| AMPLITUDE MODE | AMPLITUDE REF LEVEL | A4 pin B7 |
|-------------------|------------------------|-------------|
| LOG | 0 | - 2.50 V dc |
| LOG | - 10 | - 2.25 V dc |
| LOG | - 20 | - 2.00 V dc |
| LOG | - 30 | - 1.75 V dc |
| LOG | - 40 | - 1.50 V dc |
| LOG | - 50 | - 1.25 V dc |
| LOG | - 60 | - 1.00 V dc |
| LOG | - 70 | - 0.75 V dc |
| LINEAR | Any Setting | - 2.50 V dc |





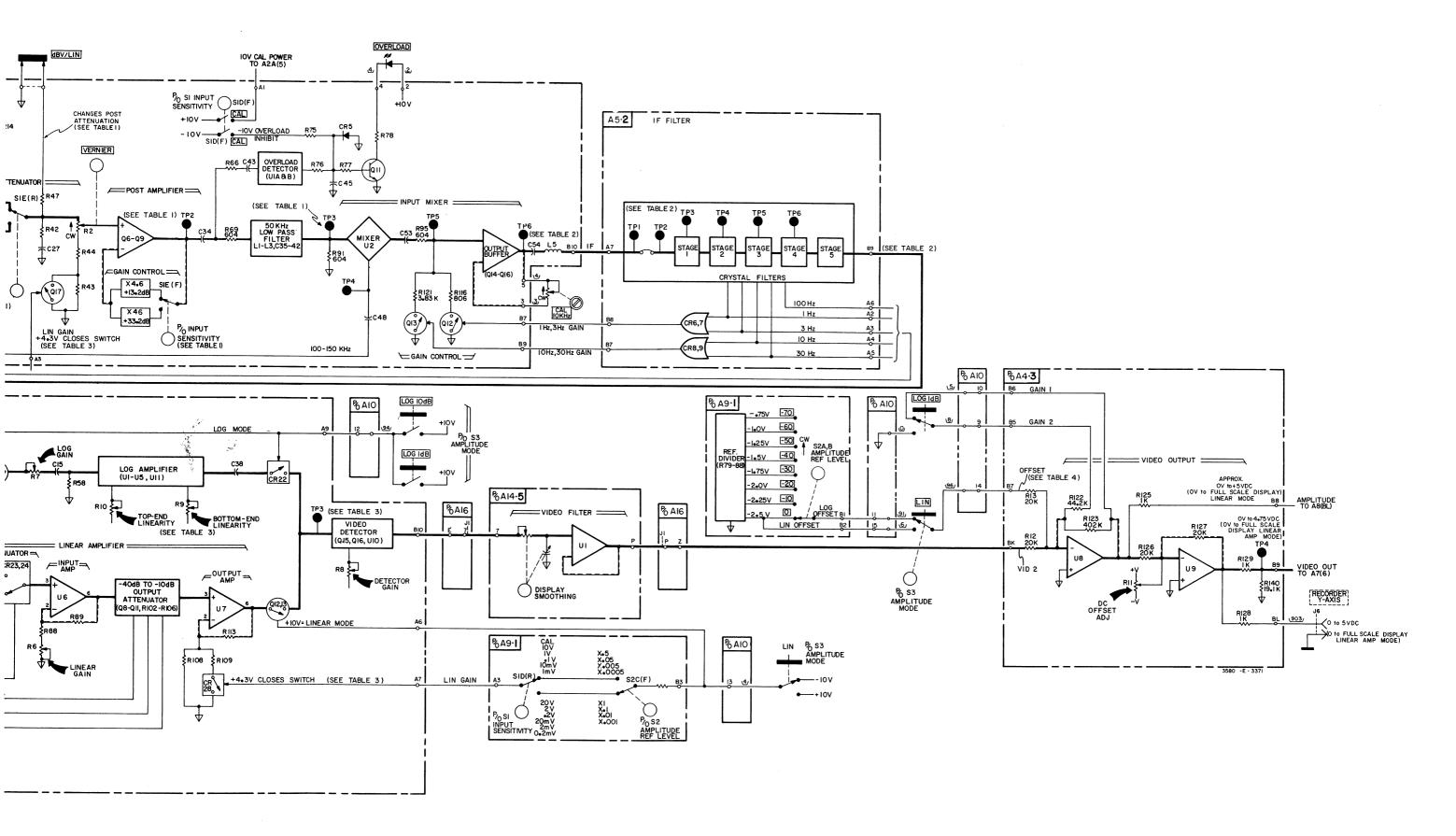
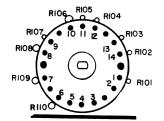


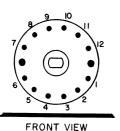
Figure 7-3. Amplitude Section Detailed Block Diagram. 8-43/8-44

Page 7-23/7-24, Figure 7-8 should appear as follows:

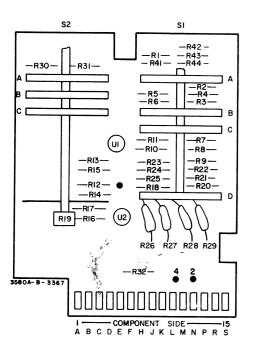
RESISTORS MOUNTED BETWEEN WAFERS B & C ON SI.



FRONT VIEW PIN POSITIONS FOR AI4SI



PIN POSITIONS FOR AI4S2 AI5SI AI5S2



NOTE 1

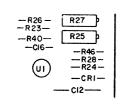
A15

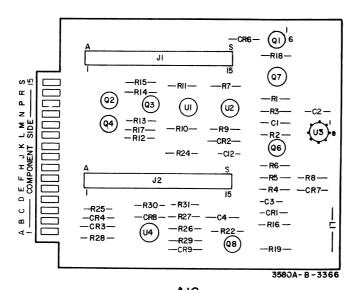
Rev B

hp Part No 03580-66515

WHENEVER REPLACING A14U1, USE hp- PART NO. 1826-0304 AND REMOVE CAPACITOR A14C16 (IF PRESENT).

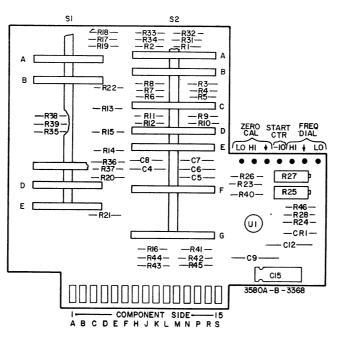




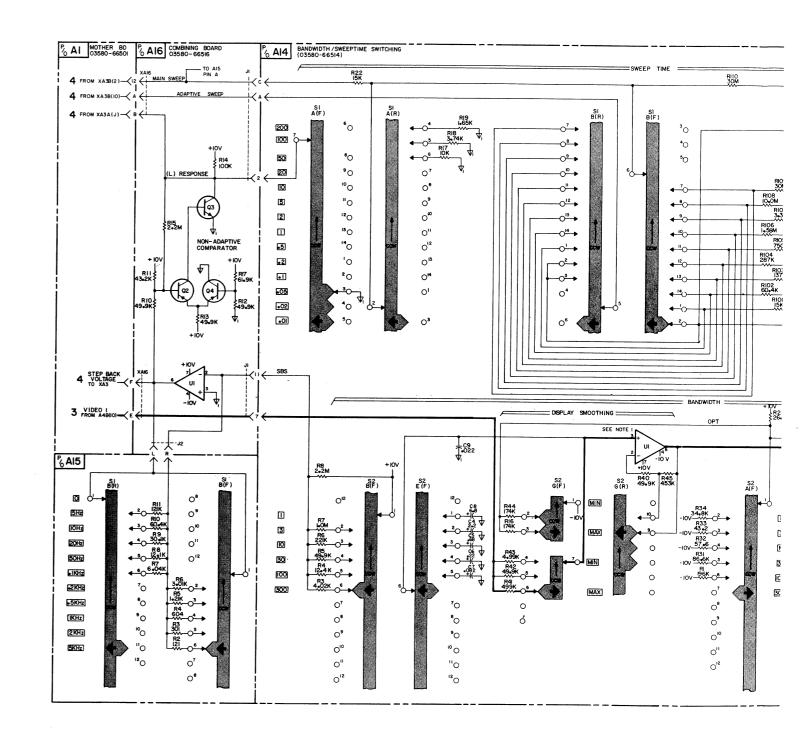


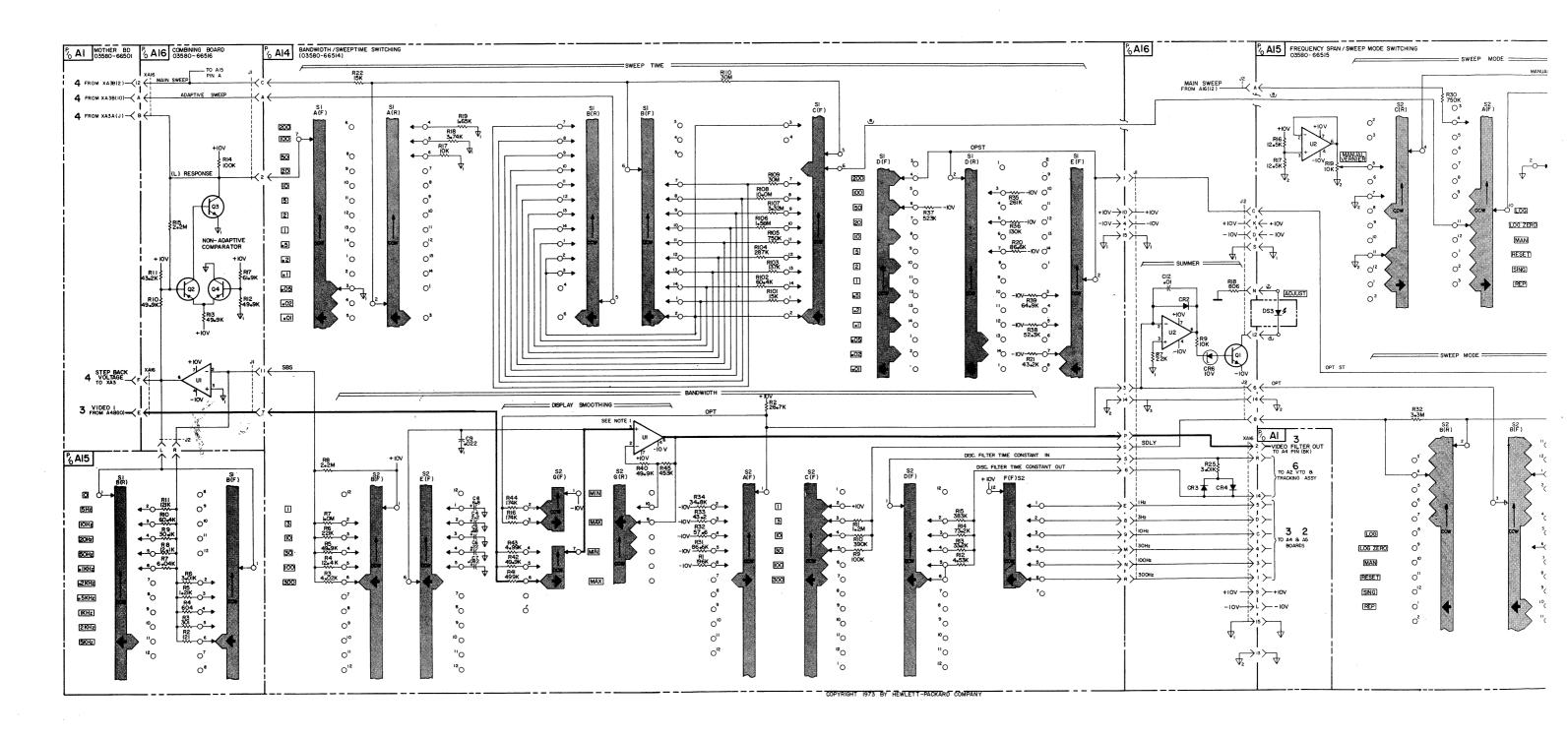
A16 hp Part No. 03580 - 66516 Rev A

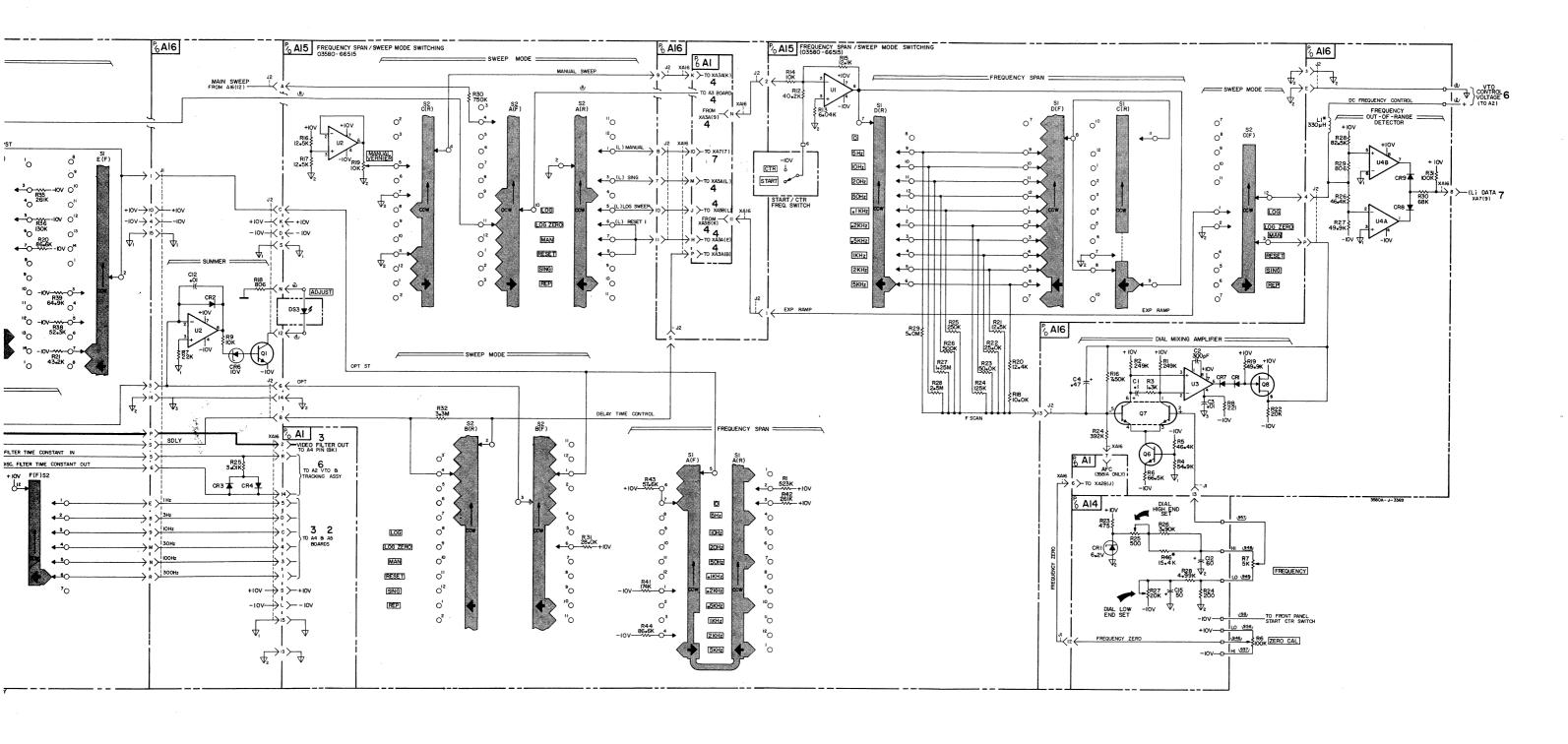
A14 plugs into J1 of A16. A15 plugs into J2 of A16.



A14 hp Part No 03580 - 66514 Rev B







CHANGE NO. Δ 17:

Applies to instruments having a serial number 2030A04745 or lower.

Page 6-18/6-19, Table 6-3

Change the following parts:

A7R13, A7R16, A7R61, A7R64, A7R127, and A7R129 to P/N 0698-3445, Resistor 348 Ω 1%

Change the following parts:

A7R14, A7R17, A7R62, A7R65, A7R73, and A7R128 to P/N 0811-1794, Resistor 99.25k .1%

Change A7R74 to P/N 0698-4456, Resistor 549 Ω 1%

CHANGE NO. Δ 18:

Applies to instruments having a serial number 2030A04461 and greater.

Page 8-7, Figure 8-6

Change: A8C2 to 470 pF A8C23 to .047 pF Add: A8C34 (.1 μ F) in parallel with A8C32.

CHANGE NO. Δ 19:

Applies to instruments having a serial number 2030A05375 and lower.

To increase the performance of the Rev. B combinning board A16 (03580-66536), A16C31 and A16C32 should be removed and A16C45 (P/N 0160-4571) should be added.

CHANGE NO. Δ 20:

Applies to instruments having serial number 2030A04796 and lower. These instruments contained Rev. A of the A16 combinning board (P/N 03580-66536). The full schematic and parts location are shown on page 8-xx.

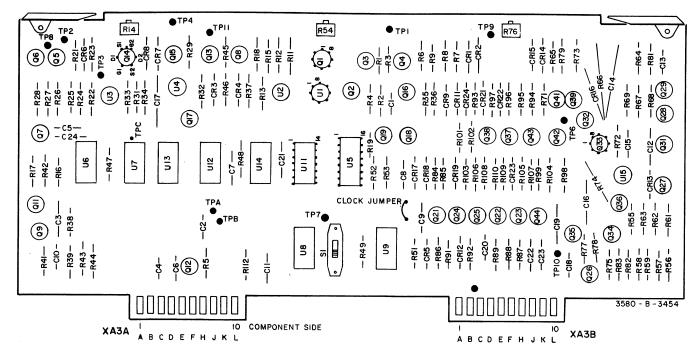
CHANGE NO. Δ 21:

Applies to instruments having a serial number 2030A06029 and lower.

These instruments contained Rev. A of the A13 Deflection amplifier (P/N 03580-66513) and no A40 Rear Panel Adjustment board. The following figure and parts list are for the A13 Rev. A board.

Should an A13 Rev. B board be used to replace an A13 Rev. A board then insert the following components into the Rev. B circuit board. The locations for these components are silkscreened on the Rev. B board.

A13R37 100 ohm 0757-0401 A13R38 100 ohm 0757-0401 A13R39 10k ohm 2100-3210

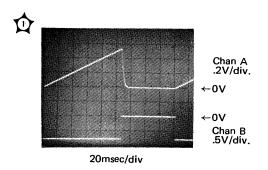


A3 -hp- Part No. 03580-66503 Rev. B

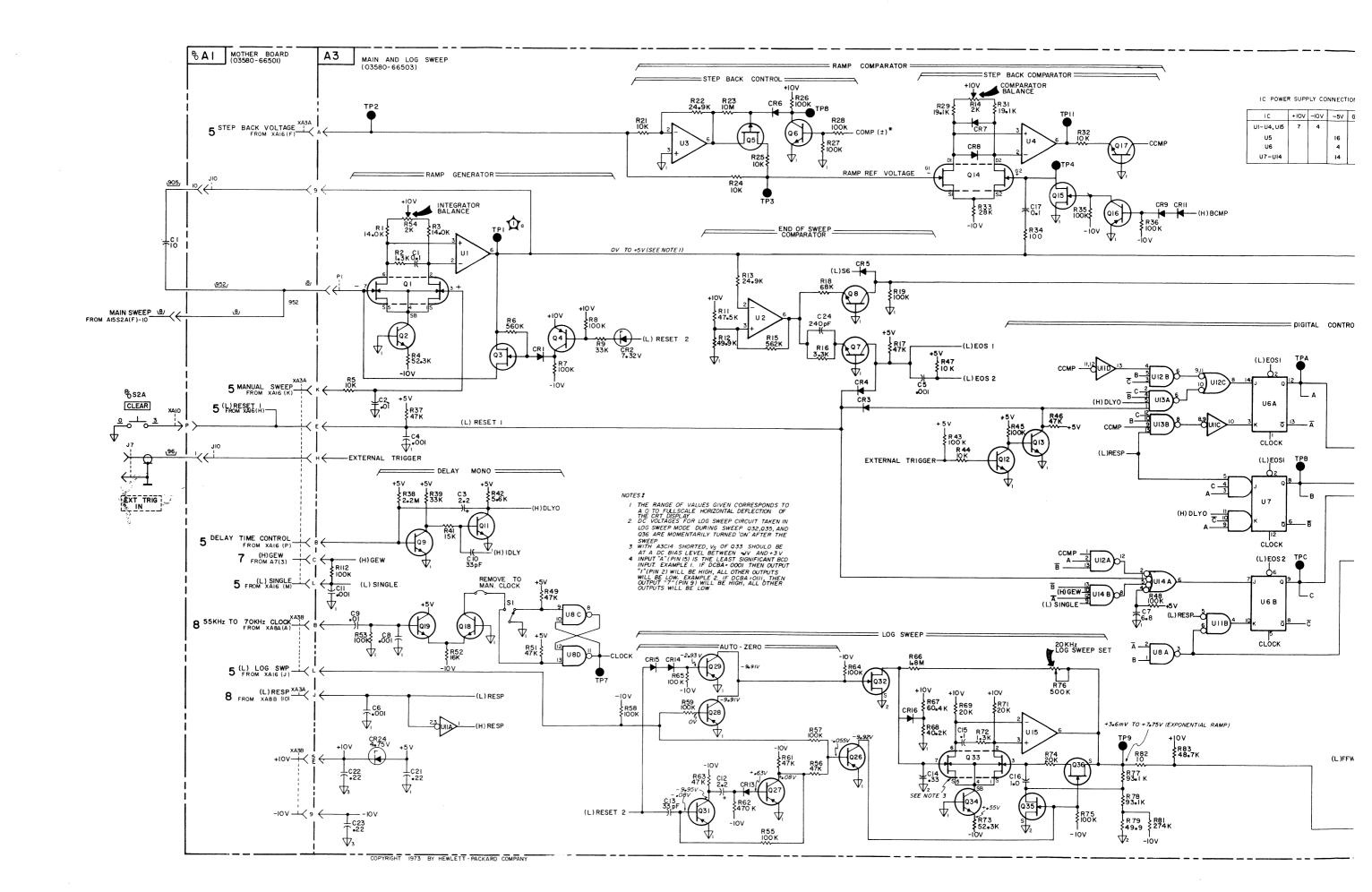
| 3580A | |
|---------------------------------|----|
| ADAPTIVE SWEEPOF | F |
| DISPLAY All pushbuttons release | d |
| FREQUENCY 00.0 kH | lz |
| START - CTRSTAR | |
| DISPLAY SMOOTHINGMII | Ν |
| RESOLUTION BANDWIDTH300 H | z |
| FREQ. SPAN/DIV 5 KH | lz |
| SWEEP TIME/DIV0.01 SE | |
| SWEEP MODERE | Ρ |
| | |

OSCILLOSCOPE DC coupled, dual trace (choppe

DC coupled, dual trace (chopped), triggered by Channel B.



A 10:1 divider probe was used on the oscilloscope input. The vertical amplitude sensitivity is the actual amplifier setting and does not reflect the X10 multiplier introduced by the probe.



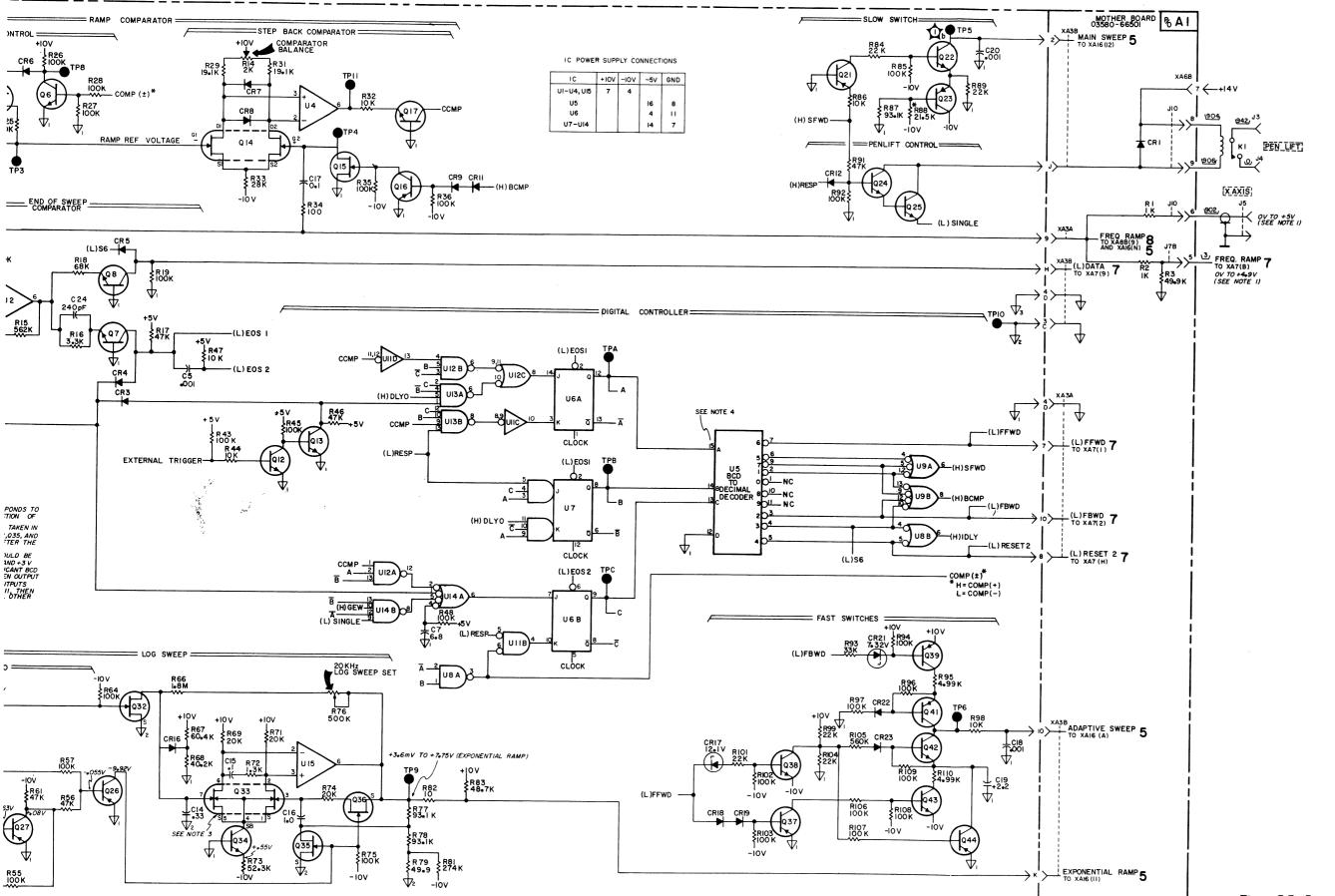
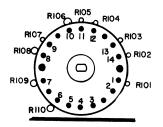
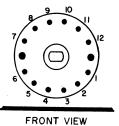


Figure 7-7. Sweep Generator (A3) Schematic and Component Location Diagram.

RESISTORS MOUNTED BETWEEN WAFERS B & C ON SI.

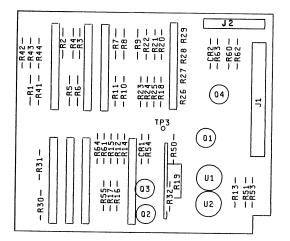


FRONT VIEW PIN POSITIONS FOR AI4SI

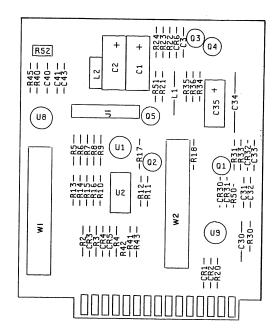


PIN POSITIONS FOR AI4S2

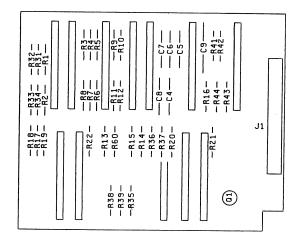
A15S1 A15S2



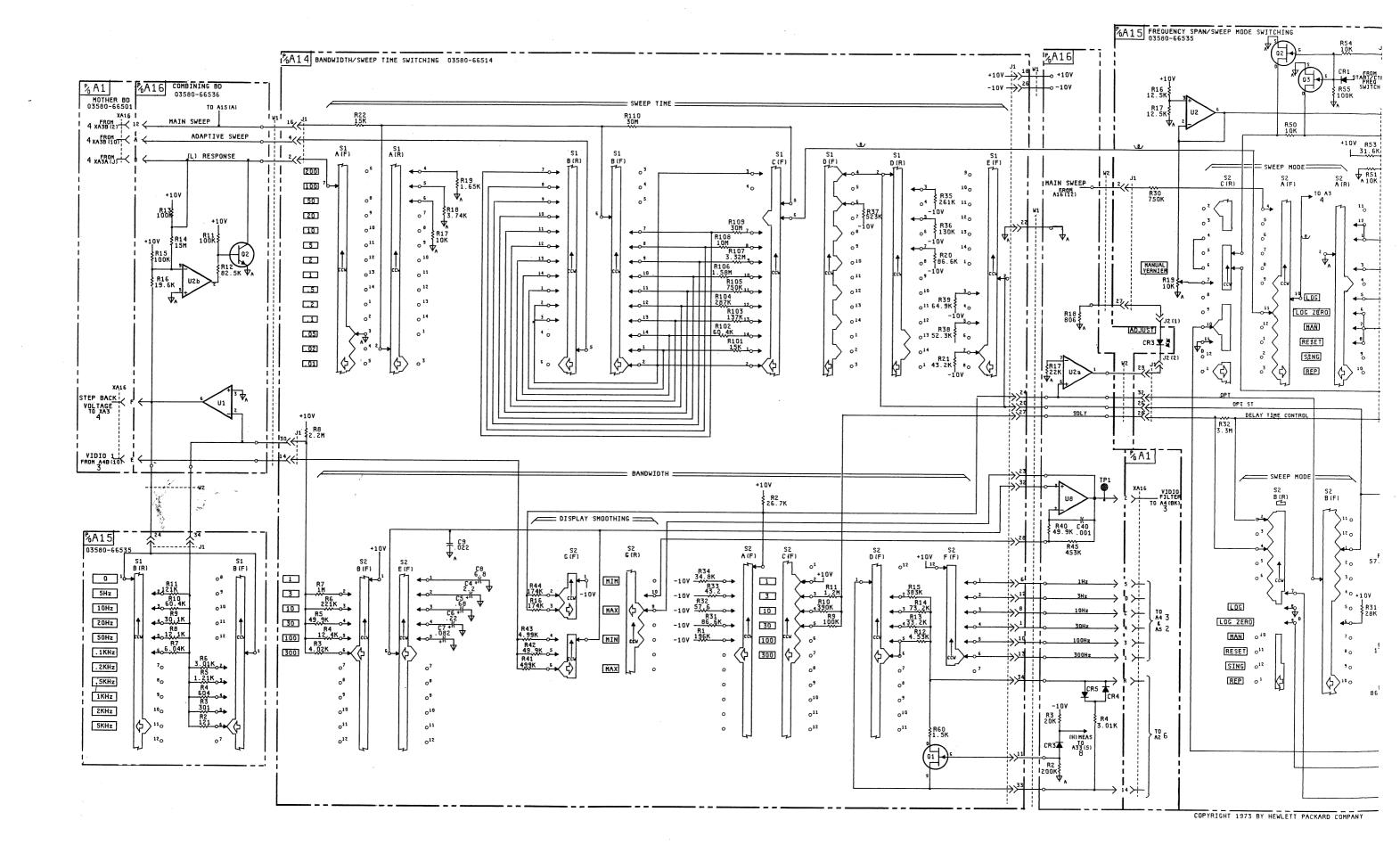
A15 H-P Part No 03580-66535



A16 H-P Part No 03580 66536 Rev A



A14 H-P Part No 3580-66534 Rev A



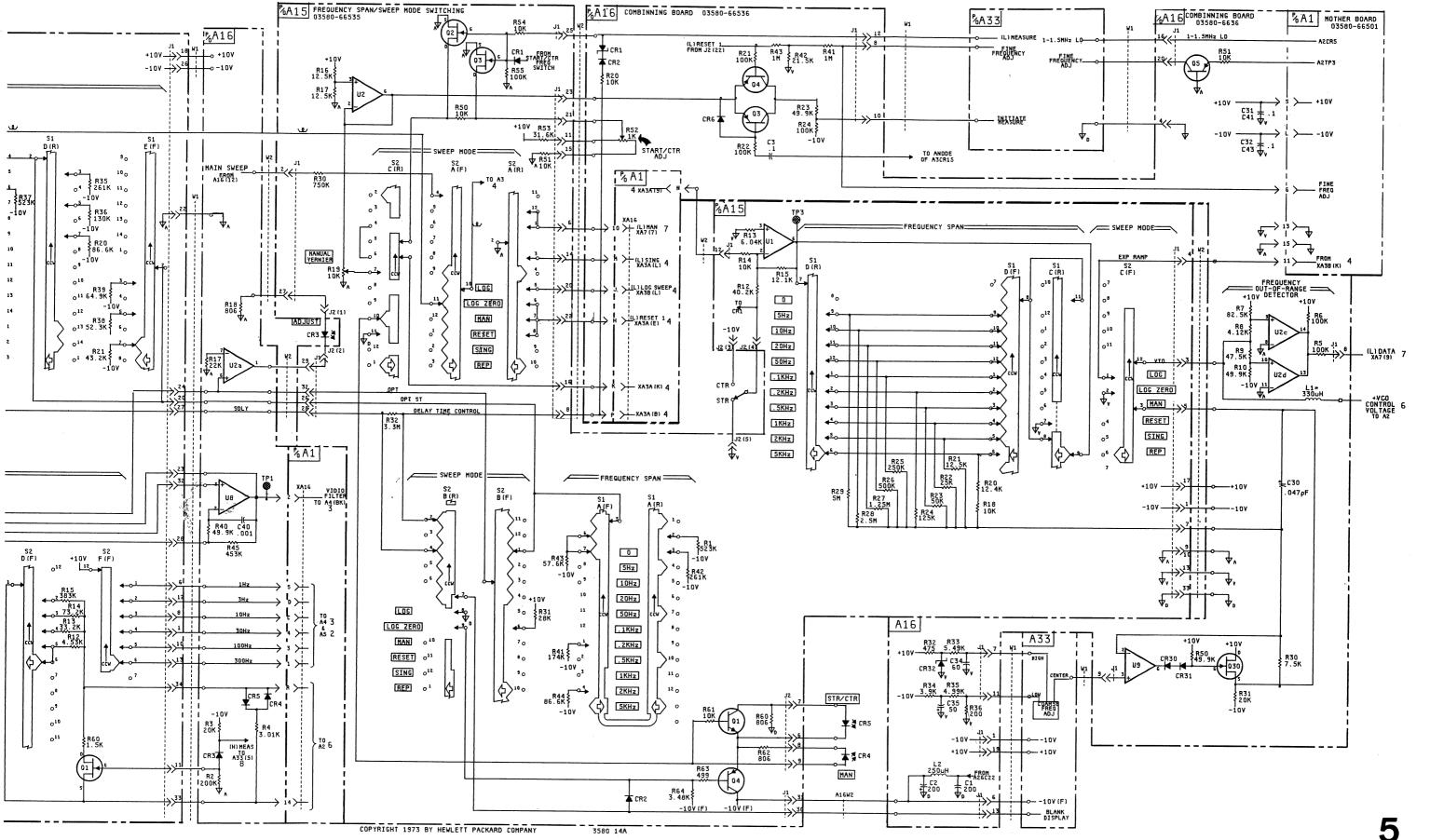
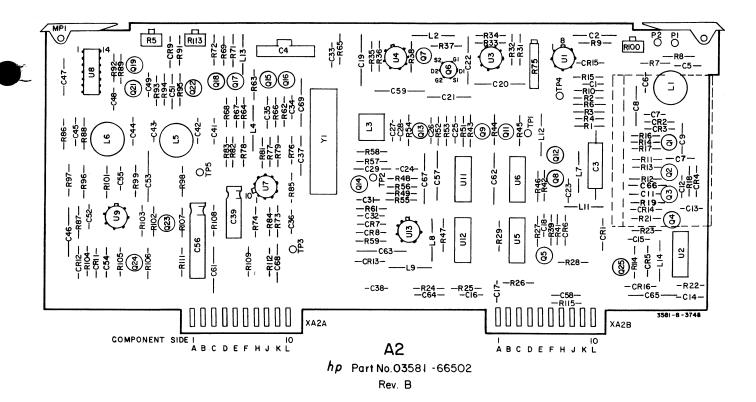


Figure 7-8. Frequency Control Circuits (A14, A15, A16) Schematic and Component Location Diagrams.



GENERAL WAVEFORM INFORMATION

3580A Control Settings

The 3580A is manually tuned to 10 kHz.

Oscilloscope Settings

All waveforms were recorded using a 10:1 divider probe on the oscilloscope inputs. The vertical amplitude sensitivity is the actual amplifier setting and does not include the X10 multiplier introduced by the probe. All dual traces were made with the oscilloscope in the chopped mode and triggered by Channel B.

Oscilloscope:
ac coupled

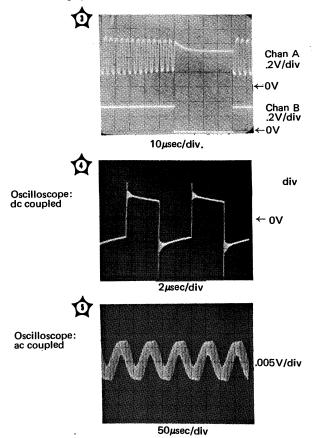
2µsec/div

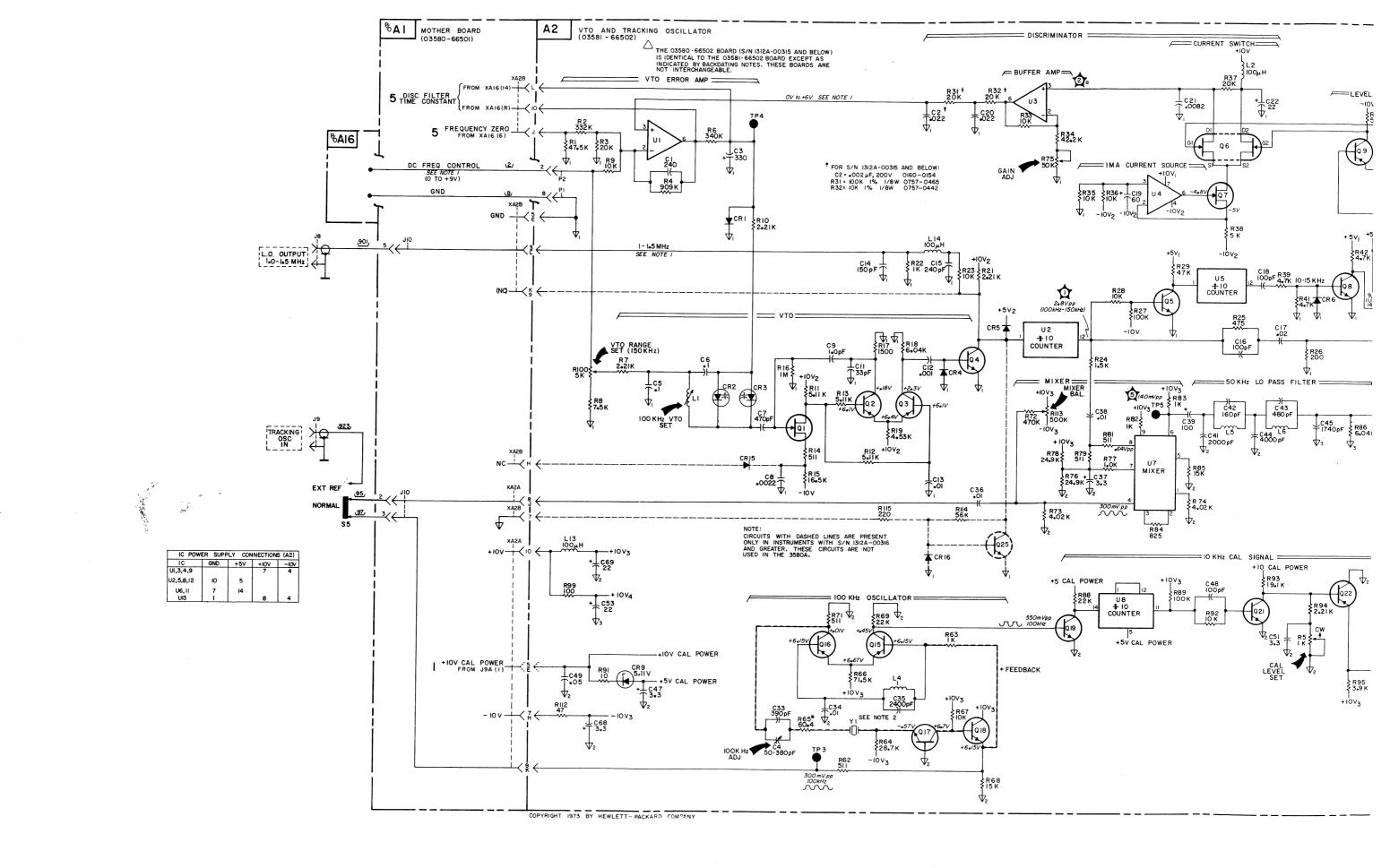
Chan A
.1V/div
←0V

Chan B
.2V/div
←0V

NOTE 1: The range of values given corresponds to a tuned frequency between 0 Hz and 50 kHz.

 ${\bf NOTE~2};$ Replacement circuit boards do not contain A2Y1 and A2R65. See Paragraph 7-25.





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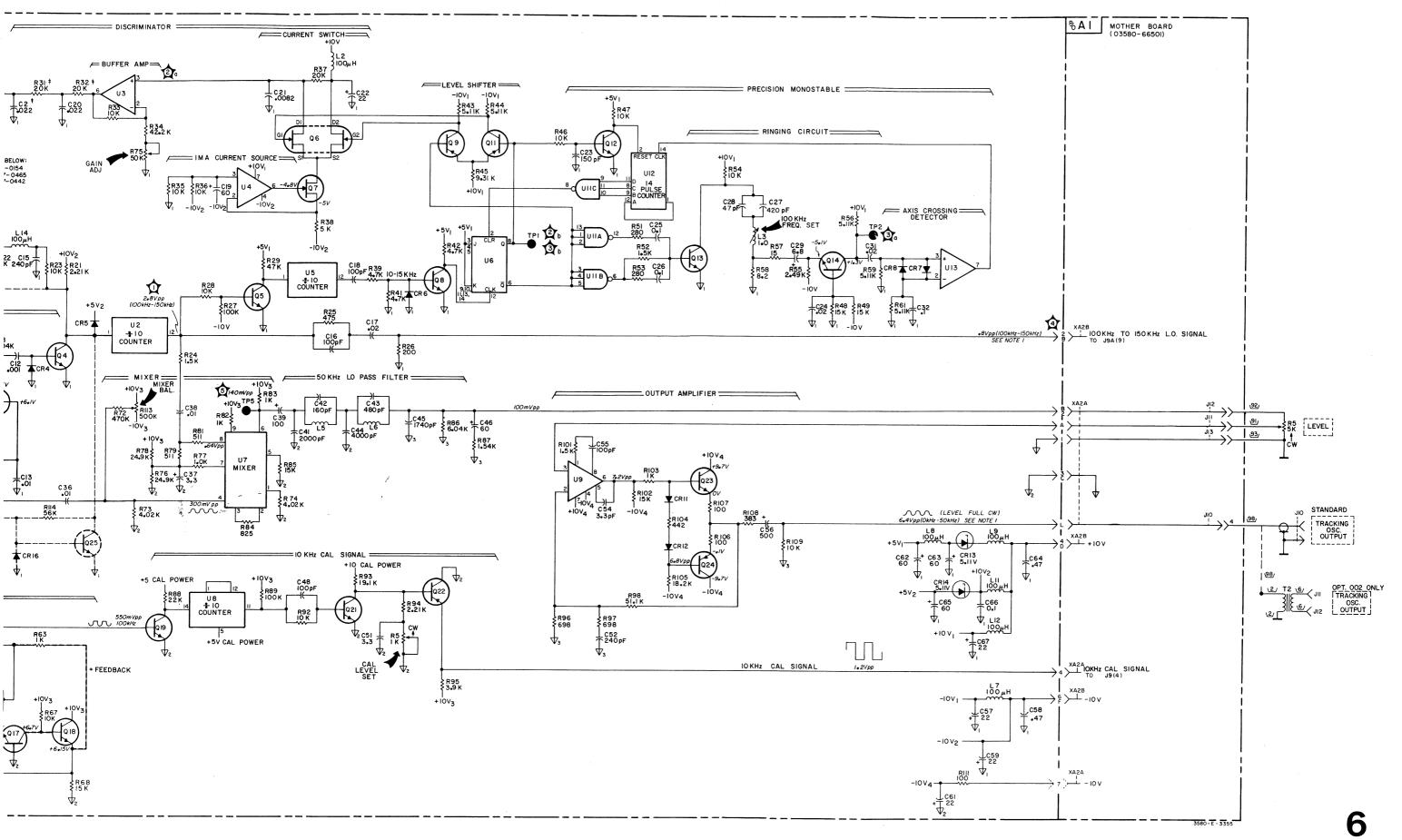
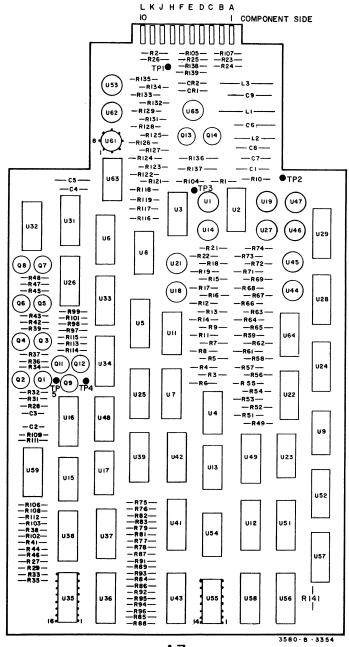


Figure 7-9. VTO and Tracking Oscillator Assembly (A2) Schematic and Component Location Diagram.



A7 hp Part No. 03580-66507 Rev. C

GENERAL WAVEFORM INFORMATION

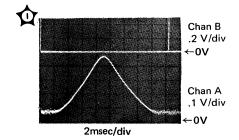
| 3580A |
|--|
| ADAPTIVE SWEEPOFF |
| DISPLAY All pushbuttons released |
| AMPLITUDE MODELOG 10 dBv/DIV |
| AMPLITUDE REF LEVELNORMAL |
| dBv/LIN - dBm 600 Ω dBv/LIN |
| INPUT SENSITIVITYCAL |
| VERNIER (Amplitude) |
| (Fully CW) |
| FREQUENCY10 kHz |
| START - CTRCTR |
| RESOLUTION BANDWIDTH 300 Hz |
| DISPLAY SMOOTHINGMIN |
| FREQ. SPAN/DIV |
| SWEEP TIME/DIV1 SEC |
| SWEEP MODEREP |
| |

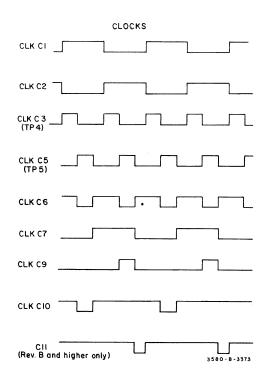
Option 002: Set dBm 900 Ω/LIN - dBm 600 Ω switch to dBm 900 Ω ; set INPUT MODE switch to UNBAL.

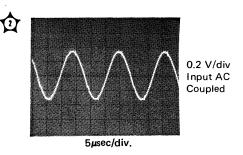
The front panel ZERO CAL and CAL 10 KHz were adjusted to give a full scale display of the 10 KHz CAL signal in the center of the screen. It is easiest to set this adjustment in the MANUAL mode, and then switch to the REPetitive sweep mode to measure the waveforms.

OSCILLOSCOPE

10:1 Probe, Dual Trace (Chopped), triggered by Channel B. The vertical amplitude sensitivity is the actual amplifier setting and does not include the X10 multiplier introduced by the probe.

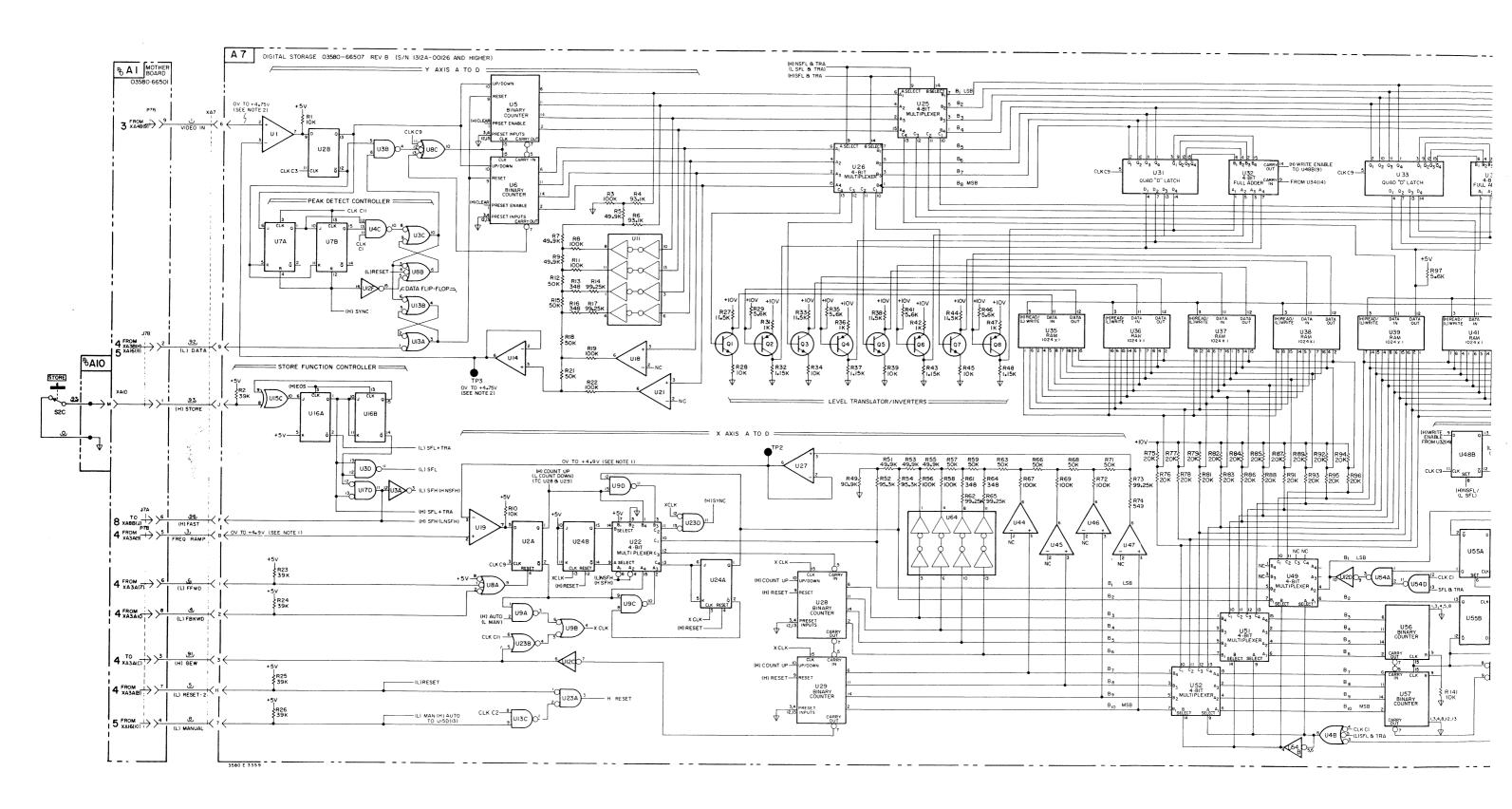






NOTE 1: The voltage range given is for a zero to full scale horizontal frequency sweep of the CRT display.

NOTE 2: The voltage range given is for a zero to full scale vertical deflection of the CRT display, LINEAR MODE. In LOG 10 dB/div Amplitude Mode, 0 volts cannot be obtained because of the - 80 dB noise floor.



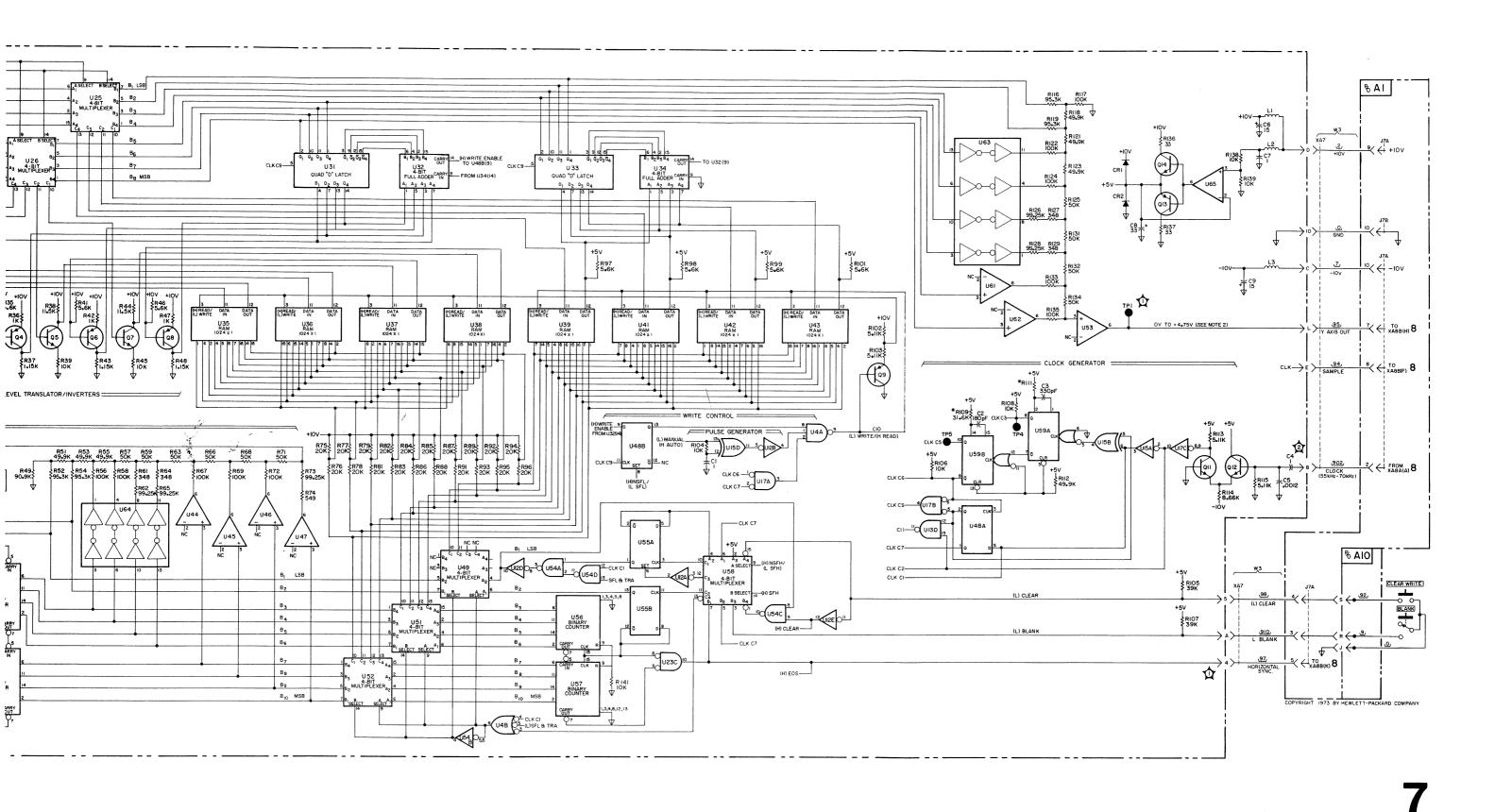
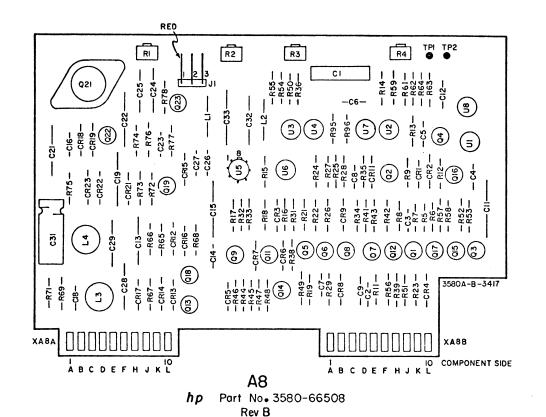


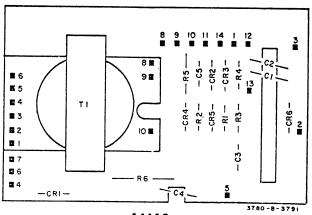
Figure 7-10. Digital Storage Assembly (A7) Schematic and Component Location Diagram.



TPI — R35— — R24— R25— — R28— — R3— — R28— — R28— — R34— — R28— — R28— — R34— — R28—
A 13 -hp- Part No. 03580-66513 Rev. A

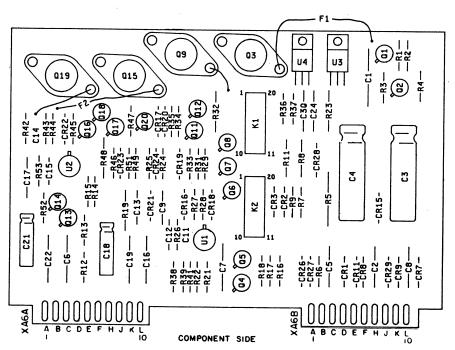
□ | s₁

RI --RII----RI2----RI6--



A11A2 -hp- Part No. 03580-66532 Rev. A

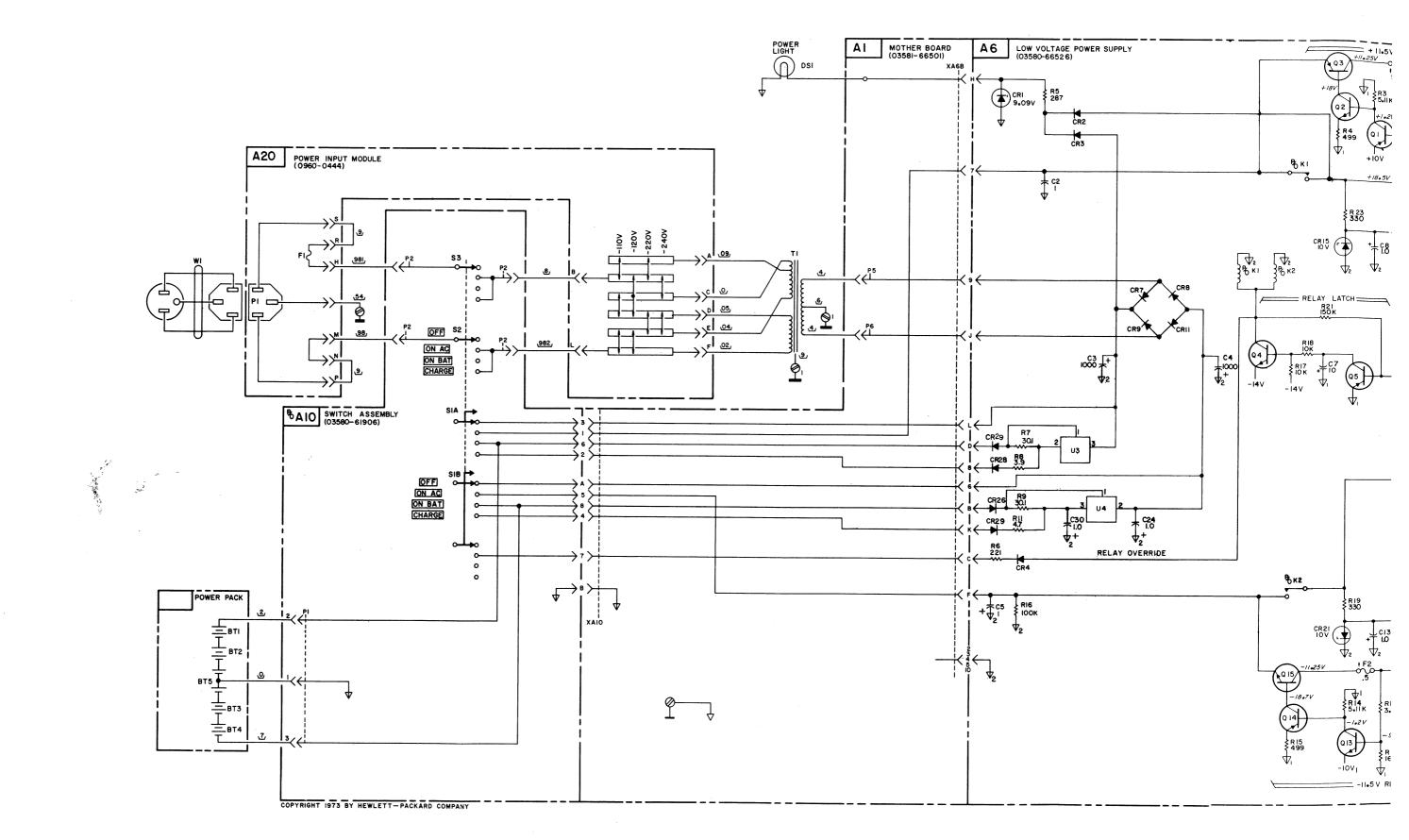
R5



A6 hp Part No. 03580-66506 Rev D

*NOTE

 \pm 11.5 V Power Nominal \pm 14 V Power Nominal Standard Instrument \approx \pm 20 V Opt 001 (Line Powered) \approx \pm 18 V Opt 001 (Battery Powered) \approx \pm 12 V



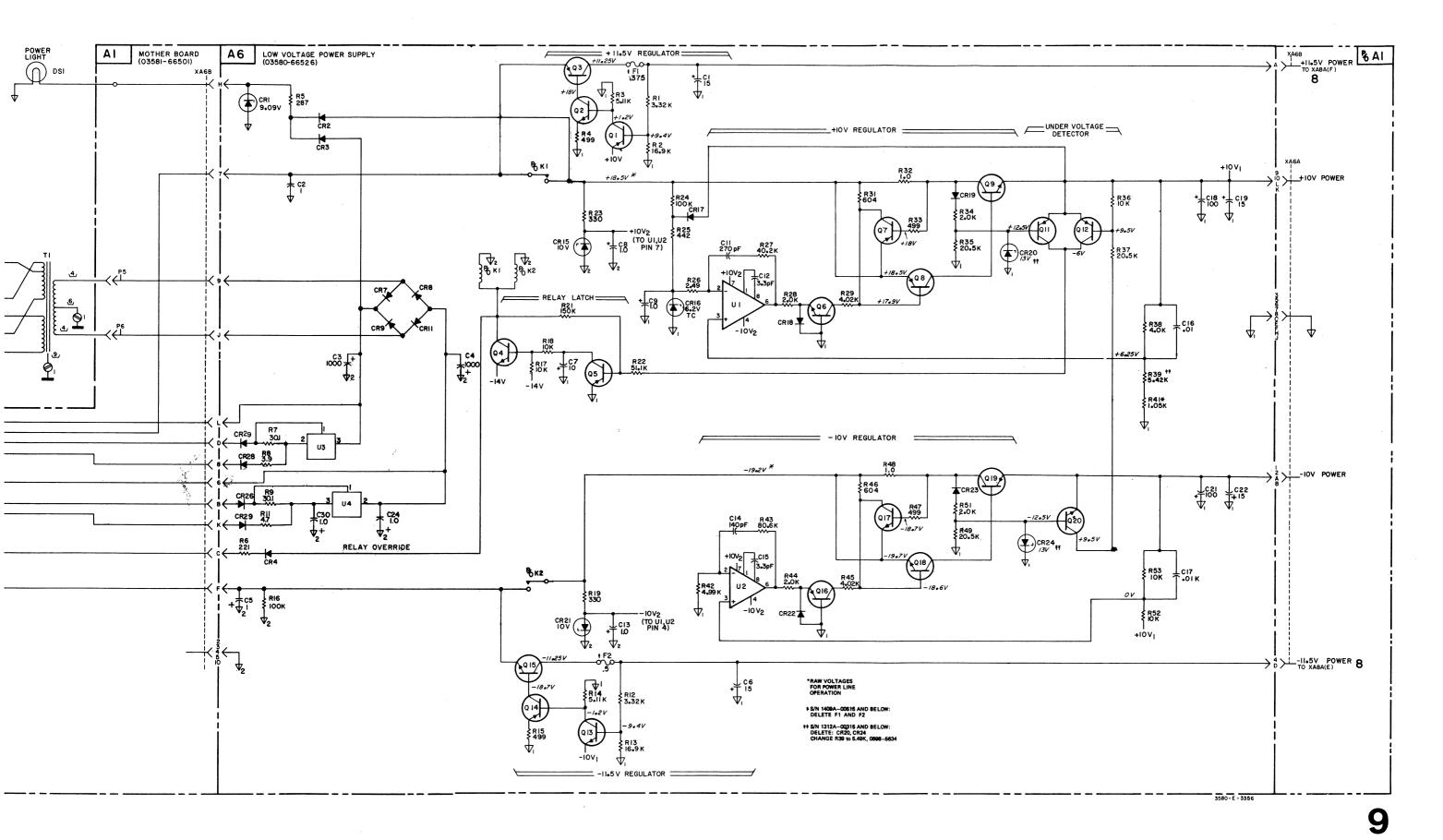


Figure 7-12. Low Voltage Power Supply (A6) Schematic and Component Location Diagram, and Power Input Module (A20) Schematic.