

# DS90UB934-Q1 12-Bit, 100-MHz FPD-Link III Deserializer for 1MP/60fps and 2MP/30fps Cameras

## 1 Features

- Qualified for Automotive Applications
- AEC-Q100 Qualified for Automotive Applications With the Following Results:
  - Device Temperature Grade 2:  $-40^{\circ}\text{C}$  to  $+105^{\circ}\text{C}$  Ambient Operating Temperature
  - Device HBM ESD Classification Level  $\pm 2$  kV
  - Device CDM ESD Classification Level C4
- Operates up to 100 MHz in 12-bit Mode to Support 1MP/60fps and 2MP/30fps Imagers as well as Satellite RADAR
- Configurable 12-bit Parallel CMOS Compatible with DS90UB913A/933 Serializers
- Adaptive Equalization Compensates for Cable Aging and Degradation Effects
- Ultra-low Latency Bi-directional Control Data Channel with Data Protection
- Cable Link Detect Diagnostics
- Supports Power-over-Coax Operation (PoC)
- ISO 10605 and IEC 61000-4-2 ESD Compliant
- Low Radiated and Conductive Emissions
- BIST (Built-In Self-Test)

## 2 Applications

- Automotive
  - Rear-View Cameras (RVC)
  - Surround View Systems (SVS)
  - Camera Monitor Systems (CMS)
  - Forward Vision Cameras (FC)
  - Driver Monitoring Systems (DMS)
  - Satellite RADAR Modules
- Security and Surveillance Cameras
- Industrial and Medical Imaging

## 3 Description

The DS90UB934-Q1 FPD-Link III deserializer, in conjunction with the DS90UB913A/933-Q1 serializers, supports the video transport needs with an ultra-high-speed forward channel and an embedded bidirectional control channel. The DS90UB934-Q1 converts the FPD-Link III stream into a parallel CMOS output interface designed to support automotive image sensors up to 12 bits at 100 MHz with resolutions including 1MP/60fps and 2MP/30fps.

The DS90UB933/934 chipset is fully AEC-Q100 qualified and designed to receive data across either 50- $\Omega$  single-ended coaxial or 100- $\Omega$  shielded-twisted pair (STP) cable assemblies. The DS90UB934-Q1 uses an advanced adaptive equalizer to allow support of various cable lengths and types with no additional programming required.

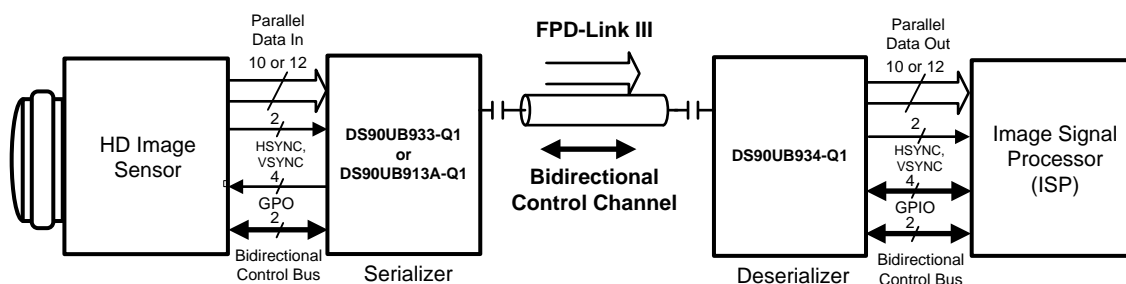
The DS90UB934-Q1 is improved over prior generations of ADAS FPD-Link III deserializer devices (such as DS90UB914A-Q1) offering higher bandwidth support with additional enhancements.

### Device Information<sup>(1)</sup>

PART NUMBER	PACKAGE	BODY SIZE (NOM)
DS90UB934-Q1	VQFN (48)	7.00 mm x 7.00 mm

(1) For all available packages, see the orderable addendum at the end of the data sheet.

### Typical Application Schematic



Copyright © 2017, Texas Instruments Incorporated



## Table of Contents

<b>1 Features</b>	<b>1</b>	<b>8 Application and Implementation</b>	<b>54</b>
<b>2 Applications</b>	<b>1</b>	8.1 Application Information	54
<b>3 Description</b>	<b>1</b>	8.2 Power Over Coax	54
<b>4 Revision History</b>	<b>2</b>	8.3 Typical Application	57
<b>5 Pin Configuration and Functions</b>	<b>4</b>	8.4 System Examples	60
<b>6 Specifications</b>	<b>7</b>	<b>9 Power Supply Recommendations</b>	<b>61</b>
6.1 Absolute Maximum Ratings	7	9.1 VDD Power Supply	61
6.2 ESD Ratings	7	9.2 Power-Up Sequencing	61
6.3 Recommended Operating Conditions	8	9.3 PDB Pin	61
6.4 Thermal Information	8	9.4 Ground	61
6.5 DC Electrical Characteristics	8	<b>10 Layout</b>	<b>62</b>
6.6 AC Electrical Characteristics	11	10.1 Layout Guidelines	62
6.7 Recommended Timing for the Serial Control Bus	12	10.2 Layout Example	63
6.8 Typical Characteristics	15	<b>11 Device and Documentation Support</b>	<b>64</b>
<b>7 Detailed Description</b>	<b>17</b>	11.1 Documentation Support	64
7.1 Overview	17	11.2 Receiving Notification of Documentation Updates	64
7.2 Functional Block Diagram	18	11.3 Community Resources	64
7.3 Feature Description	18	11.4 Trademarks	64
7.4 Device Functional Modes	20	11.5 Electrostatic Discharge Caution	64
7.5 Programming	26	11.6 Glossary	64
7.6 Register Maps	32	<b>12 Mechanical, Packaging, and Orderable Information</b>	<b>64</b>

## 4 Revision History

Changes from Revision A (January 2017) to Revision B	Page
• Added that unused GPIOs can be left open or floating	5
• Added that PDB is internal pull down enabled	5
• Added description for selecting pull up resistor for OSS_SEL	5
• Added description for selecting pull up resistor for OEN	5
• Removed S, PD type for RES (pin 44)	6
• Removed S, PD type for RES (pin 43) and added it must be tied to GND.	6
• Added PDB test conditions for the LVCMOS IO voltage parameter in the <i>Absolute Maximum Ratings</i> table	7
• Changed typical LVCMOS low-to-high transition time value from: 2.5 ns to: 2 ns	11
• Changed maximum LVCMOS low-to-high transition time value from: 4 ns to: 3 ns	11
• Changed typical LVCMOS high-to-low transition time value from: 2.5 ns to: 2 ns	11
• Changed maximum LVCMOS high-to-low transition time value from: 4 ns to: 3 ns	11
• Changed receiver clock jitter test condition from: SSCG[3:0] = OFF to: SSCG[0] = OFF	11
• Changed deserializer period jitter test condition from: SSCG[3:0] = OFF to: SSCG[0] = OFF	11
• Changed deserializer cycle-to-cycle clock jitter test condition from: SSCG[3:0] = OFF to: SSCG[0] = OFF	11
• Changed input jitter symbol from: TOL <sub>JIT</sub> to: T <sub>JIT</sub>	12
• Added reference to compatibility with DS90UB953-Q1/935-Q1 serializers	17
• Added column for DS90UB953-Q1/935-Q1	20
• Added clarification on input mode selection	20
• Fixed typo in Figure 13 supply rail text	20
• Changed pullup power supply node from VDDIO to V(I2C)	26
• Removed pullup resistor recommendation	26
• Updated description of clock frequency during BIST operation	31
• Fixed typos in register maps	32

## Revision History (continued)

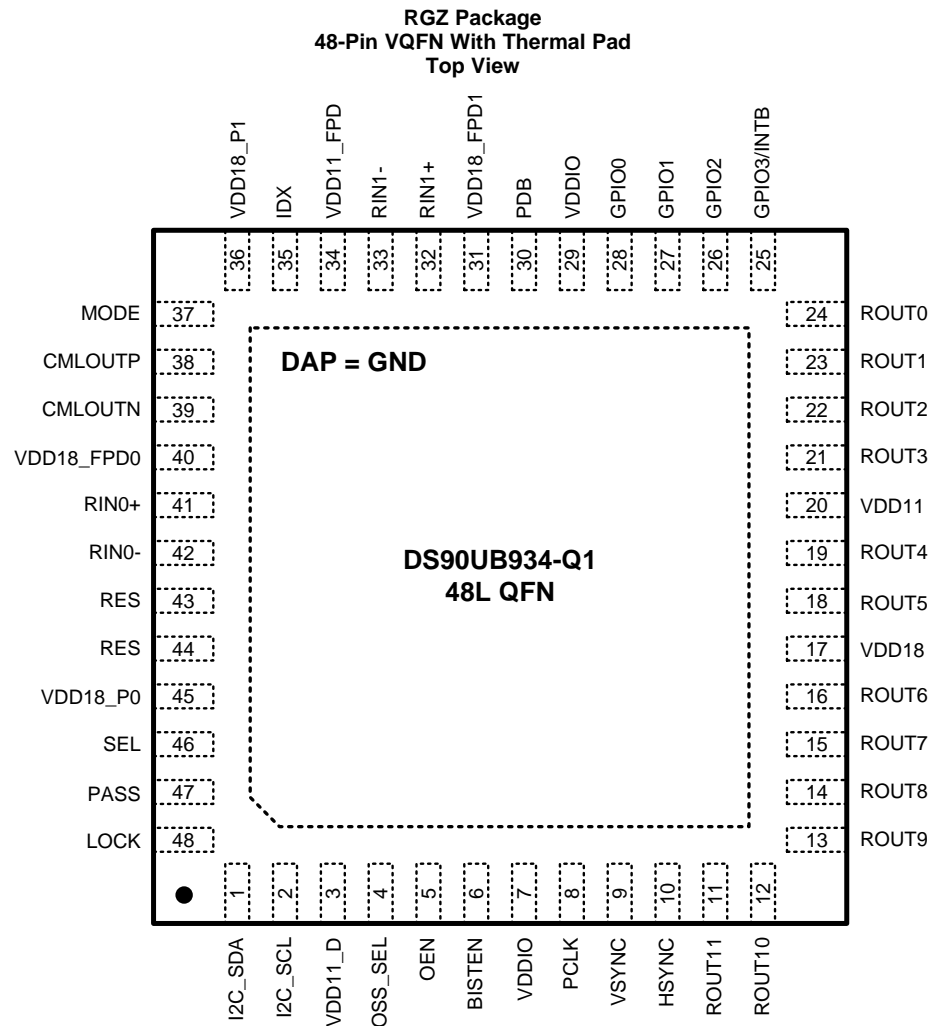
• Updated register "TYPE" column per legend .....	32
• Fixed typo in register name. ....	40
• Added Power Over Coax section .....	54
• Updated return loss S11 values .....	56
• Added STP typical connection diagram .....	58
• Updated recommendation for common ground plane .....	61
• Updated recommendation for bypass capacitors .....	62
• Updated typical bypass capacitor value from 50uF to 47uF .....	62

## Changes from Original (September 2016) to Revision A

## Page

• Changed product preview to production data .....	1
• Fixed broken link in <i>Power Over Coax</i> section.....	54

## 5 Pin Configuration and Functions



### Pin Functions

PIN		I/O TYPE	DESCRIPTION
NAME	NO.		
RECEIVE DATA PARALLEL OUTPUT			
ROUT0	24	O	RECEIVE DATA OUTPUT: This signal carries data from the FPD-LINK III deserializer to the processor. Output is parallel, configurable for up to 12 bits (ROUT0 – ROUT11) single ended outputs. VDDIO logic levels. For unused outputs leave as No Connect.
ROUT1	23		
ROUT2	22		
ROUT3	21		
ROUT4	19		
ROUT5	18		
ROUT6	16		
ROUT7	15		
ROUT8	14		
ROUT9	13		
ROUT10	12		
ROUT11	11		
HSYNC	10	O	Horizontal SYNC output. VDDIO logic levels.

### Pin Functions (continued)

PIN		I/O TYPE	DESCRIPTION
NAME	NO.		
VSYNC	9	O	Vertical SYNC output. VDDIO logic levels.
PCLK	8	O	Pixel clock (PCLK) output. VDDIO logic levels.
GPIO			
GPIO0	28	I/O, PD	General purpose input/output: Pins can be used to control and respond to various commands. They may be configured to be the input signals for the corresponding GPOs on the serializer or they may be configured to be outputs to follow local register settings. At power up the GPIO are disabled and by default include a 25-kΩ (typical) pulldown resistor. VDDIO logic levels. Unused GPIOs can be left open or floating.
GPIO1	27		
GPIO2	26		
GPIO3/INTB	25	I/O, Open Drain	General purpose input/output: Pin GPIO3 can be configured to be an input signal for GPOs on the serializer. Pin 25 is shared with INTB. Pull up with 4.7 kΩ to VDDIO. Programmable input/output pin is an active-low open drain and controlled by the status registers. Unused GPIOs can be left open or floating.
FPD-LINK III INTERFACE			
RIN0+	41	I/O	Receive input channel 0: Differential FPD-Link receiver and bidirectional control back channel output. The IO must be AC coupled. There is internal 100Ω differential termination between RIN0+ and RIN0-. For applications using single-ended coaxial channel connect RIN0+ with 100-nF, AC-coupling capacitor and terminate RIN0- to GND with a 47-nF capacitor and 50-Ω resistor. For STP applications connect both RIN0+ and RIN0- with 100-nF, AC-coupling capacitor.
RIN0-	42		
RIN1+	32	I/O	Receive input channel 1: Differential FPD-Link receiver and bidirectional control back channel output. The IO must be AC coupled. There is internal 100Ω differential termination between RIN1+ and RIN1-. For applications using single-ended coaxial channel connect RIN0+ with 100nF AC coupling capacitor and terminate RIN1- to Ground with a 47 nF capacitor and 50 ohm resistor. For STP applications connect both RIN1+ and RIN1- with 100 nF AC coupling capacitor.
RIN1-	33		
I2C PINS			
I2C_SCL	2	I/O, Open Drain	I2C serial clock: Clock line for the bidirectional control bus communication. External 2-kΩ to 4.7-kΩ pullup resistor to V <sub>I2C</sub> recommended per I2C interface standards.
I2C_SDA	1	I/O, Open Drain	I2C serial data: Data line for bidirectional control bus communication. External 2-kΩ to 4.7-kΩ pullup resistor to V <sub>I2C</sub> recommended per I2C interface standards.
CONFIGURATION and CONTROL PINS			
IDX	35	S	Input. I2C serial control bus device ID address Connect to external pullup to VDD18 (pin 17) and pull down to GND to create a voltage divider. See <a href="#">Table 7</a> .
MODE	37	S	Mode select configuration input to set operating mode based on input voltage level. Typically connected to voltage divider via external pullup to VDD18 (pin 17) and pulldown to GND See <a href="#">Table 2</a> .
PDB	30	S, PD	Power-down inverted Input Pin. This pin is internal pull down enabled. When PDB input is brought HIGH, the device is enabled. Asserting PDB signal low powers down the device and consume minimum power. The default function of this pin is PDB = LOW; POWER DOWN. This pin has a 50-kΩ (typical) internal pulldown resistor. <i>INPUT IS 3.3 V TOLERANT.</i> PDB = 1.8 V, device is enabled (normal operation) PDB = 0, device is powered down.
SEL	46	S,PD	MUX select: Digital input for selecting FPD Link input channel 0 (A) or channel 1 (B). The default state of SEL = L, selects RIN0, input A, as the active channel on the deserializer. Asserting SEL = H selects RIN1 input B as the active channel on the deserializer. This pin has a 25-kΩ (typical) internal pulldown resistor. VDDIO logic levels.
OSS_SEL	4	S, PD	Output sleep state select pin for enabling output sleep state. This pin has a 25-kΩ (typical) internal pulldown resistor. If unused, connect to VDD. If using pullup resistor to connect to VDD, the resistor value should be <= 4.3-kΩ. VDDIO logic levels. See <a href="#">DVP Output Control</a> .
OEN	5	S, PD	Output enable. This pin has a 1-MΩ (typical) internal pulldown resistor. If unused, connect to VDD. If using pullup resistor to connect to VDD, the resistor value should be <= 4.3-kΩ. VDDIO logic levels. See <a href="#">DVP Output Control</a> .

### Pin Functions (continued)

PIN		I/O TYPE	DESCRIPTION
NAME	NO.		
DIAGNOSTIC PINS			
CMLOUTP	38	O	Channel monitor loop-through (CML) driver differential output. Typically routed to test points and not connected. For monitoring terminate CMLOUT with a 100-Ω differential load.
CMLOUTN	39		
BISTEN	6	S, PD	BIST enable: BISTEN = H, BIST mode is enabled BISTEN = L, BIST mode is disabled. See <i>Built-In Self Test (BIST)</i> for more information. This pin has a 25-kΩ (typ) internal pulldown resistor. VDDIO logic levels.
PASS	47	O	PASS Output: PASS = H, ERROR FREE transmission in forward channel operation. PASS = L, one or more errors were detected in the received payload. See <i>Built-In Self Test (BIST)</i> for more information. Leave No Connect if unused. Typically route to test point for monitoring. VDDIO logic levels.
LOCK	48	O	LOCK Status: Output pin for monitoring lock status of FPD-Link III channel. LOCK = H, PLL is Locked, outputs are active. LOCK = L, PLL is unlocked, may be used as link status. VDDIO logic levels.
RES	44	-	Reserved. Must be NC or tied to GND for normal operation.
RES	43	-	Reserved. This pin has internal pull-up resistor. Must be tied to GND for normal operation.
POWER AND GROUND			
VDDIO	7,29	P	VDDIO voltage supply input: The single-ended outputs and control input are powered from VDDIO. VDDIO can be connected to a 1.8-V, ±5% or 3-V to 3.6-V power rail. Each pin requires a minimum 10-nF capacitor to GND.
VDD18	17	P	1.8-V (±5%) power supply. Requires 1-μF, 0.1-μF, and 0.01-μF capacitors to GND at each VDD pin.
VDD18_P0 VDD18_P1	45 36	P	1.8-V (±5%) PLL power supplies. Requires 1-μF, 0.1-μF, and 0.01-μF capacitors to GND at each VDD pin.
VDD18_FPD0 VDD18_FPD1	40 31	P	1.8-V (±5%) high-speed transceiver (HSTRX) analog power supplies. Requires 10-μF, 0.1-μF, and 0.01-μF capacitors to GND at each VDD pin.
VDD11_FPD	34	D	Decoupling capacitor connection for internal analog regulator. Requires a minimum 4.7-μF capacitor to GND and must not be connected to other 1.1-V supply rails.
VDD11_DVP	20	D	Decoupling capacitor connection for internal mixed signal regulator. Requires a minimum 4.7-μF capacitor to GND and must not be connected to other 1.1-V supply rails.
VDD11_D	3	D	Decoupling capacitor connection for internal digital regulator. Requires a minimum 4.7-μF capacitor to GND and must not be connected to other 1.1-V supply rails.
GND	DAP	G	DAP is the large metal contact at the bottom side, located at the center of the QFN package. Connect to the ground plane (GND).

The definitions below define the functionality of the I/O cells for each pin.

TYPE:

- I = Input
- O = Output
- I/O = Input/Output
- S = Configuration pin (All strap pins have internal pulldowns. If the default strap value needs to be changed then use an external resistor.)
- PD = Internal pulldown
- P, G = Power supply, ground
- D = Decoupling pin for internal voltage rail

## 6 Specifications

### 6.1 Absolute Maximum Ratings

Over operating free-air temperature range (unless otherwise noted)<sup>(1) (2)</sup>

		MIN	MAX	UNIT
Supply voltage	VDD18 (VDD18, VDD18_P1, VDD18_P0, VDD18_FPD0, VDD18_FPD1)	−0.3	2.16	V
	VDDIO	−0.3	3.96	
FPD-Link III input voltage	RIN0+, RIN0−, RIN1+, RIN1− Device powered up (VDD18 and VDDIO within recommended operating conditions)	−0.3	2.75	V
	RIN0+, RIN0−, RIN1+, RIN1− Device powered down (VDD18 and VDDIO below recommended operating conditions) Transient Voltage	−0.3	1.45	
	RIN0+, RIN0−, RIN1+, RIN1− Device powered down (VDD18 and VDDIO below recommended operating conditions) DC Voltage	−0.3	1.35	
LVCMOS IO voltage	ROUT[11:0], PCLK, VSYNC, HSYNC, GPIO0, GPIO1, GPIO2, SEL, OSS_SEL, OEN, BISTEN, PASS, LOCK	−0.3	$V_{(VDDIO)} + 0.3$	V
	PDB	−0.3	3.96	
Configuration input voltage	MODE, IDX	−0.3	$V_{(VDD18)} + 0.3$	V
Open-drain voltage	GPIO3/INTB, I2C_SDA, I2C_SCL	−0.3	3.96	
Junction temperature			150	°C
Storage temperature, T <sub>stg</sub>		−65	150	°C

- (1) If Military/Aerospace specified devices are required, contact the Texas Instruments Sales Office or Distributors for availability and specifications.
- (2) Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under *Recommended Operating Conditions*. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

### 6.2 ESD Ratings

			VALUE	UNIT
V <sub>(ESD)</sub> Electrostatic discharge	Human body model (HBM), per AEC Q100-002 <sup>(1)</sup>	RIN0+, RIN0−, RIN1+, RIN1−	±2000	V
		Other pins	±2000	
	Charged device model (CDM), per AEC Q100-011		±750	
	ESD Rating (IEC 61000-4-2) R <sub>D</sub> = 330 Ω, C <sub>S</sub> = 150 pF	Contact Discharge (RIN0+, RIN0−, RIN1+, RIN1−)	±8000	
		Air Discharge (RIN0+, RIN0−, RIN1+, RIN1−)	±15000	
	ESD Rating (ISO 10605) R <sub>D</sub> = 330 Ω, C <sub>S</sub> = 150 pF and 330 pF R <sub>D</sub> = 2 kΩ, C <sub>S</sub> = 150 pF and 330 pF	Contact Discharge (RIN0+, RIN0−, RIN1+, RIN1−)	±8000	
		Air Discharge (RIN0+, RIN0−, RIN1+, RIN1−)	±15000	

- (1) AEC Q100-002 indicates HBM stressing is done in accordance with the ANSI/ESDA/JEDEC JS-001 specification.

## DS90UB934-Q1

SNLS507B – SEPTEMBER 2016 – REVISED OCTOBER 2018

www.ti.com

### 6.3 Recommended Operating Conditions

Over operating free-air temperature range (unless otherwise noted)

		MIN	NOM	MAX	UNIT
Supply voltage	$V_{(VDD18)}$	1.71	1.8	1.89	V
LVCMOS supply voltage	$V_{(VDDIO)} = 1.8 \text{ V}$	1.71	1.8	1.89	V
	$V_{(VDDIO)} = 3.3 \text{ V}$	3.0	3.3	3.6	V
Operating free-air temperature, $T_A$		–40	25	105	°C
Data rate		0.7		1.87	Gbps
PCLK frequency		25		100	MHz
Local I <sup>2</sup> C frequency, $f_{I2C}$				1	MHz
Supply Noise <sup>(1)(2)</sup>	$V_{(VDD18)}$			50	mV <sub>p-p</sub>
	$V_{(VDDIO)}$			50	
Power-over-Coax noise <sup>(3)</sup>	RIN0+, RIN0–, RIN1+, RIN1–		20		

(1) DC-50 MHz

(2) Specification is ensured by design and/or characterization and is not tested in production.

(3) Measured across RIN[1:0]+ and RIN[1:0]– terminals

### 6.4 Thermal Information

THERMAL METRIC <sup>(1)</sup>		DS90UB934-Q1	UNIT
		RGZ (VQFN)	
		48 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	30.3	°C/W
$R_{\theta JC(TOP)}$	Junction-to-case (top) thermal resistance	12.3	°C/W
$R_{\theta JC(BOT)}$	Junction-to-case (bottom) thermal resistance	1.2	°C/W
$R_{\theta JB}$	Junction-to-board thermal resistance	6.9	°C/W
$\Psi_{JT}$	Junction-to-top characterization parameter	0.2	°C/W
$\Psi_{JB}$	Junction-to-board characterization parameter	6.8	°C/W

(1) For more information about traditional and new thermal metrics, see the [Semiconductor and IC Package Thermal Metrics](#) (SPRA953).

### 6.5 DC Electrical Characteristics

Over recommended operating supply and temperature ranges unless otherwise specified.

PARAMETER	TEST CONDITIONS	PIN OR FREQUENCY	MIN	TYP	MAX	UNIT
<b>TOTAL POWER CONSUMPTION</b>						
$P_T$	Total Power Consumption normal operation See <a href="#">Figure 5</a>	Worst Case pattern Default registers	$V_{(VDD18)} = V_{(VDDIO)} = 1.89 \text{ V}$	500	685	mW
			$V_{(VDD18)} = 1.89 \text{ V}, V_{(VDDIO)} = 3.6 \text{ V}$	900	1125	



## DC Electrical Characteristics (continued)

Over recommended operating supply and temperature ranges unless otherwise specified.

PARAMETER		TEST CONDITIONS		PIN OR FREQUENCY	MIN	TYP	MAX	UNIT
SUPPLY CURRENT								
I <sub>DD</sub>	Deserializer Supply Current (includes load current). See <a href="#">Figure 5</a> .	f = 100 MHz, 10-bit mode V <sub>(VDD18)</sub> = 1.89 V Worst Case Pattern, Default Registers C <sub>L</sub> = 8 pF	V <sub>(VDDIO)</sub> = 1.89 V OR 3.6 V	VDD18			250	mA
			V <sub>(VDDIO)</sub> = 1.89 V	VDDIO			60	
			V <sub>(VDDIO)</sub> = 3.6 V	VDDIO			145	
		f = 100 MHz, 12-bit HF mode V <sub>(VDD18)</sub> = 1.89 V Worst Case Pattern, Default Registers C <sub>L</sub> = 8 pF	V <sub>(VDDIO)</sub> = 1.89 V OR 3.6 V	VDD18			270	
			V <sub>(VDDIO)</sub> = 1.89 V	VDDIO			90	
			V <sub>(VDDIO)</sub> = 3.6 V	VDDIO			170	
		f = 50 MHz, 12-bit LF mode V <sub>(VDD18)</sub> = 1.89 V Worst Case Pattern, Default Registers C <sub>L</sub> = 8 pF	V <sub>(VDDIO)</sub> = 1.89 V OR 3.6 V	VDD18			240	
			V <sub>(VDDIO)</sub> = 1.89 V	VDDIO			80	
			V <sub>(VDDIO)</sub> = 3.6 V	VDDIO			155	
I <sub>DDZ</sub>	Deserializer Power Down Supply Current	V <sub>(VDD18)</sub> = 1.89 V, V <sub>(VDDIO)</sub> = 3.6V PDB = L, All other LVCMOS inputs = 0V, Default Registers		VDD18			30	mA
				VDDIO			10	
1.8-V LVCMOS I/O <sup>(1)</sup>								
V <sub>OH</sub>	High Level Output Voltage	I <sub>OH</sub> = −2 mA	V <sub>(VDDIO)</sub> = 1.71 V to 1.89 V	ROUT[11:0], HSYNC, VSYNC, LOCK, PASS	V <sub>(VDDIO)</sub> − 0.45		V <sub>(VDDIO)</sub>	V
V <sub>OL</sub>	Low Level Output Voltage	I <sub>OL</sub> = 2 mA	V <sub>(VDDIO)</sub> = 1.71 V to 1.89 V		GND		0.45	V
V <sub>IH</sub>	High Level Input Voltage	V <sub>(VDDIO)</sub> = 1.71 V to 1.89 V		GPIO[3:0], PDB, OEN, SEL, OSS_SEL, BISTEN	0.65 × V <sub>(VDDIO)</sub>		V <sub>(VDDIO)</sub>	V
V <sub>IL</sub>	Low Level Input Voltage	V <sub>(VDDIO)</sub> = 1.71 V to 1.89 V			GND		0.35 × V <sub>(VDDIO)</sub>	V
I <sub>IH</sub>	Input High Current	V <sub>IN</sub> = 1.71 V to 1.89 V		GPIO[3:0] <sup>(2)</sup> , OEN	−20		20	μA
				GPIO[2:0] <sup>(3)</sup> , SEL, PDB, OSS_SEL, BISTEN	−100		100	
I <sub>IL</sub>	Input Low Current	V <sub>IN</sub> = 0 V		GPIO[3:0], PDB, OEN, SEL, OSS_SEL, BISTEN	−20		20	μA
I <sub>OS</sub>	Output Short Circuit Current	V <sub>OUT</sub> = 0 V				−17		mA
I <sub>OZ</sub>	TRI-STATE Output Current	V <sub>OUT</sub> = 0 V or V <sub>(VDDIO)</sub> , PDB = L			−20		20	μA
3.3-V LVCMOS I/O <sup>(4)</sup>								
V <sub>OH</sub>	High Level Output Voltage	I <sub>OH</sub> = −4 mA	V <sub>(VDDIO)</sub> = 3.0 V to 3.6 V	GPIO[3:0], ROUT[11:0], HSYNC, VSYNC, LOCK, PASS	2.4		V <sub>(VDDIO)</sub>	V
V <sub>OL</sub>	Low Level Output Voltage	I <sub>OL</sub> = 4 mA	V <sub>(VDDIO)</sub> = 3.0 V to 3.6 V		GND		0.4	V

(1)  $V_{(VDDIO)} = 1.8$  V  $\pm$  5%

(2) GPIO[2:0] Pulldown disabled; Register 0xBE = 0x03

(3) GPIO[2:0] Pulldown enabled; Register 0xBE = 0x00

(4)  $V_{(VDDIO)} = 3.0$  V to 3.6 V

## DC Electrical Characteristics (continued)

Over recommended operating supply and temperature ranges unless otherwise specified.

PARAMETER		TEST CONDITIONS	PIN OR FREQUENCY	MIN	TYP	MAX	UNIT	
V <sub>IH</sub>	High Level Input Voltage	V <sub>(VDDIO)</sub> = 3.0 V to 3.6 V	GPIO[3:0], OEN, SEL, OSS_SEL, BISTEN	2		V <sub>(VDDIO)</sub>	V	
			PDB	1.17	V <sub>(VDDIO)</sub>			
V <sub>IL</sub>	Low Level Input Voltage	V <sub>(VDDIO)</sub> = 3.0 V to 3.6 V	GPIO[3:0], OEN, SEL, OSS_SEL, BISTEN	GND		0.8	V	
			PDB	GND	0.63			
I <sub>IH</sub>	Input High Current	V <sub>IN</sub> = 3.0 V to 3.6 V	GPIO[3:0] <sup>(2)</sup> , OEN, PDB	-20		20	μA	
			GPIO[2:0] <sup>(3)</sup> , SEL, OSS_SEL, BISTEN	-190		190		
I <sub>IL</sub>	Input Low Current	V <sub>IN</sub> = 0 V	GPIO[3:0], OEN, SEL, OSS_SEL, BISTEN, PDB	-20		20	μA	
I <sub>OS</sub>	Output Short Circuit Current	V <sub>OUT</sub> = 0 V			-40		mA	
I <sub>OZ</sub>	TRI-STATE Output Current	V <sub>OUT</sub> = 0 V or V <sub>(VDDIO)</sub> , PDB = LOW		-60		60	μA	
I2C SERIAL CONTROL BUS <sup>(5)</sup>								
V <sub>IH</sub>	Input High Level		I2C_SDA, I2C_SCL	0.7 × V <sub>(VDDIO)</sub>		V <sub>(VDDIO)</sub>	V	
V <sub>IL</sub>	Input Low Level			GND		0.3 × V <sub>(VDDIO)</sub>	V	
V <sub>HY</sub>	Input Hysteresis			50			mV	
V <sub>OL</sub>	Output Low Level	Standard/Fast Mode - I <sub>OL</sub> = 4 mA; Fast Plus Mode - I <sub>OL</sub> = 20 mA		0		0.4	V	
I <sub>IH</sub>	Input High Current	V <sub>IN</sub> = V <sub>(VDDIO)</sub>		-10		10	μA	
I <sub>IL</sub>	Input Low Current	V <sub>IN</sub> = 0V		-10		10	μA	
C <sub>IN</sub>	Input Capacitance <sup>(6)</sup>					5	10	pF
FPD-LINK III INPUT								
V <sub>CM</sub>	Common Mode Voltage See <a href="#">Figure 2</a> .				1.2		V	
R <sub>T</sub>	Internal Termination Resistor	Single Ended		40	50	60	Ω	
		Differential		80	100	120		
FPD-LINK III BIDIRECTIONAL CONTROL CHANNEL								
V <sub>OUT-BC</sub>	Back Channel Single-Ended Output Voltage	RL = 50 Ω, Coaxial configuration, forward channel disabled	RIN0+, RIN1+	190		260	mV	
V <sub>OD-BC</sub>	Back Channel Differential Output Voltage	RL = 100 Ω, STP configuration, forward channel disabled	RIN0+, RIN0- RIN1+, RIN1-	380		520	mV	

(5)  $V_{(VDDIO)} = 1.8\text{ V} \pm 5\%$  **OR** 3.0 V to 3.6 V

(6) Specification is ensured by design and/or characterization and is not tested in production.

## 6.6 AC Electrical Characteristics

Over recommended operating supply and temperature ranges unless otherwise specified.

PARAMETER		TEST CONDITIONS	PIN OR FREQUENCY	MIN	TYP	MAX	UNIT
<b>LVC MOS I/O</b>							
$t_{RCP}$	Receiver Output Clock Period. See <a href="#">Figure 7</a> .	10-bit Mode	PCLK, 50 - 100 MHz	10		20	ns
		12-bit HF Mode	PCLK, 37.5 - 100 MHz	10		26.7	
		12-bit LF Mode	PCLK, 25 - 50 MHz	20		40	
$t_{PDC}$	PCLK Duty Cycle <sup>(1)</sup>	10-bit Mode	PCLK	45%	50%	55%	
		12-bit HF or LF Mode		40%	50%	60%	
$t_{CLH}$	LVC MOS Low-to-High Transition Time <sup>(1)</sup> See <a href="#">Figure 1</a> .	V <sub>(VDDIO)</sub> = 1.71 V to 1.89 V OR V <sub>(VDDIO)</sub> = 3.0 V to 3.6 V CL = 8 pF (lumped load) Default Registers	PCLK		2	2.8	ns
$t_{CHL}$	LVC MOS High-to-Low Transition Time <sup>(1)</sup> See <a href="#">Figure 1</a> .		PCLK		2	2.8	ns
$t_{CLH}$	LVC MOS Low-to-High Transition Time <sup>(1)</sup> See <a href="#">Figure 1</a> .		ROUT[11:0], HSYNC, VSYNC, GPIO[2:0]		2	3	ns
$t_{CHL}$	LVC MOS High-to-Low Transition Time <sup>(1)</sup> See <a href="#">Figure 1</a> .		ROUT[11:0], HSYNC, VSYNC, GPIO[2:0]		2	3	ns
$t_{ROS}$	ROUT Setup Data to PCLK <sup>(1)</sup> See <a href="#">Figure 7</a> .		PCLK, ROUT[11:0], HSYNC, VSYNC	0.38T	0.5T		ns
$t_{ROH}$	ROUT Hold Data to PCLK <sup>(1)</sup> See <a href="#">Figure 7</a> .		PCLK, ROUT[11:0], HSYNC, VSYNC	0.38T	0.5T		ns
$t_{DD}$	Deserializer Delay <sup>(1)</sup> See <a href="#">Figure 6</a> .	Default Registers (RRFB = 1)	10-bit mode	175T		185T	ns
			12-bit HF mode	100T		115T	
			12-bit LF mode	65T		80T	
$t_{DDL T}$	Deserializer Data Lock Time See <a href="#">Figure 3</a> .	Digital Reset, or PDB = HIGH to LOCK = HIGH	10-bit mode			22	ms
			12-bit HF mode			22	
			12-bit LF mode			22	
$t_{RCJ}$	Receiver Clock Jitter <sup>(1)</sup>	PCLK, SSCG[0] = OFF	10-bit mode		40	70	ps
			12-bit HF mode		52	90	
			12-bit LF mode		45	85	
$t_{DPJ}$	Deserializer Period Jitter <sup>(1)</sup>	PCLK, SSCG[0] = OFF	10-bit mode		885	1020	ps
			12-bit HF mode		420	880	
			12-bit LF mode		400	515	
$t_{DCCJ}$	Deserializer Cycle-to-Cycle Clock Jitter <sup>(1)(2)</sup>	PCLK, SSCG[0] = OFF	10-bit mode		1360	1800	ps
			12-bit HF mode		1280	1500	
			12-bit LF mode		890	1150	
$f_{dev}$	Spread Spectrum Clocking Deviation Frequency See <a href="#">Figure 9</a> .	LVC MOS Output Bus, SSCG[0] = ON	25 - 100 MHz		±0.5% to ±2.5%		
$f_{mod}$	Spread Spectrum Clocking Modulation Frequency See <a href="#">Figure 9</a> .	LVC MOS Output Bus, SSCG[0] = ON	25 - 100 MHz		5 to 50		kHz
<b>FPD-Link III</b>							

(1) Specification is ensured by design and/or characterization and is not tested in production.

(2) Specification is ensured by characterization

## AC Electrical Characteristics (continued)

Over recommended operating supply and temperature ranges unless otherwise specified.

PARAMETER		TEST CONDITIONS	PIN OR FREQUENCY	MIN	TYP	MAX	UNIT
<b>LVCMOS I/O</b>							
$V_{IN}$	Single Ended Input Voltage See Figure 2.	Coaxial configuration. 1010 pattern applied to the far end of a 15 meter cable. $V_{IN}$ measured after the cable, at the deserializer input pins.			50		mV
$V_{ID}$	Differential Input Voltage See Figure 2.	STP Configuration. 1010 pattern applied to the far end of a 15 meter cable. $V_{ID}$ measured after the cable, at the deserializer input pins.			100		mV
$f_{BC}$	Back Channel Frequency		RIN0+, RIN0– RIN1+, RIN1–	3.5		5.5	MHz
$T_J$	Back Channel Jitter <sup>(1)</sup>				7	15	ns
$T_{IJIT}$	Input Jitter	10MHz Sinusoidal Jitter applied to FPD-Link III input				0.4	UI <sup>(3)</sup>

(3) 1UI = 1 bit time of FPD-Link Forward channel

## 6.7 Recommended Timing for the Serial Control Bus

Over I<sup>2</sup>C supply and temperature ranges unless otherwise specified.

			MIN	MAX	UNIT
<b>I<sup>2</sup>C SERIAL CONTROL BUS (Figure 4)</b>					
$f_{SCL}$	SCL Clock Frequency	Standard-mode	>0	100	kHz
		Fast-mode	>0	400	
		Fast-mode Plus	>0	1000	
$t_{LOW}$	SCL Low Period	Standard-mode	4.7		$\mu$ s
		Fast-mode	1.3		
		Fast-mode Plus	0.5		
$t_{HIGH}$	SCL High Period	Standard-mode	4		$\mu$ s
		Fast-mode	0.6		
		Fast-mode Plus	0.26		
$t_{HD;STA}$	Hold time for a start or a repeated start condition	Standard-mode	4		$\mu$ s
		Fast-mode	0.6		
		Fast-mode Plus	0.26		
$t_{SU;STA}$	Set Up time for a start or a repeated start condition	Standard-mode	4.7		$\mu$ s
		Fast-mode	0.6		
		Fast-mode Plus	0.26		
$t_{HD;DAT}$	Data Hold Time	Standard-mode	0		$\mu$ s
		Fast-mode	0		
		Fast-mode Plus	0		
$t_{SU;DAT}$	Data Set Up Time	Standard-mode	250		ns
		Fast-mode	100		
		Fast-mode Plus	50		
$t_{SU;STO}$	Set Up Time for STOP Condition	Standard-mode	4		$\mu$ s
		Fast-mode	0.6		
		Fast-mode Plus	0.26		

## Recommended Timing for the Serial Control Bus (continued)

Over I<sup>2</sup>C supply and temperature ranges unless otherwise specified.

			MIN	MAX	UNIT
<b>I<sup>2</sup>C SERIAL CONTROL BUS (Figure 4)</b>					
$t_{BUF}$	Bus Free Time Between STOP and START	Standard-mode	4.7		$\mu s$
		Fast-mode	1.3		
		Fast-mode Plus	0.5		
$t_r$	SCL and SDA Rise Time	Standard-mode		1000	ns
		Fast-mode		300	
		Fast-mode Plus		120	
$t_f$	SCL and SDA Fall Time	Standard-mode		300	ns
		Fast-mode		300	
		Fast-mode Plus		120	
$C_b$	Capacitive Load for Each Bus Line <sup>(1)</sup>	Standard-mode		400	pF
		Fast-mode		400	
		Fast-mode Plus		550	
$t_{SP}$	Input Filter <sup>(1)</sup>	Fast-mode		50	ns
		Fast-mode Plus		50	

(1) Specification is ensured by design and/or characterization and is not tested in production.

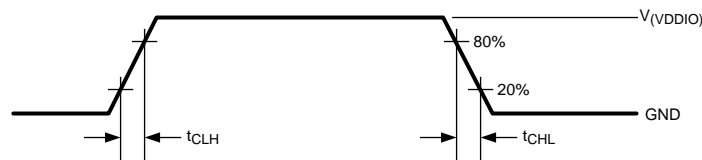


Figure 1. LVCMOS Transition Times

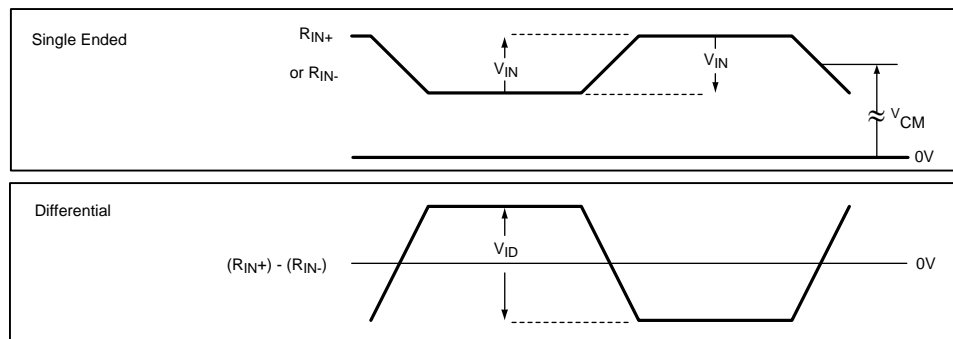


Figure 2. FPD-Link III Receiver  $V_{ID}$ ,  $V_{IN}$ ,  $V_{CM}$

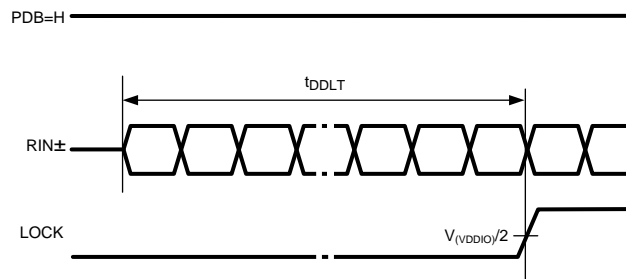
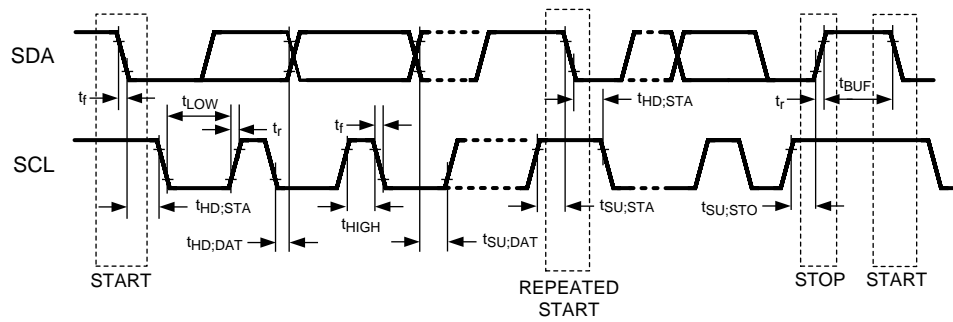
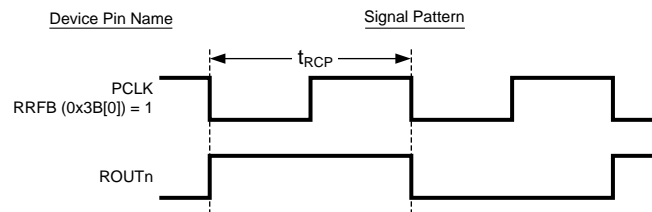
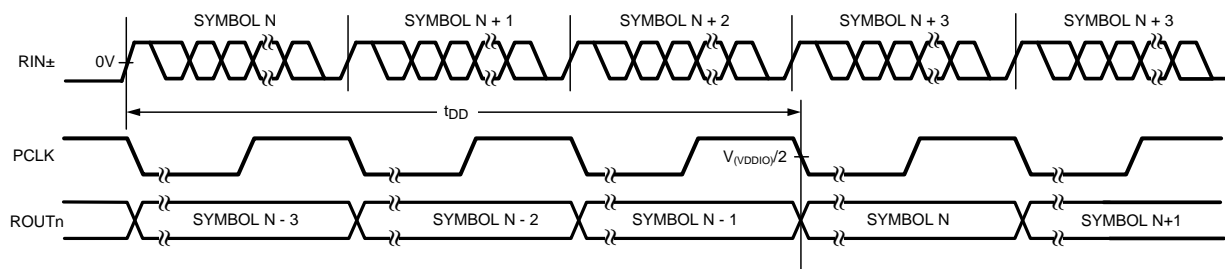
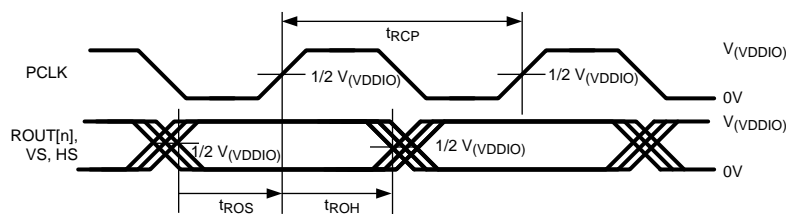
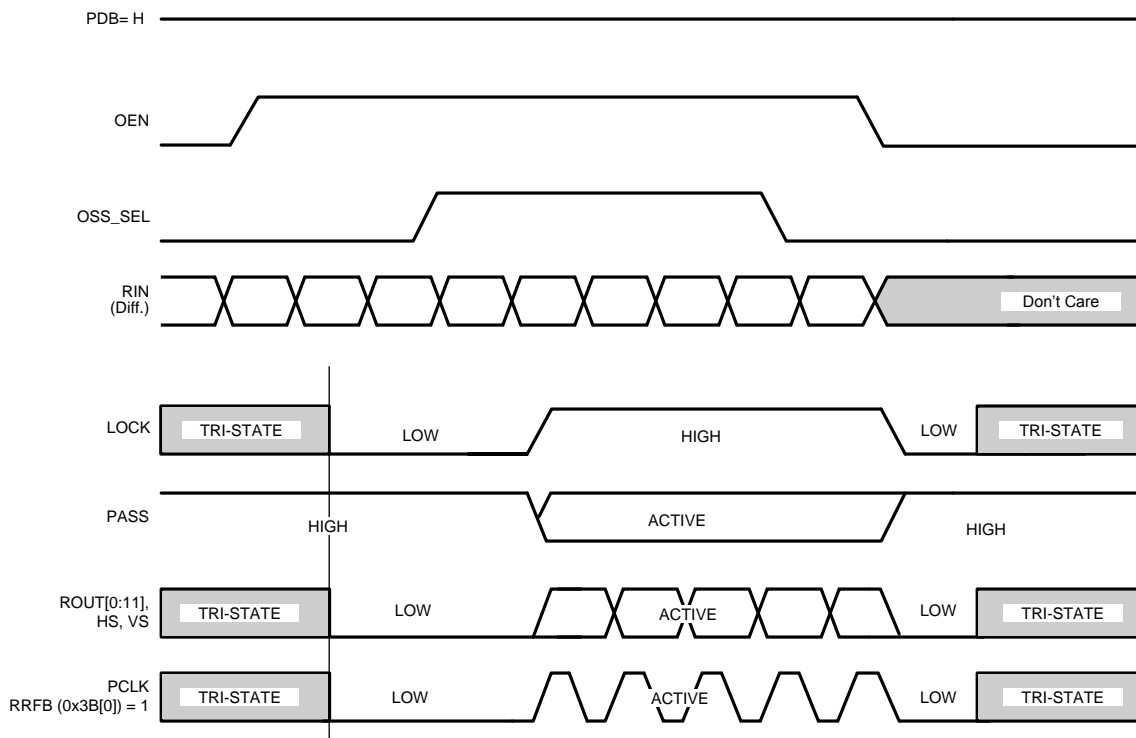
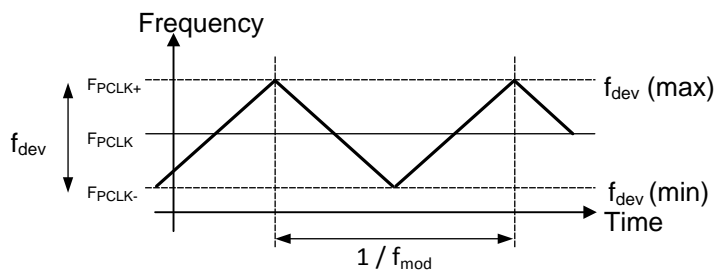


Figure 3. Deserializer Data Lock Time


**Figure 4. I2C Serial Control Bus Timing**

**Figure 5. SSO Test Pattern for Power Consumption**

**Figure 6. Deserializer Delay**

**Figure 7. Deserializer Output Setup/Hold Times**



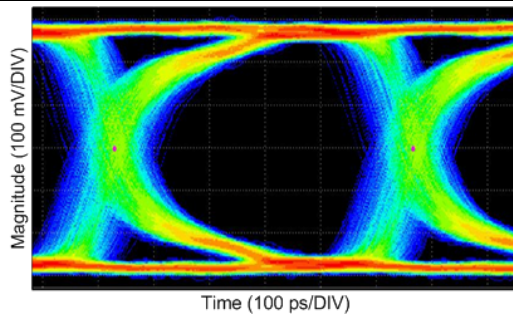
**Figure 8. Output State (Setup and Hold) Times**



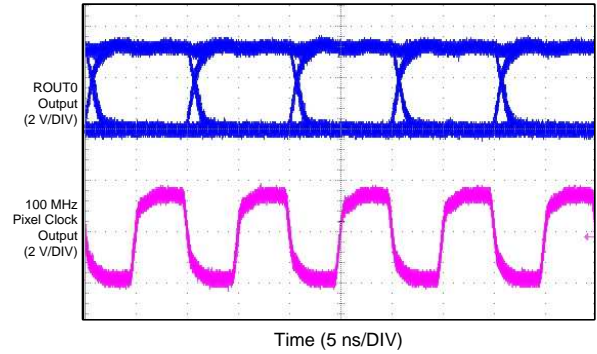
**Figure 9. Spread Spectrum Clock Output Profile**

## 6.8 Typical Characteristics

## Typical Characteristics (continued)



**Figure 10. CMLOUTP/N Loop-through Eye Diagram at 1.867 Gbps**



**Figure 11. ROUT0 Data Sampled by 100-MHz PCLK  
RRFB (0x3B[0]) = 1**



## 7 Detailed Description

### 7.1 Overview

The DS90UB934-Q1 FPD-Link III deserializer, in conjunction with the DS90UB913A/933-Q1 serializers, supports the video transport needs with a ultra-high-speed forward channel and an embedded bidirectional control channel. The DS90UB934-Q1 deserializer selects data streams from dual camera sources and outputs the recovered data onto a parallel LVCMOS output data bus. The DS90UB934-Q1 is designed to interface with the DS90UB933-Q1 device and is backwards compatible with the DS90UB913A-Q1 device using a 50-Ω coax interface. The DS90UB934-Q1 also works with the DS90UB933-Q1 or DS90UB913A-Q1 using an STP interface. The DS90UB934-Q1 can also work with the DS90UB953-Q1 or DS90UB935-Q1 in the backwards compatible mode (see the [Backwards Compatibility Modes for Operation with Parallel Output Deserializers](#) (SNLA270)). The DS90UB933/934 FPD-link III chipsets are intended to link mega-pixel camera imagers and video processors in ECUs. The serializer/deserializer chipset can operate from 25-MHz to 100-MHz pixel clock frequency.

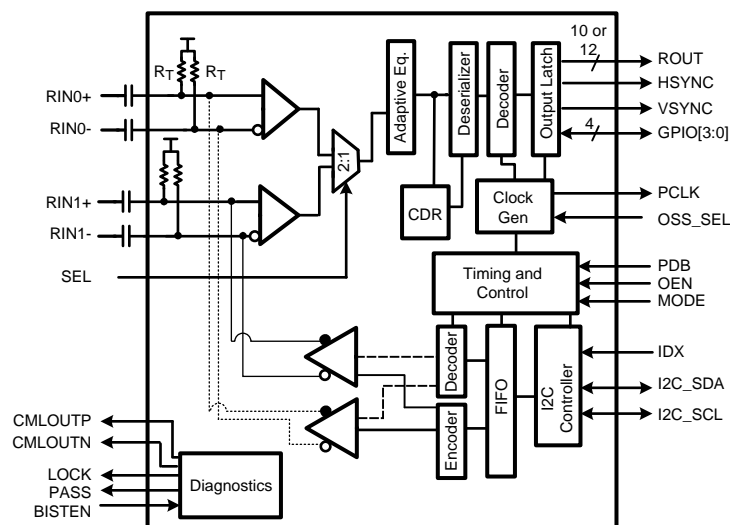
#### 7.1.1 Functional Description

The DS90UB934-Q1 converts the FPD-Link III stream into a parallel CMOS output interface designed to support automotive image sensors up to 12 bits at 100 MHz with resolutions including 1MP/60fps and 2MP/30fps. The DS90UB934-Q1 device recovers a high-speed FPD-Link III forward channel signal and outputs a 10- or 12-bit wide parallel LVCMOS data bus along with generating a bidirectional control channel control signal in the reverse channel direction. The high-speed, serial-bit stream contains an embedded clock and DC-balanced information which enhances signal quality to support AC coupling. The DS90UB934 deserializer can accept up to:

- 12 bits of DATA + 2 SYNC bits for an input PCLK range of 37.5 MHz to 100 MHz in the 12-bit high frequency mode. Note: No HS/VS restrictions (raw).
- 10 bits of DATA + 2 SYNC bits for an input PCLK range of 50 MHz to 100 MHz in the 10-bit mode. Note: HS/VS restricted to no more than one transition per 10 PCLK cycles.
- 12 bits of DATA + 2 bits SYNC for an input PCLK range of 25 MHz to 50 MHz in the 12-bit low frequency mode. Note: No HS/VS restrictions (raw).

The DS90UB934-Q1 device has a 2:1 multiplexer, which allows customers to select between two serializer inputs. The control channel function of the DS90UB933/DS90UB934-Q1 chipset provides bidirectional communication between the image sensor and ECUs. The integrated bidirectional control channel transfers data bidirectionally over the same channel used for video data interface. This interface offers advantages over other chipsets by eliminating the need for additional wires for programming and control. The bidirectional control channel bus is controlled via an I2C port. The bidirectional control channel offers asymmetrical communication and is not dependent on video blanking intervals. The DS90UB933/934 chipset offer customers the choice to work with different clocking schemes. The DS90UB933/934 chipsets can use an external oscillator as the reference clock source for the PLL or PCLK from the imager as primary reference clock to the PLL (see the [DS90UB933-Q1 data sheet](#)).

## 7.2 Functional Block Diagram



### 7.3 Feature Description

The DS90UB934-Q1 device has a 2:1 multiplexer that allows customers to select between two serializer inputs for camera applications. Frequency range operates up to 100 MHz in 12-bit mode or in 10-bit mode to support 1MP/60fps and 2MP/30fps imagers. The device accepts FPD-Link III inputs compatible to DS90UB933/913A/935/953 serializers. The received camera data stream from the selected input port is output onto the parallel interface.

### 7.3.1 Serial Frame Format

The high-speed forward channel is composed of 28 bits of data containing video data, sync signals, I2C, and parity bits. This data payload is optimized for signal transmission over an AC-coupled link. Data is randomized, DC-balanced, and scrambled. The 28-bit frame structure changes in the 12-bit, low-frequency mode, 12-bit, high-frequency mode and the 10-bit mode internally and is seamless to the customer. The bidirectional control channel data is transferred over the single serial link along with the high-speed forward data. This architecture provides a full duplex low-speed forward and backward path across the serial link together with a high-speed forward channel without the dependence on the video blanking phase.

### 7.3.2 Line Rate Calculations for the DS90UB933/934

The DS90UB933-Q1 device divides the clock internally by divide-by-1 in the 12-bit low-frequency mode, by divide-by-2 in the 10-bit mode, and by divide-by-1.5 in the 12-bit high-frequency mode. Conversely, the DS90UB934-Q1 multiplies the recovered serial clock to generate the proper pixel clock output frequency. Thus the maximum line rate in the three different modes remains 1.867 Gbps. The following are the formulae used to calculate the maximum line rate in the different modes:

- For the 12-bit mode: Line rate =  $f_{\text{PCLK}} \times (2/3) \times 28$ ; for example,  $f_{\text{PCLK}} = 100 \text{ MHz}$ , line rate =  $(100 \text{ MHz}) \times (2/3) \times 28 = 1.87 \text{ Gbps}$
- For the 10-bit mode: Line rate =  $f_{\text{PCLK}} / 2 \times 28$ ; for example,  $f_{\text{PCLK}} = 100 \text{ MHz}$ , line rate =  $(100 \text{ MHz}/2) \times 28 = 1.4 \text{ Gbps}$

### 7.3.3 Deserializer Multiplexer Input

The DS90UB934-Q1 offers a 2:1 multiplexer that can be used to select which camera is used as the input. [Figure 12](#) shows the operation of the 2:1 multiplexer in the deserializer. The selection of the camera can be pin controlled as well as register controlled. Only one deserializer input can be selected at a time. If the serializer A is selected as the active serializer, the back-channel for deserializer A turns ON and vice versa. To switch between the two cameras, first the serializer B must be selected using the SEL pin/register on the deserializer. After that the back channel driver for deserializer B has to be enabled using the register in the deserializer.

## Feature Description (continued)

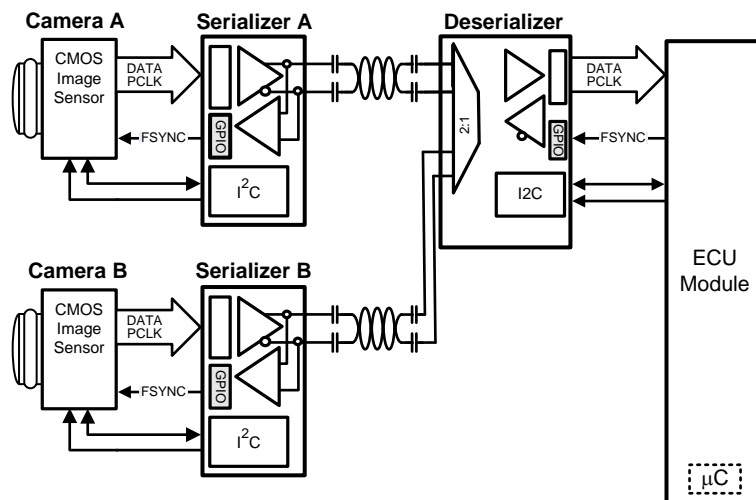


Figure 12. Using the Multiplexer on the Deserializer to Enable a Two-Camera System

## 7.4 Device Functional Modes

DS90UB934-Q1 supports the use cases shown in [Table 1](#):

**Table 1. PCLK Frequency Modes**

DS90UB934-Q1 DEVICE MODE	PCLK FREQUENCY RANGE		
	DS90UB913A-Q1 PARTNER	DS90UB933-Q1 PARTNER	DS90UB953-Q1/DS90UB935-Q1 PARTNER
RAW12 High-Frequency (HF)	37.5 MHz - 75 MHz	37.5 MHz - 100 MHz	37.5 MHz - 100 MHz
RAW12 Low-Frequency (LF)	25 MHz - 50 MHz	N/A	N/A
RAW10	50 MHz - 100 MHz	50 MHz - 100 MHz	50 MHz - 100 MHz

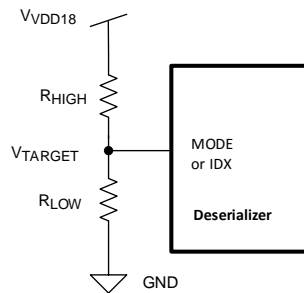
The modes control the FPD-Link III receiver operation of the device. In each of the cases, the output format for the device is parallel.

The input mode of operation is controlled by the MODE strap pin. The input mode may also be overridden and configured by FPD3\_MODE (Register 0x6D[1:0]) setting in the Port Configuration register.

### 7.4.1 RX MODE Pin

Configuration of the device may be done via the MODE input strap pin, or via the configuration register bits. A pullup resistor and a pulldown resistor of suggested values may be used to set the voltage ratio of the MODE input ( $V_{TARGET}$ ) and  $V_{(VDD18)}$  (pin 17) to select one of the 6 possible selected modes. Possible configurations are:

- FPD-Link III coax or STP
- 12-bit HF / 12-bit LF / 10-bit DVP modes



**Figure 13. Strap Pin Connection Diagram**

**Table 2. Strap Configuration Mode Select**

MODE NO.	V <sub>TARGET</sub> VOLTAGE RANGE			V <sub>TARGET</sub> STRAP VOLTAGE	SUGGESTED STRAP RESISTORS (1% TOL)		COAX /STP	RX MODE
	V <sub>MIN</sub>	V <sub>TYP</sub>	V <sub>MAX</sub>	(V); V <sub>(VDD18)</sub> = 1.8 V	R <sub>HIGH</sub> (kΩ )	R <sub>LOW</sub> (kΩ )		
0	RESERVED							
1	0.179 × V <sub>(VDD18)</sub>	0.213 × V <sub>(VDD18)</sub>	0.247 × V <sub>(VDD18)</sub>	0.374	88.7	23.2	STP	RAW12 LF
2	0.296 × V <sub>(VDD18)</sub>	0.330 × V <sub>(VDD18)</sub>	0.362 × V <sub>(VDD18)</sub>	0.582	75	35.7	STP	RAW12 HF
3	0.412 × V <sub>(VDD18)</sub>	0.443 × V <sub>(VDD18)</sub>	0.474 × V <sub>(VDD18)</sub>	0.792	71.5	56.2	STP	RAW10
4	RESERVED							
5	0.642 × V <sub>(VDD18)</sub>	0.673 × V <sub>(VDD18)</sub>	0.704 × V <sub>(VDD18)</sub>	1.202	39.2	78.7	COAX	RAW12 LF
6	0.761 × V <sub>(VDD18)</sub>	0.792 × V <sub>(VDD18)</sub>	0.823 × V <sub>(VDD18)</sub>	1.42	25.5	95.3	COAX	RAW12 HF
7	0.876 × V <sub>(VDD18)</sub>	V <sub>(VDD18)</sub>	V <sub>(VDD18)</sub>	1.8	10	OPEN	COAX	RAW10

The strapped values can be viewed and/or modified in the following locations:

- Coax – Port configuration COAX\_MODE (Register 0x6D[2])
- RX mode – Port configuration FPD3\_MODE (Register 0x6D[1:0])

#### 7.4.2 DVP Output Control

The LVCMOS outputs are controlled via the OEN and OSS\_SEL pins or via register override of these values. Register override is controlled by bits in the General Configuration register (0x02).

**Table 3. Output States**

INPUTS				OUTPUTS			
SERIAL INPUTS	PDB	OEN	OSS_SEL	LOCK	PASS	DATA	PCLK
X	0	X	X	Z	Z	Z	Z
X	1	0	0	L	L	L	L
X	1	0	1	Z	Z	Z	Z
static	1	1	0	L	L	L	L
static	1	1	1	L	previous state	L	L
active	1	1	0	H	L	L	L
active	1	1	1	H	valid	valid	valid

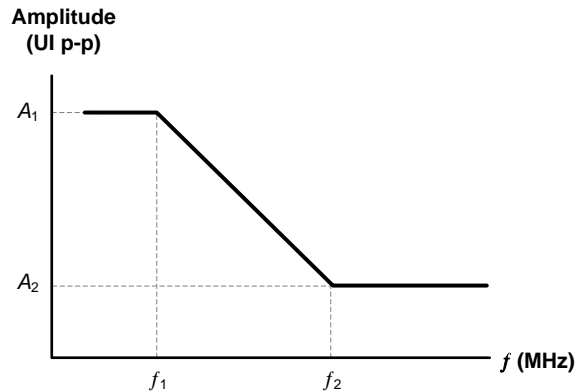
##### 7.4.2.1 LOCK Status

In 12-bit HF mode, the LOCK pin is only high if there is a link with a serializer that has an active PCLK input. LOCK is low if there is a serializer connected and there is a link established using the internal oscillator of the serializer. Therefore, when using this mode, it is preferred to use the port specific LOCK\_STS register (0x4D[0]), which is high when linked to a serializer with internal oscillator. This LOCK\_STS signal can also be output to a GPIO pin for monitoring in real time. Once LOCK\_STS is high for a specific port, remote I2C is available to that serializer.

In 12-bit LF or 10-bit modes, the LOCK pin is high when there is a link with a serializer regardless of whether there is an active PCLK input. The port specific LOCK\_STS register is also valid in either of these modes.

### 7.4.3 Input Jitter Tolerance

Input jitter tolerance is the ability of the CDR PLL of the receiver to track and recover the incoming serial data stream. Jitter tolerance at a specific frequency is the maximum jitter permissible before data errors occur. Figure 14 and Table 4 show the allowable total jitter of the receiver inputs and must be less than the values in Table 4.



**Figure 14. Input Jitter Tolerance Plot**

**Table 4. Input Jitter Tolerance Limit**

INTERFACE	JITTER AMPLITUDE (UI p-p)		FREQUENCY (MHz) <sup>(1)</sup>	
	A1	A2	f1	f2
FPD3	1	0.4	FPD3_PCLK / 80	FPD3_PCLK / 15

(1) FPD3\_PCLK is equivalent to PCLK frequency based on the operating MODE:  
10-bit mode: PCLK\_Freq. /2  
12-bit HF mode: PCLK\_Freq. x 2/3  
12-bit LF mode: PCLK\_Freq.

### 7.4.4 Adaptive Equalizer

The receiver inputs provide an adaptive equalization filter in order to compensate for signal degradation from the interconnect components. In order to determine the maximum cable reach, factors that affect signal integrity such as jitter, skew, ISI, crosstalk, etc. must be taken into consideration. The receiver incorporates an adaptive equalizer (AEQ), which continuously monitors cable characteristics for long-term cable aging and temperature changes. The AEQ attempts to optimize the equalization setting of the RX receiver.

If the deserializer loses LOCK, the adaptive equalizer resets and performs the LOCK algorithm again to reacquire the serial data stream being sent by the serializer.

### 7.4.5 Channel Monitor Loop-Through Output Driver

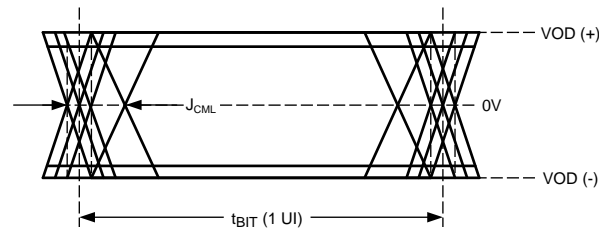
The DS90UB934-Q1 includes an internal channel monitor loop-through output on the CMLOUTP/N pins. A buffered loop-through output driver is provided on the CMLOUTP/N for observing jitter after equalization for each of the two RX receive channels. The CMLOUT monitors the post EQ stage, thus providing the recovered input of the deserializer signal. The measured serial data width on the CMLOUT loop-through is the total jitter including the internal driver, AEQ, back channel echo, etc. Each channel also has its own CMLOUT monitor and can be used for debug purposes. This CMLOUT is useful in identifying gross signal conditioning issues. The intrinsic jitter, J<sub>CML</sub>, represents the amount of jitter seen with a clean serial stream applied to the FPD-Link III input pins. When the total jitter is measured on CMLOUTP and CMLOUTN, the typical intrinsic jitter value can be subtracted to get an approximation of how much jitter is seen at the RIN[1:0]± input pins.

Table 6 includes details on selecting the corresponding RX receiver of CMLOUTP/N configuration.

**Table 5. CML Monitor Output Driver**

PARAMETER	TEST CONDITIONS	PIN	MIN	TYP	MAX	UNIT
J <sub>CML</sub> CMLOUT Differential Output Intrinsic Jitter	Clean clock fed into FPD-Link III input R <sub>L</sub> = 100 Ω (Figure 15)	CMLOUTP, CMLOUTN		0.15		UI <sup>(1)</sup>

- (1) UI – Unit interval is equivalent to one ideal serialized data bit width. The UI scales with serializer input PCLK frequency.  
 10-bit mode: 1 UI = 1 / ( PCLK\_Freq. /2 x 28 )  
 12-bit HF mode: 1 UI = 1 / ( PCLK\_Freq. x 2/3 x 28 )  
 12-bit LF mode: 1 UI = 1 / ( PCLK\_Freq. x 28 )



**Figure 15. CMLOUT Output Driver**

**Table 6. Channel Monitor Loop-Through Output Configuration**

	FPD3 RX Port 0	FPD3 RX Port 1
ENABLE MAIN LOOPTHRU DRIVER	0xB0 = 0x14 0xB1 = 0x00 0xB2 = 0x80	0xB0 = 0x14 0xB1 = 0x00 0xB2 = 0x80
SELECT CHANNEL MUX	0xB1 = 0x02 0xB2 = 0x20 0xB1 = 0x03 0xB2 = 0x28 0xB1 = 0x04 0xB2 = 0x28	0xB1 = 0x02 0xB2 = 0xA0 0xB1 = 0x03 0xB2 = 0x28 0xB1 = 0x04 0xB2 = 0x28
SELECT RX PORT	0xB0 = 0x18 0xB1 = 0x0F 0xB2 = 0x01 0xB1 = 0x10 0xB2 = 0x02	0xB0 = 0x18 0xB1 = 0x0F 0xB2 = 0x01 0xB1 = 0x10 0xB2 = 0x02

#### 7.4.5.1 Code Example for CMLOUT FPD3 RX Port 0:

```
board.WriteReg(0xB0,0x14)
board.WriteReg(0xB1,0x00)
board.WriteReg(0xB2,0x80)
board.WriteReg(0xB1,0x02)
board.WriteReg(0xB2,0x20)
board.WriteReg(0xB1,0x03)
board.WriteReg(0xB2,0x28)
board.WriteReg(0xB1,0x04)
board.WriteReg(0xB2,0x28)
board.WriteReg(0xB0,0x18)
board.WriteReg(0xB1,0x0F)
board.WriteReg(0xB2,0x01)
board.WriteReg(0xB1,0x10)
board.WriteReg(0xB2,0x02)
```

#### 7.4.6 GPIO Support

The DS90UB934-Q1 supports 4 pins programmable for use in multiple options through the GPIOx\_PIN\_CTL registers.

#### **7.4.6.1 Back Channel GPIO**

The DS90UB934-Q1 can input data on the GPIO pins to send on the back channel to remote serializers. Each GPIO pin can be programmed for input mode. In addition, the back channel for each FPD3 port can be programmed to send any of the 4 GPIO pins data. The same GPIO pin can be connected to multiple back channel GPIO signals.

In addition to sending GPIO from pins, an internally generated frame synchronization signal (FrameSync) signal may be sent on any of the back-channel GPIOs.

For each port, the following GPIO control is available through the BC\_GPIO\_CTL0 register 0x6E and BC\_GPIO\_CTL1 register 0x6F.

#### **7.4.6.2 GPIO Pin Status**

GPIO pin status may be read through the GPIO\_PIN\_STS register 0x0E. This register provides the status of the GPIO pin independent of whether the GPIO pin is configured as an input or output.

#### **7.4.6.3 Other GPIO Pin Controls**

Each GPIO pin has a input disable and a pulldown disable. By default, the GPIO pin input paths are enabled and the internal pulldown circuit in the GPIO is enabled. The GPIO\_INPUT\_CTL register 0x0F and GPIO\_PD\_CTL register 0xBE allow control of the input enable and the pulldown respectively. For most applications, there is no need to modify the default register settings.

#### **7.4.6.4 FrameSync Operation**

A FrameSync signal can be sent via the back channel using any of the back channel GPIOs. The signal can be generated in two different methods. The first option offers sending the external FrameSync using one of the available GPIO pins on the DS90UB934-Q1 and mapping that GPIO to a back channel GPIO on one of the FPD-Link III ports.

The second option is to have the DS90UB934-Q1 internally generate a FrameSync signal to send via GPIO to one of the attached serializers.

##### **7.4.6.4.1 External FrameSync Control**

In external FrameSync mode, an external signal is input to the DS90UB934-Q1 via one of the GPIO pins on the device. The external FrameSync signal may be propagated to either of the attached FPD3 serializers via a GPIO signal in the back channel.

Enabling the external FrameSync mode is done by setting the FS\_MODE control in the FS\_CTL (0x18) register to a value between 0x8 (GPIO0 pin) to 0xB (GPIO3 pin). Set FS\_GEN\_ENABLE to 0 for this mode.

To send the FrameSync signal on the BC\_GPIOx signal of a port, the BC\_GPIO\_CTL0 or BC\_GPIO\_CTL1 register must be programmed for that port to select the FrameSync signal.

##### **7.4.6.4.2 Internally Generated FrameSync**

In internal FrameSync mode, an internally generated FrameSync signal is sent to one or more of the attached FPD3 serializers via a GPIO signal in the back channel.

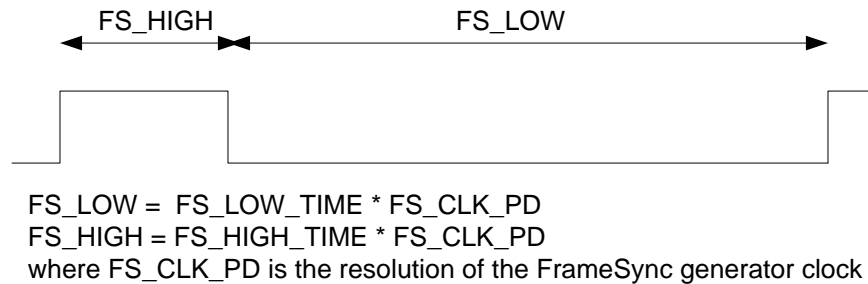
FrameSync operation is controlled by the FS\_CTL 0x18, FS\_HIGH\_TIME\_x, and FS\_LOW\_TIME\_x 0x19–0x1A registers. The resolution of the FrameSync generator clock (FS\_CLK\_PD) is derived from the back channel frame period (BC\_FREQ\_SELECT register). For 2.5-Mbps back-channel operation, the frame period is 12  $\mu$ s (30 bits  $\times$  400 ns/bit).

Once enabled, the FrameSync signal is sent continuously based on the programmed conditions.

Enabling the internal FrameSync mode is done by setting the FS\_GEN\_ENABLE control in the FS\_CTL (0x18) register to a value of 1. The FS\_MODE field controls the clock source used for the FrameSync generation. The FS\_GEN\_MODE field configures whether the duty cycle of the FrameSync is 50/50 or whether the high and low periods are controlled separately. The FrameSync high and low periods are controlled by the FS\_HIGH\_TIME and FS\_LOW\_TIME registers.



The accuracy of the internally generated FrameSync is directly dependent on the accuracy of the internal oscillator used to generate the back-channel reference clock. The internal oscillator has  $\pm 5\%$  variation over process, voltage, and temperature.



**Figure 16. Internal FrameSync Signal**

The following example shows generation of a FrameSync signal at 60 pulses per second. Mode settings:

- Programmable high/low periods: FS\_GEN\_MODE 0x18[1]=0
- Use port 0 back channel frame period: FS\_MODE 0x18[7:4]=0x0
- Back channel rate of 2.5 Mbps: BC\_FREQ\_SELECT for port 0 0x58[2:0]=0x0
- Initial FS state of 0: FS\_INIT\_STATE 0x18[2]=0

Based on mode settings, the FrameSync is generated based upon FS\_CLK\_PD of 12  $\mu$ s.

The total period of the FrameSync is (1 sec / 60 Hz) / 12  $\mu$ s or approximately 1,389 counts.

For a 10% duty cycle, set the high time to 139 (0x008A) cycles, and the low time to 1,250 (0x04E1) cycles:

- FS\_HIGH\_TIME\_1: 0x19 = 0x00
- FS\_HIGH\_TIME\_0: 0x1A = 0x8A
- FS\_LOW\_TIME\_1: 0x1B = 0x04
- FS\_LOW\_TIME\_0: 0x1C = 0xE1

#### 7.4.6.4.2.1 Code Example for Internally Generated FrameSync

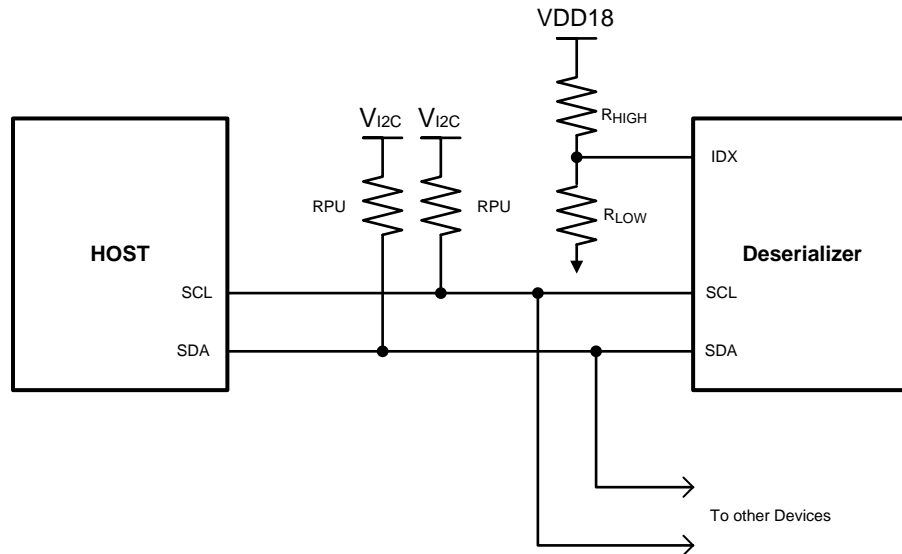
```
WriteI2C(0x4C,0x01) # RX0
WriteI2C(0x6E,0xAA) # BC_GPIO_CTL0: FrameSync signal to GPIO0/1
WriteI2C(0x4C,0x12) # RX1
WriteI2C(0x6E,0xAA) # BC_GPIO_CTL0: FrameSync signal to GPIO0/1
WriteI2C(0x10,0x91) # FrameSync signal; Device Status; Enabled
WriteI2C(0x58,0x58) # BC_FREQ_SELECT: 2.5 Mbps
WriteI2C(0x19,0x00) # FS_HIGH_TIME_1
WriteI2C(0x1A,0x8A) # FS_HIGH_TIME_0
WriteI2C(0x1B,0x04) # FS_LOW_TIME_1
WriteI2C(0x1C,0xE1) # FS_LOW_TIME_0
WriteI2C(0x18,0x01) # Enable FrameSync
```

## 7.5 Programming

### 7.5.1 Serial Control Bus

The DS90UB934-Q1 implements an I2C-compatible serial control bus. The I2C is for local device configuration and incorporates a bidirectional control channel (BCC) that allows communication with a remote serializers as well as remote I2C slave devices.

The device address is set via a resistor divider ( $R_{HIGH}$  and  $R_{LOW}$  — see Figure 17) connected to the IDX pin.



**Figure 17. Serial Control Bus Connection**

The serial control bus consists of two signals, SCL and SDA. SCL is a serial bus clock input. SDA is the serial bus data input/output signal. Both SCL and SDA signals require an external pullup resistor to 1.8-V or 3.3-V  $V_{(V_{I2C})}$ . The pullup resistor value may be adjusted for capacitive loading and data rate requirements. The signals are either pulled high or driven low.

The IDX pin configures the control interface to one of 8 possible device addresses. A pullup resistor and a pulldown resistor may be used to set the appropriate voltage ratio between the IDX input pin ( $V_{(IDX)}$ ) and  $V_{(V_{I2C})}$ , each ratio corresponding to a specific device address (see Table 7).

**Table 7. Serial Control Bus Addresses for IDX**

NO.	V <sub>IDX</sub> VOLTAGE RANGE			V <sub>IDX</sub> TARGET VOLTAGE (V); $V_{(VDD18)} = 1.8\text{ V}$	SUGGESTED STRAP RESISTORS (1% TOL)		ASSIGNED I2C ADDRESS	
	V <sub>MIN</sub>	V <sub>TYP</sub>	V <sub>MAX</sub>		R <sub>HIGH</sub> (kΩ)	R <sub>LOW</sub> (kΩ)	7-BIT	8-BIT
0	0	0	$0.131 \times V_{(VDD18)}$	0	OPEN	10.0	0x30	0x60
1	$0.179 \times V_{(VDD18)}$	$0.213 \times V_{(VDD18)}$	$0.247 \times V_{(VDD18)}$	0.374	88.7	23.2	0x32	0x64
2	$0.296 \times V_{(VDD18)}$	$0.330 \times V_{(VDD18)}$	$0.362 \times V_{(VDD18)}$	0.582	75.0	35.7	0x34	0x68
3	$0.412 \times V_{(VDD18)}$	$0.443 \times V_{(VDD18)}$	$0.474 \times V_{(VDD18)}$	0.792	71.5	56.2	0x36	0x6C
4	$0.525 \times V_{(VDD18)}$	$0.559 \times V_{(VDD18)}$	$0.592 \times V_{(VDD18)}$	0.995	78.7	97.6	0x38	0x70
5	$0.642 \times V_{(VDD18)}$	$0.673 \times V_{(VDD18)}$	$0.704 \times V_{(VDD18)}$	1.202	39.2	78.7	0x3A	0x74
6	$0.761 \times V_{(VDD18)}$	$0.792 \times V_{(VDD18)}$	$0.823 \times V_{(VDD18)}$	1.420	25.5	95.3	0x3C	0x78

## Programming (continued)

Table 7. Serial Control Bus Addresses for IDX (continued)

NO.	V <sub>IDX</sub> VOLTAGE RANGE			V <sub>IDX</sub> TARGET VOLTAGE (V); V <sub>(VDD18)</sub> = 1.8 V	SUGGESTED STRAP RESISTORS (1% TOL)		ASSIGNED I2C ADDRESS	
	V <sub>MIN</sub>	V <sub>TYP</sub>	V <sub>MAX</sub>		R <sub>HIGH</sub> (kΩ)	R <sub>LOW</sub> (kΩ)	7-BIT	8-BIT
7	$0.876 \times V_{(VDD18)}$	V <sub>(VDD18)</sub>	V <sub>(VDD18)</sub>	1.8	10	OPEN	0x3D	0x7A

The serial bus protocol is controlled by START, START-Repeated, and STOP phases. A START occurs when SDA transitions low while SCL is high. A STOP occurs when SDA transitions high while SCL is also high. See Figure 18.

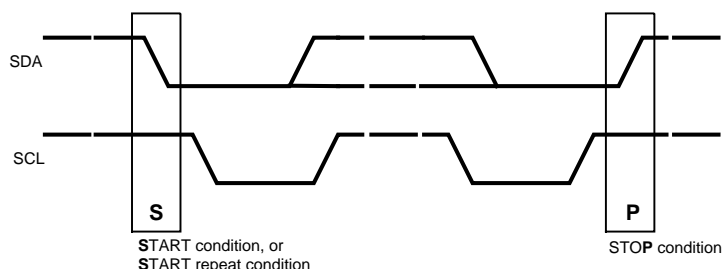


Figure 18. START and STOP Conditions

To communicate with a remote device, the host controller (master) sends the slave address and listens for a response from the slave. This response is referred to as an acknowledge bit (ACK). If a slave on the bus is addressed correctly, it acknowledges (ACKs) the master by driving the SDA bus low. If the address does not match the slave address of a device, it not-acknowledges (NACKs) the master by letting SDA be pulled High. ACKs also occur on the bus when data is being transmitted. When the master is writing data, the slave ACKs after every data byte is successfully received. When the master is reading data, the master ACKs after every data byte is received to let the slave know it wants to receive another data byte. When the master wants to stop reading, it NACKs after the last data byte and creates a stop condition on the bus. All communication on the bus begins with either a START condition or a REPEATED-START condition. All communication on the bus ends with a STOP condition. A READ is shown in Figure 19 and a WRITE is shown in Figure 20.

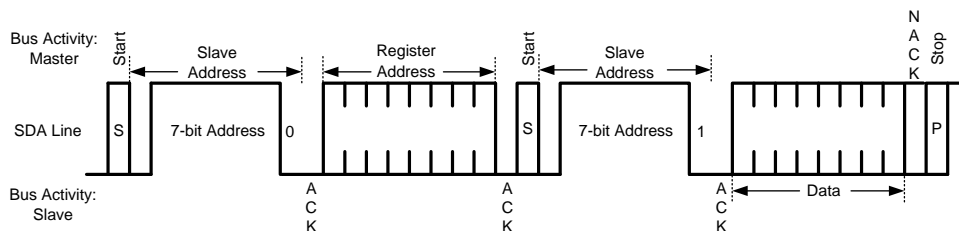


Figure 19. Serial Control Bus — READ

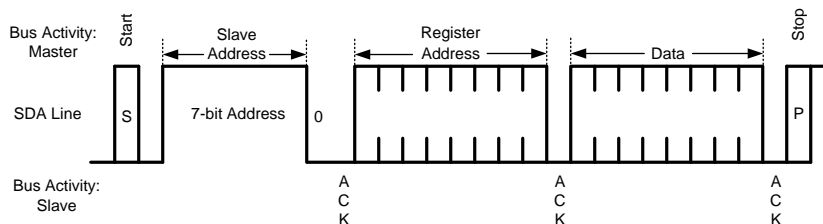
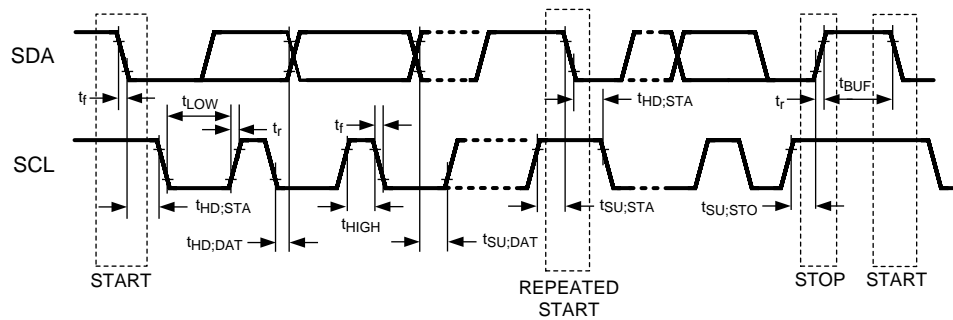


Figure 20. Serial Control Bus — WRITE



**Figure 21. Basic Operation**

The I2C master located at the deserializer must support I2C clock stretching. For more information on I2C interface requirements and throughput considerations, refer to [AN-2173 I2C Communication Over FPD-Link III with Bidirectional Control Channel](#) (SNLA131).

## 7.5.2 Interrupt Support

Interrupts can be brought out on the INTB pin as controlled by the INTERRUPT\_CTL 0x23 and INTERRUPT\_STS 0x24 registers. The main interrupt control registers provide control and status for interrupts from each of the two FPD3 receive ports. Clearing interrupt conditions requires reading the associated status register for the source. The setting of the individual interrupt status bits is not dependent on the related interrupt enable controls. The interrupt enable controls whether an interrupt is generated based on the condition, but does not prevent the interrupt status assertion.

For an interrupt to be generated based on one of the interrupt status assertions, both the individual interrupt enable and the INT\_EN control must be set in the INTERRUPT\_CTL 0x23 register. For example, to generate an interrupt if IS\_RX0 is set, both the IE\_RX0 and INT\_EN bits must be set. If IE\_RX0 is set but INT\_EN is not, the INT status is indicated in the INTERRUPT\_STS register, and the INTB pin does not indicate the interrupt condition.

See INTERRUPT\_CTL 0x23 and INTERRUPT\_STS 0x24 in [Table 9](#) for details.

### 7.5.2.1 Code Example to Enable Interrupts

```
# RX0/1 INTERRUPT_CTL enable
# "RX0 INTERRUPT_CTL enable"
WriteI2C(0x4C,0x01) # RX0
WriteI2C(0x23,0x81) # RX0 & INTB PIN EN

# "RX1 INTERRUPT_CTL enable"
WriteI2C(0x4C,0x12) # RX1
WriteI2C(0x23,0x82) # RX1 & INTB PIN EN
```

### 7.5.2.2 FPD-Link III Receive Port Interrupts

For each FPD-Link III receive port, multiple options are available for generating interrupts. Interrupt generation is controlled via the PORT\_ICR\_HI 0xD8 and PORT\_ICR\_LO 0xD9 registers. In addition, the PORT\_ISR\_HI 0xDA and PORT\_ISR\_LO 0xDB registers provide read-only status for the interrupts. Clearing of interrupt conditions is handled by reading the RX\_PORT\_STS and RX\_PORT\_STS2 registers. The status bits in the PORT\_ISR\_HI/LO registers are copies of the associated bits in the main status registers.

To enable interrupts from one of the receive port interrupt sources:

1. Enable the interrupt source by setting the appropriate interrupt enable bit in the PORT\_ICR\_HI or PORT\_ICR\_LO register
2. Set the RX port X Interrupt control bit (IE\_RXx) in the INTERRUPT\_CTL register
3. Set the INT\_EN bit in the INTERRUPT\_CTL register to allow the interrupt to assert the INTB pin low

To clear interrupts from one of the receive port interrupt sources:

1. (optional) Read the INTERRUPT\_STS register to determine which RX port caused the interrupt
2. (optional) Read the PORT\_ISR\_HI and PORT\_ISR\_LO registers to determine source of interrupt
3. Read the appropriate RX\_PORT\_STS1, RX\_PORT\_STS2 register to clear the interrupt.

The first two steps are optional. The interrupt could be determined and cleared by just reading the status registers.

### 7.5.2.3 Code Example to Readback Interrupts

```

INTERRUPT_STS = ReadI2C(0x24) # 0x24 INTERRUPT_STS

if ((INTERRUPT_STS & 0x80) >> 7):
    print "# GLOBAL INTERRUPT DETECTED "
if ((INTERRUPT_STS & 0x02) >> 1):
    print "# IS_RX1 DETECTED "
if ((INTERRUPT_STS & 0x01) ):
    print "# IS_RX0 DETECTED "

# "#####
# "RX0 status"
# "#####

WriteReg(0x4C,0x01) # RX0
PORT_ISR_LO = ReadI2C(0xDB)

print "0xDB PORT_ISR_LO : ", hex(PORT_ISR_LO) # readout; cleared by RX_PORT_STS2

if ((PORT_ISR_LO & 0x04) >> 2):
    print "# IS_FPD3_PAR_ERR DETECTED "
if ((PORT_ISR_LO & 0x02) >> 1):
    print "# IS_PORT_PASS DETECTED "
if ((PORT_ISR_LO & 0x01) ):
    print "# IS_LOCK_STS DETECTED "

#####

PORT_ISR_HI = ReadI2C(0xDA)

print "0xDA PORT_ISR_HI : ", hex(PORT_ISR_HI) # readout; cleared by RX_PORT_STS2

if ((PORT_ISR_HI & 0x04) >> 2):
    print "# IS_FPD3_ENC_ERR DETECTED "
if ((PORT_ISR_HI & 0x02) >> 1):
    print "# IS_BCC_SEQ_ERR DETECTED "
if ((PORT_ISR_HI & 0x01) ):
    print "# IS_BCC_CRC_ERR DETECTED "

#####

RX_PORT_STS1 = ReadI2C(0x4D) # R/COR

elif ((RX_PORT_STS1 & 0xc0) >> 6) == 1:
    print "# RX_PORT_NUM = RX1"
elif ((RX_PORT_STS1 & 0xc0) >> 6) == 0:
    print "# RX_PORT_NUM = RX0"

if ((RX_PORT_STS1 & 0x20) >> 5):
    print "# BCC_CRC_ERR DETECTED "
if ((RX_PORT_STS1 & 0x10) >> 4):
    print "# LOCK_STS_CHG DETECTED "
if ((RX_PORT_STS1 & 0x08) >> 3):
    print "# BCC_SEQ_ERROR DETECTED "
if ((RX_PORT_STS1 & 0x04) >> 2):
    print "# PARITY_ERROR DETECTED "
if ((RX_PORT_STS1 & 0x02) >> 1):
    print "# PORT_PASS=1 "
if ((RX_PORT_STS1 & 0x01) ):
    print "# LOCK_STS=1 "

#####

```

**DS90UB934-Q1**

SNLS507B –SEPTEMBER 2016–REVISED OCTOBER 2018

[www.ti.com](http://www.ti.com)

```

RX_PORT_STS2 = ReadI2C(0x4E)

if ((RX_PORT_STS2 & 0x20) >> 5):
    print "# FPD3_ENCODE_ERROR DETECTED "
if ((RX_PORT_STS2 & 0x04) >> 2):
    print "# FREQ_STABLE DETECTED "
if ((RX_PORT_STS2 & 0x02) >> 1):
    print "# NO_FPD3_CLK DETECTED "

#####

# "#####"
# "RX1 status"
# "#####"

WriteReg(0x4C,0x12) # RX1
PORT_ISR_LO = ReadI2C(0xDB) # PORT_ISR_LO readout; cleared by RX_PORT_STS2

if ((PORT_ISR_LO & 0x04) >> 2):
    print "# IS_FPD3_PAR_ERR DETECTED "
if ((PORT_ISR_LO & 0x02) >> 1):
    print "# IS_PORT_PASS DETECTED "
if ((PORT_ISR_LO & 0x01) ):
    print "# IS_LOCK_STS DETECTED "

#####

PORT_ISR_HI = ReadI2C(0xDA) # readout; cleared by RX_PORT_STS2

if ((PORT_ISR_HI & 0x04) >> 2):
    print "# IS_FPD3_ENC_ERR DETECTED "
if ((PORT_ISR_HI & 0x02) >> 1):
    print "# IS_BCC_SEQ_ERR DETECTED "
if ((PORT_ISR_HI & 0x01) ):
    print "# IS_BCC_CRC_ERR DETECTED "

#####

RX_PORT_STS1 = ReadI2C(0x4D) # R/COR

elif ((RX_PORT_STS1 & 0xc0) >> 6) == 1:
    print "# RX_PORT_NUM = RX1"
elif ((RX_PORT_STS1 & 0xc0) >> 6) == 0:
    print "# RX_PORT_NUM = RX0"

if ((RX_PORT_STS1 & 0x20) >> 5):
    print "# BCC_CRC_ERR DETECTED "
if ((RX_PORT_STS1 & 0x10) >> 4):
    print "# LOCK_STS_CHG DETECTED "
if ((RX_PORT_STS1 & 0x08) >> 3):
    print "# BCC_SEQ_ERROR DETECTED "
if ((RX_PORT_STS1 & 0x04) >> 2):
    print "# PARITY_ERROR DETECTED "
if ((RX_PORT_STS1 & 0x02) >> 1):
    print "# PORT_PASS=1 "
if ((RX_PORT_STS1 & 0x01) ):
    print "# LOCK_STS=1 "

#####

RX_PORT_STS2 = ReadI2C(0x4E)

if ((RX_PORT_STS2 & 0x20) >> 5):
    print "# FPD3_ENCODE_ERROR DETECTED "
if ((RX_PORT_STS2 & 0x04) >> 2):
    print "# FREQ_STABLE DETECTED "
if ((RX_PORT_STS2 & 0x02) >> 1):
    print "# NO_FPD3_CLK DETECTED "

#####

```

#### **7.5.2.4 Built-In Self Test (BIST)**

An optional at-speed BIST feature supports testing of the high-speed serial link and the back channel without external data connections. This is useful in the prototype stage, equipment production, in-system test, and system diagnostics.

##### **7.5.2.4.1 BIST Configuration and Status**

The BIST mode is enabled by BIST configuration register 0xB3. The test may select either an external PCLK or the internal oscillator clock (OSC) frequency in the serializer. In the absence of PCLK, the user can select the internal OSC frequency at the deserializer through the BIST configuration register. When BIST is activated at the deserializer, a BIST enable signal is sent to the serializer through the back channel. The serializer outputs a continuous stream of a pseudo-random sequence and drives the link at speed. The deserializer detects the test pattern and monitors it for errors. The serializer also tracks errors indicated by the CRC fields in each back channel frame. While the lock indications are required to identify the beginning of proper data reception, for any link failures or data corruption, the best indication is the contents of the error counter in the BIST\_ERR\_COUNT register 0x57 for each RX port.

The clock frequency that is output onto the PCLK pin during BIST mode is based on an internal FPD-Link III clock, and may not match the expected PCLK coming from the serializer.

## 7.6 Register Maps

In the register definitions under the *TYPE* and *DEFAULT* heading, the following definitions apply:

- R = Read only access
- R/W = Read / Write access
- R/RC = Read only access, Read to Clear
- (R/W)/SC = Read / Write access, Self-Clearing bit
- (R/W)/S = Read / Write access, Set based on strap pin configuration at startup
- LL = Latched Low and held until read
- LH = Latched High and held until read
- S = Set based on strap pin configuration at startup

### 7.6.1 Register Description

The DS90UB934-Q1 implements the following register blocks, accessible via I2C as well as the bi-directional control channel:

- Main registers
- FPD3 RX port registers (separate register block for each of the two RX ports)
- DVP port registers

**Table 8. Main Register Map Descriptions**

ADDRESS RANGE	DESCRIPTION	ADDRESS MAP	
0x00-0x31	Digital Shared Registers	Shared	
0x32-0x3A	Reserved	Reserved	
0x3B-0x3F	Digital DVP Registers	Shared	
0x4C-0x7F	Digital RX Port Registers (paged)	FPD3 RX Port 0 R: 0x4C[5:4]=00 W: 0x4C[0]=1	FPD3 RX Port 1 R: 0x4C[5:4]=01 W: 0x4C[1]=1
0x80-0xAF	Reserved	Reserved	
0xB0-0xB2	Indirect Access Registers	Shared	
0xB0-0xBF	Digital Share Registers	Shared	
0xC0-0xCF	Reserved	Reserved	
0xD0-0xDF	Digital RX Port Test Mode Registers	FPD3 RX Port 0	FPD3 RX Port 1
0xE0-0xEF	Reserved	Reserved	
0xF0-0xF5	FPD3 RX ID	Shared	
0xF6-0xF7	Reserved	Reserved	
0xF8-0xFB	Port I2C Addressing	Shared	
0xFC-0xFF	Reserved	Reserved	



**Table 9. Serial Control Bus Registers**

Page	Addr (hex)	Register Name	Bit(s)	Field	Type	Default	Description
Share	0x00	I2C Device ID	7:1	DEVICE ID	(R/W)/S	0x3D	7-bit I2C ID of deserializer Defaults to address configured by IDX strap pin. This field always indicates the current value of the I2C ID. When bit 0 of this register is 0, this field is read-only and shows the strapped ID. When bit 1 of this register is 1, this field is read/write and can be used to assign any valid I2C ID.
			0	DES ID	R/W	0x0	0: Device ID is from IDX strap pin 1: Register I2C device ID overrides strapped value
Share	0x01	Reset	7:3	RESERVED	R/W	0x0	Reserved
			2	RESTART_AUTOLOAD	(R/W)/S C	0x0	Restart ROM auto-load Setting this bit to 1 causes a re-load of the ROM. This bit is self-clearing. Software may check for auto-load complete by checking the CFG_INIT_DONE bit in the DEVICE_STS register.
			1	DIGITAL RESET1	(R/W)/S C	0x0	Digital reset Resets the entire digital block including registers. This bit is self-clearing. 1: Reset 0: Normal operation
			0	DIGITAL RESET0	(R/W)/S C	0x0	Digital reset Resets the entire digital block except registers. This bit is self-clearing. 1: Reset 0: Normal operation
Share	0x02	General Configuration	7	INPUT_PORT_OVERRIDE	R/W	0x0	Input port override bit allows control of the input port selection via the INPUT_PORT_SEL bit in this register.
			6	INPUT_PORT_SEL	R/W	0x0	Input port select. This bit either controls the input mode (if INPUT_PORT_OVERRIDE is set) or indicates the status of the SEL pin.
			5	OUTPUT_OVERRIDE	R/W	0x0	Output Control Override bit. The OUTPUT_ENABLE and OUTPUT_SLEEP_STATE_SEL values typically come from the device input pins. If this bit is set, the register values in this register will be used instead.
			4	RESERVED	R/W	0x1	Reserved
			3	OUTPUT_ENABLE	R/W	0x1	Output enable control (in conjunction with output sleep state select) If OUTPUT_SLEEP_STATE_SEL is set to 1 and this bit is set to 0, the TX outputs will be forced into a high impedance state. If OUTPUT_OVERRIDE is 0, this register indicates the value on the OEN pin. See <a href="#">Table 3</a> .
			2	OUTPUT_SLEEP_STATE_SEL	R/W	0x1	OSS Select controls the output state when LOCK is low (used in conjunction with Output Enable) When this bit is set to 0, the TX outputs is forced into a HS-0 state. If OUTPUT_OVERRIDE is 0, this register indicates the value on the OSS_SEL pin. See <a href="#">Table 3</a> .

**Table 9. Serial Control Bus Registers (continued)**

Page	Addr (hex)	Register Name	Bit(s)	Field	Type	Default	Description
			1	RX_PARITY_CHECK ER_EN	R/W	0x1	FPD3 Receiver Parity Checker Enable. When enabled, the parity check function is enabled for the FPD3 receiver. This allows detection of errors on the FPD3 receiver data bits. 0: Disable 1: Enable
			0	Reserved	R/W	0x0	Reserved
Share	0x03	Revision/Mask ID	7:4	REVISION_ID	R	0x0	Revision ID 0000: Production release
			3:0	RESERVED	R	0x0	Reserved
Share	0x04	DEVICE_STS	7	CFG_CKSUM_STS	R	0x1	Config Checksum passed This bit is set following initialization if the configuration data in the eFuse ROM had a valid checksum
			6	CFG_INIT_DONE	R	0x1	Power-up initialization complete This bit is set after Initialization is complete. Configuration from eFuse ROM has completed.
			5:4	RESERVED	R	0x0	Reserved
			3	PASS	R, LH	0x0	Device PASS status This bit indicates the PASS status for the device. The value in this register matches the indication on the PASS pin.
			2	LOCK	R, LH	0x0	Device LOCK status This bit indicates the LOCK status for the device. The value in this register matches the indication on the LOCK pin.
			1:0	RESERVED	R	0x0	Reserved
Share	0x05	PAR_ERR_THOLD_HI	7:0	PAR_ERR_THOLD_HI	R/W	0x01	FPD3 parity error threshold high byte This register provides the 8 most significant bits of the parity error threshold value. For each port, if the FPD-Link III receiver detects a number of parity errors greater than or equal to this value, the PARITY_ERROR flag is set in the RX_PORT_STS1 register.
Share	0x06	PAR_ERR_THOLD_LO	7:0	PAR_ERR_THOLD_LO	R/W	0x0	FPD3 parity error threshold low byte This register provides the 8 least significant bits of the parity error threshold value. For each port, if the FPD-Link III receiver detects a number of parity errors greater than or equal to this value, the PARITY_ERROR flag is set in the RX_PORT_STS1 register.
Share	0x07	BCC Watchdog Control	7:1	BCC WATCHDOG TIMER	R/W	0x7F	The watchdog timer allows termination of a control channel transaction if it fails to complete within a programmed amount of time. This field sets the bidirectional control channel watchdog timeout value in units of 2 milliseconds. Do not set this field to 0.
			0	BCC WATCHDOG TIMER DISABLE	R/W	0x0	Disable bidirectional control channel watchdog timer 1: Disables BCC watchdog timer operation 0: Enables BCC watchdog timer operation
Share	0x08	I2C Control 1	7	LOCAL WRITE DISABLE	R/W	0x0	Disable remote writes to local registers Setting this bit to a 1 prevents remote writes to local device registers from across the control channel. This prevents writes to the deserializer registers from an I2C master attached to the serializer. Setting this bit does not affect remote access to I2C slaves at the deserializer.

**Table 9. Serial Control Bus Registers (continued)**

Page	Addr (hex)	Register Name	Bit(s)	Field	Type	Default	Description
			6:4	I2C SDA HOLD	R/W	0x1	Internal SDA hold time This field configures the amount of internal hold time provided for the SDA input relative to the SCL input. Units are 50 nanoseconds.
			3:0	I2C FILTER DEPTH	R/W	0xC	I2C glitch filter depth This field configures the maximum width of glitch pulses on the SCL and SDA inputs that will be rejected. Units are 5 nanoseconds.
Share	0x09	I2C Control 2	7:4	SDA Output Setup	R/W	0x1	Remote Ack SDA output setup When a control channel (remote) access is active, this field configures setup time from the SDA output relative to the rising edge of SCL during ACK cycles. Setting this value will increase setup time in units of 640ns. The nominal output setup time value for SDA to SCL when this field is 0 is 80 ns.
			3:2	SDA Output Delay	R/W	0x0	SDA output delay This field configures additional delay on the SDA output relative to the falling edge of SCL. Setting this value increases output delay in units of 40 ns. Nominal output delay values for SCL to SDA are: 00: 240 ns 01: 280 ns 10: 320 ns 11: 360 ns
			1	I2C BUS TIMER SPEEDUP	R/W	0x0	Speed up I2C bus watchdog timer 1: Watchdog Timer expires after approximately 50 microseconds 0: Watchdog Timer expires after approximately 1 second.
			0	I2C BUS TIMER DISABLE	R/W	0x0	Disable I2C bus watchdog timer When enabled the I2C Watchdog Timer may be used to detect when the I2C bus is free or hung up following an invalid termination of a transaction. If SDA is high and no signaling occurs for approximately 1 second, the I2C bus is assumed to be free. If SDA is low and no signaling occurs, the device will attempt to clear the bus by driving 9 clocks on SCL.
Share	0x0A	SCL High Time	7:0	SCL HIGH TIME	R/W	0x7A	I2C Master SCL high time This field configures the high pulse width of the SCL output when the Serializer is the Master on the local I2C bus. Units are 40 ns for the nominal oscillator clock frequency. The default value is set to approximately 100 kHz with the internal oscillator clock running at nominal 25 MHz. Delay includes 4 additional oscillator clock periods. Nominal High Time = 40 ns × (TX_SCL_HIGH + 4) The internal oscillator has ±10% variation which must be taken into account when setting the SCL High and Low Time registers.

**Table 9. Serial Control Bus Registers (continued)**

Page	Addr (hex)	Register Name	Bit(s)	Field	Type	Default	Description
Share	0x0B	SCL Low Time	7:0	SCL LOW TIME	R/W	0x7A	I2C SCL low time This field configures the low pulse width of the SCL output when the serializer is the master on the local I2C bus. This value is also used as the SDA setup time by the I2C Slave for providing data prior to releasing SCL during accesses over the Bidirectional control channel. Units are 40 ns for the nominal oscillator clock frequency. The default value is set to approximately 100 kHz with the internal oscillator clock running at nominal 25 MHz. Delay includes 4 additional clock periods. Nominal low time = 40 ns × (TX_SCL_LOW + 4) The internal oscillator has ±10% variation which must be taken into account when setting the SCL High and Low Time registers.
Share	0x0C	RESERVED	7:0	RESERVED	R/W	0x0	Reserved
Share	0x0D	IO_CTL	7	SEL3P3V	R/W	0x0	3.3-V I/O Select on pins INTB, I2C 0: 1.8-V I/O Supply 1: 3.3-V I/O Supply If IO_SUPPLY_MODE_OV is 0, a read of this register returns the detected I/O voltage level.
			6	IO_SUPPLY_MODE_OV	R/W	0x0	Override I/O Supply Mode bit If set to 0, the detected voltage level is used for both SEL3P3V and IO_SUPPLY_MODE controls. If set to 1, the values written to the SEL3P3V and IO_SUPPLY_MODE fields is used.
			5:4	IO_SUPPLY_MODE	R/W	0x0	I/O supply mode 00: 1.8 V 11: 3.3 V If IO_SUPPLY_MODE_OV is 0, a read of this register returns the detected I/O voltage level.
			3:0	RESERVED	R/W	0x9	Reserved
Share	0x0E	GPIO_PIN_STS	7:4	RESERVED	R/W	0x0	Reserved
			3:0	GPIO_STS	R	0x0	GPIO pin status This register reads the current values on each of the 4 GPIO pins. Bit 3 reads GPIO3 and bit 0 reads GPIO0.
Share	0x0F	GPIO_INPUT_CTL	7:4	RESERVED	R/W	0x7	Reserved
			3	GPIO3_INPUT_EN	R/W	0x1	GPIO3 input enable 0: Disabled 1: Enabled
			2	GPIO2_INPUT_EN	R/W	0x1	GPIO2 input enable 0: Disabled 1: Enabled
			1	GPIO1_INPUT_EN	R/W	0x1	GPIO1 input enable 0: Disabled 1: Enabled
			0	GPIO0_INPUT_EN	R/W	0x1	GPIO0 input enable 0: Disabled 1: Enabled

**Table 9. Serial Control Bus Registers (continued)**

Page	Addr (hex)	Register Name	Bit(s)	Field	Type	Default	Description
Share	0x10	GPIO0_PIN_CTL	7:5	GPIO0_OUT_SEL	R/W	0x0	GPIO0 output select Determines the output data for the selected source. If GPIO0_OUT_SRC is set to 0xx (one of the RX Ports), the following selections apply: 000 : Received GPIO0 001 : Received GPIO1 010 : Received GPIO2 011 : Received GPIO3 100 : RX port lock indication 101 : RX port pass indication 110- 111 : Reserved If GPIO0_OUT_SRC is set to 100 (Device Status), the following selections apply: 000 : Value in GPIO0_OUT_VAL 001 : Logical OR of Lock indication from enabled RX ports 010 : Logical AND of Lock indication from enabled RX ports 011 : Logical AND of Pass indication from enabled RX ports 100 : FrameSync signal 101 - 111 : Reserved
			4:2	GPIO0_OUT_SRC	R/W	0x0	GPIO0 Output source select Selects output source for GPIO0 data: 000 : RX Port 0 001 : RX Port 1 01x : Reserved 100 : Device status 101 - 111 : Reserved
			1	GPIO0_OUT_VAL	R/W	0x0	GPIO0 output value This register provides the output data value when the GPIO pin is enabled to output the local register controlled value.
			0	GPIO0_OUT_EN	R/W	0x0	GPIO0 Output Enable 0: Disabled 1: Enabled
Share	0x11	GPIO1_PIN_CTL	7:5	GPIO1_OUT_SEL	R/W	0x0	GPIO1 Output Select Determines the output data for the selected source. If GPIO1_OUT_SRC is set to 0xx (one of the RX Ports), the following selections apply: 000 : Received GPIO0 001 : Received GPIO1 010 : Received GPIO2 011 : Received GPIO3 100 : RX Port Lock indication 101 : RX Port Pass indication 110- 111 : Reserved If GPIO1_OUT_SRC is set to 100 (Device Status), the following selections apply: 000 : Value in GPIO1_OUT_VAL 001 : Logical OR of Lock indication from enabled RX ports 010 : Logical AND of Lock indication from enabled RX ports 011 : Logical AND of Pass indication from enabled RX ports 100 : FrameSync signal 101 - 111 : Reserved

**Table 9. Serial Control Bus Registers (continued)**

Page	Addr (hex)	Register Name	Bit(s)	Field	Type	Default	Description
			4:2	GPIO1_OUT_SRC	R/W	0x0	GPIO1 Output Source Select Selects output source for GPIO1 data: 000 : RX port 0 001 : RX port 1 01x : Reserved 100 : Device status 101 - 111 : Reserved
			1	GPIO1_OUT_VAL	R/W	0x0	GPIO1 output value This register provides the output data value when the GPIO pin is enabled to output the local register controlled value.
			0	GPIO1_OUT_EN	R/W	0x0	GPIO1 output enable 0: Disabled 1: Enabled
Share	0x12	GPIO2_PIN_CTL	7:5	GPIO2_OUT_SEL	R/W	0x0	GPIO2 output select Determines the output data for the selected source. If GPIO2_OUT_SRC is set to 0xx (one of the RX Ports), the following selections apply: 000 : Received GPIO0 001 : Received GPIO1 010 : Received GPIO2 011 : Received GPIO3 100 : RX port lock indication 101 : RX port pass indication 110- 111 : Reserved If GPIO2_OUT_SRC is set to 100 (Device Status), the following selections apply: 000 : Value in GPIO2_OUT_VAL 001 : Logical OR of Lock indication from enabled RX ports 010 : Logical AND of Lock indication from enabled RX ports 011 : Logical AND of Pass indication from enabled RX ports 100 : FrameSync signal 101 - 111 : Reserved
			4:2	GPIO2_OUT_SRC	R/W	0x0	GPIO2 output source select Selects output source for GPIO2 data: 000 : RX port 0 001 : RX port 1 01x : Reserved 100 : Device status 101 - 111 : Reserved
			1	GPIO2_OUT_VAL	R/W	0x0	GPIO2 output value This register provides the output data value when the GPIO pin is enabled to output the local register controlled value.
			0	GPIO2_OUT_EN	R/W	0x0	GPIO2 output enable 0: Disabled 1: Enabled

**Table 9. Serial Control Bus Registers (continued)**

Page	Addr (hex)	Register Name	Bit(s)	Field	Type	Default	Description
Share	0x13	GPIO3_PIN_CTL	7:5	GPIO3_OUT_SEL	R/W	0x0	GPIO3 output select Determines the output data for the selected source. If GPIO3_OUT_SRC is set to 0xx (one of the RX Ports), the following selections apply: 000 : Received GPIO0 001 : Received GPIO1 010 : Received GPIO2 011 : Received GPIO3 100 : RX port lock indication 101 : RX port pass indication 110- 111 : Reserved If GPIO2_OUT_SRC is set to 100 (Device Status), the following selections apply: 000 : Value in GPIO3_OUT_VAL 001 : Logical OR of lock indication from enabled RX ports 010 : Logical AND of lock indication from enabled RX ports 011 : Logical AND of pass indication from enabled RX ports 100 : FrameSync signal 101 - 111 : Reserved
			4:2	GPIO3_OUT_SRC	R/W	0x0	GPIO3 output source select Selects output source for GPIO3 data: 000 : RX port 0 001 : RX port 1 01x : Reserved 100 : Device Status 101 - 111 : Reserved
			1	GPIO3_OUT_VAL	R/W	0x0	GPIO3 output value This register provides the output data value when the GPIO pin is enabled to output the local register controlled value.
			0	GPIO3_OUT_EN	R/W	0x0	GPIO3 output enable 0: Disabled 1: Enabled
Share	0x14 - 0x17	RESERVED	7:0	RESERVED	R/W	0x0	Reserved
Share	0x18	FS_CTL	7:4	FS_MODE	R/W	0x0	FrameSync mode 0000: Internal generated FrameSync, use back-channel frame clock from port 0 0001: Internal generated FrameSync, use back-channel frame clock from port 1 0010 : Reserved 0011: Reserved 01xx: Internal generated FrameSync, use 25-MHz (typical) clock 1000: External FrameSync from GPIO0 1001: External FrameSync from GPIO1 1010: External FrameSync from GPIO2 1011: External FrameSync from GPIO3 1100 - 1111: Reserved
			3	FS_SINGLE	(R/W)/S C	0x0	Generate single FrameSync pulse When this bit is set, a single FrameSync pulse is generated. The system waits for the full duration of the desired pulse before generating another pulse. When using this feature, the FS_GEN_ENABLE bit remains set to 0. This bit is self-clearing and always returns to 0.
			2	FS_INIT_STATE	R/W	0x0	Initial State This register controls the initial state of the FrameSync signal. 0: FrameSync initial state is 0 1: FrameSync initial state is 1

**Table 9. Serial Control Bus Registers (continued)**

Page	Addr (hex)	Register Name	Bit(s)	Field	Type	Default	Description
			1	FS_GEN_MODE	R/W	0x0	FrameSync Generation Mode This control selects between Hi/Lo and 50/50 modes. In High/Lo mode, the FrameSync generator uses the FS_HIGH_TIME and FS_LOW_TIME register values to separately control the high and low periods for the generated FrameSync signal. In 50/50 mode, the FrameSync generator uses the values in the FS_HIGH_TIME_0, FS_LOW_TIME_1 and FS_LOW_TIME_0 registers as a 24-bit value for both the high and low periods of the generated FrameSync signal. 0: Hi/Lo 1: 50/50
			0	FS_GEN_ENABLE	R/W	0x0	FrameSync generation enable 0: Disabled 1: Enabled
Share	0x19	FS_HIGH_TIME_1	7:0	FRAMESYNC_HIGH_TIME_1	R/W	0x0	FrameSync high time bits 15:8 The value programmed to the FS_HIGH_TIME register should be reduced by 1 from the desired delay. For example, a value of 0 in the FRAMESYNC_HIGH_TIME field results in a 1 cycle high pulse on the FrameSync signal.
Share	0x1A	FS_HIGH_TIME_0	7:0	FRAMESYNC_HIGH_TIME_0	R/W	0x0	FrameSync High Time bits 7:0 The value programmed to the FS_HIGH_TIME register should be reduced by 1 from the desired delay. For example, a value of 0 in the FRAMESYNC_HIGH_TIME field results in a 1 cycle high pulse on the FrameSync signal.
Share	0x1B	FS_LOW_TIME_1	7:0	FRAMESYNC_LOW_TIME_1	R/W	0x0	FrameSync Low Time bits 15:8 The value programmed to the FS_LOW_TIME register should be reduced by 1 from the desired delay. For example, a value of 0 in the FRAMESYNC_LOW_TIME field results in a 1 cycle low pulse on the FrameSync signal.
Share	0x1C	FS_LOW_TIME_0	7:0	FRAMESYNC_LOW_TIME_0	R/W	0x0	FrameSync Low Time bits 7:0 The value programmed to the FS_LOW_TIME register should be reduced by 1 from the desired delay. For example, a value of 0 in the FRAMESYNC_LOW_TIME field results in a 1 cycle low pulse on the FrameSync signal.
Share	0x1D - 0x22	RESERVED	7:0	RESERVED	R	0x00	Reserved
Share	0x23	INTERRUPT_CTL	7	INT_EN	R/W	0x0	Global interrupt enable Enables interrupt on the interrupt signal to the controller.
			6:2	RESERVED	R/W	0x0	Reserved
			1	IE_RX1	R/W	0x0	RX port 1 Interrupt: Enable interrupt from receiver port 1.
			0	IE_RX0	R/W	0x0	RX Port 0 Interrupt: Enable interrupt from receiver port 0.
Share	0x24	INTERRUPT_STS	7	INT	R	0x0	Global Interrupt: Set if any enabled interrupt is indicated in the individual status bits in this register. The setting of this bit is not dependent on the INT_EN bit in the INTERRUPT_CTL register but does depend on the IE_XXX bits. For example, if IE_RX0 and IS_RX0 are both asserted, the INT bit is set to 1.
			6:2	RESERVED	R	0x0	Reserved



**Table 9. Serial Control Bus Registers (continued)**

Page	Addr (hex)	Register Name	Bit(s)	Field	Type	Default	Description
			1	IS_RX1	R	0x0	RX port 1 interrupt: An interrupt has occurred for receive port 1. This interrupt is cleared by reading the associated status register(s) for the event(s) that caused the interrupt. The status registers are RX_PORT_STS1 and RX_PORT_STS2.
			0	IS_RX0	R	0x0	RX Port 0 Interrupt: An interrupt has occurred for receive port 0. This interrupt is cleared by reading the associated status register(s) for the event(s) that caused the interrupt. The status registers are RX_PORT_STS1 and RX_PORT_STS2.
Share	0x25	FS_CONFIG	7	RESERVED	R/W	0x0	Reserved
			6	FS_POLARITY	R/W	0x0	Framesync Polarity Indicates active edge of FrameSync signal 0: Rising edge 1: Falling edge
			5:0	RESERVED	R/W	0x00	Reserved
Share	0x26 - 0x3A	RESERVED	7:0	RESERVED	R/W	0x00	Reserved
DVP	0x3B	DVP_CLK_CTL	7:1	RESERVED	R/W	0x00	Reserved
			0	RRFB	R/W	0x1	Pixel clock edge select (relative to the sink) 1: Parallel interface data is driven on the falling clock edge and sampled on the rising clock edge 0: Parallel interface data is driven on the rising clock edge and sampled on the falling clock edge
DVP	0x3C	DVP_FREQ_DET0	7:5	RESERVED	R/W	0x0	Reserved
			4:0	FPD3_FREQ_LO_TH R	R/W	0x14	Frequency low threshold Sets the low threshold for the CDR Clock frequency detect circuit in MHz. This value is used to determine if the clock frequency is too low for proper operation.
DVP	0x3E	DVP_SSCG_CTL	7:6	RESERVED	R	0x0	Reserved
			5	RESERVED	R/W	0x0	Reserved
			4	SSCG_ENABLE	R/W	0x0	Enable SSCG modulation 0 : SSCG modulation is disabled 1 : SSCG modulation is enabled Prior to enabling SSCG, the SSCG_MOD_RATE must be set. This requires a separate write to set the SSCG_MOD_RATE with SSCG disabled, then a write to set the SSCG_ENABLE with the same SSCG_MOD_RATE setting. In addition, when changing the SSCG_MOD_RATE, disable the SSCG first.
			3:1	RESERVED	R/W	0x0	Reserved
			0	SSCG_MOD_RATE	R/W	0x0	SSCG modulation frequency with its deviation 0: Reserved 1: frequency modulation PCLK/3168 ±1%
Share	0x44 - 0x4B	RESERVED	7:0	RESERVED	R/W	0x00	Reserved

**Table 9. Serial Control Bus Registers (continued)**

Page	Addr (hex)	Register Name	Bit(s)	Field	Type	Default	Description
Share	0x4C	FPD3_PORT_SEL	7:6	PHYS_PORT_NUM	R	0x0	Physical port number This field provides the physical port connection when reading from a remote device via the bidirectional control channel. When accessed via local I2C interfaces, the value returned is always 0. When accessed via bidirectional control channel, the value returned is the port number of the receive port connection.
			5	RESERVED			Reserved
			4	RX_READ_PORT	R/W	0x0	Select RX port for register read This bit selects one of the two RX port register blocks for readback. This applies to all paged FPD3 receiver port registers. 0: Port 0 registers 1: Port 1 registers When accessed via local I2C interfaces, the default setting is 0. When accessed via bidirectional control channel, the default value is the port number of the receive port connection.
			3:2	RESERVED	R/W	0x0	Reserved
			1	RX_WRITE_PORT_1	R/W	0x0	Write Enable for RX port 1 registers This bit enables writes to RX port 1 registers. Any combination of RX port registers can be written simultaneously. This applies to all paged FPD3 Receiver port registers. 0: Writes disabled 1: Writes enabled When accessed via bidirectional control channel, the default value is 1 if accessed over RX port 1.
			0	RX_WRITE_PORT_0	R/W	0x0	Write Enable for RX port 0 registers This bit enables writes to RX port 0 registers. Any combination of RX port registers can be written simultaneously. This applies to all paged FPD3 receiver port registers. 0: Writes disabled 1: Writes enabled When accessed via Bidirectional Control Channel, the default value is 1 if accessed over RX port 0.
RX	0x4D	RX_PORT_STS1	7	RESERVED	R	0x0	Reserved
			6	RX_PORT_NUM	R	0x0	RX port number This read-only field indicates the number of the currently selected RX read port.
			5	BCC_CRC_ERROR	R, LH	0x0	Bidirectional control channel CRC error detected This bit indicates a CRC error has been detected in the forward control channel. If this bit is set, an error may have occurred in the control channel operation. This bit is cleared on read.
			4	LOCK_STS_CHG	R, LH	0x0	Lock status changed This bit is set if a change in receiver lock status has been detected since the last read of this register. Current lock status is available in the LOCK_STS bit of this register This bit is cleared on read.

**Table 9. Serial Control Bus Registers (continued)**

Page	Addr (hex)	Register Name	Bit(s)	Field	Type	Default	Description
			3	BCC_SEQ_ERROR	R, LH	0x0	Bidirectional control channel sequence error detected This bit indicates a sequence error has been detected in the forward control channel. If this bit is set, an error may have occurred in the control channel operation. This bit is cleared on read.
			2	PARITY_ERROR	R, LH	0x0	FPD3 parity errors detected This flag is set when the number of parity errors detected is greater than the threshold programmed in the PAR_ERR_THOLD registers. 1: Number of FPD3 parity errors detected is greater than the threshold 0: Number of FPD3 parity errors is below the threshold. This bit is cleared when the RX_PAR_ERR_HI/LO registers are cleared.
			1	PORT_PASS	R	0x0	Receiver PASS indication This bit indicates the current status of the Receiver PASS indication. The requirements for setting the Receiver PASS indication are controlled by the PORT_PASS_CTL register. 1: Receive input has met PASS criteria 0: Receive input does not meet PASS criteria
			0	LOCK_STS	R	0x0	FPD-Link III receiver is locked to incoming data 1: Receiver is locked to incoming data 0: Receiver is not locked
RX	0x4E	RX_PORT_STS2	7:6	RESERVED	R	0x0	Reserved
			5	FPD3_ENCODE_ERROR	R, LH	0x0	FPD3 encoder error detected If set, this flag indicates an error in the FPD-Link III encoding has been detected by the FPD-Link III receiver. This bit is cleared on read.
			4:3	RESERVED	R	0x0	Reserved
			2	FREQ_STABLE	R	0x0	Frequency measurement stable
			1	NO_FPD3_CLK	R	0x0	No FPD-Link III input clock detected
			0	RESERVED	R	0x0	Reserved
RX	0x4F	RX_FREQ_HIGH	7:0	FREQ_CNT_HIGH	R	0x0	FPD Link-III frequency measurement high byte (MHz) The frequency counter reports the measured frequency for the FPD3 receiver. This portion of the field is the integer value in MHz. Frequency measurements scales with reference clock frequency.
RX	0x50	RX_FREQ_LOW	7:0	FREQ_CNT_LOW	R	0x0	FPD Link-III frequency measurement low byte (1/256 MHz) The Frequency counter reports the measured frequency for the FPD3 Receiver. This portion of the field is the fractional value in 1/256 MHz. Values scales with reference clock frequency.
RX	0x51	RESERVED	7:0	RESERVED	R	0x0	Reserved
RX	0x52	RESERVED	7:0	RESERVED	R	0x0	Reserved
RX	0x53	RESERVED	7:0	RESERVED	R	0x0	Reserved
RX	0x54	RESERVED	7:0	RESERVED	R	0x0	Reserved

**Table 9. Serial Control Bus Registers (continued)**

Page	Addr (hex)	Register Name	Bit(s)	Field	Type	Default	Description
RX	0x55	RX_PAR_ERR_HI	7:0	PAR ERROR BYTE 1	R	0x0	Number of FPD3 parity errors – 8 most significant bits. The parity error counter registers return the number of data parity errors that have been detected on the FPD3 Receiver data since the last detection of valid lock or last read of the RX_PAR_ERR_LO register. For accurate reading of the parity error count, disable the RX_PARITY_CHECKER_ENABLE bit in register 0x02 prior to reading the parity error count registers. This register is cleared upon reading the RX_PAR_ERR_LO register.
RX	0x56	RX_PAR_ERR_LO	7:0	PAR ERROR BYTE 0	R	0x0	Number of FPD3 parity errors – 8 least significant bits. The parity error counter registers return the number of data parity errors that have been detected on the FPD3 Receiver data since the last detection of valid lock or last read of the RX_PAR_ERR_LO register. For accurate reading of the parity error count, disable the RX_PARITY_CHECKER_ENABLE bit in register 0x02 prior to reading the parity error count registers. This register will be cleared on read.
RX	0x57	BIST_ERR_COUNT	7:0	BIST ERROR COUNT	R	0x0	BIST error count Returns BIST error count
RX	0x58	BCC_CONFIG	7	I2C PASS THROUGH ALL	R/W	0x0	I2C pass-through all transactions 0: Disabled 1: Enabled
			6	I2C PASS THROUGH	R/W	0x0	I2C pass-through to serializer if decode matches 0: Pass-through disabled 1: Pass-through enabled
			5	AUTO ACK ALL	R/W	0x0	Automatically acknowledge all I2C writes independent of the forward channel lock state or status of the remote acknowledge 1: Enable 0: Disable
			4	BACK CHANNEL ENABLE FOR CAMERA MODE	R/W	0x1	Back channel enable for camera mode (display mode BC is always enabled) 1: Enable 0: Disable
			3	BC CRC GENERATOR ENABLE	R/W	0x1	Back Channel CRC Generator Enable 0: Disable 1: Enable
			2	RESERVED	R/W	0x0	Reserved
			1:0	BC FREQ SELECT	(R/W)/S	0x0	Back channel frequency select 00: 2.5 Mbps (default) 01: 1.5625 Mbps 10 - 11 : Reserved Note that changing this setting results in some errors on the back channel for a short period of time. If set over the control channel, first program the deserializer to Auto-Ack operation to avoid a control channel timeout due to lack of response from the serializer.
RX	0x59	RESERVED	7:0	RESERVED	R/W	0x0	Reserved
RX	0x5A	RESERVED	7:0	RESERVED	R/W	0x0	Reserved
RX	0x5B	SER_ID	7:1	SER ID	R/W	0x00	Remote serializer ID This field is normally loaded automatically from the remote serializer.

**Table 9. Serial Control Bus Registers (continued)**

Page	Addr (hex)	Register Name	Bit(s)	Field	Type	Default	Description
			0	FREEZE DEVICE ID	R/W	0x0	Freeze serializer device ID Prevent auto-loading of the serializer device ID from the forward channel. The ID is frozen at the value written.
RX	0x5C	SER_ALIAS_ID	7:1	SER ALIAS ID	R/W	0x0	7-bit remote serializer alias ID Configures the decoder for detecting transactions designated for an I2C slave device attached to the remote deserializer. The transaction will be remapped to the address specified in the slave ID register. A value of 0 in this field disables access to the remote I2C slave.
			0	SER AUTO ACK	R/W	0x0	Automatically acknowledge all I2C writes to the remote serializer independent of the forward channel lock state or status of the remote serializer acknowledge 1: Enable 0: Disable
RX	0x5D	SlaveID[0]	7:1	SLAVE ID0	R/W	0x0	7-bit remote slave device ID 0 Configures the physical I2C address of the remote I2C slave device attached to the remote serializer. If an I2C transaction is addressed to the slave alias ID0, the transaction is remapped to this address before passing the transaction across the bidirectional control channel to the serializer.
			0	RESERVED	R	0x0	Reserved
RX	0x5E	SlaveID[1]	7:1	SLAVE ID1	R/W	0x0	7-bit remote slave device ID 1 Configures the physical I2C address of the remote I2C Slave device attached to the remote Serializer. If an I2C transaction is addressed to the slave alias ID1, the transaction is remapped to this address before passing the transaction across the bidirectional control channel to the serializer.
			0	RESERVED	R	0x0	Reserved
RX	0x5F	SlaveID[2]	7:1	SLAVE ID2	R/W	0x0	7-bit remote slave device ID 2 Configures the physical I2C address of the remote I2C Slave device attached to the remote Serializer. If an I2C transaction is addressed to the Slave Alias ID2, the transaction is remapped to this address before passing the transaction across the bidirectional control channel to the serializer.
			0	RESERVED	R	0x0	Reserved
RX	0x60	SlaveID[3]	7:1	SLAVE ID3	R/W	0x0	7-bit remote slave device ID 3 Configures the physical I2C address of the remote I2C slave device attached to the remote serializer. If an I2C transaction is addressed to the slave alias ID3, the transaction is remapped to this address before passing the transaction across the bidirectional control channel to the serializer.
			0	RESERVED	R	0x0	Reserved
RX	0x61	SlaveID[4]	7:1	SLAVE ID4	R/W	0x0	7-bit remote slave device ID 4 Configures the physical I2C address of the remote I2C slave device attached to the remote Serializer. If an I2C transaction is addressed to the Slave Alias ID4, the transaction is remapped to this address before passing the transaction across the bidirectional control channel to the serializer.

**Table 9. Serial Control Bus Registers (continued)**

Page	Addr (hex)	Register Name	Bit(s)	Field	Type	Default	Description
			0	RESERVED	R	0x0	Reserved
RX	0x62	SlaveID[5]	7:1	SLAVE ID5	R/W	0x0	7-bit remote slave device ID 5 Configures the physical I2C address of the remote I2C slave device attached to the remote serializer. If an I2C transaction is addressed to the slave alias ID5, the transaction is remapped to this address before passing the transaction across the bidirectional control channel to the serializer.
			0	RESERVED	R	0x0	Reserved
RX	0x63	SlaveID[6]	7:1	SLAVE ID6	R/W	0x0	7-bit remote slave device ID 6 Configures the physical I2C address of the remote I2C slave device attached to the remote serializer. If an I2C transaction is addressed to the slave alias ID6, the transaction is remapped to this address before passing the transaction across the bidirectional control channel to the serializer.
			0	RESERVED	R	0x0	Reserved
RX	0x64	SlaveID[7]	7:1	SLAVE ID7	R/W	0x0	7-bit remote slave device ID 7 Configures the physical I2C address of the remote I2C slave device attached to the remote serializer. If an I2C transaction is addressed to the slave alias ID7, the transaction is remapped to this address before passing the transaction across the bidirectional control channel to the serializer.
			0	RESERVED	R	0x0	Reserved
RX	0x65	SlaveAlias[0]	7:1	SLAVE ALIAS ID0	R/W	0x0	7-bit remote slave device alias ID 0 Configures the decoder for detecting transactions designated for an I2C slave device attached to the remote serializer. The transaction is remapped to the address specified in the slave ID0 register. A value of 0 in this field disables access to the remote I2C slave.
			0	SLAVE AUTO ACK 0	R/W	0x0	Automatically acknowledge all I2C writes to the remote slave 0 independent of the forward channel lock state or status of the remote serializer acknowledge. 1: Enable 0: Disable
RX	0x66	SlaveAlias[1]	7:1	SLAVE ALIAS ID1	R/W	0x0	7-bit remote slave device alias ID 1 Configures the decoder for detecting transactions designated for an I2C slave device attached to the remote serializer. The transaction is remapped to the address specified in the slave ID1 register. A value of 0 in this field disables access to the remote I2C slave.
			0	SLAVE AUTO ACK 1	R/W	0x0	Automatically acknowledge all I2C writes to the remote slave 1 independent of the forward channel lock state or status of the remote serializer acknowledge 1: Enable 0: Disable

**Table 9. Serial Control Bus Registers (continued)**

Page	Addr (hex)	Register Name	Bit(s)	Field	Type	Default	Description
RX	0x67	SlaveAlias[2]	7:1	SLAVE ALIAS ID2	R/W	0x0	7-bit remote slave device alias ID 2 Configures the decoder for detecting transactions designated for an I2C slave device attached to the remote serializer. The transaction is remapped to the address specified in the slave ID2 register. A value of 0 in this field disables access to the remote I2C slave.
			0	SLAVE AUTO ACK 2	R/W	0x0	Automatically acknowledge all I2C writes to the remote slave 2 independent of the forward channel lock state or status of the remote serializer acknowledge 1: Enable 0: Disable
RX	0x68	SlaveAlias[3]	7:1	SLAVE ALIAS ID3	R/W	0x0	7-bit remote slave device alias ID 3 Configures the decoder for detecting transactions designated for an I2C slave device attached to the remote serializer. The transaction is remapped to the address specified in the slave ID3 register. A value of 0 in this field disables access to the remote I2C slave.
			0	SLAVE AUTO ACK 3	R/W	0x0	Automatically acknowledge all I2C writes to the remote slave 3 independent of the forward channel lock state or status of the remote serializer acknowledge. 1: Enable 0: Disable
RX	0x69	SlaveAlias[4]	7:1	SLAVE ALIAS ID4	R/W	0x0	7-bit remote slave device alias ID 4 Configures the decoder for detecting transactions designated for an I2C slave device attached to the remote serializer. The transaction is remapped to the address specified in the slave ID4 register. A value of 0 in this field disables access to the remote I2C slave.
			0	SLAVE AUTO ACK 4	R/W	0x0	Automatically acknowledge all I2C writes to the remote slave 4 independent of the forward channel lock state or status of the remote serializer acknowledge. 1: Enable 0: Disable
RX	0x6A	SlaveAlias[5]	7:1	SLAVE ALIAS ID5	R/W	0x0	7-bit remote slave device alias ID 5 Configures the decoder for detecting transactions designated for an I2C slave device attached to the remote serializer. The transaction is remapped to the address specified in the slave ID5 register. A value of 0 in this field disables access to the remote I2C slave.
			0	SLAVE AUTO ACK 5	R/W	0x0	Automatically acknowledge all I2C writes to the remote slave 5 independent of the forward channel lock state or status of the remote serializer acknowledge. 1: Enable 0: Disable
RX	0x6B	SlaveAlias[6]	7:1	SLAVE ALIAS ID6	R/W	0x0	7-bit remote slave device alias ID 6 Configures the decoder for detecting transactions designated for an I2C slave device attached to the remote serializer. The transaction is remapped to the address specified in the slave ID6 register. A value of 0 in this field disables access to the remote I2C slave.

**Table 9. Serial Control Bus Registers (continued)**

Page	Addr (hex)	Register Name	Bit(s)	Field	Type	Default	Description
			0	SLAVE AUTO ACK 6	R/W	0x0	Automatically acknowledge all I2C writes to the remote slave 6 independent of the forward channel lock state or status of the remote serializer acknowledge. 1: Enable 0: Disable
RX	0x6C	SlaveAlias[7]	7:1	SLAVE ALIAS ID7	R/W	0x0	7-bit remote slave device alias ID 7 Configures the decoder for detecting transactions designated for an I2C slave device attached to the remote serializer. The transaction is remapped to the address specified in the slave ID7 register. A value of 0 in this field disables access to the remote I2C slave.
			0	SLAVE AUTO ACK 7	R/W	0x0	Automatically acknowledge all I2C writes to the remote slave 7 independent of the forward channel lock state or status of the remote serializer acknowledge. 1: Enable 0: Disable
RX	0x6D	PORT_CONFIG	7:3	RESERVED	R/W	0x0F	Reserved
			2	COAX_MODE	(R/W)/S	0x0	Enable coax cable mode 0: Shielded twisted pair (STP) mode 1: Coax mode This bit is loaded from the MODE pin strap at power-up.
			1:0	FPD3_MODE	(R/W)/S	0x0	FPD3 input mode 00: Reserved 01: RAW12 LF mode 10: RAW12 HF mode 11: RAW10 mode This field is loaded from the MODE pin strap at power-up.
RX	0x6E	BC_GPIO_CTL0	7:4	BC_GPIO1_SEL	R/W	0x8	Back channel GPIO1 select: Determines the data sent on GPIO1 for the port back channel. 0000 : GPIO Pin 0 0001 : GPIO Pin 1 0010 : GPIO Pin 2 0011 : GPIO Pin 3 0100 - 0111 : Reserved 1000 : Constant value of 0 1001 : Constant value of 1 1010 : FrameSync signal 1011 - 1111 : Reserved
			3:0	BC_GPIO0_SEL	R/W	0x8	Back channel GPIO0 Select: Determines the data sent on GPIO0 for the port back channel. 0000 : GPIO Pin 0 0001 : GPIO Pin 1 0010 : GPIO Pin 2 0011 : GPIO Pin 3 0100 - 0111 : Reserved 1000 : Constant value of 0 1001 : Constant value of 1 1010 : FrameSync signal 1011 - 1111 : Reserved



**Table 9. Serial Control Bus Registers (continued)**

Page	Addr (hex)	Register Name	Bit(s)	Field	Type	Default	Description
RX	0x6F	BC_GPIO_CTL1	7:4	BC_GPIO3_SEL	R/W	0x8	Back channel GPIO3 select: Determines the data sent on GPIO3 for the port back channel. 0000 : GPIO Pin 0 0001 : GPIO Pin 1 0010 : GPIO Pin 2 0011 : GPIO Pin 3 0100 - 0111 : Reserved 1000 : Constant value of 0 1001 : Constant value of 1 1010 : FrameSync signal 1011 - 1111 : Reserved
			3:0	BC_GPIO2_SEL	R/W	0x8	Back channel GPIO2 select: Determines the data sent on GPIO2 for the port back channel. 0000 : GPIO Pin 0 0001 : GPIO Pin 1 0010 : GPIO Pin 2 0011 : GPIO Pin 3 0100 - 0111 : Reserved 1000 : Constant value of 0 1001 : Constant value of 1 1010 : FrameSync signal 1011 - 1111 : Reserved
RX	0x70 - 0x76	RESERVED	7:0	RESERVED	R/W	0x00	Reserved
RX	0x77	FREQ_DET_CTL	7:6	FREQ_HYST	R/W	0x3	Frequency detect hysteresis: The frequency detect hysteresis controls reporting of the FPD3 Clock frequency stability via the FREQ_STABLE status in the RX_PORT_STS2 register. The frequency is considered stable when the frequency remains within a range of +/- the FREQ_HYST value from the previous measurement. The FREQ_HYST setting is in MHz.
			5:4	FREQ_STABLE_THR	R/W	0x0	Frequency stability threshold: The frequency detect circuit can be used to detect a stable clock frequency. The stability threshold determines the amount of time required for the clock frequency to stay within the FREQ_HYST range to be considered stable: 00 : 40 $\mu$ s 01 : 80 $\mu$ s 10 : 320 $\mu$ s 11 : 1.28 ms
			3:0	FREQ_LO_THR	R/W	0x5	Frequency low threshold: Sets the low threshold for the clock frequency detect circuit in MHz. If the input clock is below this threshold, the NO_FPD3_CLK status is set to 1.
RX	0x78	MAILBOX_1	7:0	MAILBOX_0	R/W	0x0	Mailbox register This register is an unused read/write register that can be used for any purpose such as passing messages between I2C masters on opposite ends of the link.
RX	0x79	MAILBOX_2	7:0	MAILBOX_1	R/W	0x01	Mailbox register This register is an unused read/write register that can be used for any purpose such as passing messages between I2C masters on opposite ends of the link.
RX	0x7A - 0x7F	RESERVED	7:0	RESERVED	R	0x0	Reserved

**Table 9. Serial Control Bus Registers (continued)**

Page	Addr (hex)	Register Name	Bit(s)	Field	Type	Default	Description
Share	0xB0	IND_ACC_CTL	7:6	RESERVED	R	0x0	Reserved
			5:2	IA_SEL	R/W	0x0	Indirect Access register select: Selects target for register access 0000 : Reserved 0001 : FPD3 RX Port 0 registers 0010 : FPD3 RX Port 1 registers 0011 : Reserved 0100 : Reserved 0101 : FPD3 RX Shared registers 0110 : Simultaneous write to FPD3 RX Port 0-1 registers 0111 : Reserved
			1	IA_AUTO_INC	R/W	0x0	Indirect access auto increment: Enables auto-increment mode. Upon completion of a read or write, the register address automatically increments by 1
			0	IA_READ	R/W	0x0	Indirect access read: Setting this allows generation of a read strobe to the selected register block upon setting of the IND_ACC_ADDR register. In auto-increment mode, read strobes is also asserted following a read of the IND_ACC_DATA register. This function is only required for blocks that need to pre-fetch register data.
Share	0xB1	IND_ACC_ADDR	7:0	IA_ADDR	R/W	0x0	Indirect access register offset: This register contains the 8-bit register offset for the indirect access.
Share	0xB2	IND_ACC_DATA	7:0	IA_DATA	R/W	0x0	Indirect access data: Writing this register causes an indirect write of the IND_ACC_DATA value to the selected analog block register. Reading this register returns the value of the selected block register
Share	0xB3	BIST Control	7:6	BIST_OUT_MODE	R/W	0x0	BIST output mode 00 : No toggling 01 : Alternating 1/0 toggling 1x : Toggle based on BIST data
			5:4	RESERVED	R/W	0x0	Reserved
			3	BIST PIN CONFIG	R/W	0x1	BIST Configured through pin 1: BIST configured through pin 0: BIST configured through bits 2:0 in this register
			2:1	BIST CLOCK SOURCE	R/W	0x0	BIST Clock Source This register field selects the BIST clock source at the Serializer. These register bits are automatically written to the CLOCK SOURCE bits (register offset 0x14) in the serializer after BIST is enabled. See the appropriate serializer register descriptions for details. Note: When connected to a DS90UB913A or DS90UB933, a setting of 0x3 may result in a clock frequency that is too slow for proper recovery.
			0	BIST_EN	R/W	0x0	BIST Control 1: Enabled 0: Disabled
Share	0xB8	MODE_IDX_STS	7	IDX_DONE	R	0x1	IDX Done: If set, indicates the IDX decode has completed and latched into the IDX status bits.
			6:4	IDX	R	0x0	IDX Decode 3-bit decode from IDX pin

**Table 9. Serial Control Bus Registers (continued)**

Page	Addr (hex)	Register Name	Bit(s)	Field	Type	Default	Description
			3	MODE_DONE	R	0x1	MODE Done: If set, indicates the MODE decode has completed and latched into the MODE status bits.
			2:0	MODE	R	0x0	MODE Decode 3-bit decode from MODE pin
Share	0xBE	GPIO_PD_CTL	7:3	RESERVED	R/W	0x0	Reserved
			2	GPIO2_PD_DIS	R/W	0x0	GPIO2 pulldown resistor disable: The GPIO pins by default include a pulldown resistor that is automatically enabled when the GPIO is not in an output mode. When this bit is set, the pulldown resistor is also disabled when the GPIO pin is in an input only mode. 1 : Disable GPIO pulldown resistor 0 : Enable GPIO pulldown resistor
			1	GPIO1_PD_DIS	R/W	0x0	GPIO1 pulldown resistor disable: The GPIO pins by default include a pulldown resistor that is automatically enabled when the GPIO is not in an output mode. When this bit is set, the pulldown resistor is also disabled when the GPIO pin is in an input only mode. 1 : Disable GPIO pulldown resistor 0 : Enable GPIO pulldown resistor
			0	GPIO0_PD_DIS	R/W	0x0	GPIO0 pulldown resistor disable: The GPIO pins by default include a pulldown resistor that is automatically enabled when the GPIO is not in an output mode. When this bit is set, the pulldown resistor is also disabled when the GPIO pin is in an input only mode. 1 : Disable GPIO pulldown resistor 0 : Enable GPIO pulldown resistor
RX	0xD0	PORT_DEBUG	7:6	RESERVED	R/W	0x0	Reserved
			5	SER_BIST_ACT	R	0x0	Serializer BIST Active This register indicates whether the serializer is in BIST mode. 0: BIST mode not active 1: BIST mode active If the deserializer is not in BIST mode, this bit being 1 could indicate an error condition.
			4:2	RESERVED	R/W	0x0	Reserved
			1	FORCE_BC_ERRORS	R/W	0x0	This bit introduces continuous errors into the back channel frame.
			0	FORCE_1_BC_ERROR	(R/W)/S C	0x0	This bit introduces typically one, worst case two, errors into the back channel frame. Self clearing bit.
RX	0xD8	PORT_ICR_HI	7:3	Reserved	R	0x0	Reserved
			2	IE_FPD3_ENC_ERR	R/W	0x0	Interrupt on FPD-Link III receiver encoding error When enabled, an interrupt is generated on detection of an encoding error on the FPD-Link III interface for the receive port as reported in the FPD3_ENC_ERROR bit in the RX_PORT_STS2 register
			1	IE_BCC_SEQ_ERR	R/W	0x0	Interrupt on BCC SEQ sequence error When enabled, an interrupt is generated if a sequence error is detected for the bidirectional control channel forward channel receiver as reported in the BCC_SEQ_ERROR bit in the RX_PORT_STS1 register.

**Table 9. Serial Control Bus Registers (continued)**

Page	Addr (hex)	Register Name	Bit(s)	Field	Type	Default	Description
			0	IE_BCC_CRC_ERR	R/W	0x0	Interrupt on BCC CRC error detect When enabled, an interrupt is generated if a CRC error is detected on a bidirectional control channel frame received over the FPD-Link III forward channel as reported in the BCC_CRC_ERROR bit in the RX_PORT_STS1 register.
RX	0xD9	PORT_ICR_LO	7:3	RESERVED	R/W	0x0	Reserved
			2	IE_FPD3_PAR_ERR	R/W	0x0	Interrupt on FPD-Link III receiver parity error When enabled, an interrupt is generated on detection of parity errors on the FPD-Link III interface for the receive port. Parity error status is reported in the PARITY_ERROR bit in the RX_PORT_STS1 register.
			1	IE_PORT_PASS	R/W	0x0	Interrupt on change in port PASS status When enabled, an interrupt is generated on a change in receiver port valid status as reported in the PORT_PASS bit in the PORT_STS1 register.
			0	IE_LOCK_STS	R/W	0x0	Interrupt on change in lock status When enabled, an interrupt is generated on a change in lock status. Status is reported in the LOCK_STS_CHG bit in the RX_PORT_STS1 register.
RX	0xDA	PORT_ISR_HI	7:3	Reserved	R	0x0	Reserved
			2	IS_FPD3_ENC_ERR	R	0x0	FPD-Link III receiver encode error interrupt status An encoding error on the FPD-Link III interface for the receive port has been detected. Status is reported in the FPD3_ENC_ERROR bit in the RX_PORT_STS2 register. This interrupt condition is cleared by reading the RX_PORT_STS2 register.
			1	IS_BCC_SEQ_ERR	R	0x0	BCC CRC sequence error interrupt status A sequence error has been detected for the bidirectional control channel forward channel receiver. Status is reported in the BCC_SEQ_ERROR bit in the RX_PORT_STS1 register. This interrupt condition is cleared by reading the RX_PORT_STS1 register.
			0	IS_BCC_CRC_ERR	R	0x0	BCC CRC error detect interrupt status A CRC error has been detected on a bidirectional control channel frame received over the FPD-Link III forward channel. Status is reported in the BCC_CRC_ERROR bit in the RX_PORT_STS1 register. This interrupt condition is cleared by reading the RX_PORT_STS1 register.
RX	0xDB	PORT_ISR_LO	7:3	Reserved	R	0x0	Reserved
			2	IS_FPD3_PAR_ERR	R	0x0	FPD-Link III receiver parity error interrupt status A parity error on the FPD-Link III interface for the receive port has been detected. Parity error status is reported in the PARITY_ERROR bit in the RX_PORT_STS1 register. This interrupt condition is cleared by reading the RX_PORT_STS1 register.

**Table 9. Serial Control Bus Registers (continued)**

Page	Addr (hex)	Register Name	Bit(s)	Field	Type	Default	Description
			1	IS_PORT_PASS	R	0x0	Port valid interrupt status A change in receiver port valid status as reported in the PORT_PASS bit in the PORT_STS1 register. This interrupt condition is cleared by reading the RX_PORT_STS1 register.
			0	IS_LOCK_STS	R	0x0	Lock interrupt status A change in lock status has been detected. Status is reported in the LOCK_STS_CHG bit in the RX_PORT_STS1 register. This interrupt condition is cleared by reading the RX_PORT_STS1 register.
Share	0xF0	FPD3_RX_ID0	7:0	FPD3_RX_ID0	R	0x5F	FPD3_RX_ID0: First byte ID code: '_'
Share	0xF1	FPD3_RX_ID1	7:0	FPD3_RX_ID1	R	0x55	FPD3_RX_ID1: 2nd byte of ID code: 'U'
Share	0xF2	FPD3_RX_ID2	7:0	FPD3_RX_ID2	R	0x42	FPD3_RX_ID2: 3rd byte of ID code: 'B'
Share	0xF3	FPD3_RX_ID3	7:0	FPD3_RX_ID3	R	0x39	FPD3_RX_ID3: 4th byte of ID code: '9'
Share	0xF4	FPD3_RX_ID4	7:0	FPD3_RX_ID4	R	0x33	FPD3_RX_ID4: 5th byte of ID code: '3'
Share	0xF5	FPD3_RX_ID5	7:0	FPD3_RX_ID5	R	0x34	FPD3_RX_ID5: 6th byte of ID code: '4'
Share	0xF8	I2C_RX0_ID	7:1	RX_PORT0_ID	R/W	0x00	7-bit Receive Port 0 I2C ID Configures the decoder for detecting transactions designated for Receiver port 0 registers. This provides a simpler method of accessing device registers specifically for port 0 without having to use the paging function to select the register page. A value of 0 in this field disables the Port0 decoder.
			0	RESERVED	R	0x0	Reserved
Share	0xF9	I2C_RX1_ID	7:1	RX_PORT1_ID	R/W	0x00	7-bit Receive Port 1 I2C ID Configures the decoder for detecting transactions designated for Receiver port 1 registers. This provides a simpler method of accessing device registers specifically for port 1 without having to use the paging function to select the register page. A value of 0 in this field disables the Port1 decoder.
			0	RESERVED	R	0x0	Reserved

## 8 Application and Implementation

### NOTE

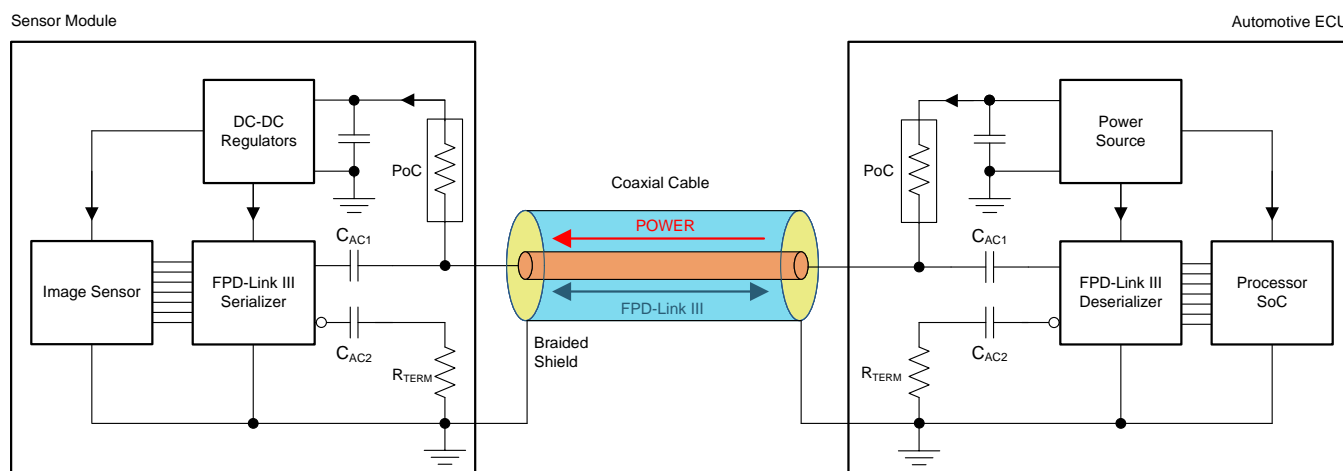
Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

### 8.1 Application Information

The DS90UB933/934 chipset supports video transport and bidirectional control over a single coaxial or STP cable targeted at ADAS applications, such as front, rear, and surround-view cameras, camera monitoring systems, and sensor fusion.

### 8.2 Power Over Coax

The DS90UB34-Q1 is designed to support the Power-over-Coax (PoC) method of powering remote sensor systems. With this method, the power is delivered over the same medium (a coaxial cable) used for high-speed digital video data and bidirectional control and diagnostics data transmission. The method utilizes passive networks or filters that isolate the transmission line from the loading of the DC-DC regulator circuits and their connecting power traces on both sides of the link as shown in Figure 22.

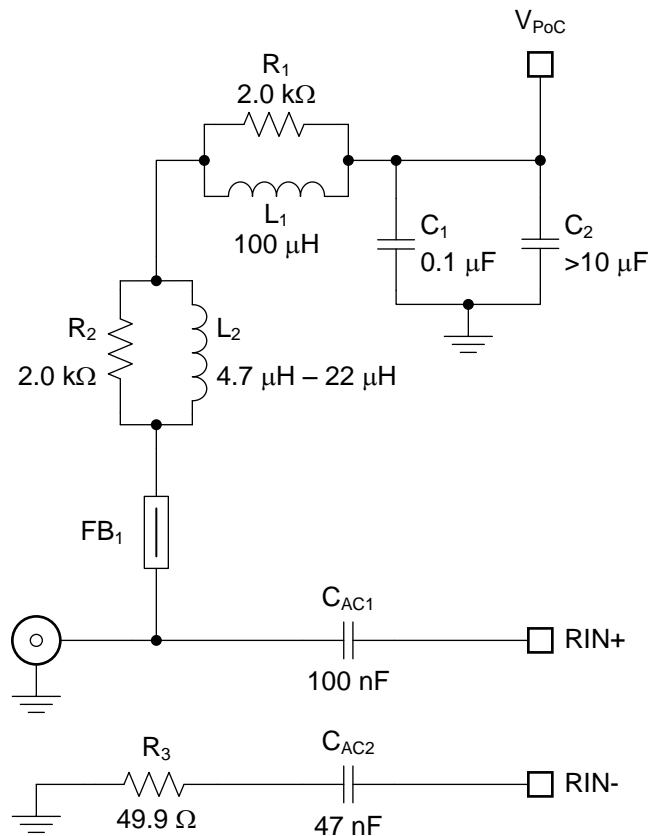


**Figure 22. Power Over Coax (PoC) System Diagram**

The PoC networks' impedance of  $\geq 2 \text{ k}\Omega$  over a specific frequency band is typically sufficient to isolate the transmission line from the loading of the regulator circuits. The lower limit of the frequency band is defined as  $\frac{1}{2}$  of the frequency of the bidirectional control channel,  $f_{BCC}$ . The upper limit of the frequency band is the frequency of the forward high-speed channel,  $f_{FC}$ .

Figure 23 shows a PoC network recommended for a FPD-Link III consisting of DS90UB913A-Q1/DS90UB933-Q1 and DS90UB934-Q1 pair with the bidirectional channel operating at 5 Mbps ( $\frac{1}{2} f_{BCC} = 2.5 \text{ MHz}$ ) and the forward channel operating at 1.87 Gbps ( $f_{FC} = 1 \text{ GHz}$ ).

## Power Over Coax (continued)



**Figure 23. Typical PoC Network for a 2G FPD-Link III**

Table 10 lists essential components for this particular PoC network.

**Table 10. Suggested Components for a 2G FPD-Link III PoC Network**

COUNT	REF DES	DESCRIPTION	PART NUMBER	MFR
1	L1	Inductor, 100 $\mu$ H, 0.310 $\Omega$ maximum, 710 mA minimum (Isat, Itemp) 7.2-MHz SRF typical, 6.6 mm $\times$ 6.6 mm, AEC-Q200	MSS7341-104ML	Coilcraft
1	L2	Inductor, 4.7 $\mu$ H, 0.350 $\Omega$ maximum, 700 mA minimum (Isat, Itemp) 160-MHz SRF typical, 3.8 mm $\times$ 3.8 mm, AEC-Q200	1008PS-472KL	Coilcraft
		Inductor, 4.7 $\mu$ H, 0.130 $\Omega$ maximum, 830 mA minimum (Isat, Itemp), 70-MHz SRF typical, 3.2 mm $\times$ 2.5 mm, AEC-Q200	CBC3225T4R7MRV	Taiyo Yuden
1	FB1	Ferrite Bead, 1500 k $\Omega$ at 1 GHz, 0.5 $\Omega$ maximum at DC 500-mA at 85°C, SM0603, General-Purpose	BLM18HE152SN1	Murata
		Ferrite Bead, 1500 k $\Omega$ at 1 GHz, 0.5 $\Omega$ maximum at DC 500-mA at 85°C, SM0603, AEC-Q200	BLM18HE152SZ1	Murata

Application report [Sending Power over Coax in DS90UB913A Designs](#) (SNLA224) discusses defining PoC networks in more detail.

In addition to the PoC network components selection, their placement and layout play a critical role as well.

- Place the smallest component, typically a ferrite bead or a chip inductor, as close to the connector as possible. Route the high-speed trace through one of its pads to avoid stubs.
- Use the smallest component pads as allowed by manufacturer's design rules. Add anti-pads in the inner planes below the component pads to minimize impedance drop.
- Consult with connector manufacturer for optimized connector footprint. If the connector is mounted on the same side as the IC, minimize the impact of the thru-hole connector stubs by routing the high-speed signal traces on the opposite side of the connector mounting side.
- Use coupled 100- $\Omega$  differential signal traces from the device pins to the AC-coupling caps. Use 50- $\Omega$  single-ended traces from the AC-coupling capacitors to the connector.
- Terminate the inverting signal traces close to the connectors with standard 49.9- $\Omega$  resistors.

The suggested characteristics for single-ended PCB traces (microstrips or striplines) for serializer or deserializer boards are detailed in [Table 11](#). The effects of the PoC networks need to be accounted for when testing the traces for compliance to the suggested limits.

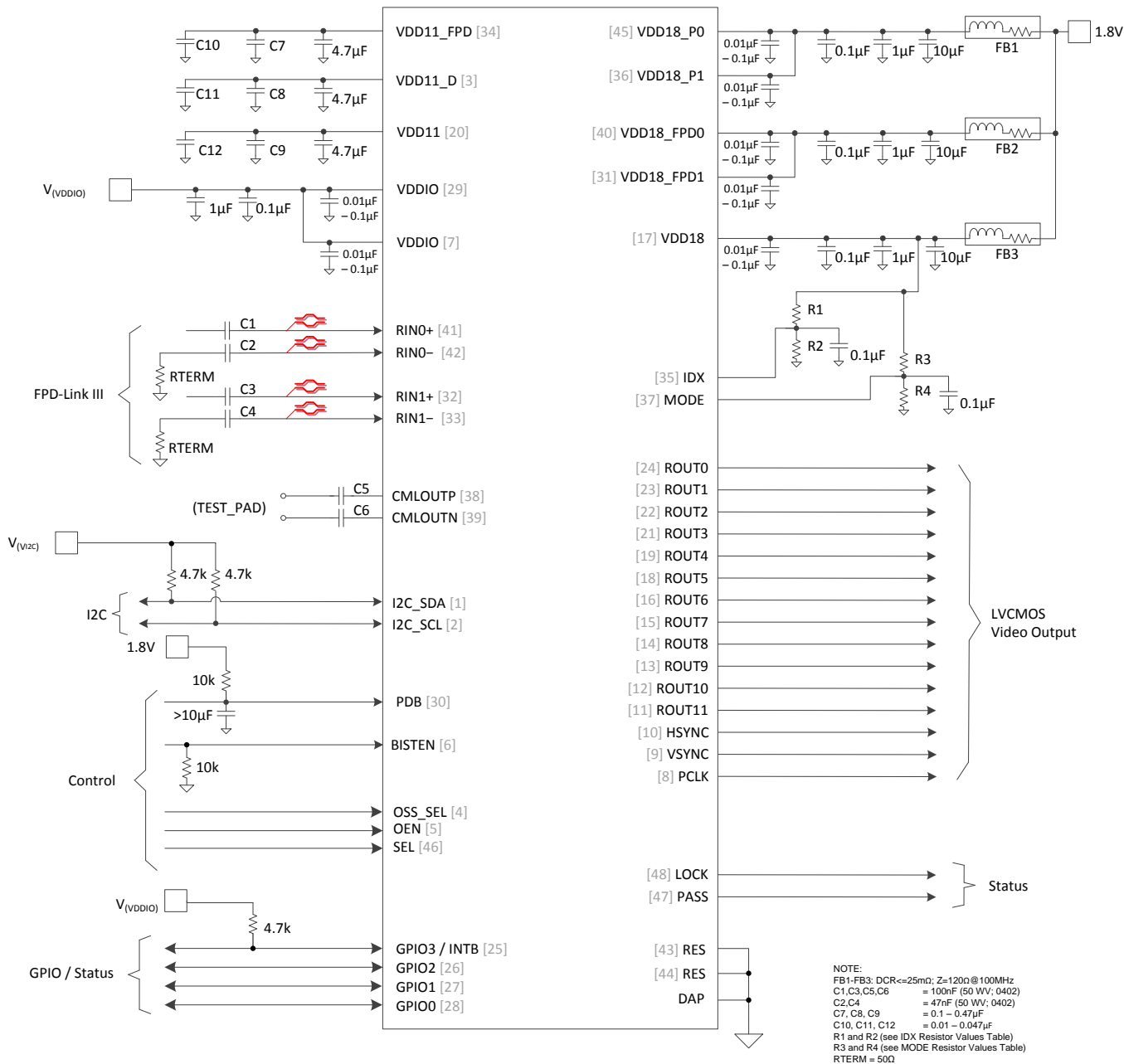
**Table 11. Suggested Characteristics for Single-Ended PCB Traces With Attached PoC Networks**

PARAMETER		MIN	TYP	MAX	UNIT
$L_{\text{trace}}$	Single-ended PCB trace length from the device pin to the connector pin			5	cm
$Z_{\text{trace}}$	Single-ended PCB trace characteristic impedance	45	50	55	$\Omega$
$Z_{\text{con}}$	Connector (mounted) characteristic impedance	40	50	60	$\Omega$
RL	Return Loss, S11	$\frac{1}{2} f_{\text{BCC}} < f < 0.1 \text{ GHz}$		-20	
		$0.1 \text{ GHz} < f < 1 \text{ GHz}$ (f in GHz)		$-12 + 8 \cdot \log(f)$	
IL	Insertion Loss, S12	$f < 0.5 \text{ GHz}$		-0.35	
		$f = 1 \text{ GHz}$		-0.6	



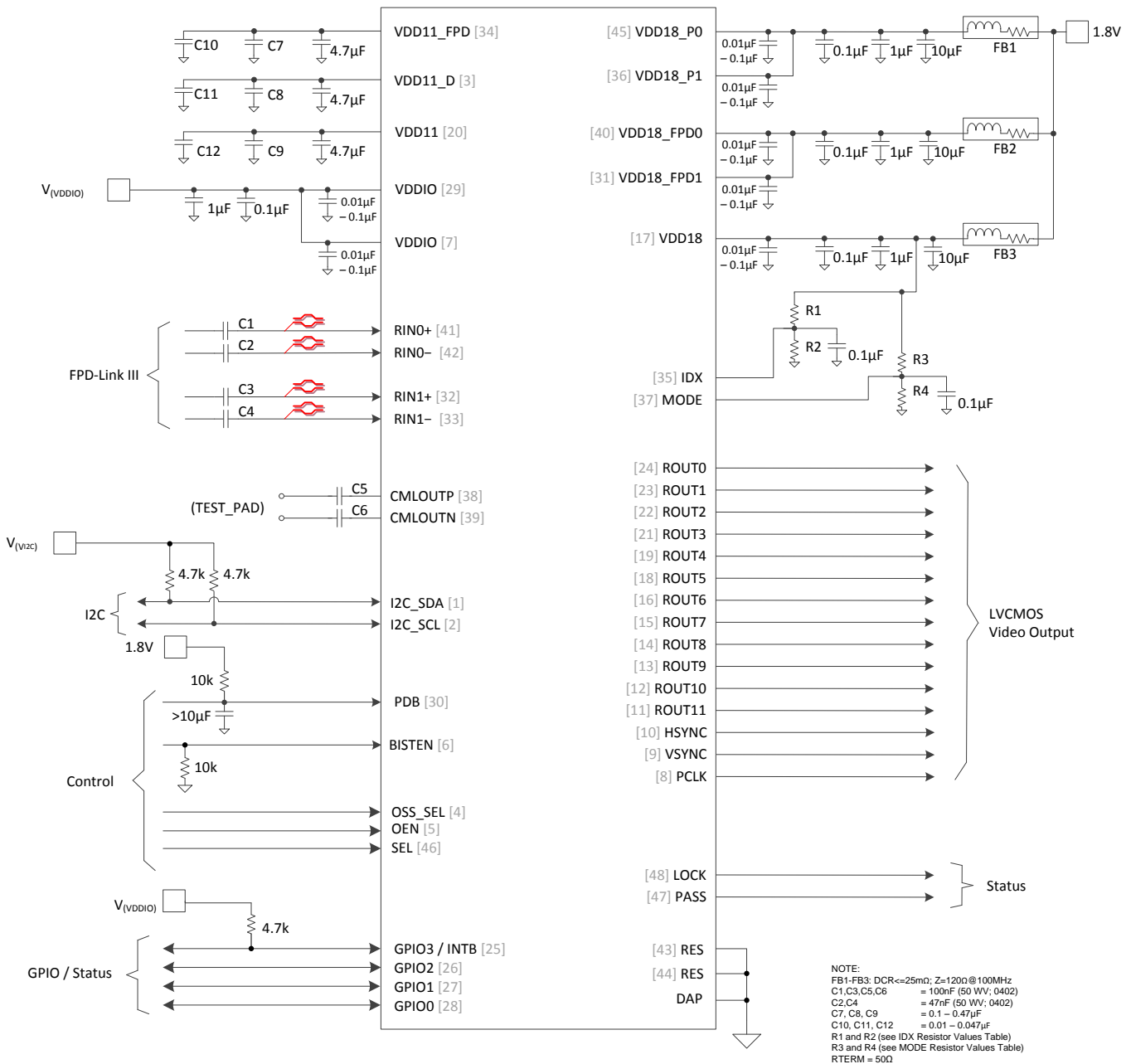
The  $V_{POC}$  noise needs to be kept to 10 mVp-p or lower on the source / deserializer side of the system. The  $V_{POC}$  fluctuations on the serializer side, caused by the transient current draw of the sensor and the DC resistance of cables and PoC components, need to be kept at minimum as well. Increasing the  $V_{POC}$  voltage and adding extra decoupling capacitance ( $> 10 \mu\text{F}$ ) help reduce the amplitude and slew rate of the  $V_{POC}$  fluctuations.

### 8.3 Typical Application



Copyright © 2017, Texas Instruments Incorporated

Figure 24. Typical Connection Diagram Coaxial

**Typical Application (continued)**

**Figure 25. Typical Connection Diagram STP**

## Typical Application (continued)

### 8.3.1 Design Requirements

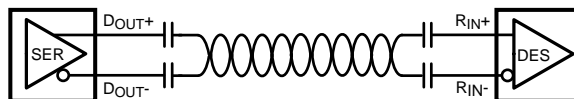
For the typical FPD-Link III serializer and deserializer applications, use the input parameters in [Table 12](#).

**Table 12. Design Parameters**

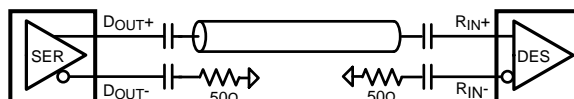
DESIGN PARAMETER	EXAMPLE VALUE
$V_{(VI2C)}$	1.8 V or 3.3 V
$V_{(VDD18)}$	1.8 V
AC-coupling capacitor for STP: $RIN[1:0]\pm$	100 nF (50 WV 0402)
AC-coupling capacitor for coaxial: $RIN[1:0]+$	100 nF (50 WV 0402)
AC-coupling capacitor for coaxial: $RIN[1:0]-$	47 nF (50 WV 0402)

### 8.3.2 Detailed Design Procedure

The serializer and deserializer support only AC-coupled interconnects through an integrated DC-balanced decoding scheme. External AC-coupling capacitors must be placed in series in the FPD-Link III signal path as shown in [Figure 26](#). For applications utilizing single-ended 50-Ω coaxial cable, terminate the unused data pins ( $RIN0-$ ,  $RIN1-$ ,  $RIN2-$ ,  $RIN3-$ ) with AC coupling capacitor and a 50-Ω resistor.



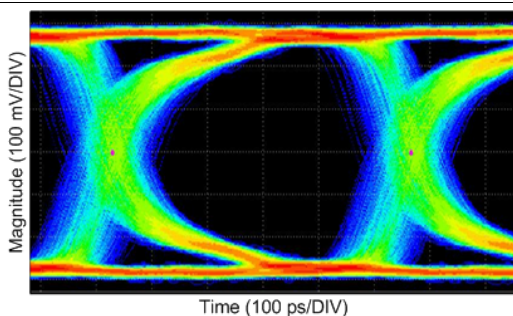
**Figure 26. AC-Coupled Connection (STP)**



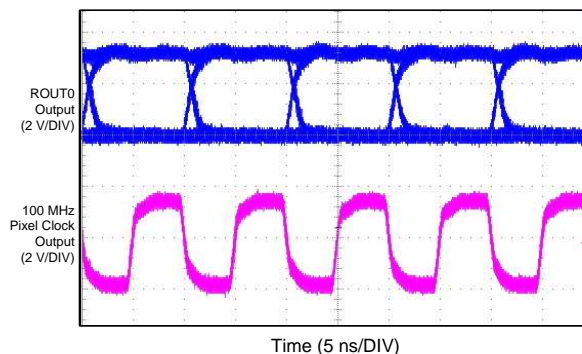
**Figure 27. AC-Coupled Connection (Coaxial)**

For high-speed FPD-Link III transmissions, use the smallest available package for the AC-coupling capacitor. This helps minimize degradation of signal quality due to package parasitics.

### 8.3.3 Application Curves



**Figure 28. CMLOUTP/N Loop-through Eye Diagram at  
1.867 Gbps  
15 Meters of DACAR 462 Cable**

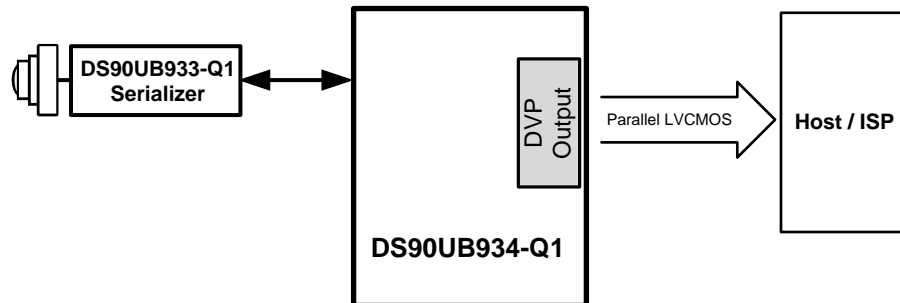


**Figure 29. ROUT0 Data Sampled by 100-MHz PCLK  
RRFB (0x3B[0]) = 1**

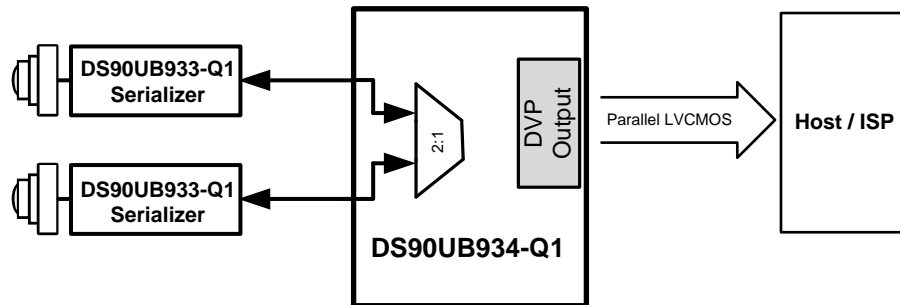
## 8.4 System Examples

The DS90UB934-Q1 has two input ports that operate as a multiplexer controlled by the SEL pin. A single camera can be connected to either Rx input port 0 or Rx input port 1 ([Figure 30](#)).

Two cameras can be connected simultaneously, but only one is active at a time ([Figure 31](#)). The SEL pin can be toggled on-the-fly to select which camera is forwarded to the DVP output.



**Figure 30. DS90UB933-Q1 Camera Data to 1 Rx Port**



Copyright © 2016, Texas Instruments Incorporated

**Figure 31. Two DS90UB933-Q1 Camera Data to 2 Rx Ports**

## 9 Power Supply Recommendations

This device provides separate power and ground pins for different portions of the circuit. This is done to isolate switching noise effects between different sections of the circuit. Separate planes on the PCB are typically not required. provide guidance on which circuit blocks are connected to which power pin pairs. In some cases, an external filter may be used to provide clean power to sensitive circuits such as PLLs.

### 9.1 VDD Power Supply

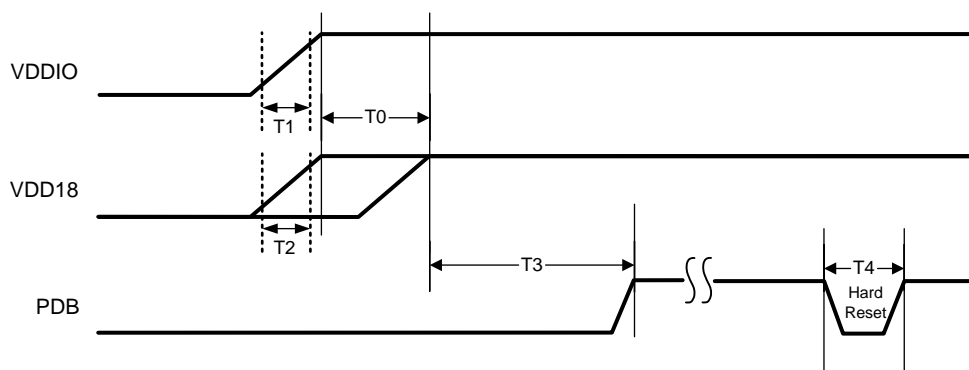
Each VDD power supply pin must have a 10-nF capacitor to ground connected as close as possible to the DS90UB934-Q1 device. TI recommends having additional decoupling capacitors (0.1  $\mu$ F, 1  $\mu$ F, and 10  $\mu$ F) on it. It is also recommended to have the pins connected to a solid power plane.

### 9.2 Power-Up Sequencing

All inputs must not be driven until both power supplies have reached steady state. The power-up sequence for the DS90UB934-Q1 is as follows:

**Table 13. Timing Diagram for the Power-Up Sequence**

PARAMETER	MIN	TYP	MAX	UNIT	NOTES
T0	$V_{(VDDIO)}$ to $V_{(VDD18)}$	0		ms	$V_{(VDDIO)}$ must come before (or at the same time as) $V_{(VDD18)}$
T1	$V_{(VDDIO)}$ rise time	1		ms	rise time = 10/90%
T2	$V_{(VDD18)}$ rise time	1		ms	rise time = 10/90%
T3	$V_{(VDDIO)}$ / $V_{(VDD18)}$ stable to PDB	0		ms	PDB = H must come after supplies are stable
T4	PDB pulse width	2		ms	Hard reset



**Figure 32. Power-Up Sequencing**

### 9.3 PDB Pin

The PDB pin is internal pull down enabled with 50k Ohm resistor. It is active HIGH and must remain LOW until the power supplies are within the recommended operating conditions. An external RC network on the PDB pin may be connected to ensure PDB arrives after all the supply pins have settled to the recommended operating voltage. When PDB pin is pulled up to VDD18, a 10-k $\Omega$  pullup and a >10- $\mu$ F capacitor to GND are required to delay the PDB input signal rise.

### 9.4 Ground

TI recommends that common ground plane be used in the design. This provides the best image plane for signal traces running above the plane. Connect the thermal pad of the DS90UB934-Q1 to this plane with vias.

## 10 Layout

### 10.1 Layout Guidelines

Circuit board layout and stack-up for the FPD-Link III devices must be designed to provide low-noise power feed to the device. Good layout practice also separates high-frequency or high-level inputs and outputs to minimize unwanted stray noise pickup, feedback, and interference. Power system performance may be greatly improved by using thin dielectrics (2 to 4 mils) for power/ground sandwiches. This arrangement provides plane capacitance for the PCB power system with low-inductance parasitics, which has proven especially effective at high frequencies and makes the value and placement of external bypass capacitors less critical. External bypassing should be low-ESR ceramic capacitors with high-quality dielectric. Voltage rating of the tantalum capacitors must be at least 5× the power supply voltage being used.

TI recommends surface mount capacitors due to their smaller parasitics. When using multiple capacitors per supply pin, locate the smaller value closer to the pin. A large bulk capacitor is recommended at the point of power entry. This is typically in the 47- $\mu$ F to 100- $\mu$ F range and smooths low frequency switching noise. TI recommends connecting power and ground pins directly to the power and connecting ground planes with bypass capacitors to the plane with via on both ends of the capacitor. Connecting power or ground pins to an external bypass capacitor increases the inductance of the path.

A small body size X7R chip capacitor, such as 0603 or 0402, is recommended for external bypass. Its small body size reduces the parasitic inductance of the capacitor. The user must pay attention to the resonance frequency of these external bypass capacitors, usually in the range of 20 to 30 MHz. To provide effective bypassing, multiple capacitors are often used to achieve low impedance between the supply rails over the frequency of interest. At high frequency, it is also a common practice to use two vias from power and ground pins to the planes, reducing the impedance at high frequency.

Some devices provide separate power and ground pins for different portions of the circuit. This is done to isolate switching noise effects between different sections of the circuit. Separate planes on the PCB are typically not required. Pin function tables typically provide guidance on which circuit blocks are connected to which power pin pairs. In some cases, an external filter may be used to provide clean power to sensitive circuits such as PLLs.

Use at least a four-layer board with a power and ground plane. Locate LVCMOS signals away from the differential lines to prevent coupling from the LVCMOS lines to the differential lines. Differential impedance of 100  $\Omega$  are typically recommended for STP interconnect and single-ended impedance of 50  $\Omega$  for coax interconnect. The closely coupled lines help to ensure that coupled noise appears as common-mode and thus is rejected by the receivers. The tightly coupled lines also radiate less.

#### 10.1.1 DVP Interface Guidelines

1. Route  $R_{OUT}[11:0]$  with controlled 50- $\Omega$  single-ended impedance ( $\pm 15\%$ ).
2. Keep away from other high speed signals.
3. Keep lengths to within 5 mils of each other.
4. Length matching must be near the location of mismatch.
5. Separate each signal by at least by 3 times the signal trace width.
6. Keep the use of bends in traces to a minimum. When bends are used, the number of left and right bends must be as equal as possible, and the angle of the bends must be  $\geq 135$  degrees. This arrangement minimizes any length mismatch caused by the bends, and therefore minimizes the impact that bends have on EMI.
7. Route all signals on the same layer
8. The number of vias should be kept to a minimum. TI recommends keeping the via count to 2 or fewer.
9. Keep traces on layers adjacent to ground plane.
10. Do NOT route signals over any GND plane split.
11. Adding test points causes impedance discontinuity and therefore negatively impacts signal performance. If test points are used, place them in series and symmetrically. They must not be placed in a manner that causes a stub.

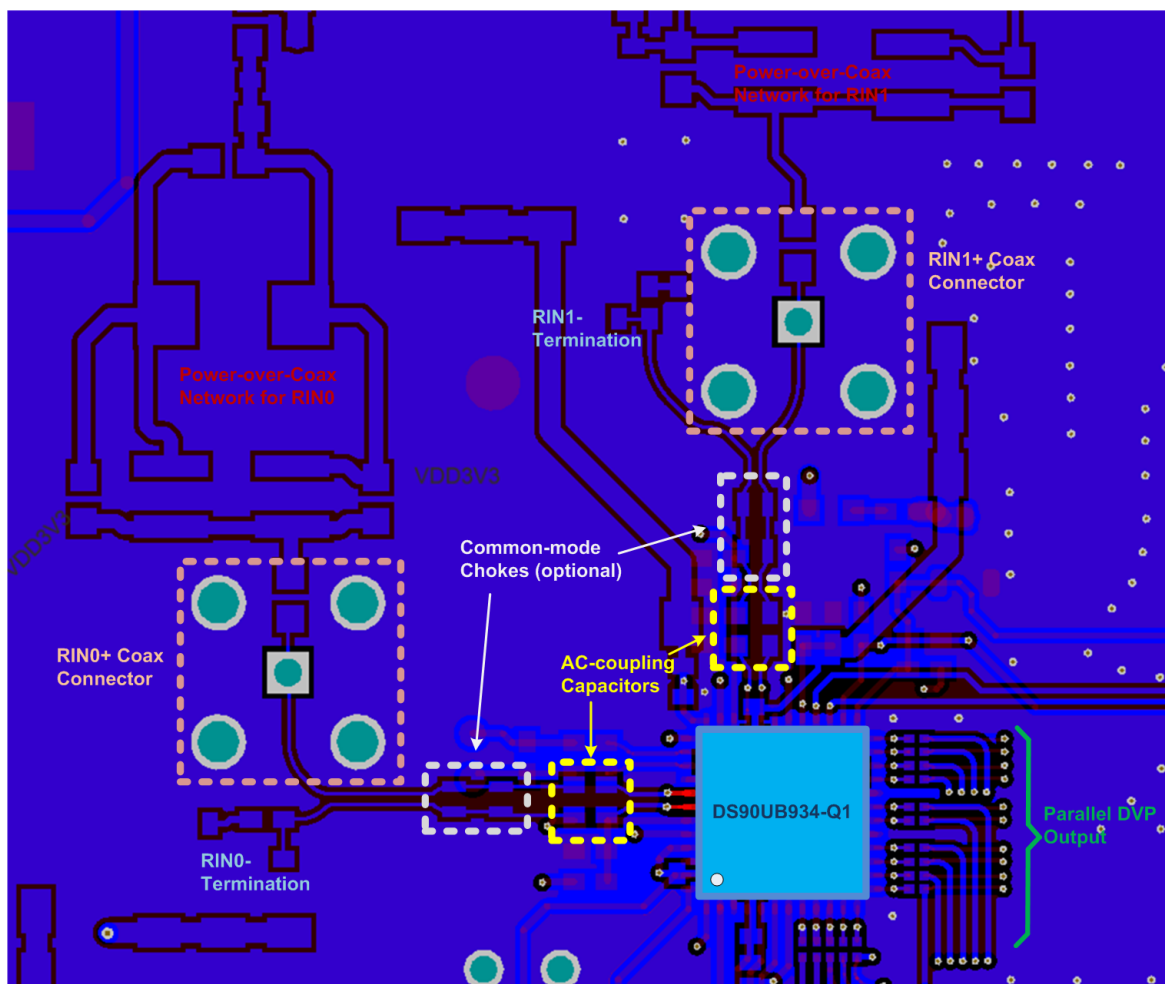
## 10.2 Layout Example

Stencil parameters such as aperture area ratio and the fabrication process have a significant impact on paste deposition. Inspection of the stencil prior to placement of the VQFN package is highly recommended to improve board assembly yields. If the via and aperture openings are not carefully monitored, the solder may flow unevenly through the DAP.

Figure 33 shows a PCB layout example derived from the layout design of the DS90UB934-Q1EVM Evaluation Board. The graphic and layout description are used to determine proper routing when designing the board. The FPD-Link III traces leading to RIN0+, RIN0–, RIN1+, RIN1– carry critical high-speed signals, and have highest priority in routing.

For STP applications, the positive and negative traces are tightly coupled with differential 100-Ω characteristic impedance.

For coaxial applications, the FPD-Link III traces must have 50-Ω characteristic impedance. As a secondary priority, loosely couple the traces with differential 100-Ω characteristic impedance.



**Figure 33. DS90UB934-Q1 Example PCB Layout**

1. Place vias, AC-coupling capacitors, and common-mode chokes (if used) on the FPD-Link III traces closely together so that the impedance discontinuity appears as tightly grouped as possible.
2. If PoC is used, place a ferrite bead placed as close as possible to the FPD-Link III trace to minimize the stub seen due to the filter network.
3. The high-speed FPD-Link III traces are routed differentially up to the connector. For the layout of a coaxial interconnects, use coupled traces with the RINx– termination near to the connector.



## 11 Device and Documentation Support

### 11.1 Documentation Support

#### 11.1.1 Related Documentation

For related documentation see the following:

- [DS90UB934-Q1EVM User's Guide](#)
- [FPD-Link Learning Center](#)
- [Backwards Compatibility Modes for Operation with Parallel Output Deserializers](#)
- [I2C over DS90UB913/4 FPD-Link III with Bidirectional Control Channel](#)
- [Sending Power Over Coax in DS90UB913A Designs](#)
- [I2C Bus Pullup Resistor Calculation](#)
- [Soldering Specifications Application Report](#)
- [Semiconductor and IC Package Thermal Metrics Application Report](#)
- [Leadless Leadframe Package \(LLP\) Application Report](#)
- [LVDS Owner's Manual](#)
- [An EMC/EMI System-Design and Testing Methodology for FPD-Link III SerDes](#)
- [Ten Tips for Successfully Designing with Automotive EMC/EMI Requirements](#)

#### 11.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

#### 11.3 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's [Terms of Use](#).

**TI E2E™ Online Community** *TI's Engineer-to-Engineer (E2E) Community*. Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

**Design Support** *TI's Design Support* Quickly find helpful E2E forums along with design support tools and contact information for technical support.

#### 11.4 Trademarks

E2E is a trademark of Texas Instruments.

All other trademarks are the property of their respective owners.

#### 11.5 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

#### 11.6 Glossary

[SLYZ022](#) — *TI Glossary*.

This glossary lists and explains terms, acronyms, and definitions.

## 12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.



## PACKAGING INFORMATION

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
DS90UB934TRGZRQ1	ACTIVE	VQFN	RGZ	48	2500	RoHS & Green	Call TI   NIPDAU	Level-3-260C-168 HR	-40 to 105	UB934Q	<a href="#">Samples</a>
DS90UB934TRGZTQ1	ACTIVE	VQFN	RGZ	48	250	RoHS & Green	Call TI   NIPDAU	Level-3-260C-168 HR	-40 to 105	UB934Q	<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

(3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.

(4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.

(5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.

(6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

**Important Information and Disclaimer:** The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.



**TAPE AND REEL INFORMATION**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
DS90UB934TRGZRQ1	VQFN	RGZ	48	2500	330.0	16.4	7.3	7.3	1.1	12.0	16.0	Q2
DS90UB934TRGZTQ1	VQFN	RGZ	48	250	180.0	16.4	7.3	7.3	1.1	12.0	16.0	Q2

## TAPE AND REEL BOX DIMENSIONS



\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
DS90UB934TRGZRQ1	VQFN	RGZ	48	2500	367.0	367.0	38.0
DS90UB934TRGZTQ1	VQFN	RGZ	48	250	210.0	185.0	35.0

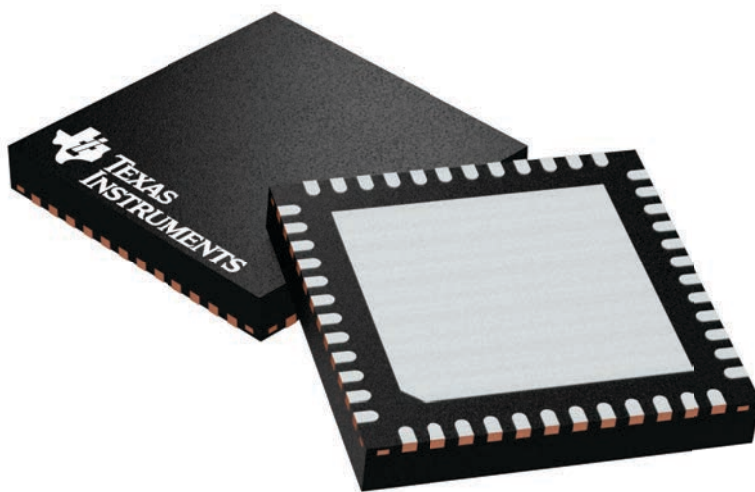
## GENERIC PACKAGE VIEW

**RGZ 48**

**VQFN - 1 mm max height**

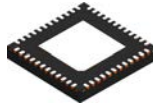
7 x 7, 0.5 mm pitch

PLASTIC QUADFLAT PACK- NO LEAD

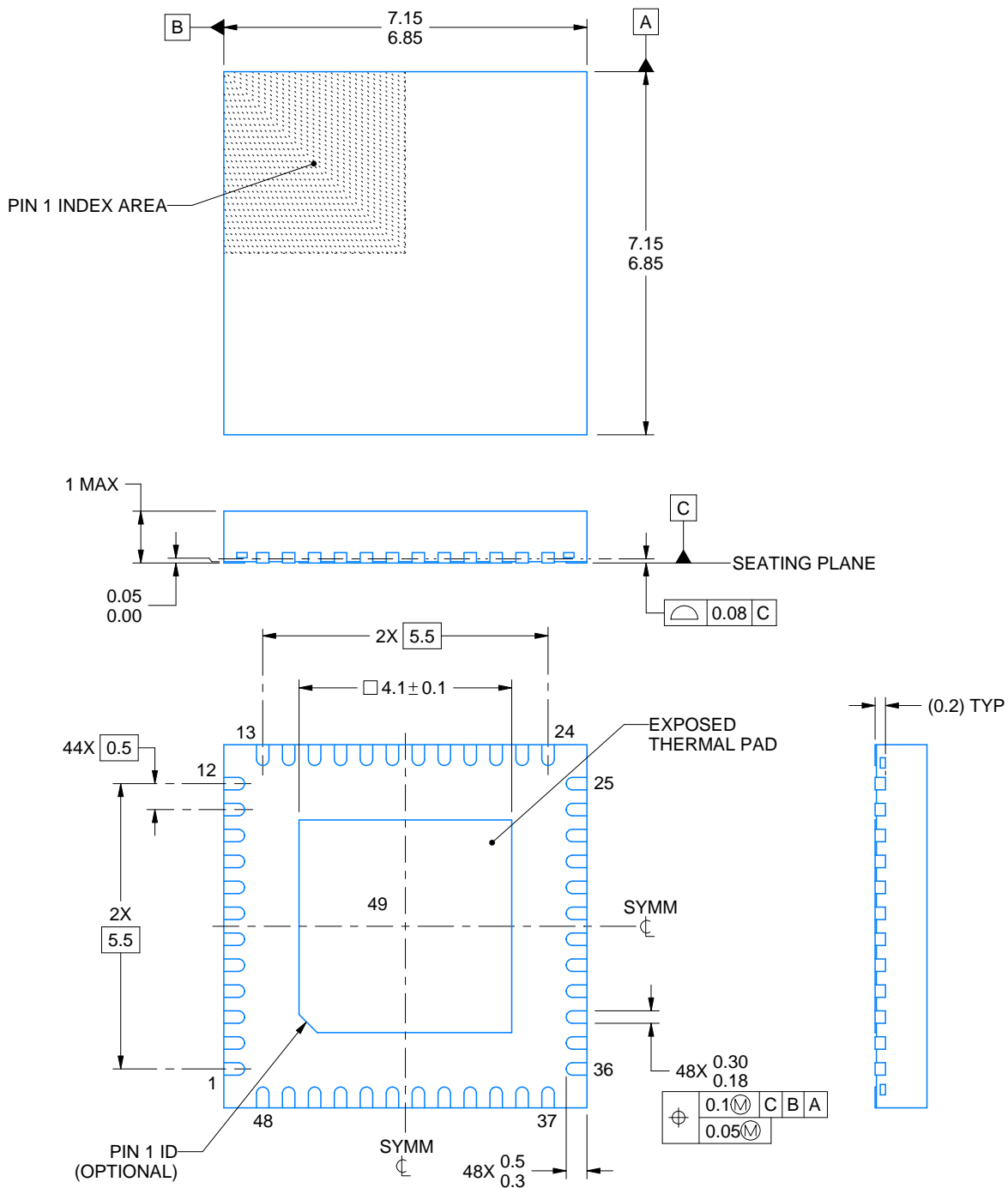


Images above are just a representation of the package family, actual package may vary.  
Refer to the product data sheet for package details.

4224671/A

**RGZ0048B****PACKAGE OUTLINE****VQFN - 1 mm max height**

PLASTIC QUAD FLATPACK - NO LEAD



4218795/B 02/2017

**NOTES:**

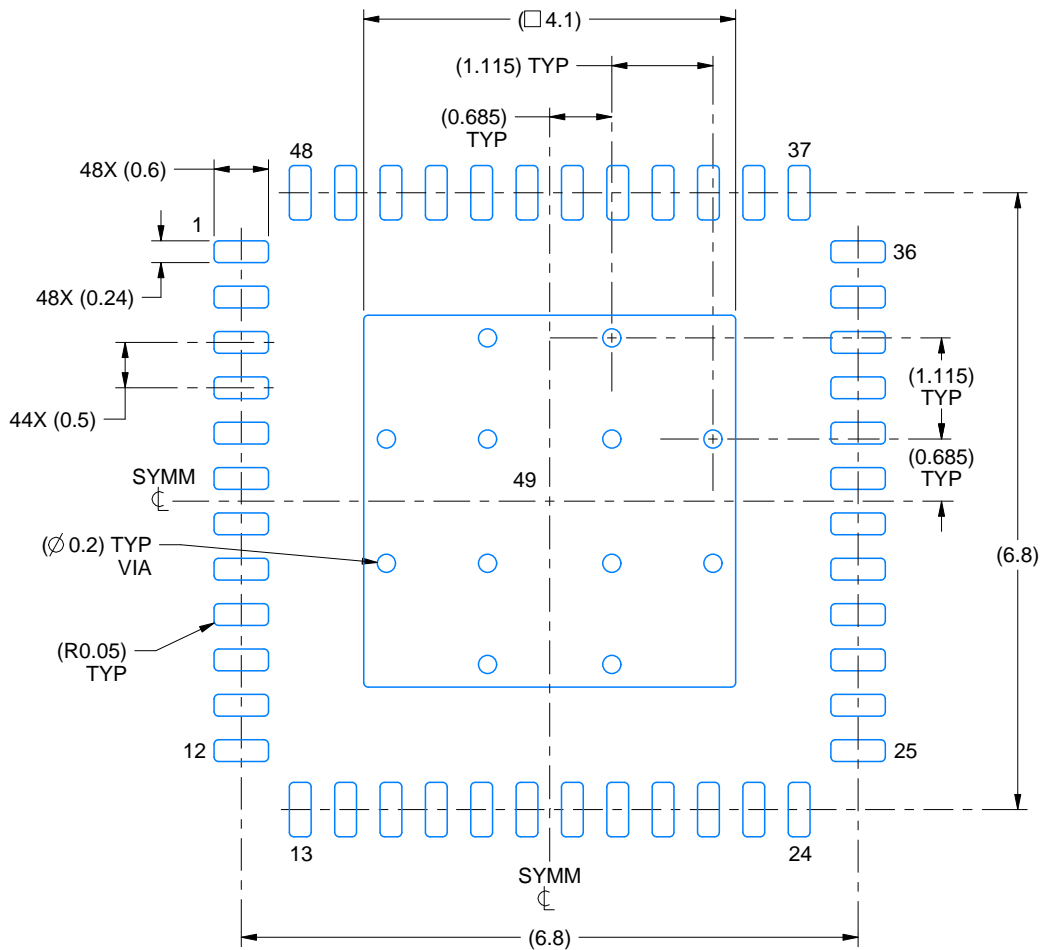
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.
3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.

# EXAMPLE BOARD LAYOUT

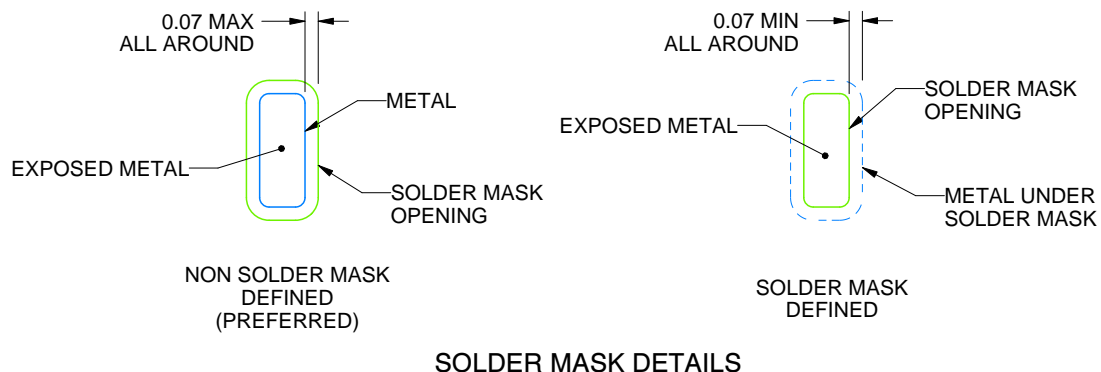
RGZ0048B

VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



LAND PATTERN EXAMPLE  
EXPOSED METAL SHOWN  
SCALE:12X



4218795/B 02/2017

NOTES: (continued)

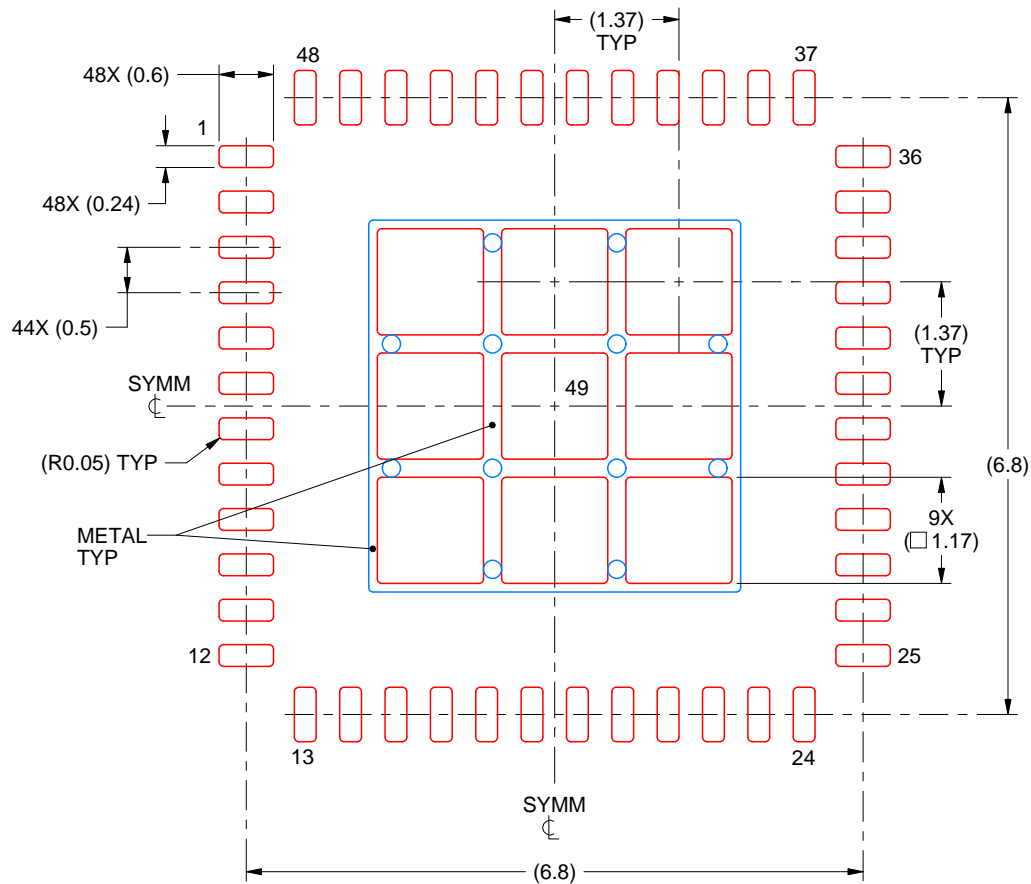
- This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 ([www.ti.com/lit/slue271](http://www.ti.com/lit/slue271)).
- Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.

# EXAMPLE STENCIL DESIGN

RGZ0048B

VQFN - 1 mm max height

PLASTIC QUAD FLATPACK - NO LEAD



## SOLDER PASTE EXAMPLE BASED ON 0.125 mm THICK STENCIL

EXPOSED PAD 49  
73% PRINTED SOLDER COVERAGE BY AREA UNDER PACKAGE  
SCALE:12X

4218795/B 02/2017

NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.



## IMPORTANT NOTICE AND DISCLAIMER

TI PROVIDES TECHNICAL AND RELIABILITY DATA (INCLUDING DATA SHEETS), DESIGN RESOURCES (INCLUDING REFERENCE DESIGNS), APPLICATION OR OTHER DESIGN ADVICE, WEB TOOLS, SAFETY INFORMATION, AND OTHER RESOURCES "AS IS" AND WITH ALL FAULTS, AND DISCLAIMS ALL WARRANTIES, EXPRESS AND IMPLIED, INCLUDING WITHOUT LIMITATION ANY IMPLIED WARRANTIES OF MERCHANTABILITY, FITNESS FOR A PARTICULAR PURPOSE OR NON-INFRINGEMENT OF THIRD PARTY INTELLECTUAL PROPERTY RIGHTS.

These resources are intended for skilled developers designing with TI products. You are solely responsible for (1) selecting the appropriate TI products for your application, (2) designing, validating and testing your application, and (3) ensuring your application meets applicable standards, and any other safety, security, regulatory or other requirements.

These resources are subject to change without notice. TI grants you permission to use these resources only for development of an application that uses the TI products described in the resource. Other reproduction and display of these resources is prohibited. No license is granted to any other TI intellectual property right or to any third party intellectual property right. TI disclaims responsibility for, and you will fully indemnify TI and its representatives against, any claims, damages, costs, losses, and liabilities arising out of your use of these resources.

TI's products are provided subject to [TI's Terms of Sale](#) or other applicable terms available either on [ti.com](#) or provided in conjunction with such TI products. TI's provision of these resources does not expand or otherwise alter TI's applicable warranties or warranty disclaimers for TI products.

TI objects to and rejects any additional or different terms you may have proposed.

Mailing Address: Texas Instruments, Post Office Box 655303, Dallas, Texas 75265  
Copyright © 2022, Texas Instruments Incorporated