

## LMV225/LMV226/LMV228 RF Power Detector for CDMA and WCDMA

Check for Samples: [LMV225](#), [LMV226](#), [LMV228](#)

### FEATURES

- 30 dB Linear in dB Power Detection Range
- Output Voltage Range 0.2 to 2V
- Logic Low Shutdown
- Multi-Band Operation from 450 MHz to 2000 MHz
- Accurate Temperature Compensation
- Packages:
  - DSBGA Thin 1.0 mm x 1.0 mm x 0.6 mm
  - DSBGA Ultra Thin 1.0 mm x 1.0 mm x 0.35 mm
  - WSON 2.2 mm x 2.5 mm x 0.8 mm
  - (LMV225 and LMV228)

### APPLICATIONS

- CDMA RF Power Control
- WCDMA RF Power Control
- CDMA2000 RF Power Control
- PA Modules

### DESCRIPTION

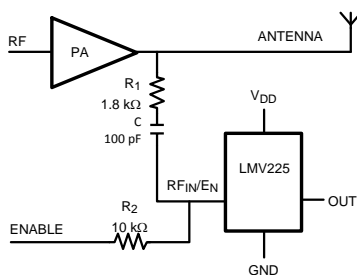
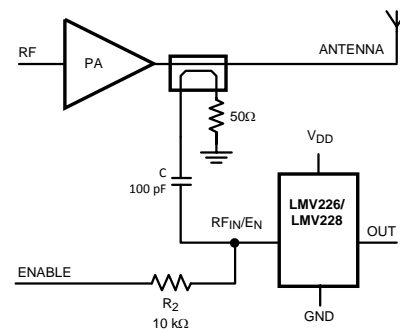
The LMV225/LMV226/LMV228 are 30 dB RF power detectors intended for use in CDMA and WCDMA applications. The device has an RF frequency range from 450 MHz to 2 GHz. It provides an accurate temperature and supply compensated output voltage that relates linearly to the RF input power in dBm. The circuit operates with a single supply from 2.7V to 5.5V. The LMV225/LMV226/LMV228 have an integrated filter for low-ripple average power detection of CDMA signals with 30 dB dynamic range. Additional filtering can be applied using a single external capacitor.

The LMV225 has an RF power detection range from –30 dBm to 0 dBm and is ideally suited for direct use in combination with resistive taps. The LMV226/LMV228 have a detection range from –15 dBm to 15 dBm and are intended for use in combination with a directional coupler. The LMV226 is equipped with a buffered output which makes it suitable for GSM, EDGE, GPRS and TDMA applications.

The device is active for Enable = HI, otherwise it is in a low power consumption shutdown mode. During shutdown the output will be LOW. The output voltage ranges from 0.2V to 2V and can be scaled down to meet ADC input range requirements.

The LMV225/LMV226/LMV228 power detectors are offered in the thin 1.0 mm x 1.0 mm x 0.6 mm DSBGA package and the ultra thin 1.0 mm x 1.0 mm x 0.35 mm DSBGA package. The LMV225 and the LMV228 are also offered in the 2.2 mm x 2.5 mm x 0.8 mm WSON package.

### Typical Application


**Figure 1. LMV225**

**Figure 2. LMV226/LMV228**


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These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

### ABSOLUTE MAXIMUM RATINGS <sup>(1)(2)</sup>

Supply Voltage	
$V_{DD} - GND$	6.0V Max
ESD Tolerance <sup>(3)</sup>	
Human Body Model	2000V
Machine Model	200V
Storage Temperature Range	
	-65°C to 150°C
Junction Temperature <sup>(4)</sup>	
	150°C Max
Mounting Temperature, Infrared or convection (20 sec)	
Tin/Lead	235°C
Lead-Free	260°C

- (1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not specified. For specifications and the test conditions, see the Electrical Characteristics.
- (2) If Military/Aerospace specified devices are required, please contact the Texas Instruments Sales Office/ Distributors for availability and specifications.
- (3) Human body model: 1.5 k $\Omega$  in series with 100 pF. Machine model, 0 $\Omega$  in series with 100 pF.
- (4) The maximum power dissipation is a function of  $T_{J(MAX)}$ ,  $\theta_{JA}$  and  $T_A$ . The maximum allowable power dissipation at any ambient temperature is  $P_D = (T_{J(MAX)} - T_A)/\theta_{JA}$ . All numbers apply for packages soldered directly into a PC board

### OPERATING RATINGS <sup>(1)</sup>

Supply Voltage	2.7V to 5.5V
Temperature Range	-40°C to +85°C
RF Frequency Range	450 MHz to 2 GHz

- (1) Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but specific performance is not specified. For specifications and the test conditions, see the Electrical Characteristics.

## 2.7 DC AND AC ELECTRICAL CHARACTERISTICS

Unless otherwise specified, all limits are specified to  $V_{DD} = 2.7V$ ;  $T_J = 25^\circ C$ . **Boldface** limits apply at temperature extremes. <sup>(1)</sup>

Symbol	Parameter	Condition	Min	Typ	Max	Units
$I_{DD}$	Supply Current	Active Mode: $RF_{IN}/E_N = V_{DD}$ (DC), No RF Input Power Present	LMV225	4.8	<b>7.8</b>	mA
			LMV226	4.9	<b>6.28</b>	
			LMV228	4.9	<b>6.28</b>	
		Shutdown: $RF_{IN}/E_N = GND$ (DC), No RF Input Power Present		0.44	<b>4.5</b>	$\mu A$
$V_{LOW}$	$E_N$ Logic Low Input Level <sup>(2)</sup>				<b>0.8</b>	V
$V_{HIGH}$	$E_N$ Logic High Input Level <sup>(2)</sup>		<b>1.8</b>			V
$t_{on}$	Turn-on-Time <sup>(3)</sup>	No RF Input Power Present, Output Loaded with 10 pF	LMV225	2.1		$\mu s$
			LMV226	1.2		
			LMV228	1.7		
$t_r$	Rise Time <sup>(4)</sup>	Step from no Power to 0 dBm Applied, Output Loaded with 10 pF	LMV225	4.5		$\mu s$
			LMV226	1.8		
		Step from no Power to 15 dBm Applied, Output Loaded with 10 pF	LMV228	4.8		
$I_{EN}$	Current into $RF_{IN}/E_N$ Pin				<b>1</b>	$\mu A$
$P_{IN}$	Input Power Range <sup>(5)</sup>	LMV225		-30 0		dBm
				-43 -13		dBV
		LMV226		-15 15		dBm
				-28 2		dBV
		LMV228		-15 15		dBm
				-28 2		dBV

(1) Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that  $T_J = T_A$ . No specification of parametric performance is indicated in the electrical tables under conditions of internal self-heating where  $T_J > T_A$ .

(2) All limits are specified by design or statistical analysis

(3) Turn-on time is measured by connecting a 10 k $\Omega$  resistor to the  $RF_{IN}/E_N$  pin. Be aware that in the actual application on the front page, the RC-time constant of resistor  $R_2$  and capacitor C adds an additional delay.

(4) Typical values represent the most likely parametric norm.

(5) Power in dBV = dBm + 13 when the impedance is 50 $\Omega$ .

### 2.7 DC AND AC ELECTRICAL CHARACTERISTICS (continued)

Unless otherwise specified, all limits are specified to  $V_{DD} = 2.7V$ ;  $T_J = 25^{\circ}C$ . **Boldface** limits apply at temperature extremes. <sup>(1)</sup>

Symbol	Parameter	Condition	Min	Typ	Max	Units
	Logarithmic Slope <sup>(6)</sup>	900 MHz	LMV225	44.0		mV/dB
			LMV226	44.5		
			LMV228 DSBGA	44.0		
			LMV228 WSON	48.5		
		1800 MHz	LMV225	39.4		
			LMV226	41.6		
			LMV228 DSBGA	41.9		
			LMV228 WSON	47.4		
		1900 MHz	LMV225	38.5		
			LMV226	41.2		
			LMV228 DSBGA	41.6		
			LMV228 WSON	46.6		
		2000 MHz	LMV225	38.5		
			LMV226	41.0		
			LMV228 DSBGA	41.2		
			LMV228 WSON	45.4		
	Logarithmic Intercept <sup>(6)</sup>	900 MHz	LMV225	-45.5		dBm
			LMV226	-24.5		
			LMV228 DSBGA	-27.2		
			LMV228 WSON	-23.7		
		1800 MHz	LMV225	-46.6		
			LMV226	-25.1		
			LMV228 DSBGA	-28.2		
			LMV228 WSON	-23.8		
		1900 MHz	LMV225	-46.3		
			LMV226	-24.9		
			LMV228 DSBGA	-28.0		
			LMV228 WSON	-23.7		
		2000 MHz	LMV225	-46.7		
			LMV226	-24.7		
			LMV228 DSBGA	-28.0		
			LMV228 WSON	-23.6		
V <sub>OUT</sub>	Output Voltage	No RF Input Power Present	LMV225	214	<b>350</b>	mV
			LMV226	223	<b>350</b>	
			LMV228	228	<b>350</b>	
I <sub>OUT</sub>	Output Current Sourcing/Sinking	LMV226 Only	<b>4.5</b>	5.3		mA
R <sub>OUT</sub>	Output Impedance	LMV225/LMV228 only, no RF Input Power Present		19.8	29 <b>34</b>	kΩ

(6) Device is set in active mode with a 10 kΩ resistor from  $V_{DD}$  to RF<sub>IN</sub>/E<sub>N</sub>. RF signal is applied using a 50Ω RF signal generator AC coupled to the RF<sub>IN</sub>/E<sub>N</sub> pin using a 100 pF coupling capacitor.

**2.7 DC AND AC ELECTRICAL CHARACTERISTICS (continued)**

 Unless otherwise specified, all limits are specified to  $V_{DD} = 2.7V$ ;  $T_J = 25^\circ C$ . **Boldface** limits apply at temperature extremes. <sup>(1)</sup>

Symbol	Parameter	Condition	Min	Typ	Max	Units
$e_n$	Output Referred Noise	RF Input = 1800 MHz, -10 dBm for LMV225 and 5 dBm for LMV226/LMV228, Measured at 10 kHz		700		$nV/\sqrt{Hz}$
	Variation Due to Temperature	900 MHz, $RF_{IN} = 0$ dBm Referred to 25°C	LMV225	+0.64 -1.07		dB
		900 MHz, $RF_{IN} = 15$ dBm Referred to 25°C	LMV226	+0.05 -0.02		
			LMV228 DSBGA	+0.22 -0.36		
			LMV228 WSON	+0.87 -0.87		
		1800 MHz, $RF_{IN} = 0$ dBm Referred to 25°C	LMV225	+0.09 -0.86		
		1800 MHz, $RF_{IN} = 15$ dBm Referred to 25°C	LMV226	+0.07 -0.10		
			LMV228 DSBGA	+0.29 -0.57		
			LMV228 WSON	+1.04 -1.23		
		1900 MHz, $RF_{IN} = 0$ dBm Referred to 25°C	LMV225	+0 -0.69		
		1900 MHz, $RF_{IN} = 15$ dBm Referred to 25°C	LMV226	+0 -0.10		
			LMV228 DSBGA	+0.23 -0.64		
			LMV228 WSON	+1.05 -1.45		
		2000 MHz, $RF_{IN} = 0$ dBm Referred to 25°C	LMV225	+0 -0.86		
		2000 MHz, $RF_{IN} = 15$ dBm Referred to 25°C	LMV226	+0 -0.29		
			LMV228 DSBGA	+0.27 -0.65		
			LMV228 WSON	+1.04 -2.02		

### 5.0 DC AND AC ELECTRICAL CHARACTERISTICS

Unless otherwise specified, all limits are specified to  $V_{DD} = 5.0V$ ;  $T_J = 25^\circ C$ . **Boldface** limits apply at temperature extremes. <sup>(1)</sup>

Symbol	Parameter	Condition	Min	Typ	Max	Units
$I_{DD}$	Supply Current	Active Mode: $RF_{IN}/E_N = V_{DD}$ (DC), no RF Input Power Present.	LMV225	5.3	<b>7.5</b> <b>9</b>	mA
			LMV226	5.3	<b>6.8</b> <b>9</b>	
			LMV228	5.4	<b>6.8</b> <b>9</b>	
		Shutdown: $RF_{IN}/E_N = GND$ (DC), no RF Input Power Present.			0.32	<b>4.5</b>
$V_{LOW}$	$E_N$ Logic Low Input Level <sup>(2)</sup>				<b>0.8</b>	V
$V_{HIGH}$	$E_N$ Logic High Input Level <sup>(2)</sup>		<b>1.8</b>			V
$t_{on}$	Turn-on-Time <sup>(3)</sup>	No RF Input Power Present, Output Loaded with 10 pF	LMV225	2.1		$\mu s$
			LMV226	1.0		
			LMV228	1.7		
$t_r$	Rise Time <sup>(4)</sup>	Step from no Power to 0 dBm Applied, Output Loaded with 10 pF	LMV225	4.5		$\mu s$
			LMV226	1.4		
		Step from no Power to 15 dBm Applied, Output Loaded with 10 pF	LMV228	4.8		
$I_{EN}$	Current Into $RF_{IN}/E_N$ Pin				<b>1</b>	$\mu A$
$P_{IN}$	Input Power Range <sup>(5)</sup>	LMV225		-30 0		<b>dBm</b>
				-43 -13		<b>dBV</b>
		LMV226		-15 15		<b>dBm</b>
				-28 2		<b>dBV</b>
		LMV228		-15 15		<b>dBm</b>
				-28 2		<b>dBV</b>

- (1) Electrical Table values apply only for factory testing conditions at the temperature indicated. Factory testing conditions result in very limited self-heating of the device such that  $T_J = T_A$ . No specification of parametric performance is indicated in the electrical tables under conditions of internal self-heating where  $T_J > T_A$ .
- (2) All limits are specified by design or statistical analysis
- (3) Turn-on time is measured by connecting a 10 k $\Omega$  resistor to the  $RF_{IN}/E_N$  pin. Be aware that in the actual application on the front page, the RC-time constant of resistor  $R_2$  and capacitor C adds an additional delay.
- (4) Typical values represent the most likely parametric norm.
- (5) Power in dBV = dBm + 13 when the impedance is 50 $\Omega$ .

**5.0 DC AND AC ELECTRICAL CHARACTERISTICS (continued)**

 Unless otherwise specified, all limits are specified to  $V_{DD} = 5.0V$ ;  $T_J = 25^\circ C$ . **Boldface** limits apply at temperature extremes. <sup>(1)</sup>

Symbol	Parameter	Condition	Min	Typ	Max	Units
	Logarithmic Slope <sup>(6)</sup>	900 MHz	LMV225	44.6		mV/dB
			LMV226	44.6		
			LMV228 DSBGA	44.2		
			LMV228 WSON	48.4		
		1800 MHz	LMV225	40.6		
			LMV226	42.2		
			LMV228 DSBGA	42.4		
			LMV228 WSON	48.3		
		1900 MHz	LMV225	39.6		
			LMV226	41.8		
			LMV228 DSBGA	42.2		
			LMV228 WSON	47.8		
		2000 MHz	LMV225	39.7		
			LMV226	41.6		
			LMV228 DSBGA	41.8		
			LMV228 WSON	47.2		
	Logarithmic Intercept <sup>(6)</sup>	900 MHz	LMV225	-47.0		dBm
			LMV226	-25.0		
			LMV228 DSBGA	-27.7		
			LMV228 WSON	-23.9		
		1800 MHz	LMV225	-48.5		
			LMV226	-25.7		
			LMV228 DSBGA	-28.9		
			LMV228 WSON	-23.6		
		1900 MHz	LMV225	-48.2		
			LMV226	-25.6		
			LMV228 DSBGA	-28.7		
			LMV228 WSON	-23.1		
		2000 MHz	LMV225	-48.9		
			LMV226	-25.5		
			LMV228 DSBGA	-28.7		
			LMV228 WSON	-23.0		
$V_{OUT}$	Output Voltage	No RF Input Power Present	LMV225	222	<b>400</b>	mV
			LMV226	231	<b>400</b>	
			LMV228	244	<b>400</b>	
$I_{OUT}$	Output Current Sourcing/Sinking	LMV226 Only	<b>4.5</b>	5.3		mA
$R_{OUT}$	Output Impedance	No RF Input Power Present		23.7	29 <b>31</b>	k $\Omega$

(6) Device is set in active mode with a 10 k $\Omega$  resistor from  $V_{DD}$  to  $RF_{IN}/E_N$ . RF signal is applied using a 50 $\Omega$  RF signal generator AC coupled to the  $RF_{IN}/E_N$  pin using a 100 pF coupling capacitor.

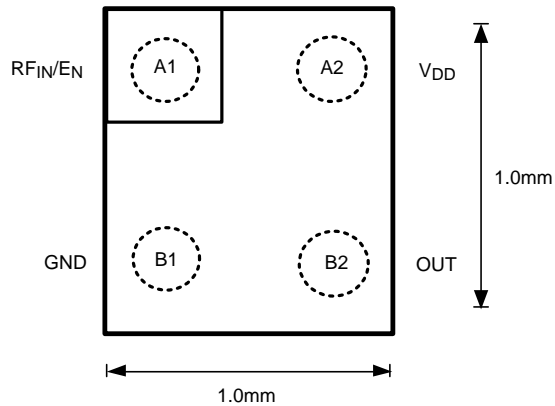
### 5.0 DC AND AC ELECTRICAL CHARACTERISTICS (continued)

Unless otherwise specified, all limits are specified to  $V_{DD} = 5.0V$ ;  $T_J = 25^\circ C$ . **Boldface** limits apply at temperature extremes. <sup>(1)</sup>

Symbol	Parameter	Condition	Min	Typ	Max	Units
$e_n$	Output Referred Noise	RF Input = 1800 MHz, -10 dBm for LMV225 and 5 dBm for LMV226/LMV228, Measured at 10 kHz		700		$nV/\sqrt{Hz}$
	Variation Due to Temperature	900 MHz, $RF_{IN} = 0$ dBm Referred to 25°C	LMV225	+0.89 -1.16		dB
		900 MHz, $RF_{IN} = 15$ dBm Referred to 25°C	LMV226	+0.25 -0.16		
			LMV228 DSBGA	+0.46 -0.62		
			LMV228 WSON	+1.39 -1.19		
		1800 MHz, $RF_{IN} = 0$ dBm Referred to 25°C	LMV225	+0.3 -0.82		
		1800 MHz, $RF_{IN} = 15$ dBm Referred to 25°C	LMV226	+0.21 -0.09		
			LMV228 DSBGA	+0.55 -0.78		
			LMV228 WSON	+1.39 -1.43		
		1900 MHz, $RF_{IN} = 0$ dBm Referred to 25°C	LMV225	+0.34 -0.63		
		1900 MHz, $RF_{IN} = 15$ dBm Referred to 25°C	LMV226	+0.21 -0.19		
			LMV228 DSBGA	+0.55 -0.93		
			LMV228 WSON	+1.54 -1.64		
		2000 MHz, $RF_{IN} = 0$ dBm Referred to 25°C	LMV225	+0.22 -0.75		
		2000 MHz $RF_{IN} = 15$ dBm Referred to 25°C	LMV226	+0.25 -0.34		
			LMV228 DSBGA	+0.61- 0.91		
	LMV228 WSON		+0.89 -0.99			



CONNECTION DIAGRAM



BUMP PITCH 500µm  
 BUMP DIAMETER 300µm  
 SOLDER DOT DIAMETER/  
 PASSIVATION OPENING 125µm

Figure 3. 4-Bump DSBGA – Top View  
 See Package Number YZR0004 or YPD0004

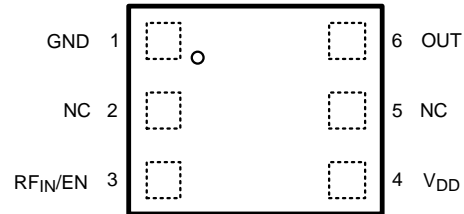
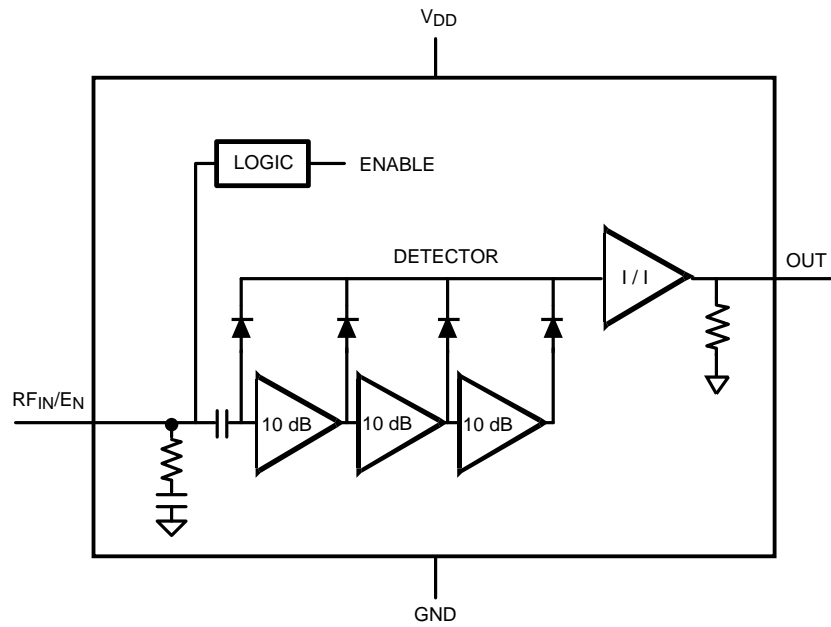


Figure 4. 6-pin WSON – Top View  
 See Package Number NGF0006A

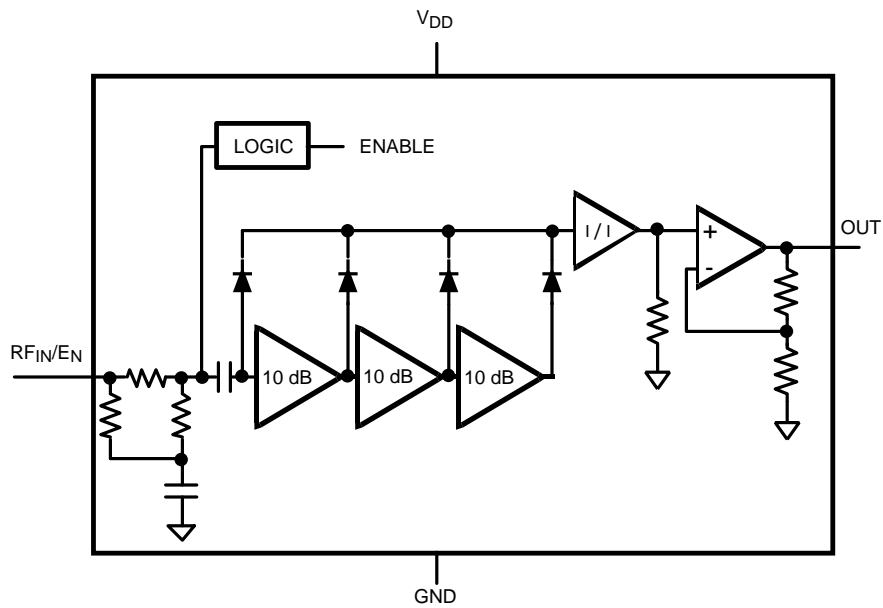
PIN DESCRIPTIONS

	Pin		Name	Description
	DSBGA	WSON6		
Power Supply	A2	4	V <sub>DD</sub>	Positive Supply Voltage
	B1	1	GND	Power Ground
	A1	3	RF <sub>IN</sub> /E <sub>N</sub>	DC voltage determines enable state of the device (HIGH = device active). AC voltage is the RF input signal to the detector (beyond 450 MHz). The RF <sub>IN</sub> /E <sub>N</sub> pin is internally terminated with 50Ω in series with 45 pF.
Output	B2	6	Out	Ground referenced detector output voltage (linear in dBm)

**Block Diagrams**



**Figure 5. LMV225**



**Figure 6. LMV226**

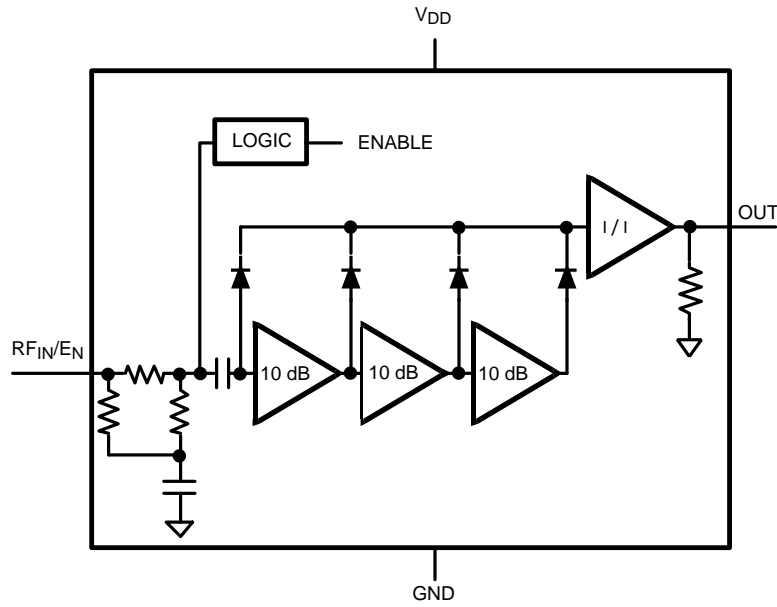
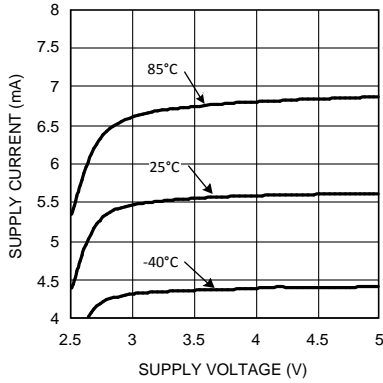


Figure 7. LMV228

### TYPICAL PERFORMANCE CHARACTERISTICS LMV225

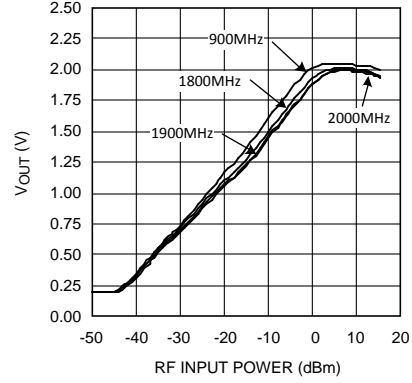
Unless otherwise specified,  $V_{DD} = 2.7V$ ,  $T_J = 25^\circ C$ .

**Supply Current vs. Supply Voltage (LMV225)**



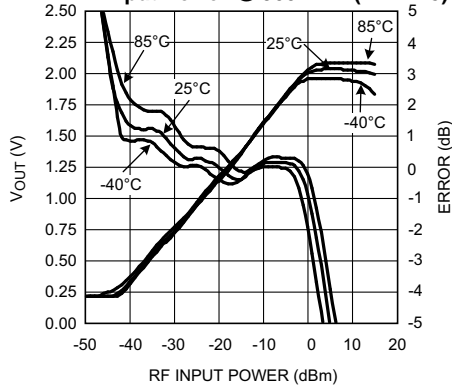
**Figure 8.**

**Output Voltage vs. RF Input Power (LMV225)**



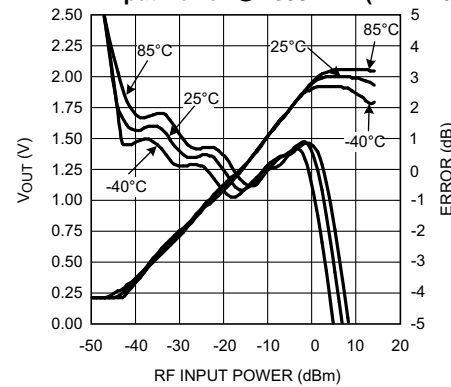
**Figure 9.**

**Output Voltage and Log Conformance vs. RF Input Power @ 900 MHz (LMV225)**



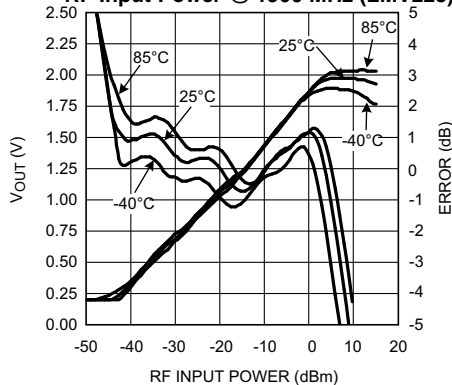
**Figure 10.**

**Output Voltage and Log Conformance vs. RF Input Power @ 1800 MHz (LMV225)**



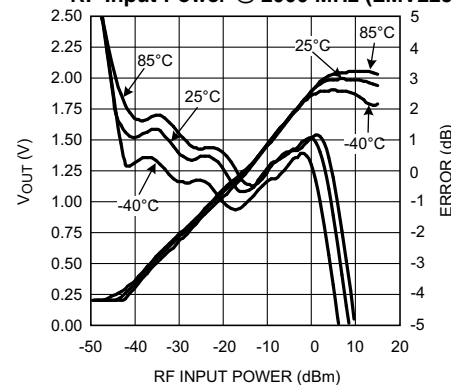
**Figure 11.**

**Output Voltage and Log Conformance vs. RF Input Power @ 1900 MHz (LMV225)**



**Figure 12.**

**Output Voltage and Log Conformance vs. RF Input Power @ 2000 MHz (LMV225)**

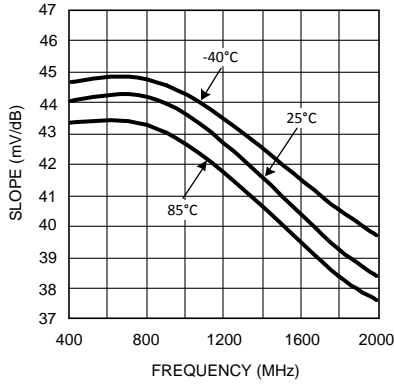


**Figure 13.**

**TYPICAL PERFORMANCE CHARACTERISTICS LMV225**  
(continued)

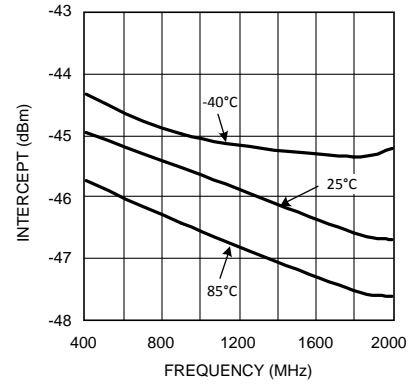
Unless otherwise specified,  $V_{DD} = 2.7V$ ,  $T_J = 25^\circ C$ .

**Logarithmic Slope vs. Frequency (LMV225)**



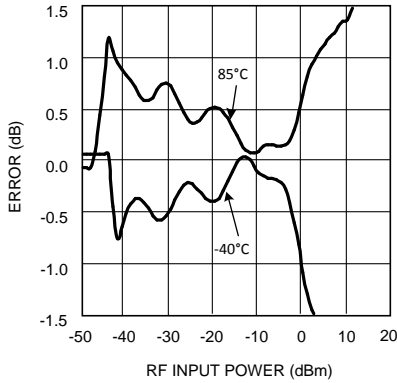
**Figure 14.**

**Logarithmic Intercept vs. Frequency (LMV225)**



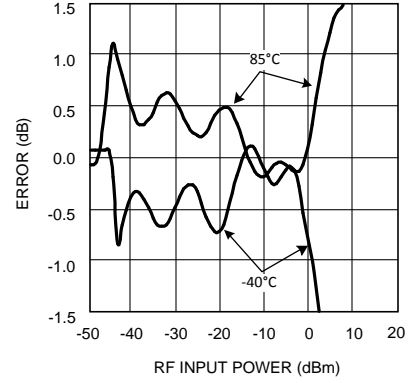
**Figure 15.**

**Output Variation vs. RF Input Power Normalized to 25°C @ 900 MHz (LMV225)**



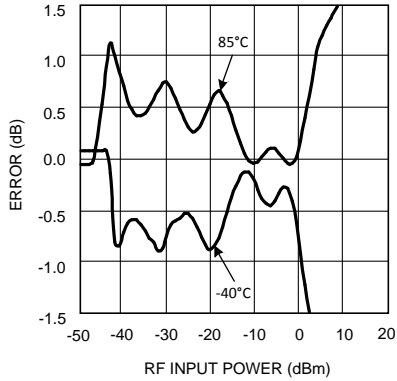
**Figure 16.**

**Output Variation vs. RF Input Power Normalized to 25°C @ 1800 MHz (LMV225)**



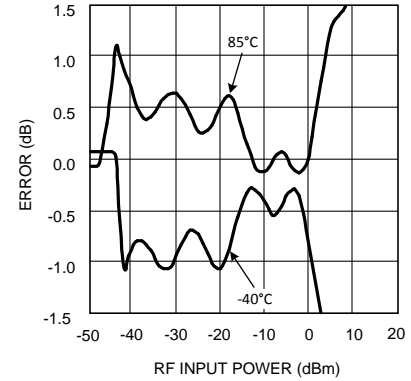
**Figure 17.**

**Output Variation vs. RF Input Power Normalized to 25°C @ 1900 MHz (LMV225)**



**Figure 18.**

**Output Variation vs. RF Input Power Normalized to 25°C @ 2000 MHz (LMV225)**



**Figure 19.**

**TYPICAL PERFORMANCE CHARACTERISTICS LMV225**  
(continued)

Unless otherwise specified,  $V_{DD} = 2.7V$ ,  $T_J = 25^\circ C$ .

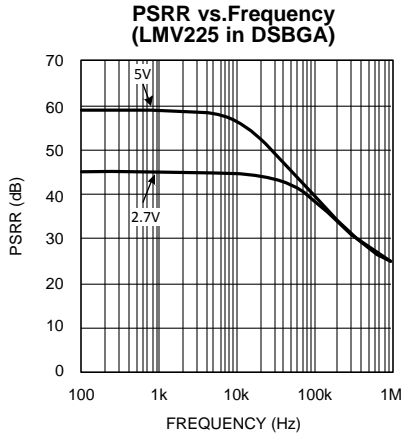


Figure 20.

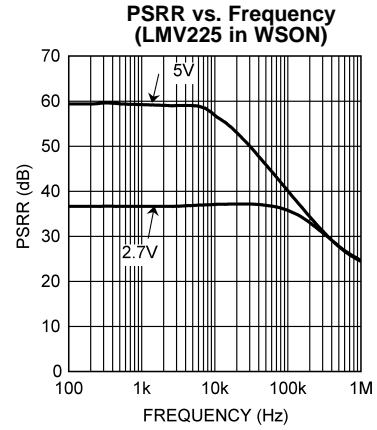


Figure 21.

**RF Input Impedance vs. Frequency @ Resistance and Reactance (LMV225 in DSBGA)**

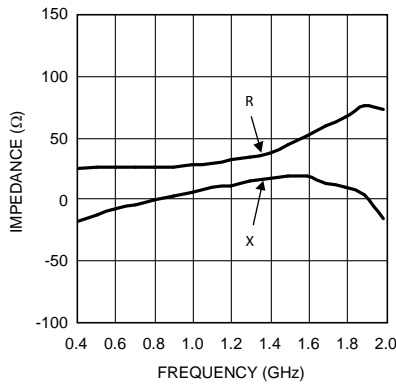


Figure 22.

**RF Input Impedance vs. Frequency @ Resistance and Reactance (LMV225 in WSON)**

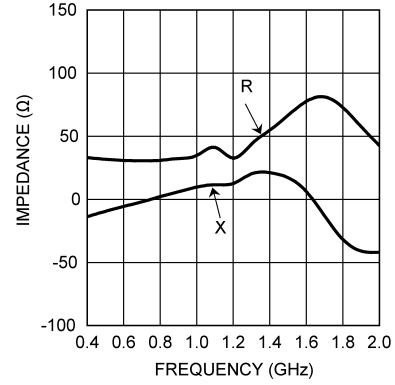


Figure 23.

**TYPICAL PERFORMANCE CHARACTERISTICS LMV226**

Unless otherwise specified,  $V_{DD} = 2.7V$ ,  $T_J = 25^\circ C$ .

**Supply Current vs. Supply Voltage (LMV226)**

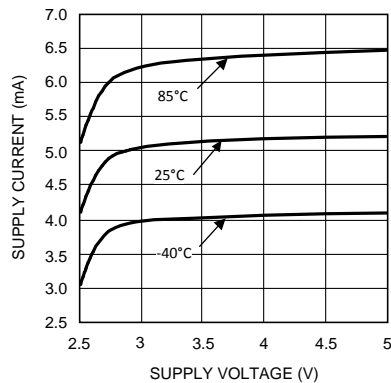


Figure 24.

**Output Voltage vs. RF Input Power (LMV226)**

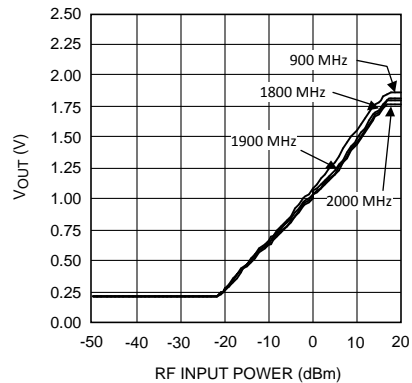


Figure 25.

**TYPICAL PERFORMANCE CHARACTERISTICS LMV226**  
(continued)

Unless otherwise specified,  $V_{DD} = 2.7V$ ,  $T_J = 25^\circ C$ .

**Output Voltage and Log Conformance vs. RF Input Power @ 900 MHz (LMV226)**

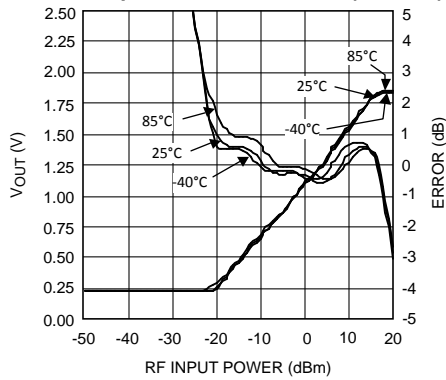


Figure 26.

**Output Voltage and Log Conformance vs. RF Input Power @ 1800 MHz (LMV226)**

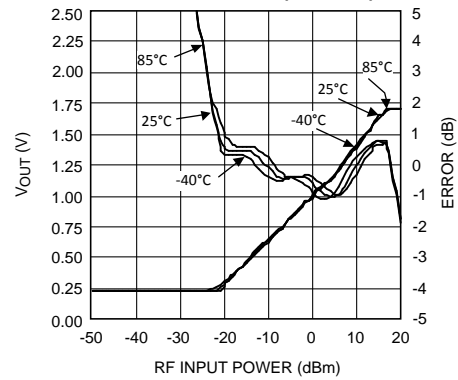


Figure 27.

**Output Voltage and Log Conformance vs. RF Input Power @ 1900 MHz (LMV226)**

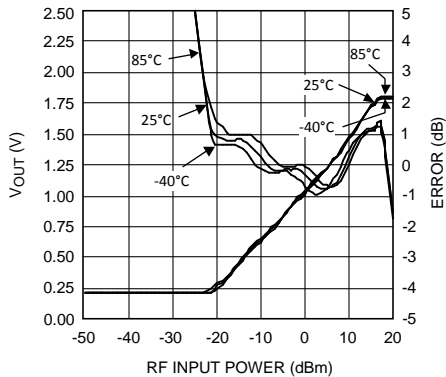


Figure 28.

**Output Voltage and Log Conformance vs. RF Input Power @ 2000 MHz (LMV226)**

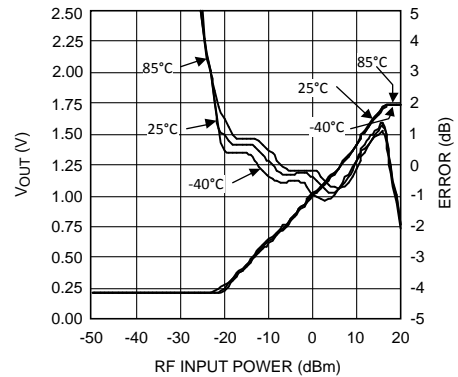


Figure 29.

**Logarithmic Slope vs. Frequency (LMV226)**

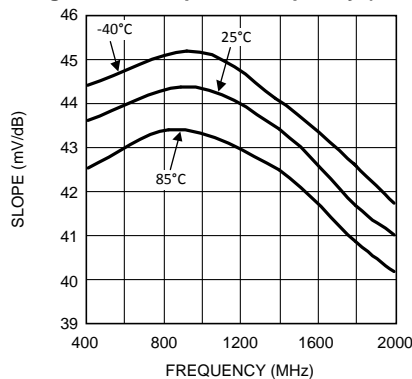


Figure 30.

**Logarithmic Intercept vs. Frequency (LMV226)**

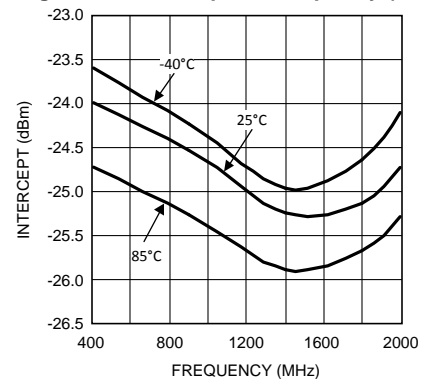
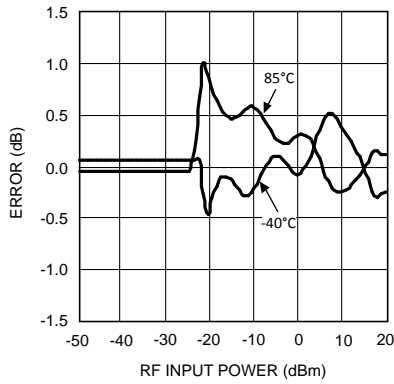


Figure 31.

**TYPICAL PERFORMANCE CHARACTERISTICS LMV226**  
(continued)

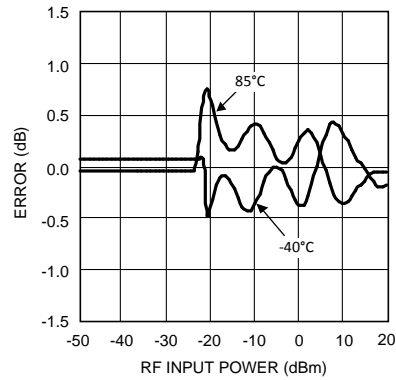
Unless otherwise specified,  $V_{DD} = 2.7V$ ,  $T_J = 25^\circ C$ .

**Output Variation vs. RF Input Power  
Normalized to 25°C @ 900 MHz (LMV226)**



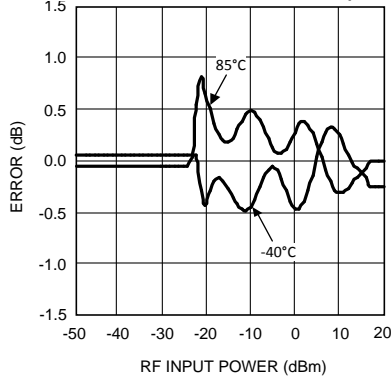
**Figure 32.**

**Output Variation vs. RF Input Power  
Normalized to 25°C @ 1800 MHz (LMV226)**



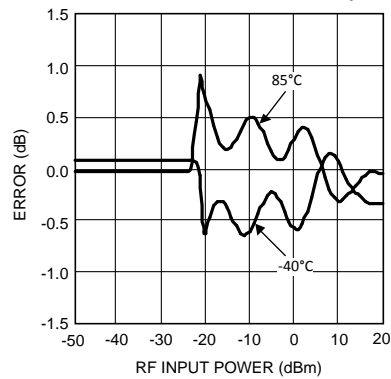
**Figure 33.**

**Output Variation vs. RF Input Power  
Normalized to 25°C @ 1900 MHz (LMV226)**



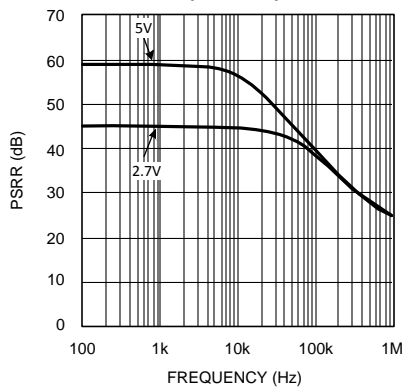
**Figure 34.**

**Output Variation vs. RF Input Power  
Normalized to 25°C @ 2000 MHz (LMV226)**



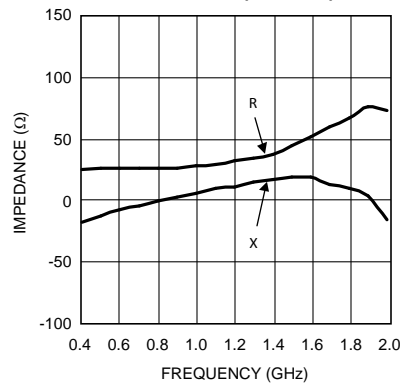
**Figure 35.**

**PSRR vs. Frequency  
(LMV226)**



**Figure 36.**

**RF Input Impedance vs. Frequency @ Resistance and Reactance (LMV226)**



**Figure 37.**



TYPICAL PERFORMANCE CHARACTERISTICS LMV228 IN DSBGA

Unless otherwise specified,  $V_{DD} = 2.7V$ ,  $T_J = 25^\circ C$ .

Supply Current vs. Supply Voltage (LMV228 in DSBGA)

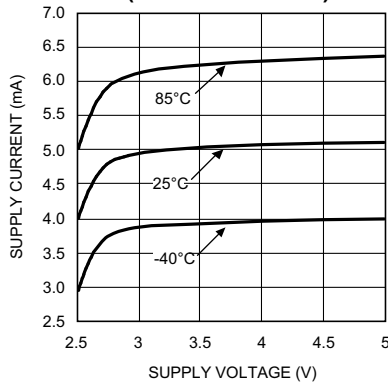


Figure 38.

Output Voltage vs. RF Input Power (LMV228 in DSBGA)

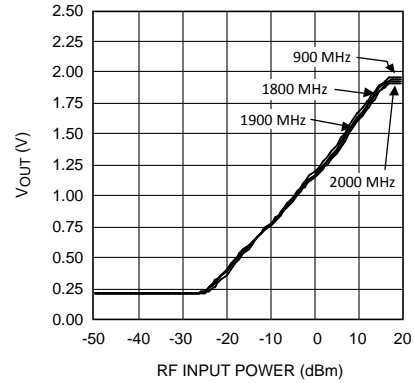


Figure 39.

Output Voltage and Log Conformance vs. RF Input Power @ 900 MHz (LMV228 in DSBGA)

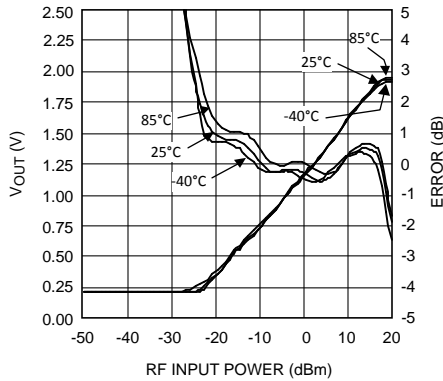


Figure 40.

Output Voltage and Log Conformance vs. RF Input Power @ 1800 MHz (LMV228 in DSBGA)

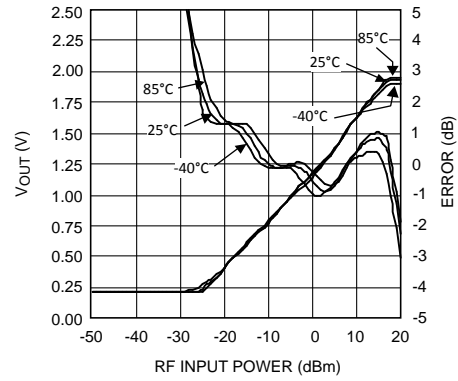


Figure 41.

Output Voltage and Log Conformance vs. RF Input Power @ 1900 MHz (LMV228 in DSBGA)

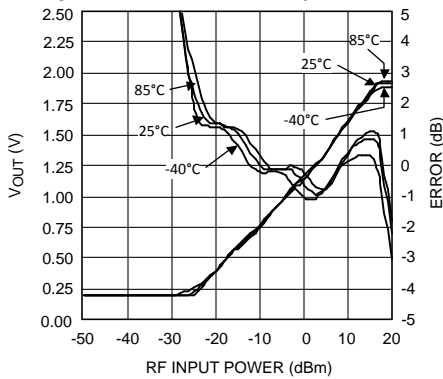


Figure 42.

Output Voltage and Log Conformance vs. RF Input Power @ 2000 MHz (LMV228 in DSBGA)

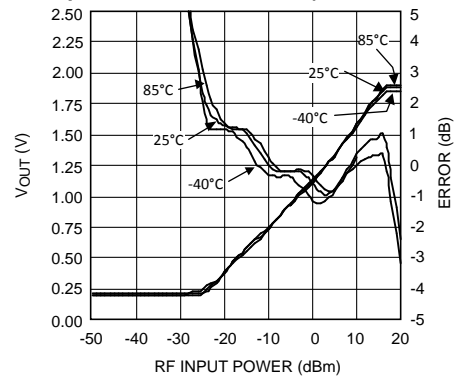


Figure 43.

## TYPICAL PERFORMANCE CHARACTERISTICS LMV228 IN DSBGA (continued)

Unless otherwise specified,  $V_{DD} = 2.7V$ ,  $T_J = 25^\circ C$ .

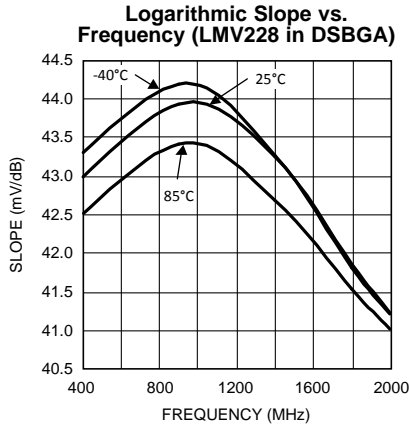


Figure 44.

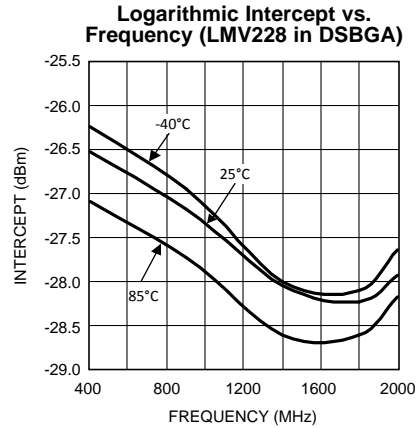


Figure 45.

**Output Variation vs.  
RF Input Power Normalized to 25°C @ 900 MHz (LMV228 in DSBGA)**

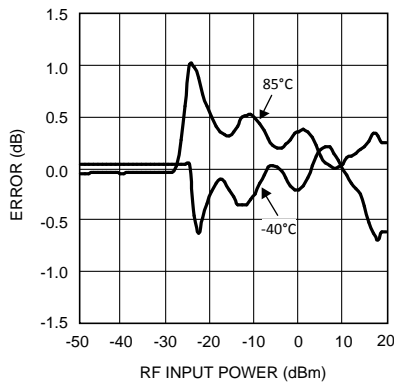


Figure 46.

**Output Variation vs.  
RF Input Power Normalized to 25°C @ 1800 MHz (LMV228 in DSBGA)**

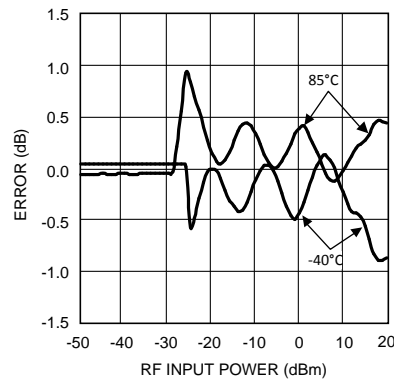


Figure 47.

**Output Variation vs.  
RF Input Power Normalized to 25°C @ 1900 MHz (LMV228 in DSBGA)**

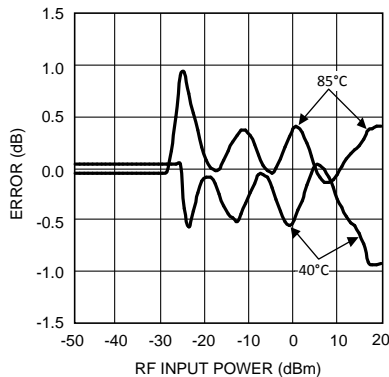


Figure 48.

**Output Variation vs.  
RF Input Power Normalized to 25°C @ 2000 MHz (LMV228 in DSBGA)**

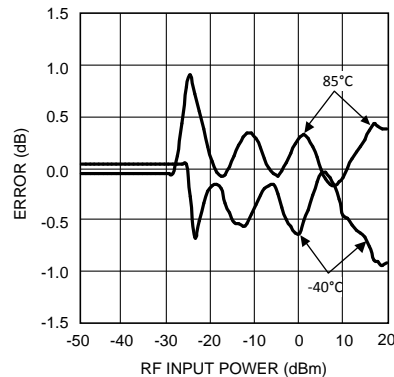


Figure 49.

**TYPICAL PERFORMANCE CHARACTERISTICS LMV228 IN DSBGA**  
(continued)

Unless otherwise specified,  $V_{DD} = 2.7V$ ,  $T_J = 25^\circ C$ .

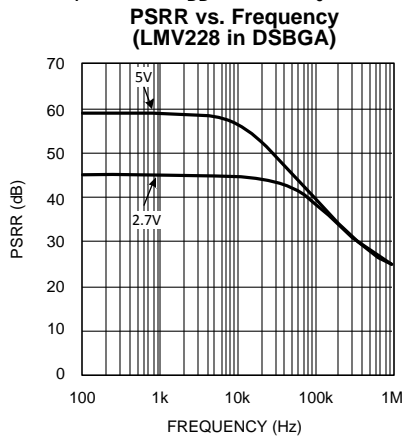


Figure 50.

**RF Input Impedance vs. Frequency @ Resistance and Reactance (LMV228 in DSBGA)**

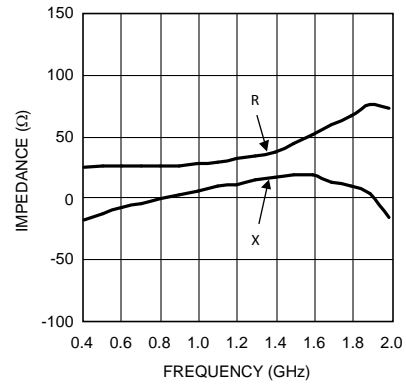


Figure 51.

**TYPICAL PERFORMANCE CHARACTERISTICS LMV228 IN WSON**

Unless otherwise specified,  $V_{DD} = 2.7V$ ,  $T_J = 25^\circ C$ .

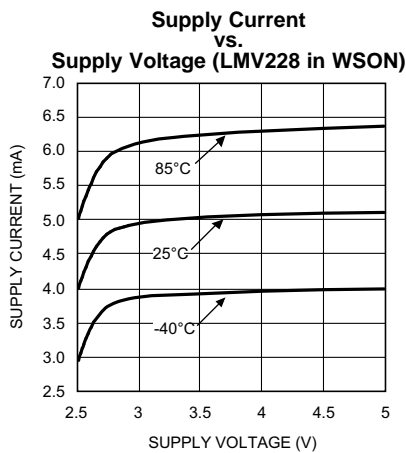


Figure 52.

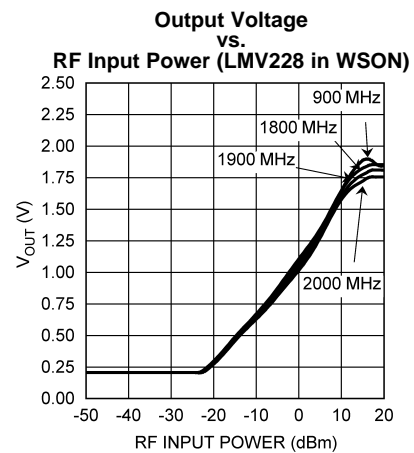


Figure 53.

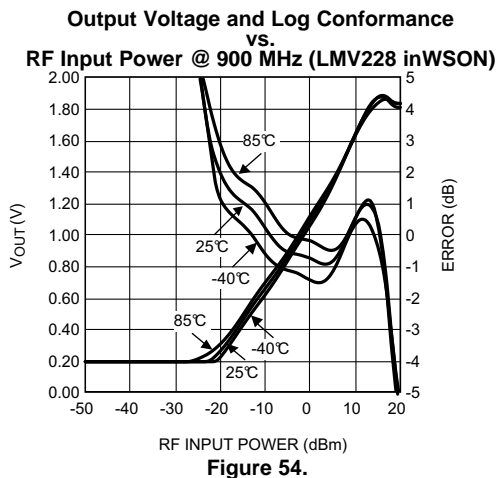


Figure 54.

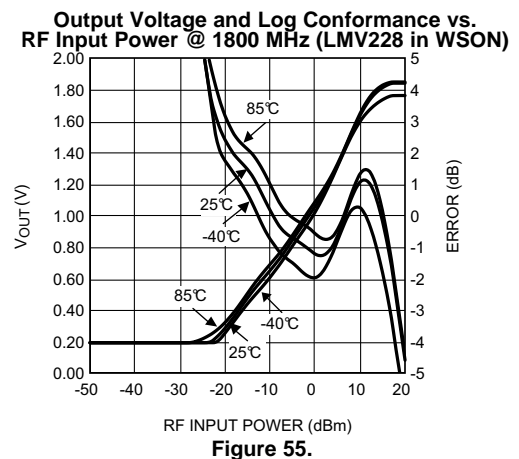


Figure 55.

**TYPICAL PERFORMANCE CHARACTERISTICS LMV228 IN WSON**  
(continued)

Unless otherwise specified,  $V_{DD} = 2.7V$ ,  $T_J = 25^\circ C$ .

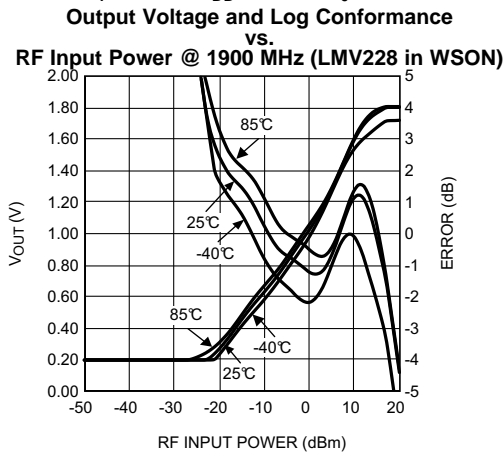


Figure 56.

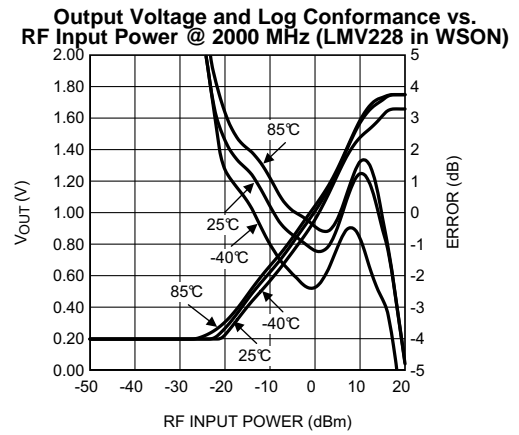


Figure 57.

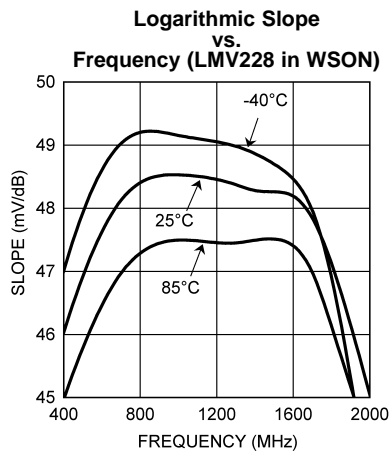


Figure 58.

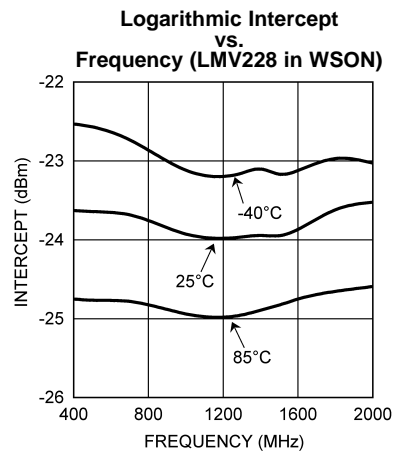


Figure 59.

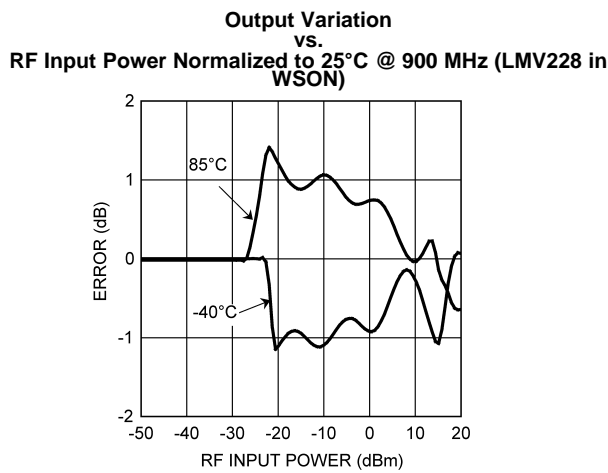


Figure 60.

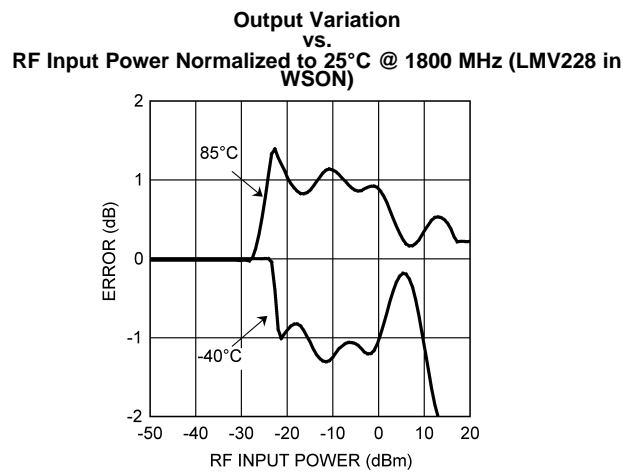


Figure 61.

**TYPICAL PERFORMANCE CHARACTERISTICS LMV228 IN WSON**  
**(continued)**

Unless otherwise specified,  $V_{DD} = 2.7V$ ,  $T_J = 25^\circ C$ .

**Output Variation vs. RF Input Power Normalized to 25°C @ 1900 MHz (LMV228 in WSON)**

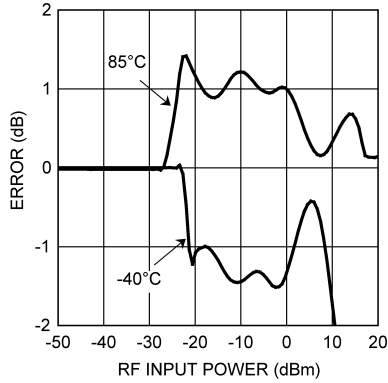


Figure 62.

**Output Variation vs. RF Input Power Normalized to 25°C @ 2000 MHz (LMV228 in WSON)**

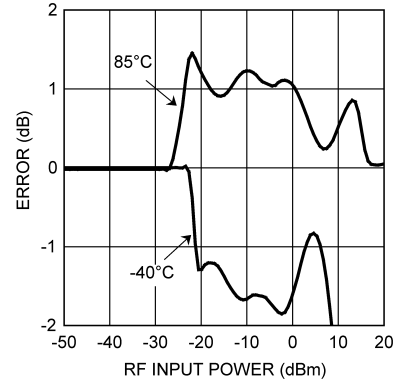


Figure 63.

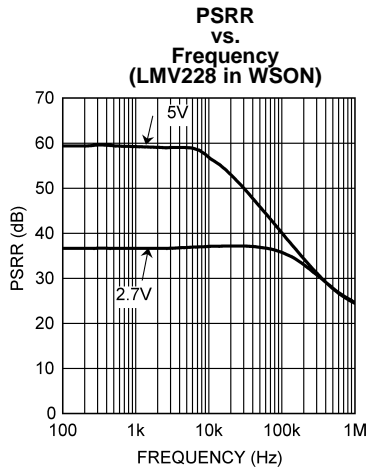


Figure 64.

**RF Input Impedance vs. Frequency @ Resistance and Reactance (LMV228 in WSON)**

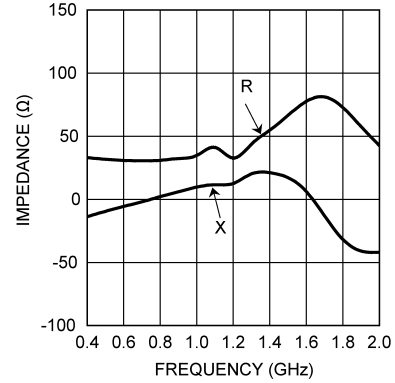


Figure 65.

## APPLICATION NOTES

### CONFIGURING A TYPICAL APPLICATION

The LMV225/LMV226/LMV228 are power detectors intended for CDMA and WCDMA applications. Power applied at its input translates to a DC voltage on the output through a linear-in-dB response. The LMV225 detector is especially suited for power measurements via a high-resistive tap, while the LMV226/LMV228 are designed to be used in combination with a directional coupler. The LMV226 has an additional output voltage buffer and therefore a low output impedance. The key features of the devices are shown in .

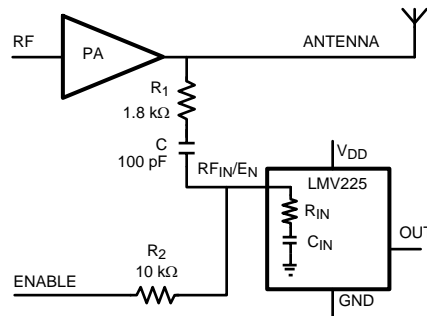
**Table 1. DEVICE CHARACTERISTICS**

	Input Range (dBm)	Output Buffer	Application
LMV225	-30 / 0	No	High Resistive Tap
LMV226	-15 / 15	Yes	Directional Coupler
LMV228	-15 / 15	No	Directional Coupler

In order to match the output power range of the power amplifier (PA) with the range of the LMV225’s input, the high resistive tap needs to be configured correctly. In case of the LMV226/LMV228 the coupling factor of the directional coupler needs to be chosen correctly.

### HIGH RESISTIVE TAP APPLICATION

The constant input impedance of the device enables the realization of a frequency independent input attenuation to adjust the LMV225’s range to the range of the PA. Resistor R<sub>1</sub> and the 50Ω input resistance (R<sub>IN</sub>) of the device realize this attenuation (Figure 66). To minimize insertion loss, resistor R<sub>1</sub> needs to be sufficiently large. The following example demonstrates how to determine the proper value for R<sub>1</sub>.



**Figure 66. Typical LMV225 Application with High Resistive Tap**

Suppose the useful output power of the PA ranges up to +31 dBm. As the LMV225 can handle input power levels up to 0 dBm. R<sub>1</sub> should realize a minimum attenuation of 31 - 0 = 31 dB. The attenuation realized by R<sub>1</sub> and the effective input resistance R<sub>IN</sub> of the detector equals:

$$A_{dB} = 20 \cdot \text{LOG} \left[ 1 + \frac{R_1}{R_{IN}} \right] = 31 \text{ dB} \tag{1}$$

Solving this expression for R<sub>1</sub>, using that R<sub>IN</sub> = 50Ω, yields:

$$R_1 = \left[ 10^{\frac{A_{dB}}{20}} - 1 \right] \cdot R_{IN} = \left[ 10^{\frac{31}{20}} - 1 \right] \cdot 50 = 1724 \Omega \tag{2}$$

In Figure 66, R<sub>1</sub> is set to 1800Ω resulting in an attenuation of 31.4 dB.

## DIRECTIONAL COUPLER APPLICATION

The LMV226/LMV228 also has a 50Ω input resistance. However, its input range differs compared to the LMV225, i.e. -15 dBm to +15 dBm. If a typical attenuation of a directional coupler is 20 dB, the LMV226/LMV228 can be directly connected via the directional coupler to the PA without the need of additional external attenuator (Figure 67). Different PA ranges can be configured using couplers with other coupling factors.

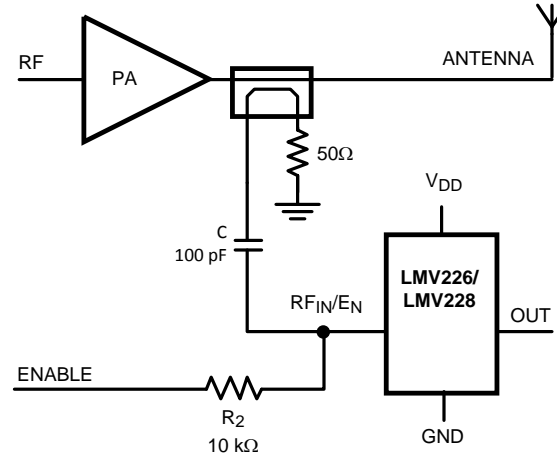


Figure 67. Typical LMV226/LMV228 Application with Directional Coupler

## SHUTDOWN FUNCTIONALITY

The LMV225/LMV226/LMV228 RF<sub>IN</sub>/E<sub>N</sub> pins have 2 functions combined:

- Enable/Shutdown
- Power input

The capacitor C and the resistor R<sub>2</sub> (Figure 66 and Figure 67) separate the DC shutdown functionality from the AC power measurement. The device is active when Enable = HI, otherwise it is in a low power consumption shutdown mode. During shutdown the output will be LOW.

Capacitor C should be chosen sufficiently large to ensure a corner frequency far below the lowest input frequency to be measured. In case of the LMV225 the corner frequency can be calculated using:

$$f = \frac{1}{2\pi(R_1 + R_{IN}) \frac{C \cdot C_{IN}}{C + C_{IN}}}$$

where

- R<sub>IN</sub> = 50Ω, C<sub>IN</sub> = 45 pF typical (3)

With R<sub>1</sub> = 1800Ω and C = 100 pF, this results in a corner frequency of 2.8 MHz. This corner frequency is an indicative number. The goal is to have a magnitude transfer, which is sufficiently flat in the used frequency range; capacitor C should be chosen significantly larger than capacitor C<sub>IN</sub> to assure a proper performance of the high resistive tap. Capacitor C shouldn't be chosen excessively large since the RC-time, it introduces in combination with resistor R<sub>2</sub>, adds to the turn-on time of the device.

The LMV226/LMV228 do not use a resistor R<sub>1</sub> like the LMV225. Though a resistor is seen on the coupler side (R<sub>COUPLER</sub>). Therefore a similar equation holds for the LMV226/LMV228 LF corner frequency, where R<sub>1</sub> is replaced with the coupler output impedance (R<sub>COUPLER</sub>).

With R<sub>COUPLER</sub> = 50Ω and C = 100 pF, the resulting corner frequency is 50 MHz.

The output voltage is proportional to the logarithm of the input power, often called "linear-in-dB". Figure 68 shows the typical output voltage versus PA output power of the LMV225 setup as depicted in Figure 66.

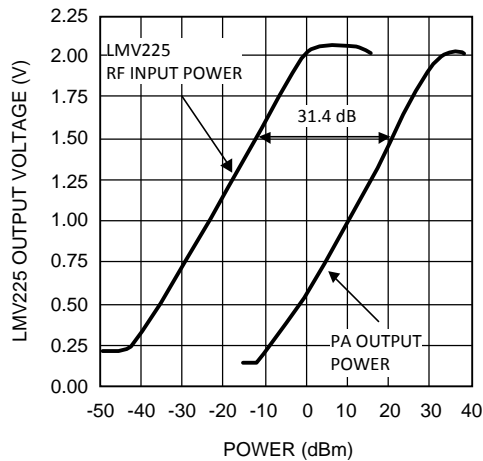


Figure 68. Typical power detector response,  $V_{OUT}$  vs. PA output Power

### OUTPUT RIPPLE DUE TO AM MODULATION

A CDMA modulated carrier wave generally contains some amplitude modulation that might disturb the RF power measurement used for controlling the PA. This section explains the relation between amplitude modulation in the RF signal and the ripple on the output of the LMV225/LMV228. Expressions are provided to estimate this ripple on the output. The ripple can be further reduced by lowpass filtering at the output. This is realized by connecting an capacitor from the output of the LMV225/LMV228 to ground.

#### Estimating Output Ripple

The CDMA modulated RF input signal of Figure 68 can be described as:

$$V_{IN}(t) = V_{IN} [1 + \mu(t)] \cos(2 \cdot \pi \cdot f \cdot t)$$

where

- $V_{IN}$  is the amplitude of the carrier frequency
- Amplitude modulation  $\mu(t)$  can be between -1 and 1

(4)

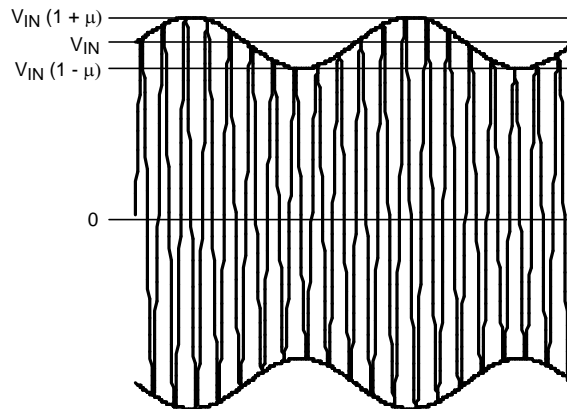


Figure 69. AM Modulated RF Signal



The ripple observed at the output of the detector equals the detectors response to the power variation at the input due to AM modulation (Figure 69). This signal has a maximum amplitude  $V_{IN} \cdot (1+\mu)$  and a minimum amplitude  $V_{IN} \cdot (1-\mu)$ , where  $1+\mu$  can be maximum 2 and  $1-\mu$  can be minimum 0. The amplitude of the ripple can be described with the formula:

$$V_{RIPPLE} = V_Y \left[ 10 \text{ LOG} \left[ \frac{V_{IN}^2 (1 + \mu)^2}{2R_{IN}} \right] + 30 \right] - V_Y \left[ 10 \text{ LOG} \left[ \frac{V_{IN}^2 (1 - \mu)^2}{2R_{IN}} \right] + 30 \right]$$

$P_{INMAX}$  IN dBm
 $P_{INMIN}$  IN dBm

where

- $V_Y$  is the slope of the detection curve (Figure 70)
  - $\mu$  is the modulation index
- (5)

Equation 5 can be reduced to:

$$V_{RIPPLE} = V_Y \cdot 20 \text{ LOG} \left[ \frac{1 + \mu}{1 - \mu} \right]$$

(6)

Consequently, the ripple is independent of the average input power of the RF input signal and only depends on the logarithmic slope  $V_Y$  and the ratio of the maximum and the minimum input signal amplitude.

For CDMA, the ratio of the maximum and the minimum input signal amplitude modulation is typically in the order of 5 to 6 dB, which is equivalent to a modulation index  $\mu$  of 0.28 to 0.33.

A further understanding of the equation above can be achieved via the knowledge that the output voltage  $V_{OUT}$  of the LMV225/LMV228 is linear in dB, or proportional to the input power  $P_{IN}$  in dBm. As discussed earlier, CDMA has a modulation in the order of 5 to 6 dB. Since the transfer is linear in dB, the output voltage  $V_{OUT}$  will vary linearly over about 5 to 6 dB in the curve (Figure 70).

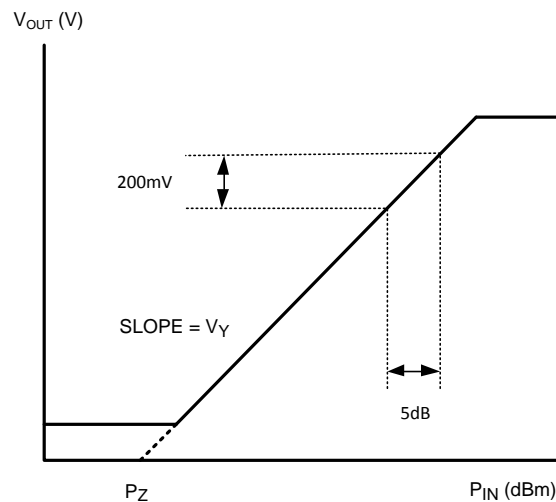


Figure 70.  $V_{OUT}$  vs. RF Input Power  $P_{IN}$

The output voltage variation  $\Delta V_{OUT}$  is thus identical for RF input signals that fall within the linear range (in dB) of the detector. In other words, the output variation is independent of the absolute RF input signal:

$$\Delta V_O = V_Y \cdot \Delta P_{IN}$$

(7)

In which  $V_Y$  is the slope of the curve. The log-conformance error is usually much smaller than the ripple due to AM modulation. In case of the LMV225/LMV228,  $V_Y = 40$  mV/dB. With  $\Delta P_{IN} = 5$  dB for CDMA,  $\Delta V_{OUT} = 200$  mV<sub>PP</sub>. This is valid for all  $V_{OUT}$ .

### Output Ripple with Additional Filtering

The calculated result above is for an unfiltered configuration. When a low pass filter is used by shunting a capacitor of e.g.  $C_{OUT} = 1.5 \text{ nF}$  at the output of the LMV225/LMV228 to ground, this ripple is further attenuated. The cut-off frequency follows from:

$$f_c = \frac{1}{2 \pi C_{OUT} R_O} \tag{8}$$

With the output resistance of the LMV225/LMV228  $R_O = 19.8 \text{ k}\Omega$  typical and  $C_{OUT} = 1.5 \text{ nF}$ , the cut-off frequency equals  $f_c = 5.36 \text{ kHz}$ . A 100 kHz AM signal then gets attenuated by  $5.36/100$  or 25.4 dB. The remaining ripple will be less than 20 mV. With a slope of 40 mV/dB this translates into an error of less than  $\pm 0.5 \text{ dB}$ . Since the LMV226 has a low output impedance buffer, a capacitor to reduce the ripple will not be effective.

### Output Ripple Measurement

Figure 71 shows the ripple reduction that can be achieved by adding additional capacitance at the output of the LMV225/LMV228. The RF signal of 900 MHz is AM modulated with a 100 kHz sinewave and a modulation index of 0.3. The RF input power is swept while the modulation index remains unchanged. Without the output capacitor the ripple is about 200 mV<sub>PP</sub>. Connecting a capacitor of 1.5 nF at the output to ground, results in a ripple of 12 mV<sub>PP</sub>. The attenuation with a 1.5 nF capacitor is then  $20 \cdot \log(200/12) = 24.4 \text{ dB}$ . This is very close to the calculated number of the previous paragraph.

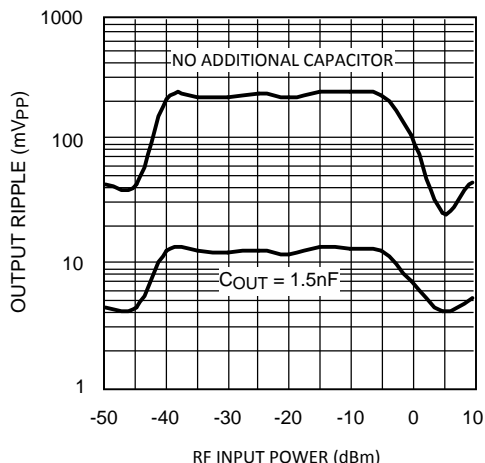


Figure 71. Output Ripple vs. RF Input Power

### PRINCIPLE OF OPERATION

The logarithmic response of the LMV225/LMV226/LMV228 is implemented by a logarithmic amplifier as shown in Figure 72. The logarithmic amplifier consists of a number of cascaded linear gain cells. With these gain cells, a piecewise approximation of the logarithmic function is constructed.

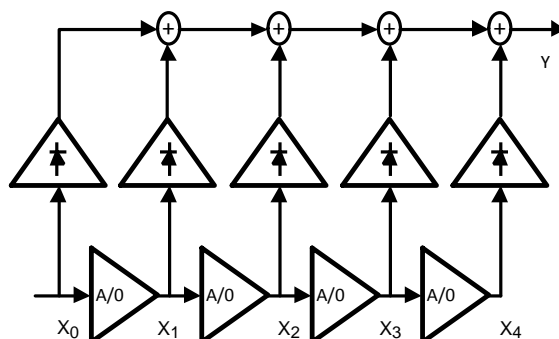


Figure 72. Logarithmic Amplifier

Every gain cell has a response according to Figure 73. At a certain threshold ( $E_K$ ), the gain cell starts to saturate, which means that the gain drops to zero. The output of gain cell 1 is connected to the input of gain cell 2 and so on.

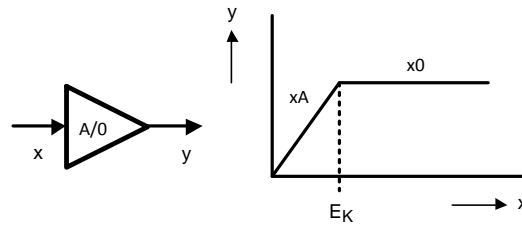


Figure 73. Gain Cell

All gain cell outputs are AM-demodulated with a peak detector and summed together. This results in a logarithmic function. The logarithmic range is about:

$$20 \cdot n \cdot \log(A)$$

where

- $n$  = number of gain cells
  - $A$  = gain per gaincell
- (9)

Figure 74 shows a logarithmic function on a linear scale and the piecewise approximation of the logarithmic function.

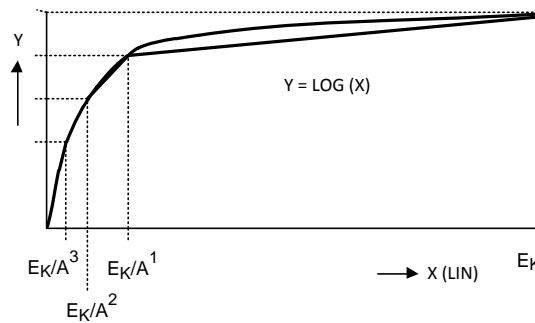


Figure 74. Log-Function on Lin Scale

Figure 75 shows a logarithmic function on a logarithmic scale and the piecewise approximation of the logarithmic function.

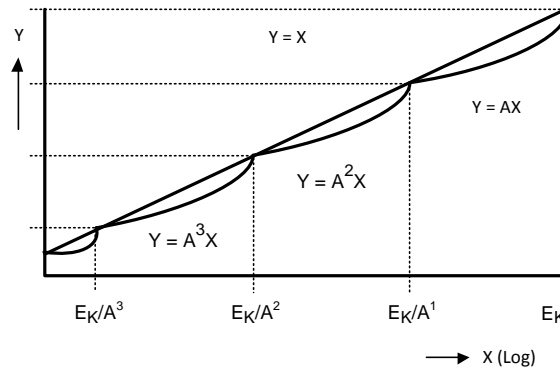


Figure 75. Log-Function on Log Scale

The maximum error for this approximation occurs at the geometric mean of a gain section, which is e.g. for the third segment:

$$\sqrt{\frac{E_K}{A^2} \cdot \frac{E_K}{A^1}} = \frac{E_K}{A\sqrt{A}} \quad (10)$$

The size of the error increases with distance between the thresholds.

## LAYOUT CONSIDERATIONS

For a proper functioning part a good board layout is necessary. Special care should be taken for the series resistance  $R_1$  ([Figure 66](#)) that determines the attenuation. For high resistor values the parasitic capacitance of the resistor may significantly impact the realized attenuation. The effective attenuation will be lower than intended. To reduce the parasitic capacitance across resistor  $R_1$ , this resistor can be composed of several components in series instead of using a single component.

---

**REVISION HISTORY**

<b>Changes from Revision K (March 2013) to Revision L</b>	<b>Page</b>
<hr/> <ul style="list-style-type: none"><li>• Changed layout of National Data Sheet to TI format .....</li></ul>	<hr/> <a href="#">28</a>

**PACKAGING INFORMATION**

Orderable Device	Status (1)	Package Type	Package Drawing	Pins	Package Qty	Eco Plan (2)	Lead finish/ Ball material (6)	MSL Peak Temp (3)	Op Temp (°C)	Device Marking (4/5)	Samples
LMV225SD/NOPB	ACTIVE	WSO	NGF	6	1000	RoHS & Green	SN	Level-1-260C-UNLIM	-40 to 85	A90	<a href="#">Samples</a>
LMV225SDX/NOPB	NRND	WSO	NGF	6	4500	RoHS & Green	SN	Level-1-260C-UNLIM	-40 to 85	A90	
LMV225TL/NOPB	ACTIVE	DSBGA	YZR	4	250	RoHS & Green	SNAGCU	Level-1-260C-UNLIM	-40 to 85		<a href="#">Samples</a>
LMV225TLX/NOPB	ACTIVE	DSBGA	YZR	4	3000	RoHS & Green	SNAGCU	Level-1-260C-UNLIM	-40 to 85		<a href="#">Samples</a>
LMV225UR/NOPB	ACTIVE	DSBGA	YPD	4	250	RoHS & Green	SNAGCU	Level-1-260C-UNLIM			<a href="#">Samples</a>
LMV225URX/NOPB	ACTIVE	DSBGA	YPD	4	3000	RoHS & Green	SNAGCU	Level-1-260C-UNLIM			<a href="#">Samples</a>
LMV226TL/NOPB	ACTIVE	DSBGA	YZR	4	250	RoHS & Green	SNAGCU	Level-1-260C-UNLIM	-40 to 85		<a href="#">Samples</a>
LMV226TLX/NOPB	ACTIVE	DSBGA	YZR	4	3000	RoHS & Green	SNAGCU	Level-1-260C-UNLIM	-40 to 85		<a href="#">Samples</a>
LMV226UR/NOPB	ACTIVE	DSBGA	YPD	4	250	RoHS & Green	SNAGCU	Level-1-260C-UNLIM			<a href="#">Samples</a>
LMV228SD/NOPB	ACTIVE	WSO	NGF	6	1000	RoHS & Green	SN	Level-1-260C-UNLIM		A89	<a href="#">Samples</a>
LMV228TL/NOPB	ACTIVE	DSBGA	YZR	4	250	RoHS & Green	SNAGCU	Level-1-260C-UNLIM	-40 to 85		<a href="#">Samples</a>
LMV228TLX/NOPB	ACTIVE	DSBGA	YZR	4	3000	RoHS & Green	SNAGCU	Level-1-260C-UNLIM	-40 to 85		<a href="#">Samples</a>

(1) The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

**LIFEBUY:** TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

**NRND:** Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW:** Device has been announced but is not in production. Samples may or may not be available.

**OBSELETE:** TI has discontinued the production of the device.

(2) **RoHS:** TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".

**RoHS Exempt:** TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

**Green:** TI defines "Green" to mean the content of Chlorine (Cl) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. - The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead finish/Ball material - Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead finish/Ball material values may wrap to two lines if the finish value exceeds the maximum column width.

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**TAPE AND REEL INFORMATION**

**QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
LMV225SD/NOPB	WSON	NGF	6	1000	178.0	12.4	2.8	2.5	1.0	8.0	12.0	Q1
LMV225SDX/NOPB	WSON	NGF	6	4500	330.0	12.4	2.8	2.5	1.0	8.0	12.0	Q1
LMV225TL/NOPB	DSBGA	YZR	4	250	178.0	8.4	1.09	1.09	0.76	4.0	8.0	Q1
LMV225TLX/NOPB	DSBGA	YZR	4	3000	178.0	8.4	1.09	1.09	0.76	4.0	8.0	Q1
LMV225UR/NOPB	DSBGA	YPD	4	250	178.0	8.4	1.04	1.04	0.56	4.0	8.0	Q1
LMV225URX/NOPB	DSBGA	YPD	4	3000	178.0	8.4	1.04	1.04	0.56	4.0	8.0	Q1
LMV226TL/NOPB	DSBGA	YZR	4	250	178.0	8.4	1.09	1.09	0.76	4.0	8.0	Q1
LMV226TLX/NOPB	DSBGA	YZR	4	3000	178.0	8.4	1.09	1.09	0.76	4.0	8.0	Q1
LMV226UR/NOPB	DSBGA	YPD	4	250	178.0	8.4	1.04	1.04	0.56	4.0	8.0	Q1
LMV228SD/NOPB	WSON	NGF	6	1000	178.0	12.4	2.8	2.5	1.0	8.0	12.0	Q1
LMV228TL/NOPB	DSBGA	YZR	4	250	178.0	8.4	1.09	1.09	0.76	4.0	8.0	Q1
LMV228TLX/NOPB	DSBGA	YZR	4	3000	178.0	8.4	1.09	1.09	0.76	4.0	8.0	Q1

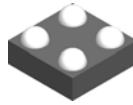


**TAPE AND REEL BOX DIMENSIONS**


\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
LMV225SD/NOPB	WSON	NGF	6	1000	208.0	191.0	35.0
LMV225SDX/NOPB	WSON	NGF	6	4500	367.0	367.0	35.0
LMV225TL/NOPB	DSBGA	YZR	4	250	208.0	191.0	35.0
LMV225TLX/NOPB	DSBGA	YZR	4	3000	208.0	191.0	35.0
LMV225UR/NOPB	DSBGA	YPD	4	250	208.0	191.0	35.0
LMV225URX/NOPB	DSBGA	YPD	4	3000	208.0	191.0	35.0
LMV226TL/NOPB	DSBGA	YZR	4	250	208.0	191.0	35.0
LMV226TLX/NOPB	DSBGA	YZR	4	3000	208.0	191.0	35.0
LMV226UR/NOPB	DSBGA	YPD	4	250	208.0	191.0	35.0
LMV228SD/NOPB	WSON	NGF	6	1000	208.0	191.0	35.0
LMV228TL/NOPB	DSBGA	YZR	4	250	208.0	191.0	35.0
LMV228TLX/NOPB	DSBGA	YZR	4	3000	208.0	191.0	35.0

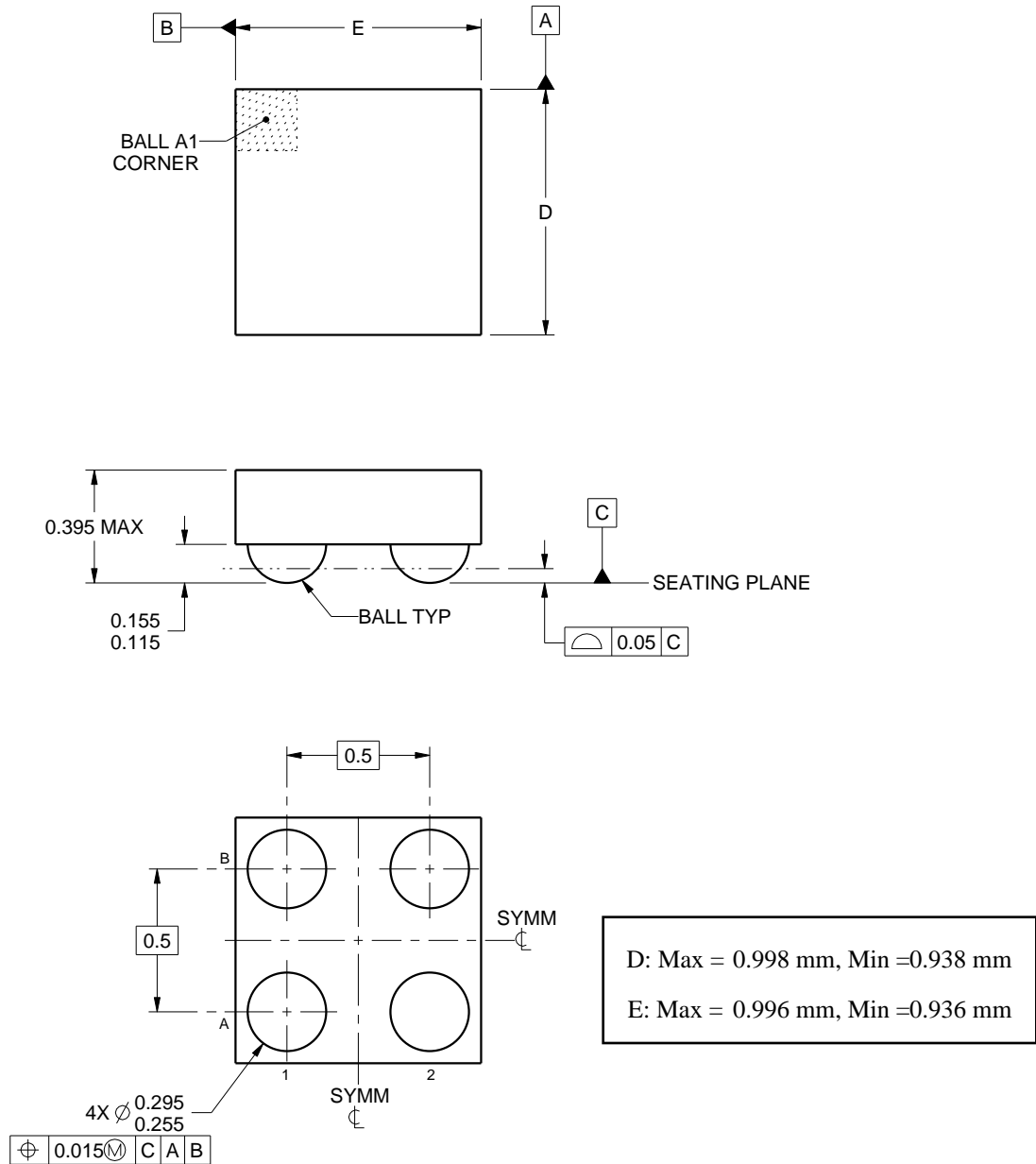
YPD0004



# PACKAGE OUTLINE

DSBGA - 0.395 mm max height

DIE SIZE BALL GRID ARRAY



4215141/B 08/2016

NOTES:

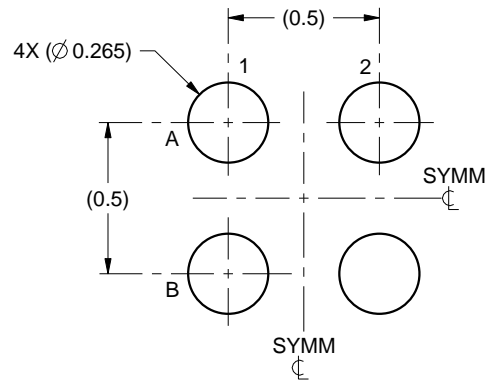
1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
2. This drawing is subject to change without notice.

# EXAMPLE BOARD LAYOUT

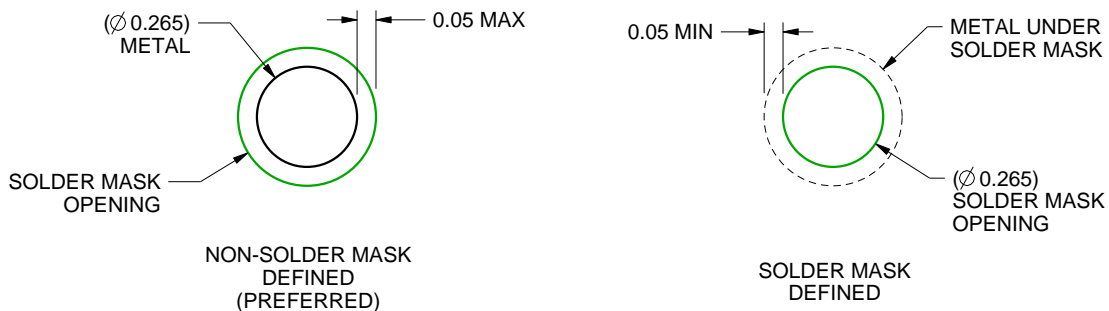
YPD0004

DSBGA - 0.395 mm max height

DIE SIZE BALL GRID ARRAY



LAND PATTERN EXAMPLE  
SCALE:40X



SOLDER MASK DETAILS  
NOT TO SCALE

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NOTES: (continued)

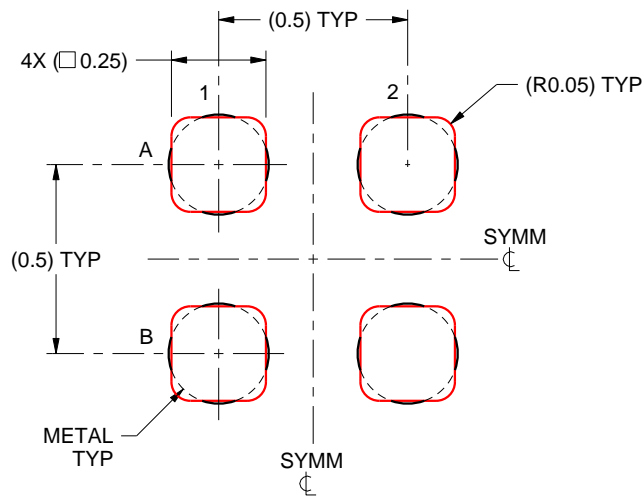
- 3. Final dimensions may vary due to manufacturing tolerance considerations and also routing constraints. See Texas Instruments Literature No. SNVA009 ([www.ti.com/lit/snva009](http://www.ti.com/lit/snva009)).

# EXAMPLE STENCIL DESIGN

YPD0004

DSBGA - 0.395 mm max height

DIE SIZE BALL GRID ARRAY



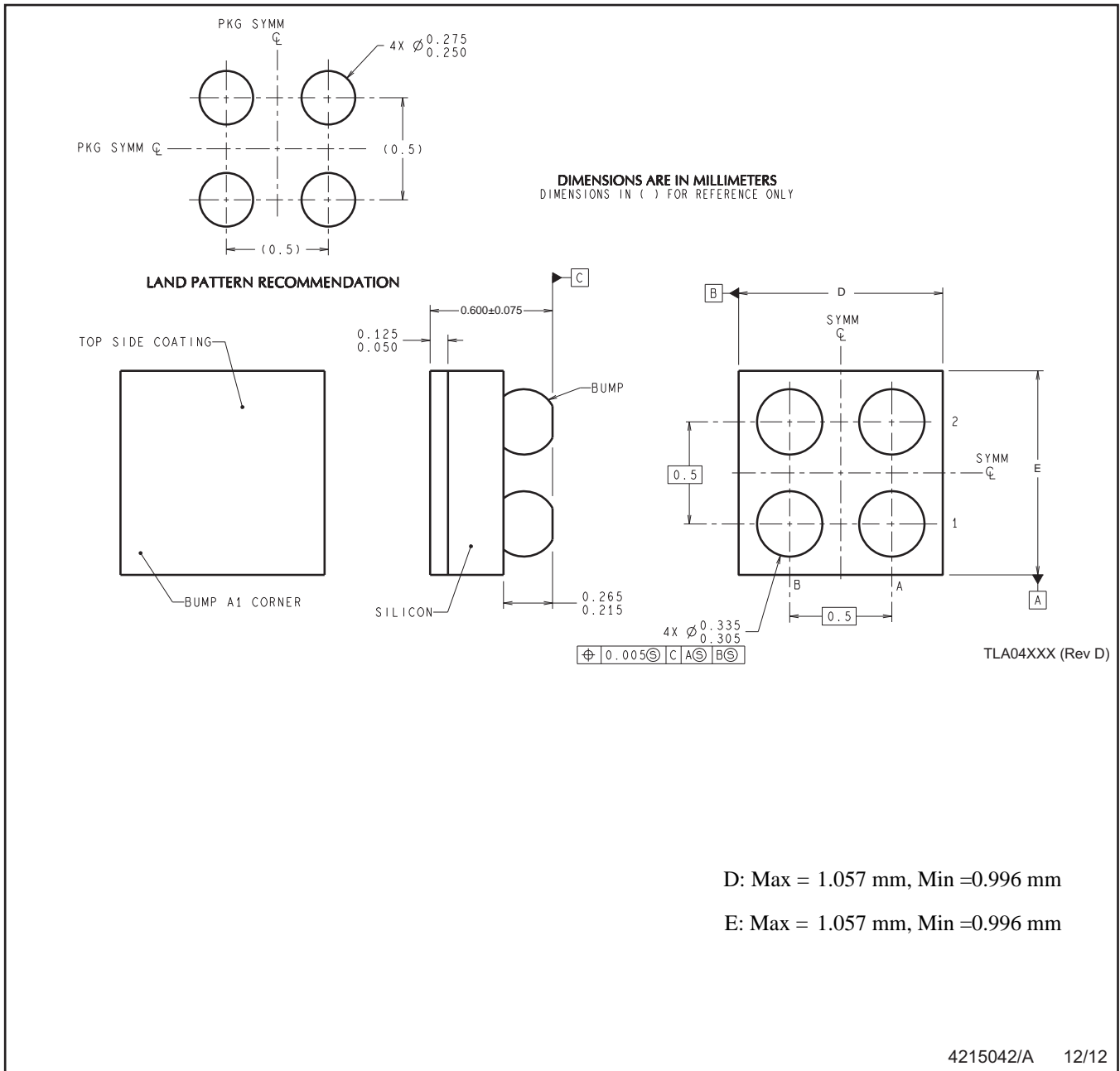
SOLDER PASTE EXAMPLE  
BASED ON 0.1 mm THICK STENCIL  
SCALE:50X

4215141/B 08/2016

NOTES: (continued)

4. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release.

YZR0004



NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.  
B. This drawing is subject to change without notice.

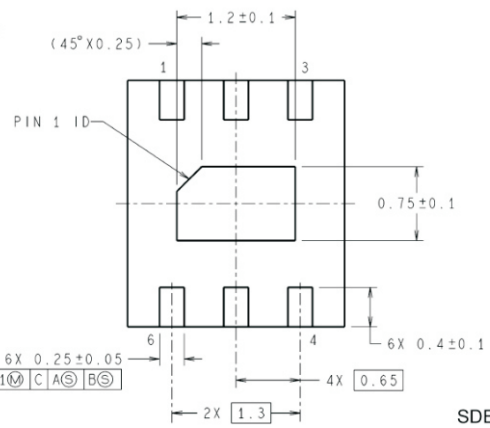
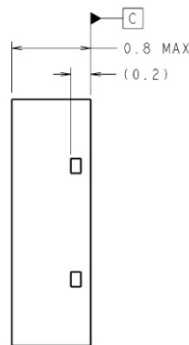
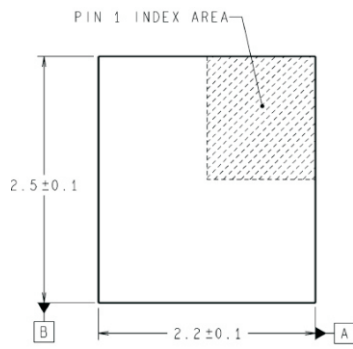
NGF0006A



DIMENSIONS ARE IN MILLIMETERS  
DIMENSIONS IN ( ) FOR REFERENCE ONLY



RECOMMENDED LAND PATTERN



SDB06A (Rev A)

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